

Discrete Fourier transform (general)

In mathematics, the **discrete Fourier transform** over an arbitrary ring generalizes the discrete Fourier transform of a function whose values are complex numbers.

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Definition

Let ***R*** be any ring, let ***n*** ≥ 1 be an integer, and let ***α*** ∈ ***R*** be a principal *n*th root of unity, defined by:^[1]

$$\begin{aligned}\alpha^n &= 1 \\ \sum_{j=0}^{n-1} \alpha^{jk} &= 0 \text{ for } 1 \leq k < n\end{aligned}\quad (1)$$

The discrete Fourier transform maps an *n*-tuple (***v***₀, ..., ***v***_{*n*−1}) of elements of ***R*** to another *n*-tuple (***f***₀, ..., ***f***_{*n*−1}) of elements of ***R*** according to the following formula:

$$f_k = \sum_{j=0}^{n-1} v_j \alpha^{jk}. \quad (2)$$

By convention, the tuple (***v***₀, ..., ***v***_{*n*−1}) is said to be in the *time domain* and the index ***j*** is called *time*. The tuple (***f***₀, ..., ***f***_{*n*−1}) is said to be in the *frequency domain* and the index ***k*** is called *frequency*. The tuple (***f***₀, ..., ***f***_{*n*−1}) is also called the spectrum of (***v***₀, ..., ***v***_{*n*−1}). This terminology derives from the applications of Fourier transforms in signal processing.

If \mathbf{R} is an integral domain (which includes fields), it is sufficient to choose α as a primitive n th root of unity, which replaces the condition (1) by:^[1]

$$\alpha^k \neq 1 \text{ for } 1 \leq k < n$$

Proof: take $\beta = \alpha^k$ with $1 \leq k < n$. Since $\alpha^n = 1$, $\beta^n = (\alpha^n)^k = 1$, giving:

$$\beta^n - 1 = (\beta - 1) \left(\sum_{j=0}^{n-1} \beta^j \right) = 0$$

where the sum matches (1). Since α is a primitive root of unity, $\beta - 1 \neq 0$. Since \mathbf{R} is an integral domain, the sum must be zero. ■

Another simple condition applies in the case where n is a power of two: (1) may be replaced by $\alpha^{n/2} = -1$.^[1]

Inverse

The inverse of the discrete Fourier transform is given as:

$$v_j = \frac{1}{n} \sum_{k=0}^{n-1} f_k \alpha^{-jk}. \quad (3)$$

where $1/n$ is the multiplicative inverse of n in \mathbf{R} (if this inverse does not exist, the DFT cannot be inverted).

Proof: Substituting (2) into the right-hand-side of (3), we get

$$\begin{aligned} & \frac{1}{n} \sum_{k=0}^{n-1} f_k \alpha^{-jk} \\ &= \frac{1}{n} \sum_{k=0}^{n-1} \sum_{j'=0}^{n-1} v_{j'} \alpha^{j'k} \alpha^{-jk} \\ &= \frac{1}{n} \sum_{j'=0}^{n-1} v_{j'} \sum_{k=0}^{n-1} \alpha^{(j'-j)k}. \end{aligned}$$

This is exactly equal to v_j , because $\sum_{k=0}^{n-1} \alpha^{(j'-j)k} = 0$ when $j' \neq j$ (by (1) with $k = j' - j$), and

$$\sum_{k=0}^{n-1} \alpha^{(j'-j)k} = n \text{ when } j' = j. \quad \blacksquare$$

Matrix formulation

Since the discrete Fourier transform is a linear operator, it can be described by matrix multiplication. In matrix notation, the discrete Fourier transform is expressed as follows:

$$\begin{bmatrix} f_0 \\ f_1 \\ \vdots \\ f_{n-1} \end{bmatrix} = \begin{bmatrix} 1 & 1 & 1 & \cdots & 1 \\ 1 & \alpha & \alpha^2 & \cdots & \alpha^{n-1} \\ 1 & \alpha^2 & \alpha^4 & \cdots & \alpha^{2(n-1)} \\ \vdots & \vdots & \vdots & & \vdots \\ 1 & \alpha^{n-1} & \alpha^{2(n-1)} & \cdots & \alpha^{(n-1)(n-1)} \end{bmatrix} \begin{bmatrix} v_0 \\ v_1 \\ \vdots \\ v_{n-1} \end{bmatrix}.$$

The matrix for this transformation is called the DFT matrix.

Similarly, the matrix notation for the inverse Fourier transform is

$$\begin{bmatrix} v_0 \\ v_1 \\ \vdots \\ v_{n-1} \end{bmatrix} = \frac{1}{n} \begin{bmatrix} 1 & 1 & 1 & \cdots & 1 \\ 1 & \alpha^{-1} & \alpha^{-2} & \cdots & \alpha^{-(n-1)} \\ 1 & \alpha^{-2} & \alpha^{-4} & \cdots & \alpha^{-2(n-1)} \\ \vdots & \vdots & \vdots & & \vdots \\ 1 & \alpha^{-(n-1)} & \alpha^{-2(n-1)} & \cdots & \alpha^{-(n-1)(n-1)} \end{bmatrix} \begin{bmatrix} f_0 \\ f_1 \\ \vdots \\ f_{n-1} \end{bmatrix}.$$

Polynomial formulation^[2]

Sometimes it is convenient to identify an n -tuple (v_0, \dots, v_{n-1}) with a formal polynomial

$$p_v(x) = v_0 + v_1x + v_2x^2 + \cdots + v_{n-1}x^{n-1}.$$

By writing out the summation in the definition of the discrete Fourier transform (2), we obtain:

$$f_k = v_0 + v_1\alpha^k + v_2\alpha^{2k} + \cdots + v_{n-1}\alpha^{(n-1)k}.$$

This means that f_k is just the value of the polynomial $p_v(x)$ for $x = \alpha^k$, i.e.,

$$f_k = p_v(\alpha^k).$$

The Fourier transform can therefore be seen to relate the *coefficients* and the *values* of a polynomial: the coefficients are in the time-domain, and the values are in the frequency domain. Here, of course, it is important that the polynomial is evaluated at the n th roots of unity, which are exactly the powers of α .

Similarly, the definition of the inverse Fourier transform (3) can be written:

$$v_j = \frac{1}{n}(f_0 + f_1\alpha^{-j} + f_2\alpha^{-2j} + \cdots + f_{n-1}\alpha^{-(n-1)j}). \quad (5)$$

With

$$p_f(x) = f_0 + f_1x + f_2x^2 + \cdots + f_{n-1}x^{n-1},$$

this means that

$$v_j = \frac{1}{n}p_f(\alpha^{-j}).$$

We can summarize this as follows: if the *values* of $p(x)$ are the *coefficients* of $q(x)$, then the *values* of $q(x)$ are the *coefficients* of $p(x)$, up to a scalar factor and reordering.

Special cases

Complex numbers

If $F = \mathbb{C}$ is the field of complex numbers, then the n th roots of unity can be visualized as points on the unit circle of the complex plane. In this case, one usually takes

$$\alpha = e^{\frac{-2\pi i}{n}},$$

which yields the usual formula for the complex discrete Fourier transform:

$$f_k = \sum_{j=0}^{n-1} v_j e^{\frac{-2\pi i}{n} jk}.$$

Over the complex numbers, it is often customary to normalize the formulas for the DFT and inverse DFT by using the scalar factor $\frac{1}{\sqrt{n}}$ in both formulas, rather than **1** in the formula for the DFT and $\frac{1}{n}$ in the formula for the inverse DFT. With this normalization, the DFT matrix is then unitary. Note that \sqrt{n} does not make sense in an arbitrary field.

Finite fields

If $F = GF(q)$ is a finite field, where q is a prime power, then the existence of a primitive n th root automatically implies that n divides $q - 1$, because the multiplicative order of each element must divide the size of the multiplicative group of F , which is $q - 1$. This in particular ensures that $n = \underbrace{1 + 1 + \dots + 1}_{n \text{ times}}$ is invertible, so that the notation $\frac{1}{n}$ in (3) makes sense.

An application of the discrete Fourier transform over $GF(q)$ is the reduction of Reed–Solomon codes to BCH codes in coding theory. Such transform can be carried out efficiently with proper fast algorithms, for example, cyclotomic fast Fourier transform.

Number-theoretic transform

The **number-theoretic transform (NTT)** is obtained by specializing the discrete Fourier transform to $F = \mathbb{Z}/p$, the integers modulo a prime p . This is a finite field, and primitive n th roots of unity exist whenever n divides $p - 1$, so we have $p = \xi n + 1$ for a positive integer ξ . Specifically, let ω be a primitive $(p - 1)$ th root of unity, then an n th root of unity α can be found by letting $\alpha = \omega^\xi$.

e.g. for $p = 5$, $\alpha = 2$

$$\begin{aligned}
2^1 &= 2 \pmod{5} \\
2^2 &= 4 \pmod{5} \\
2^3 &= 3 \pmod{5} \\
2^4 &= 1 \pmod{5}
\end{aligned}$$

when $N = 4$

$$\begin{bmatrix} F(0) \\ F(1) \\ F(2) \\ F(3) \end{bmatrix} = \begin{bmatrix} 1 & 1 & 1 & 1 \\ 1 & 2 & 4 & 3 \\ 1 & 4 & 1 & 4 \\ 1 & 3 & 4 & 2 \end{bmatrix} \begin{bmatrix} f(0) \\ f(1) \\ f(2) \\ f(3) \end{bmatrix}$$

The number theoretic transform may be meaningful in the ring \mathbb{Z}/m , even when the modulus m is not prime, provided a principal root of order n exists. Special cases of the number theoretic transform such as the Fermat Number Transform ($m = 2^k + 1$), used by the [Schönhage–Strassen algorithm](#), or Mersenne Number Transform ($m = 2^k - 1$) use a composite modulus.

Discrete weighted transform

The **discrete weighted transform (DWT)** is a variation on the discrete Fourier transform over arbitrary rings involving [weighting](#) the input before transforming it by multiplying elementwise by a weight vector, then weighting the result by another vector.^[3] The [Irrational base discrete weighted transform](#) is a special case of this.

Properties

Most of the important attributes of the [complex DFT](#), including the inverse transform, the [convolution theorem](#), and most [fast Fourier transform \(FFT\)](#) algorithms, depend only on the property that the kernel of the transform is a principal root of unity. These properties also hold, with identical proofs, over arbitrary rings. In the case of fields, this analogy can be formalized by the [field with one element](#), considering any field with a primitive n th root of unity as an algebra over the extension field \mathbf{F}_{1^n} .

In particular, the applicability of $O(n \log n)$ [fast Fourier transform](#) algorithms to compute the NTT, combined with the convolution theorem, mean that the [number-theoretic transform](#) gives an efficient way to compute exact [convolutions](#) of integer sequences. While the complex DFT can perform the same task, it is susceptible to [round-off error](#) in finite-precision [floating point](#) arithmetic; the NTT has no round-off because it deals purely with fixed-size integers that can be exactly represented.

Fast algorithms

For the implementation of a "fast" algorithm (similar to how [FFT](#) computes the [DFT](#)), it is often desirable that the transform length is also highly composite, e.g., a power of two. However, there are specialized fast Fourier transform algorithms for finite fields, such as Wang and Zhu's algorithm,^[4] that are efficient regardless of whether the transform length factors.

See also
