MoDE: Mixture-of-Decision-Experts for DRL in High-Dimensional Coupled Wireless Optimization

Shiyi Lin, Hongyang Du

***Abstract*—Next-generation wireless networks require ultra reliable connectivity, extreme data rates and low latency for applications such as immersive communications and intelligent transportation. Achieving these targets is challenging because design optimization problems are tightly coupled, high-dimensional and nonconvex. Existing approaches struggle to balance computational efficiency, stability, and scalability in such settings. To address this challenge, we propose the Mixture-of-Decision-Experts (MoDE) integrated with Deep Reinforcement Learning (DRL), which effectively decomposes complex wireless control tasks into heterogeneous expert modules with adaptive routing, reducing cross-task interference and improving specialization. Evaluated on representative wireless tasks, MoDE-aided DRL outperforms conventional Multilayer Perceptron (MLP)-based DRL and heuristic baselines. Furthermore, allocating more experts to higher-dimensional decision layers yields additional efficiency and robustness over uniform-expert-allocations design.**

***Index Terms*—Wireless Optimization, DRL, Mixture-of-Experts (MoE)**

I. Introduction

The relentless growth of wireless applications, from immersive extended-reality services to intelligent transportation, has propelled evolution toward Sixth-Generation (6G) networks. These systems aim to deliver ubiquitous connectivity, extreme data rates, and low-latency services [1], whose deployments are becoming increasingly heterogeneous and ultra-dense, incorporating new components such as Reconfigurable Intelligent Surfaces (RIS) and Unmanned Aerial Vehicles (UAVs) [2]. Controlling these components introduces a optimization problem whose decision space becomes high-dimensional: it scales with key system parameters, e.g., roughly linearly with the number of scheduled users and antenna elements [3]. Furthermore, the problem is tightly coupled, as performance objectives are non-separable due to significant cross-terms arising from inter-user interference, shared resource constraints, and intricate component interactions [4]. Addressing such high-dimensional, coupled optimization is essential to unlock diversified Quality-of-Service (QoS) and superior performance metrics like Spectral Efficiency (SE) [5]. Meanwhile, practical deployments demand strict time constraints, imposing severe computational pressure and requiring algorithms that deliver high-quality joint decisions within tight latency budgets [6].

Prior work has pursued model-based and learning-based approaches. For near-field beam training, the authors in [7] propose a multi-resolution hierarchical beam-training framework that designs polar-domain codebooks and applies a coarse-to-fine search to reduce overhead in antenna positioning and beam alignment. However, the scheme depends on precomputation, iterative solves and careful initialization, which limits adaptability and real-time use in large, dynamic, heterogeneous systems. As for learning-based approaches, Deep Reinforcement Learning (DRL) avoid explicit modeling by learning policies from interaction and therefore have been applied to many nonconvex and coupled optimization in wireless network such as beam management and UAVs control [8], but they often suffer from sample inefficiency, instability, and difficulty scaling to very large action spaces. These limitations motivate the following research question:

How can we obtain data-efficient, scalable and stable policies that solve deeply coupled wireless optimization problems while avoiding interference between heterogeneous subtasks?

Architectures that decompose joint decisions into specialized subproblems are a natural remedy [9]. Mixture-of-experts (MoE) models complement this structure by allocating model capacity within each task through learnable routing, enabling specialized experts and thereby increasing representational power [10]. Recent advances in Large Language Models (LLMs) further demonstrate that dense-activation MoE variants help stabilize expert initialization and encourage richer specialization [11]. We propose the MoDE-aided DRL framework tailored for tightly-coupled wireless optimization problems, which has three design features: (i) ***Layered task decomposition*** partitions the policy into sequential stages, where the output of an earlier stage serves as context for the next, naturally reducing exploration difficulty and mitigating the burden on routing decisions by clarifying clarifying learning targets of each stage. (ii) ***Expert specialization*** [12] allows heterogeneous expert counts and internal segmentation, where high-complexity layers receive more experts, which improves representational fidelity and stabilizes learning in composite action spaces. (iii) ***Adaptive routing [13] with dense activation*** dynamically selects and combines experts based on state, which intelligently allocates computation to the most relevant experts, promoting specialization and stable training. Algorithmically, MoDE is unified and agnostic to deployment. It can run on a single device or be partitioned across platforms.

To demonstrate the practical relevance and generality of MoDE, we evaluate it on three representative classes of tightly-coupled wireless control problems. The main contributions of this paper are summarized as follows.

- We propose MoDE, a general-purpose MoE module for structured decision making in DRL. MoDE targets tightly coupled, high-dimensional wireless optimization problems by factorizing the policy into staged expert modules.
- We instantiate MoDE in an off-policy actor-critic setting by integrating it with Soft Actor-Critic (SAC), and we detail three wireless optimizers, including Pinching Antenna Systems (PASS), Multi-User Multiple-Input Multiple-Output (MU–MIMO), and RIS-aided MIMO cases.
- Extensive simulations show that MoDE improves conver-

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gence speed and final utility over MLP-based DRL and low-complexity heuristic baselines. We also demonstrate that allocating more experts to higher-dimensional layers yields additional gains.

II. Proposed MoDE-aided DRL

A. MoDE-Based Actor Network Architecture

Our two-stage MoDE jointly predicts configuration action and high-dimensional signal action. As shown in Fig. 1, the MoDE-based actor network consists of:

1) Layered MoDE with Heterogeneous Expert Allocation

Applying a single-layer mixture-of-experts actor to high-dimensional coupled wireless optimization results in unstable learning and slow convergence because of the conflicting feature sensitivities and inductive biases required for configuration and signal-level decisions. Since many wireless tasks exhibit a natural decision hierarchy, where a low-dimensional structural choice sets the operating context for subsequent high-dimensional signal optimization, we adopt a two-stage MoDE design that aligns with this structure:

- Configuration MoDE layer: Routes the state to a small set of experts to produce a structural configuration action.
- Signal MoDE layer: Takes an enriched state $\tilde{s} = [s; a_{\text{config}}]$, where a_{config} denotes the configuration action. Then we rout the enriched state $\tilde{s}$ to a larger set of experts for final signal-level optimization.

This layered design allows each stage to specialize independently. We allocate more experts to the higher-dimensional and nonlinear-mapping signal layer to increase representational capacity, while using fewer experts for the simpler configuration task to avoid unnecessary overparameterization and reduce routing overhead. Although we focus on a two-stage design of the considered wireless problems, the proposed MoDE framework can be extended to more than two stages when additional hierarchical decisions are present.

2) Fine-Grained Expert Segmentation with Shared Experts

In conventional MoE, limited experts model diverse patterns, reducing specialization and affecting performance [14]. We therefore employ ***fine-grained segmentation***, with each expert split into smaller sub-experts, allowing finer representation learning [12]. To prevent redundancy from overlapping common knowledge, we also introduce a ***dedicated shared expert*** in each layer to capture general features, freeing the segmented experts to focus on specialized patterns [15]. Each Expert Feed-Forward Network (ExpertFFN) is a two-hidden-layer network outputting the mean μ and log-standard-deviation $\log\sigma$ of a Gaussian policy:

$$\begin{cases} h_1 = \text{ReLU}(W_1 x + b_1), \\ h_2 = \text{ReLU}(W_2 h_1 + b_2), \\ \mu = W_\mu h_2 + b_\mu, \\ log\sigma = W_\sigma h_2 + b_\sigma, \end{cases} \tag{1}$$

where x is the input vector, parameters $W_1, W_2, W_\mu, W_\sigma$ are trainable weight matrices, and $b_1, b_2, b_\mu, b_\sigma$ are their associated bias vectors. Hence, fine-grained experts outputs are

$$\begin{cases} (\mu_i^{\text{config}}, \log\sigma_i^{\text{config}}) = E_i^{\text{config}}(s), & i = 1 \dots n, \\ (\mu_j^{\text{signal}}, \log\sigma_j^{\text{signal}}) = E_j^{\text{signal}}(\tilde{s}), & j = 1 \dots m, \end{cases} \tag{2}$$

where n denote the segmentation factors for the configuration experts and m for signal experts. Each MoDE layer employs a lightweight router computing normalized gating weights $\mathbf{w} = \frac{\exp((R\mathbf{x}+\mathbf{r})_i)}{\sum_{j=1}^{E}\exp((R\mathbf{x}+\mathbf{r})_j)}$, where x is the layer input, R and $\mathbf{r}$ are router parameters, E is the number of fine-grained experts in that layer. Using the per-sample weights w_i, the mixed distribution parameters for configuration layer are computed by adding the shared expert's output with a weighted sum of fine experts:

$$\begin{cases} \mu^{\text{config}}(s) = \mu_{\text{shared}}^{\text{config}}(s) + \sum_{i=1}^{n_{\text{fine}}} w_i^{\text{config}}(s)\mu_i^{\text{config}}(s), \\ \log\sigma^{\text{config}}(s) = \log\sigma_{\text{shared}}^{\text{config}}(s) + \sum_{i=1}^{n_{\text{fine}}} w_i^{\text{config}}(s)\log\sigma_i^{\text{config}}(s). \end{cases} \tag{3}$$

We then sample the intermediate configuration action as $a_{\textbf{config}} = \tanh\left(\mu^{\text{config}}(s) + \sigma^{\text{config}}(s)\epsilon\right), \quad \epsilon \sim \mathcal{N}(0, I)$, and similarly for the signal layer using enriched state $\tilde{s}$. The final action vector a is the concatenation of configuration and signal actions as $a = \text{scale} \cdot [a_{\textbf{config}}\ ;\ a_{\textbf{signal}}] + \text{bias}$.

3) Load Balance Consideration

Automatically learned routers may suffer from routing collapse by few experts monopolizing most samples and persistent load imbalance by unequal expert utilization, both of which reduce effective capacity and slow convergence [14]. We therefore add two lightweight regularizers as follows. A ***diversity loss*** encourages high entropy in the per-sample routing distribution to mitigate routing collapse:

$$\begin{cases} \bar{H}_{\text{config}} = \frac{1}{B}\sum_{b=1}^{B}\left(-\sum_{i=1}^{E_{\text{config}}} w_{b,i}^{\text{config}} \log w_{b,i}^{\text{config}}\right), \\ \bar{H}_{\text{signal}} = \frac{1}{B}\sum_{b=1}^{B}\left(-\sum_{i=1}^{E_{\text{signal}}} w_{b,i}^{\text{signal}} \log w_{b,i}^{\text{signal}}\right), \\ \mathcal{L}_{\text{diversity}} = -\left(\bar{H}_{\text{config}} + \bar{H}_{\text{signal}}\right), \end{cases} \tag{4}$$

where $\bar{H}$ represents mean entropy, B is batch size, E is the number of experts, and $w_{b,i}$ is normalized routing weight of each layer in MoDE for sample b and expert i. This term is minimized when the router assigns weights more evenly within each sample, reducing the chance of collapse to one expert. A ***balance loss*** promotes equal expert utilization by regulating the long-term average usage of experts across a batch:

$$\begin{cases} \mu_i^{\text{config}} = \frac{1}{B}\sum_{b=1}^{B} w_{b,i}^{\text{config}}, \\ \mu_i^{\text{signal}} = \frac{1}{B}\sum_{b=1}^{B} w_{b,i}^{\text{signal}}, \\ \mathcal{L}_{\text{balance}} = \frac{1}{E}\sum_{i=1}^{E_{\text{config}}}\left(\mu_i^{\text{config}}\right)^2 + \frac{1}{E}\sum_{i=1}^{E_{\text{signal}}}\left(\mu_i^{\text{signal}}\right)^2. \end{cases} \tag{5}$$

This loss reaches its minimum when all experts are equally utilized on average, i.e., $\mu_i = \frac{1}{E}$. The total expert loss is

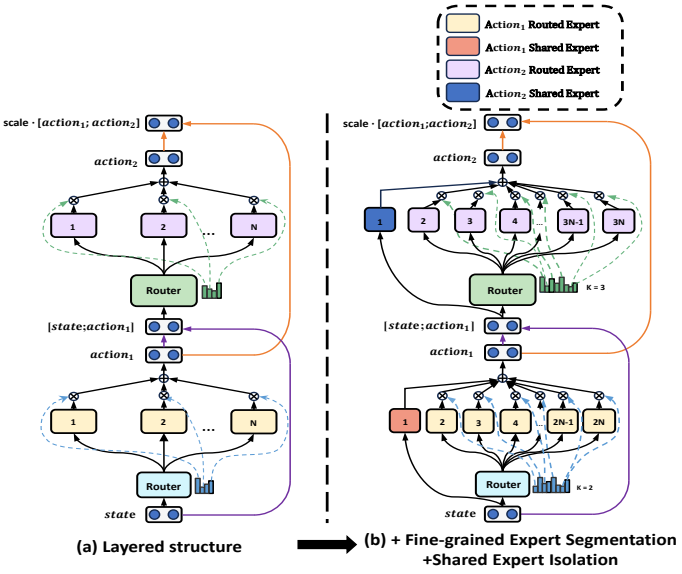


Fig. 1: Illustration of general MoDE. Subfigure (a) shows the layered MoDE structure. Subfigure (b) shows the fine-grained expert segmentation and the shared expert isolation strategy.

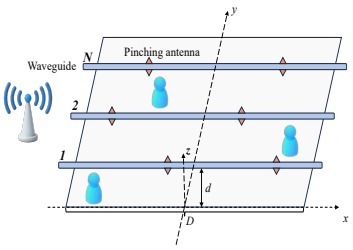
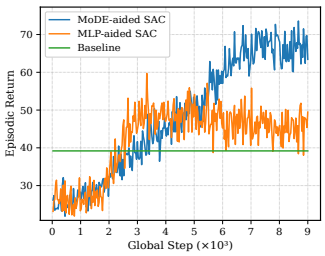
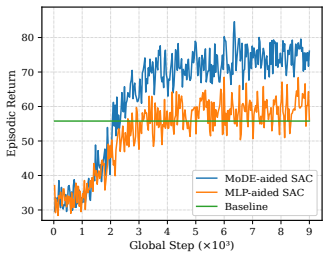


Fig. 2: PASS system model



(a) N = 2, M = 2, K = 4



(b) N = 7, M = 3, K = 13

Fig. 3: Average return versus training steps in PASS.

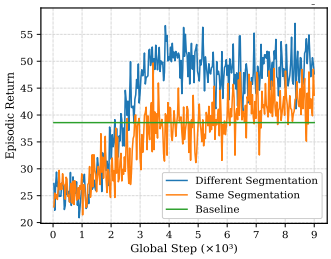


Fig. 4: MoDE with different experts-configuration.

$\mathcal{L}_{\text{expert}} = \mathcal{L}_{\text{diversity}} + \lambda \mathcal{L}_{\text{balance}}$, with $\lambda = 10^{-3}$ in our experiments to prevent collapse while encouraging global balance.

B. DRL Feedback Framework

The MoDE-based actor network serves as a structured, modular policy module yielding $\pi_\theta(s)$, which can be integrated with a variety of DRL algorithms for continuous control. We adopt SAC for its entropy regularization and stability. The framework employs two independent critics network Q_ϕ, Q_ψ to estimate state-action values. Training follows the off-policy loop: at each step, the actor samples an action a from $\pi_\theta(s)$, the environment returns a user-centric reward r, and the transitions (s, a, r, s') is stored in a replay buffer for subsequent updates of both actor and critics. Through iterative training across episodes, the policy progressively refines its decision-making capacity, converging toward a robust and effective controller for complex wireless optimization tasks.

C. Overall Complexity

We analyze the parameter complexity of the proposed MoDE-based actor relative to a standard two-layer MLP baseline. Let O be the state dimension, P and B the configuration and signal action dimensions, and action $A = P+B$. For MoDE, let H be the shared hidden dimension of an expert, a the base number of experts, and b, c the segmentation factors for signal and configuration layers respectively. This yields configuration experts $K_p = a \cdot c$ and signal experts $K_b = a \cdot b$, each with a reduced hidden dimension of $H/2$. As each ExpertFFN consists of two dense internal linear layers plus two output heads, with input I, hidden width H and output O, the parameter count is $\mathrm{P}_{\text{expert}}(I, H, O) = (I \cdot H + H) + (H \cdot H + H) + (H \cdot O + O) + (H \cdot O + O)$. Using this building block, the total parameters for MoDE are the sum of its components, $\mathrm{P}_{\text{MoDE}} = K_p \cdot \mathrm{P}_{\text{expert}}(O, H_{e,p}, P) + K_b \cdot \mathrm{P}_{\text{expert}}(O + P, H_{e,b}, B) + \mathrm{P}_{\text{shared\_pos}}(O, H, P) + \mathrm{P}_{\text{shared\_bf}}(O + P, H, B) + \mathrm{P}_{\text{routers}}$, where each router is a simple linear layer. The dominant terms are quadratic in H expressed as $\mathrm{P}_{\text{MoDE}} = O\left(a\left(OH + H(P+B) + H^2\left(\frac{1}{b} + \frac{1}{c}\right)\right) + H^2\right)$. For the two-layer MLP actor with hidden width H_{MLP} plus two output heads, we have $\mathrm{P}_{\text{MLP}} = O(H_{\text{MLP}}^2 + H_{\text{MLP}}(O+A))$. Retaining only the leading quadratic terms, the parameter ratio approximates as $\frac{\mathrm{P}_{\text{MoDE}}}{\mathrm{P}_{\text{MLP}}} \approx \frac{C_{\text{MoDE}} H^2}{H_{\text{MLP}}^2}$, with $C_{\text{MoDE}} = a\left(\frac{1}{b} + \frac{1}{c}\right) + O\left(\frac{O}{H}, \frac{A}{H}\right)$. Thus, the relative overhead depends primarily on (i) the base expert count a, the segmentation factors b, c and (ii) the choice of the MoDE shared hidden width H and the MLP hidden width H_{MLP}. In practice, MoDE intentionally increase representational capacity in a controlled way, enhancing representational power without compromising convergence.

III. Cases

In this section, we evaluate the MoDE-aided SAC framework on three tightly coupled, high-dimensional wireless optimization, i.e., PASS, MU-MIMO, and RIS-aided MIMO.

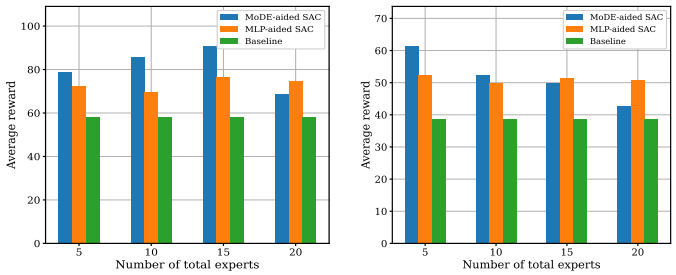
A. Joint Pinching ositions and Beamforming in PASS

1) MoDE-Aided SAC In PASS System Model

We consider a downlink multi-user PASS as Fig. 2 with N waveguides at Base Station (BS), M pinching antennas per waveguide and K users deployed in a $100 \times 100\,\mathrm{m}^2$ region. The waveguide height is $d = 3\,\mathrm{m}$, and the carrier frequency is $f_c = 28\,\mathrm{GHz}$. For an antenna at location $\psi_{m,n}^p$ and user at ψ_k, the channel is $h_{n,m,k} = \frac{\sqrt{\eta}\exp\left(-j\frac{2\pi}{\lambda}\|\psi_k - \psi_{m,n}^p\|\right)}{\|\psi_k - \psi_{m,n}^p\|}$, where $\lambda = \frac{c}{f_c}$ is the free-space wavelength, and $\eta = \frac{c}{2\pi f_c}$ is the propagation constant. The per-waveguide feed-induced phases are aggregated into $\mathbf{G} = \mathrm{blockdiag}(\mathbf{g}_1, \ldots, \mathbf{g}_N)$. With precoder $\mathbf{W} = [\mathbf{w}_1, \ldots, \mathbf{w}_K]$ and transmitted symbols x_k, the received signal is $y_k = \mathbf{h}_k^H \mathbf{G}\,\mathbf{w}_k\,x_k + \sum_{j \neq k} \mathbf{h}_k^H \mathbf{G}\,\mathbf{w}_j\,x_j + n_k$, where $n_k \sim \mathcal{CN}(0, \sigma_0^2)$ is Gaussian noise, yielding $\mathrm{SINR}_k(\Phi^p, \mathbf{W}) = \frac{|\mathbf{h}_k^H \mathbf{G} \mathbf{w}_k|^2}{\sum_{j \neq k} |\mathbf{h}_k^H \mathbf{G} \mathbf{w}_j|^2 + \sigma_0^2}$. Our objective is to maximize the sum SE defined as $\mathrm{SE}_t = \sum_{k=1}^K \log_2(1 + \mathrm{SINR}_k)$ across steps. Given the equivalent channel $\mathbf{H} = \mathbf{H}\mathbf{G}$, the precoder is $\mathbf{W} = \widetilde{\mathbf{H}}^{\mathsf{H}}\left(\Lambda \widetilde{\mathbf{H}}\widetilde{\mathbf{H}}^{\mathsf{H}} + \sigma^2 \mathbf{I}_K\right)^{-1} \mathbf{P}^{1/2}$, with $\Lambda = \mathrm{diag}(\boldsymbol{\lambda})$ and $\mathbf{P}^{1/2} = \mathrm{diag}(\sqrt{p_1}, \ldots, \sqrt{p_K})$, where per-user positive uplink factor vector $\boldsymbol{\lambda} = [\lambda_1, \ldots, \lambda_K]^{\mathsf{T}}$ and normalized downlink power fraction vector $\mathbf{p} = [p_1, \ldots, p_K]^{\mathsf{T}}$ are decoded by the beamforming latent z_w via lightweight neural network heads, ensuring $\lambda_k > 0$ and $\sum_k p_k = 1$. Finally $\mathbf{W}$ is normalized to satisfy the total transmit-power constraint. The coupling in PASS stems from the mutual interference in the SINR denominators, which jointly depends on both the pinching positions Φ^p and the beamforming matrix $\mathbf{W}$, resulting in a non-separable objective. To address this, MoDE adopts a two-stage hierarchy: the configuration layer first outputs the pinching-antenna positions, after which the signal layer produces the beamforming latent vector z_w. As for SAC, the state s concatenates user coordinates $(x_1, y_1 \ldots, x_k, y_k)$ of length $2K$, current antenna positions of length NM, and the channel $\mathbf{H}$ represented by real/imag parts flattened to length $2KMN$. The continuous action a comprises normalized first-antenna positions per waveguide of length N, relative increments for the remaining antennas Δ of length $N(M-1)$, and the beamforming latent z_w, whose dimension scales softly with the number of users and antennas. The reward is $r_t = \alpha\left(\mathrm{SE}_t + \beta(\mathrm{SE}_t - \mathrm{SE}_{t-1})\right)$, where $\alpha > 0$ balances the reward scale and β is a small improvement bonus.

2) Experimental Results

We compare the MoDE-aided SAC agent against an MLP-aided SAC baseline and a heuristic baseline with nearest-user antenna placement and Zero-Forcing precoding. Fig. 3 plots the average episodic return versus training iterations under two



(a) N = 5, M = 3, K = 10 (b) N = 3, M = 2, K = 6

Fig. 5: Impact of total expert count on late-stage episodic return under different environment complexities.

environment configurations: (a) $N = 2, M = 2, K = 4$ and (b) $N = 7, M = 3, K = 13$. MoDE-aided SAC attains higher final return than MLP-aided SAC, with both significantly outperforming the heuristic baseline after sufficient training.

Fig. 4 shows the impact of expert allocation between position and beamforming layers under $N = 3, M = 2, K = 6$. We fix the basic expert count to 3 and compare two configurations: Configuration A uses segmentation factor 1 for position layer and factor 4 for beamforming layer. Configuration B uses segmentation factor 2 for both layers. The results show that allocating more experts to the beamforming layer produces higher converged episodic return than the symmetric allocation, which itself improves mildly over the heuristic baseline.

Fig. 5 summarizes late-training average against the total number of experts for a lower- and higher-complexity environments. The ratio of the position-segmentation factor to the beamforming-segmentation factor keeps 2:3 in training. Performance peaks at a moderate expert count for the setting with $N = 5$, $M = 3$, $K = 10$, indicating a trade-off where additional experts initially provide useful specialization but eventually incur higher routing complexity and reduced per-expert sample efficiency. For the simpler setting with $N = 3$, $M = 2$, $K = 6$, performance declines with more experts, suggesting overparameterization when task demands are modest.

Fig. 6 visualizes router gating activations produced by MoDE-based actor network under varying user positions and SNR conditions in environment $N = 3$, $M = 2$, $K = 6$ with base expert count of 2, position segmentation of 2, and beamforming segmentation of 3. Early in training, routing is concentrated on a few experts, e.g., 'pos3'≈ 0.287 and 'bf5'≈ 0.236, reflecting undeveloped specialization. Post-training, the router exhibits context-aware specialization: position experts provide broadly useful spatial encodings with more uniform activation, while beamforming experts develop distinct fingerprints correlated with specific SNR regimes or user layouts, e.g., user-position set 2 relies more on 'bf5' and 'bf6' after training, whereas set 4 shows near-uniform beamforming allocation. This demonstrates the adaptive, interpretable task decomposition and functional specialization of MoDE.

B. Joint Power Allocation and Beamforming in MU-MIMO

1) MoDE-Aided SAC In MU-MIMO System Model

We consider a downlink MU-MIMO setup depicted in Fig. 7, where a BS with N antennas serves K users distributed in a $D \times D\text{m}^2$ area. Operating at $f_c = 28$ GHz with a pathloss exponent of 2.5, the aggregate channel matrix follows a deterministic LoS model comprising large-scale pathloss,

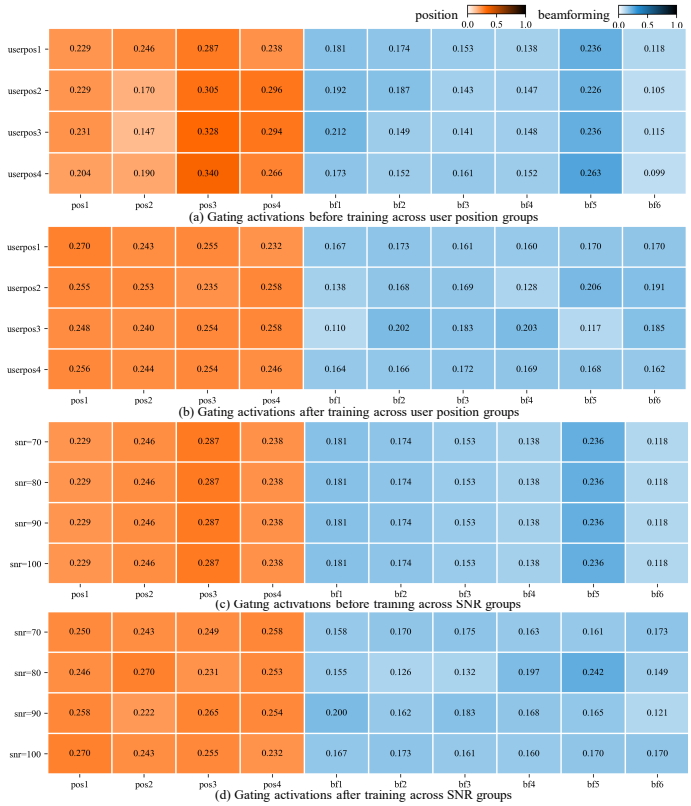


Fig. 6: Gating activations across environment settings of user-position and SNR variations.

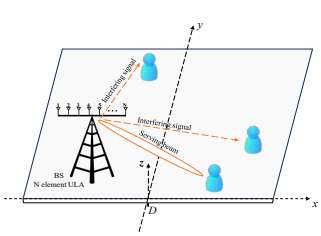


Fig. 7: MU-MIMO system model

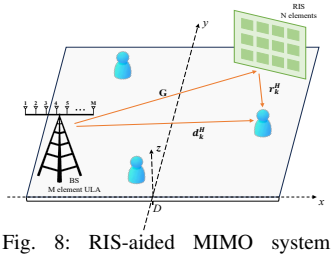
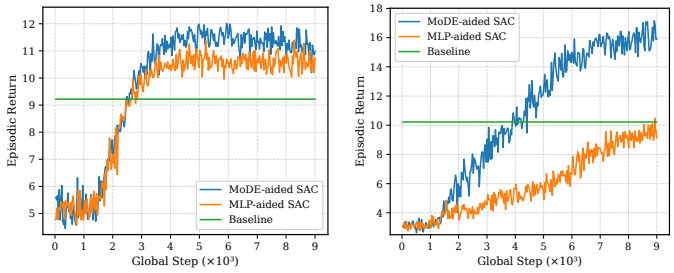


Fig. 8: RIS-aided MIMO system model

distance-dependent phase delay, and array steering response. Beamforming follows the same practical design used for PASS with the direct physical channel as the equivalent channel. The optimization objective is to maximize SE as defined in Sec. III-A1. Here, coupling arises because user SINRs are interdependent via shared channel directions and inter-user interference, linking power allocation and beamforming decisions. MoDE handles this through the layered architecture: The configuration layer outputs a K-dimensional power allocation factor from the state s containing channel state information flattened to length $2NK$, user positions of length $2K$ and antenna array geometry of length $2N$. The subsequent signal layer takes the enriched state $[s; a_{\text{power}}]$ to produce beamforming latent z_w. The continuous action a encodes both power allocation and beamforming parameters. The reward follows the same form as in PASS.

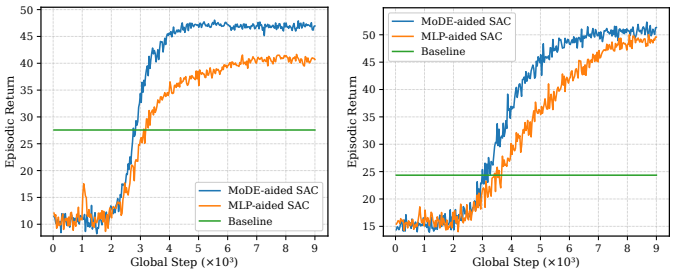
2) Experimental Results

Fig. 9 compares the average return during training for MoDE-aided SAC, MLP-aided SAC, and a heuristic baseline using Minimum Mean-Square Error (MMSE) precoding with water-filling power allocation. Results show when $N = 7$, $K = 13$, $D = 50$ and $N = 11$, $K = 18$, $D = 40$,



(a) N = 11, K = 18, D = 40 (b) N = 7, K = 13, D = 50

Fig. 9: Comparison of average return versus training steps in MU-MIMO.



(a) N = 24, M = 2, K = 3 (b) N = 36, M = 4, K = 6

Fig. 10: Comparison of average return versus training steps in RIS-aided MIMO.

MoDE-aided SAC achieves faster convergence and a higher final performance compared to MLP-aided SAC, with both significantly surpassing the baseline after sufficient training.

C. Joint RIS Phase and Beamforming in RIS-Aided MIMO

1) MoDE-Aided SAC In RIS-Aided MIMO System Model

We examine a RIS-aided MIMO downlink as illustrated in Fig. 8, where a BS equipped with M transmit antennas serves K users distributed in a $30\times30\,\mathrm{m}^2$ region, through both direct links and reflections from an N-elements RIS. The system operates at a carrier frequency of $f_c = 3.5$ GHz, with path loss exponents of 3.0 for direct links and 2 for RIS-reflected paths. The reference path loss is set to $P_{lo} = -20$ dB at $d_o = 1$ m. The composite channel for user k is $\mathbf{h}_k^H = \mathbf{r}_k^H\mathbf{\Theta}\mathbf{G} + \mathbf{d}_k^H$, where $\mathbf{G}$ is the BS-to-RIS link, $\mathbf{r}_k$ the RIS-to-user k link, $\mathbf{d}_k$ the direct BS-to-user k link, and $\Theta = \mathrm{diag}(e^{j\theta_1},\ldots,e^{j\theta_N})$ is the RIS phase-shift matrix. We use the same beamforming design in MU-MIMO using effective channel constructed from $\mathbf{h}_k^H(\Theta)$. Our objective of maximizing the total SE is defined the same as Sec. III-A1. Coupling is explicit since the effective channel directly depends on the RIS phase-shift matrix, tightly binding phase optimization and beamforming design. MoDE reflects this dependency by employing hierarchy: the first configuration layer outputs both sine and cosine components of phase shifts, which then condition the second signal layer to produce beamforming latent z_w. State s includes previous RIS phases of length $2N$, effective channel of length $2KM$, and user locations of length $2K$. The continuous action a encodes both RIS phase parameters and beamforming latent variables.

2) Experimental Results

Fig. 10 presents the average return versus training steps under $N = 24$, $M = 2$, $K = 3$, and $N = 36$, $M = 4$, and $K = 6$. The baseline operates on local copies of environment channels and alternately updates The results show that MoDE-aided SAC demonstrates faster convergence and higher final performance compared to MLP-aided SAC. Both learning-based methods substantially outperform a conventional alternating optimization baseline using alternate MMSE precoder and closed-form phase updates after sufficient training.

IV. Conclusion

We proposed MoDE-aided DRL framework for tightly coupled, high-dimensional optimization in wireless networks to enhance system performance. It enables effectively decoupling, specializing, and coordinated adaptation of both system configuration and signal parameters using real-time, sample-efficient feedback. This capability is validated across three representative scenarios, demonstrating effective joint optimization in each case. Our methodology highlights the potential of combining MoDE architectures with DRL for scalable, real-time resource optimization in 6G wireless systems.

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