

Accelerator Interface Framework for a Linear Time-Multiplexed Overlay

Abstract—Coarse-grained FPGA overlays have shown promising advantages in improved design productivity, through software-like programmability and fast compilation. However, the effectiveness of overlays is dependent on suitable interface and programming integration into a complete accelerator system, which has often been neglected in evaluations of overlays. We explore the integration of a time-multiplexed overlay into server-class PCI Express and embedded AXI connected FPGA platforms. We show how this integration can be optimised to maximise performance, and evaluate the area overhead. We also compare this complete system to an alternative custom processor overlay.

Index Terms—FPGA overlay, memory interface, framework

I. INTRODUCTION

Coarse-grained overlays, implemented on top of conventional fine-grained FPGAs, represent a promising solution to the design productivity challenge. Coarse-grained architectures enable faster compilation and software-like programmability, compared to the challenge of increasingly complex applications being mapped to very fine grained resources. FPGA overlays can be broadly categorised as spatially configured (SC) or time-multiplexed (TM). In an SC overlay, a functional unit (FU) is allocated to a single computational operation in the accelerated kernel, with FUs connected by a routing network which is essentially static during kernel execution. A TM overlay, on the other hand, shares both the FUs and the interconnect across kernel operations, allowing improved usage of limited FPGA resources. However, TM overlays suffer from relatively large area overheads, due to their underlying processor-like architecture [1], [3], [6], [13], [15] or, for CGRA-like overlays, due to the routing resources and instruction storage requirements [10], [12]. Reducing the area overhead for CGRA-like overlays, specifically for the routing network, and utilizing the fast context switching capabilities is likely to result in improved usability with corresponding improvements in design productivity. In one design, a streaming architecture based on feed-forward pipelined datapaths, connected through a simple linear interconnect of lightweight functional units, was shown to drastically reduce interconnect requirements [9]. However, performance analysis of such overlays often ignores the system interfacing aspects that are crucial to usability, as the interface between a main processor/memory subsystems and the overlay must be able to provide high-bandwidth (and large scale) data transmission. Existing FPGA interfacing frameworks fall into one of two categories, AXI bus-based solutions for FPGA SoC systems (like the Xilinx Zynq) and PCIe-based solutions for stand-alone systems connected to a host machine. Some of these frameworks abstract the interfaces with simple

FIFO connections and provide streaming data transfers, which matches the requirements of our proposed linear TM overlay.

In this paper, we implement an accelerator framework for Xilinx Zynq-based FPGA systems using the AXI interface and for standalone FPGA systems with PCIe connectivity, based on the Xillybus [20], RIFFA [8] and DyRACT [18] communication frameworks. The architecture of a TM overlay and its essential control requirements are presented in detail. We present a full working implementation integrated with a memory subsystem, enabling practical usage. We evaluate the performance of the proposed AXI bus-based and PCIe-based overlay accelerators for a range of benchmarks. The proposed AXI bus-based accelerator is compared to a state-of-art VectorBlox MXP processor overlay [15], in terms of throughput and resource usage. We compare the three proposed PCIe-based implementations and determine the most suitable interface in terms of supporting configuration of and streaming through the overlay.

The main contributions of this paper can be summarized as follows:

- Examining memory interfaces for accelerator overlays, and provide a comprehensive analysis of embedded AXI and server PCIe interfaces for a range of benchmarks.
- A comparison with the state-of-the-art VectorBlox MXP processor overlay, showing 50% throughput achieved using half of the bandwidth and less than 20% of the hardware resource.

II. RELATED WORK

A. Coarse-grained TM Overlays

A number of overlays have been proposed which share functional units among kernel operations in an attempt to reduce overlay resource requirements [10], [15], [16]. This time-multiplexing of the overlay means that it can change its behavior on a cycle-by-cycle basis while the compute kernel is executing, thus allowing sharing of the limited FPGA resources.

ADRES [11] appeared to be the first integration of a very long instruction word (VLIW) processor tightly-coupled with a coarse-grained reconfigurable matrix. Some of the functional units (FUs) and register files (RFs) are shared by the VLIW processor and the reconfigurable matrix. Thus, there is no need to transfer the data between the processor and the reconfigurable array compared to the traditional CGRAs. Taras [16] implemented the ADRES as overlays on Intel and Xilinx FPGAs using CGRA-ME modeling framework [2]. The

ADRES overlay can achieve up to $1.6\times$ speedup on Fmax and a reduction of 41% LUT usage, with the FPGA architecture optimizations.

QuickDough [10] was proposed to address FPGA design productivity issues. It is comprised of an array of nearest neighbor interconnected processing elements (PEs), referred to as SCGRA overlay. Application specific SCGRA overlay was implemented on Zynq, achieving a speedup of up to $9\times$ compared to the same application running on the Zynq ARM processor. The 250 MHz FU consists of an ALU, multiport data memory (256×32 bits) and customizable depth instruction ROM (Supporting 72-bit instructions) resulting in significant BRAM utilization. Due to the excessive usage of BRAMs, it has limited scalability as a maximum implementation of 5×5 array can fit on Zynq, despite of the frequency degradation caused by the tight placement and route.

VectorBlox MXP [15] is a commercial soft vector processor overlay targetting Altera or Xilinx FPGAs via the Avalon or AXI interfaces. Data transfer is handled by a double-buffered vector scratchpad and a dedicated DMA engine communicating with the scalar host processor via the AXI HP port. MXP can operate at a maximum frequency of 110 MHz on a Xilinx XC7Z020 device with 16 vector lanes. It demonstrates up to $1000\times$ speedup over MicroBlaze. The programming model is user-friendly as it is a combination of ANSI-C and VectorBlox C extensions. However, it is difficult to implement the applications which cannot be easily vectorized.

B. Accelerator Interfacing Frameworks

Two of the more common solutions for interfacing FPGA based hardware accelerators are AXI bus-based solutions [14], [19], [20] for FPGA SoC systems (like the Xilinx Zynq) and PCIe-based solutions [4], [8], [18], [20] for stand-alone systems connected to a host CPU.

Infrastructure to evaluate the data processing bandwidth and the energy consumption between the ARM processor subsystem and the DRAM on Zynq using the AXI HP/ACP interface was developed in [14]. The implementation achieved a maximum full-duplex data processing bandwidth of over 1.6 GB/s for a single HP/ACP port, running at 125 MHz on a Zynq XC7Z020-1C device.

ZyCAP [19] was developed as an open source soft DMA controller for high performance partial reconfiguration on the Xilinx Zynq. It provides two interfaces, an AXI-Lite interface connected to general purpose (GP) port and another AXI4 interface connected to the high performance (HP) port, and achieves a maximum throughput of 382 MB/s between the external memory and the partial reconfigurable region (PRR).

Xillybus [20] is a representative commercial solution which supports both AXI (baseline) and PCIe (Revision XL) interfaces while providing portability across different FPGA devices. While Xillybus is provided free of charge for research and teaching purposes, users have full access to customize the complete Xillybus project, including the hardware design, the device driver and the software applications, with good documentation provided. The baseline Xillybus has a data rate

TABLE I: Theoretical bandwidth of typical memory interfaces.

Interface	Frequency	Upstream BW	Downstream BW	Ports/Lanes	Total BW ¹
AXI HP	150 MHz	1200 MB/s	1200 MB/s	4	9600 MB/s
AXI ACP	150 MHz	1200 MB/s	1200 MB/s	1	2400 MB/s
PCIe Gen2	250 MHz	500 MB/s	500 MB/s	8	8000 MB/s
PCIe Gen3	250 MHz	984 MB/s	984 MB/s	8	15800 MB/s

¹ In a streaming interface, the throughput is limited by either the upstream or downstream BW, depending on the number of inputs or outputs, and so the maximum theoretical throughput would be half of the total bandwidth reported here.

of ~ 300 MB/s running at 100 MHz on a Xilinx Zynq, and a maximum data rate of ~ 3.5 GB/s for PCIe Gen 2×8 interfaces in Revision XL.

There are a number of academic PCIe interfacing solutions, such as DyRACT [18] and EPEE [4].

RIFFA [8] is an open source reusable framework for the integration of FPGA accelerators with workstations supporting PCIe Gen 2 and Gen 3 standards. A scatter-gather DMA-based design bridges the vendor-specific PCIe Endpoint core and multiple communication channels for user defined IP cores. The latest release of RIFFA (version 2.2.2) achieves a unidirectional maximum bandwidth of 3.6 GB/s for the Gen2 $\times 8$ configuration on the Xilinx VC707 platform and 3.5 GB/s for the Gen3 $\times 4$ configuration on the Terasic DE5-Net.

DyRACT focuses on dynamic partial reconfiguration over the PCIe Gen2 $\times 4$ interface, along with a configuration controller and clock management, while EPEE was developed as a general purpose PCIe communication library, targeting a wide range of FPGA devices. JetStream [17] is a PCIe Gen3 solution which supports not only FPGA-to-Host communication but also multiple FPGA boards interfaced together.

Though overlays have shown potential in improving FPGA design productivity, few of them have been developed as full accelerator systems except. Most of the FPGA SoC designs use AXI interface as the bridge between the embedded ARM (or soft MicroBlaze/NIOS) processor and the programmable logic. However, PCIe interfaces are used in larger systems where communication between a host PC and accelerator card is needed. Table I shows the theoretical bandwidth of the AXI and PCIe interfaces. Although the total theoretical bandwidth of an AXI interface is comparable to that of a PCIe interface, many of the AXI bus-based solutions are not able to make full use of all available high performance HP/ACP ports, while the PCIe-based counterparts can support up to 8 lanes and some of them achieve a bandwidth which is close to the theoretical maximum. Among the existing implementations, Xillybus, RIFFA and DyRACT appear to be the most promising solutions due to their availability, ease-of-use and portability. In subsequent sections we develop high performance overlay-based accelerators designs using these solutions and analyse the overlay capabilities with different interfaces.

III. LINEAR TM OVERLAY

A. Architecture Description

We consider a linear array of FUs as a time-multiplexed (TM) overlay similar to the design in [9] where each FU

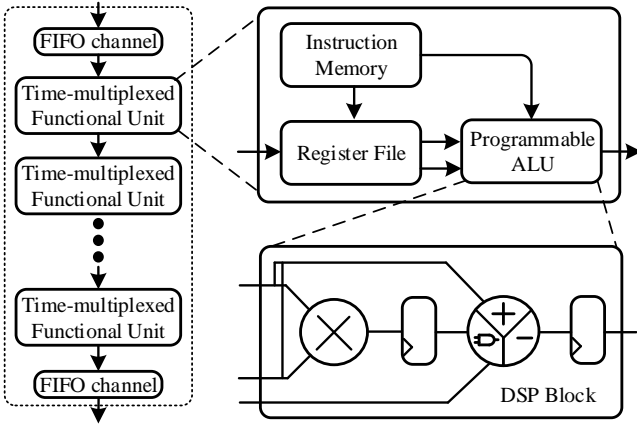


Fig. 1: A linear TM overlay.

can be time multiplexed among operations present in a single scheduling stage of a directed acyclic graph (DAG). The 32-bit linear TM overlay is comprised of a quasi-streaming data interface made up of two FIFO channels implemented using Block RAMs, which transmit data through daisy-chained fully pipelined time-multiplexed FUs, as shown in Figure 1. By eliminating the fully flexible routability of CGRA-like overlays, this structure achieves a much more area-efficient design, using fewer than 6% of the logic and DSP resources on a Xilinx Zynq device. The initiation interval (II) can be significantly reduced by making minor architectural enhancements, such as adding a rotating register file and replicating the data stream. These changes result in a peak throughput of 1.8 GOPS running at a frequency of 335 MHz. Adding write-back capabilities to the FU design reduces the overlay depth requirement by allowing multiple nodes on the critical path to be combined. This also eliminates the need to reconfigure the overlay whenever the application kernel changes, making the overlay suitable for more general purpose applications.

B. Overlay Control

A back-pressure control circuit is built around the input FIFO channel to manage the functionality of the overlay, as shown in Figure 2. There are three control signals which indicate the duration for instruction load, overlay setup and data write respectively, referred to as *inst_load*, *reg_wren*, and *data_wren* respectively. Initially, FU instructions are read from the memory and streamed through the daisy-chained FUs. During instruction load (when *inst_load* is asserted), both the write enable port of the FIFO and the valid signal (*valid_out*) for the data output are disabled. After instruction load, two integers are written to the back-pressure control circuit. The first represents the number of data words to be input to the first FU for a specific compute kernel while the other is equal to the II minus one ($II-1$) and determines the interval between data loads. These values are written into the controller when the *reg_wren* signal is asserted.

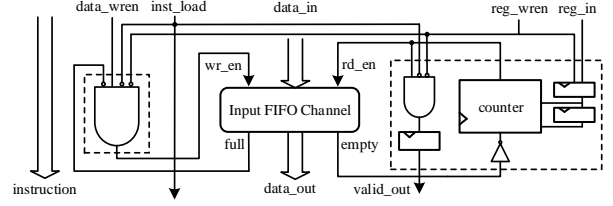


Fig. 2: Back-pressure control circuit.

The dashed box on the left of Figure 2 represents the control module for the write enable port of the FIFO, while the dashed box on the right contains the logic to control the read enable port of the FIFO. Data is written into the FIFO when *wr_en* is high. The read enable signal (*rd_en*) for the FIFO is generated from a counter which determines the number of cycles needed to load input data into the FU. The counter starts counting from 0 when the *empty* signal goes low (indicating that data is available in the FIFO). The counter counts up until the count value equals $II-1$, at which point it rolls over back to zero. The *rd_en* signal is valid only while the counter is less than the initial number loaded into the back-pressure circuit, limiting the amount of data to be loaded into the first FU. Once the counter value is greater than or equal to the data load number (in the back-pressure circuit), the *rd_en* signal is forced low and the input data is buffered in the FIFO.

C. Complete Overlay Accelerator Framework

The linear TM overlay has shown its advantages to implement an area efficient design with fast context switching and high throughput. However, to demonstrate the suitability of the overlay as an FPGA accelerator, it is important to develop an interface between the main processor/memory subsystem and the overlay which is able to provide high-bandwidth data transmission. As discussed in the previous section, two stream interfaces are required for transmitting the data between the host processor and the overlay accelerator. Additionally, it is necessary to set up memory arrays or registers so that the user has access to control the overlay system, i.e. the instruction load, overlay setup, and data write.

IV. OVERLAY ACCELERATOR FRAMEWORK

A block diagram of the proposed overlay accelerator system, based on an array of linear TM overlays as described in the previous section, which supports both Xilinx Zynq-based SoC and standalone PCIe configurations, is shown in Figure 3. The FPGA memory subsystem provides the interface between the overlay on the FPGA fabric and the ARM processor on a Zynq SoC via the AXI bus (or directly to the memory via the PCIe link). For the AXI bus-based overlay, two 32-bit FIFOs are used to connect a single 32-bit linear TM overlay with the memory subsystem. For the PCIe-based overlay, we propose replicating four 32-bit linear TM overlays to make full use of the 128-bit data bandwidth. The four overlay instances

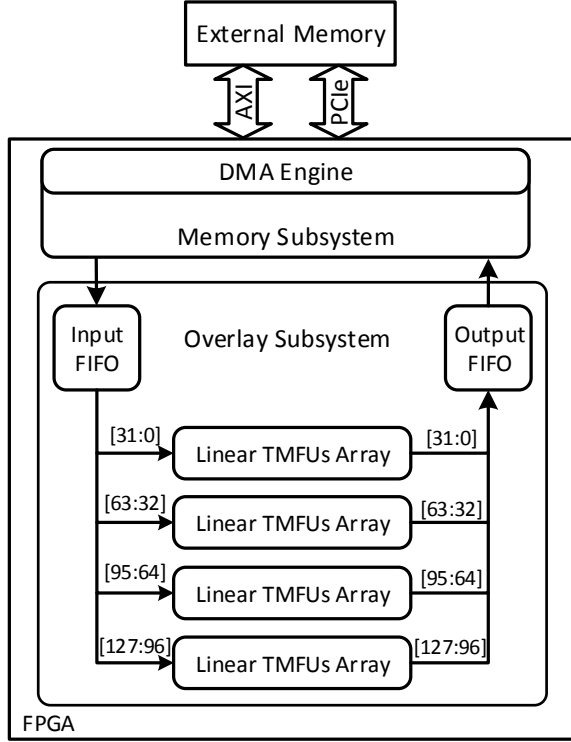


Fig. 3: The proposed overlay accelerator system.

can implement multiple compute kernels at runtime and thus reduce the II to a quarter of that of a single overlay. Data transfers between the internal memory subsystem, the DDR SDRAM and the offchip DRAM are under the control of a scatter-gather DMA engine.

A. Interfacing with Xillybus

Xillybus is a portable, easy to use DMA-based data transfer solution which provides a simple abstraction of the AXI/PCIe interfaces. One side of the Xillybus IP core is connected to the host processor, while the other side communicates with one standard FIFO, providing six signals to control the data flow. The host processor can be either the ARM processor of the Xilinx Zynq SoC (AXI-Xillybus: for an AXI bus-based solution) or a PC based x86 CPU (PCIe-Xillybus: for a PCIe-based solution). Xillybus provides a simple data loopback demo design, along with its driver for the default device files and the software code for testing purpose. When connecting Xillybus to the linear TM overlay, two standard FIFOs need be used to buffer the input/output data.

B. Interfacing with RIFFA

RIFFA has a more complicated framework which generally consists of three layers. On the bottom layer, an RX engine and a TX engine are connected to the vendor-specific PCIe Endpoint IP core, which bridges the connection between

RIFFA and the host CPU. The middle layer supports up to 12 channels which send packets to the TX engine and receive packets from the RX engine through a scatter-gather DMA engine. The overlay which connects to the RX/TX FIFOs from the communication channels, is located on the top layer. An example design of one channel for data transfer, along with driver and software code, is provided. Integration of the linear TM overlay with RIFFA is quite similar to that of Xillybus, except that the standard FIFOs are replaced with the first word fall through (FWFT) FIFOs to meet the timing closure.

C. Interfacing with DyRACT

The original version of DyRACT communication infrastructure enables PCIe-based high-speed communication between a host computer and FPGA-based user logic with support for partial reconfiguration (PR). It provisions multiple AXI4-stream backend interface, enabling seamless integration with vendor-supported IP cores. We have modified the original version to make it lite-weight and tailored to support the proposed overlay architecture. Since the present implementation does not exploit PR, the reconfiguration control logic is disabled, and a single backend AXI4-stream interface is enabled. The stream interface width is configurable through width-conversion FIFOs. A new AXI-Lite interface is added to support command-data interface between the host and the overlay. The software architecture follows similar structure of that of RIFFA. The low-level communication protocols and interrupts are managed by the driver and the user library provides APIs for integration with application programmes.

D. Programming Model

While Xillybus and RIFFA provide RTL design along with the driver support and software code demos, they generally do not have direct access to the FPGA RAMs (Xillybus AXI version has limited access to the registers). In comparison, DyRACT offers simple register access via the AXI-Lite interface. Thus, user can easily send the instructions and streaming data from the host PC to the FPGA at runtime using dedicated APIs, instead of reconfiguring the overlay. A user-friendly programming model is proposed for the linear TM overlay integrated with DyRACT system. There are three register files developed for overlay control purpose. One register file (0x30) is written with a 4-bit tag, which is followed by a register file (0x34) to restore the instruction with its corresponding FU. Another register file (0x38) is used to store the number of data words to be input to the first FU for a specific compute kernel (based on the instructions) and the value of (II-1), when the instruction load has been done. Afterwards, a bunch of user-defined data can be transferred to the FPGA, and will be sent back to the host PC after the processing of the overlay. A snippet example code shows how to load the instructions, setup the overlay, and write the data to the FPGA is presented in Table II.

The 32-bit instructions are written in hexadecimal style in Table II, which are used as configurations for the DSP-based FUs. A description of the instruction format, using *ADD R3*,

TABLE II: Example code.

```

fpga_reg_wr(0x30,0x0); //Tag of FU0
fpga_reg_wr(0x34,0x3033D080); // Instruction 0

fpga_reg_wr(0x30,0x1); //Tag of FU1
fpga_reg_wr(0x34,0x8852000); // Instruction 1

fpga_reg_wr(0x38,5); // No. of input data
fpga_reg_wr(0x38,5); // (II-1)

dyract_send_data((unsigned char *)mydata,
    sendSize*sizeof(int)); //Send data
dyract_rcv_data((unsigned char *) rcvdata,
    rcvSize*sizeof(4)); //Receive data

```

R5 (WB) operation as an example, is shown in Table III. From this table, it can be seen that a 32-bit instruction has four sections, the 18-bit ALU control, the 2-bit input map multiplexing, two 5-bit source operand addresses, with the remaining 2-bit reserved for future use.

V. EXPERIMENTAL EVALUATION

We conducted experiments for both the AXI bus-based and PCIe-based overlay accelerators. For the AXI bus-based system, the experimental software runs on the ARM processor of the Xilinx ZedBoard. For the PCIe-based accelerators (implemented using PCIe-Xillybus, RIFFA and DyRACT respectively), the experiments are run on an HP Z420 workstation (six 3.5 GHz Intel Xeon E5-1650 cores) with a Xilinx VC707 board plugged into the PCIe Gen2 slot of the motherboard.

A. AXI-Xillybus System

The AXI-Xillybus system is using one 32-bit wide AXI ACP port for data transmission, running at a frequency of 100 MHz on the Xilinx ZedBoard. Hence the theoretical total bandwidth of this system is 800 MB/s, and the maximum streaming throughput is 400 MB/s in a full-duplex system.

1) *AXI-Xillybus Loopback Test*: Before the accelerator system can be developed, we first ensure that the demo bundle provided by the Xillybus vendor operates correctly on our Zynq-based system. We evaluate the performance of Xillybus by testing the round-trip data transmission throughput with the default settings, using sequential and parallel (multi-threading) programming for *read* and *write* functions respectively.

The bandwidth of the loopback test for AXI-Xillybus using different data block sizes is shown in Figure 4. As seen from this diagram, the single-threaded loopback test saturates at a throughput of around 150 MB/s (maximum of 160 MB/s) when the transfer size exceeds 8K words, while the multi-thread loopback test saturates (with a maximum throughput of 330 MB/s) when the transfer size exceeds 500K words. The two dashed lines indicate the theoretical throughput of the 32-bit AXI-ACP port running at 100 MHz. The throughput of the multi-thread test surpasses that of the single-thread test for data block sizes exceeding 64K words, and for large block sizes achieves approximately twice the throughput. Creating Pthreads introduces a significant overhead for smaller transfer

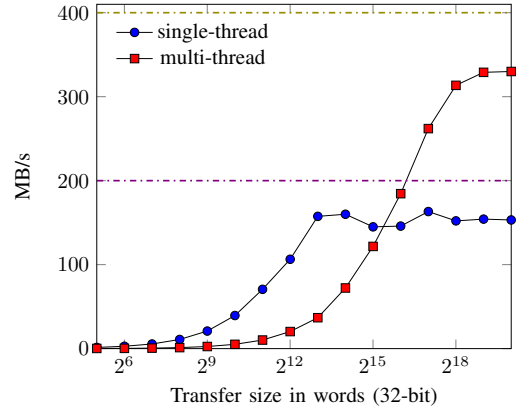


Fig. 4: AXI-Xillybus loopback test evaluation.

sizes, thus the multi-threading version is only applicable for large data transfers.

2) *AXI-Xillybus based Overlay Accelerator*: Xillybus provides an online IP core factory for users to customise the interface on demands. In our design, two 32-bit streaming device files are instantiated for data acquisition and one 32-bit seekable device file is responsible for the instruction load and overlay setup. We evaluate the performance of the overlay accelerator which is an integration of the linear TM overlay as discussed in Section III and AXI-Xillybus interface, and compare to VectorBlox MXP [15], in terms of bandwidth and area consumption. To perform this comparison, we use the set of kernels shown in Table IV, which are extracted from compute intensive applications available from [5], [7]. Both the overlay accelerator and MXP are running on Xillinux-2.0, which is the latest Linux distribution released for ZedBoard, with AXI-Xillybus using the Pthread multi-threading technique with a data block transfer size of 1M words.

Figure 5 shows the performance of an AXI-Xillybus based 32-bit linear TM overlay system and a 16-lane 32-bit MXP (MXP-V16) implemented on a ZedBoard for the benchmarks given in Table IV. As seen from the figure, MXP-V16 outperforms the AXI-Xillybus based system over all benchmarks (with the exception of ‘chebyshev’). Table V shows the hardware resource usage of the AXI-Xillybus based system and MXP-V16. As seen from this table, the AXI-Xillybus based system is much more area efficient, and consumes only 19.8% LUTs, 26.1% FFs, 7.6% BRAMs and 7.1% DSP blocks, compared to MXP-V16.

The MXP memory system is based on one 64-bit AXI HP port running at a frequency of 110 MHz while AXI-Xillybus is based on a single 32-bit AXI ACP port running at a frequency of 100 MHz. This corresponds to a theoretical maximum throughput of 880 MB/s for MXP-V16 and 400 MB/s for the AXI-Xillybus based system. As seen from Figure 5, MXP-V16 achieves a throughput of 460.6 MB/s (52.3% of theoretical maximum) on average while AXI-Xillybus has a throughput of 226.6 MB/s (56.6% of theoretical maximum) on average. The throughput of AXI-Xillybus is about half of that of MXP-V16,

TABLE III: Instruction format.

FU part	ALU Control								Input Map Control		RF Control		Reserved
Signals	<i>NDF</i>	<i>WB</i>	alumode	inmode	opmode	cea2	ceb2	usemult	split	immop	src1	src2	
No. of bits	1	1	4	2	7	1	1	1	1	1	5	5	2
Locations	[30]	[29]	[28:25]	[24:23]	[22:16]	15	14	13	12	11	[10:6]	[5:1]	[31][0]
ADD R3, R5 (<i>WB</i>)	0	1	0000	00	0110011	1	1	0	1	0	00011	00101	0 0

TABLE IV: DFG characteristics of benchmark set.

No.	Benchmark	Characteristics				
		I/O nodes	graph edges	op nodes	graph depth	graph width
1.	chebyshev	1/1	12	7	7	1
2.	mibench	3/1	22	13	6	3
3.	qspline	7/1	50	25	8	6
4.	fft	6/4	24	10	3	4
5.	kmeans	16/1	39	23	5	8
6.	mm	16/1	31	15	4	8
7.	spmv	16/2	30	14	4	8
8.	stencil	15/2	30	14	5	6

TABLE V: Area overhead of AXI bus-based systems.

System	Resource Usage			
	LUTs	FFs	BRAMs	DSPs
AXI-Xillybus-Interface	5,747	5,798	5	8
MXP-V16	28,974	22,174	66	112
Available	53,200	106,400	140	220

mainly due to AXI-Xillybus using only 32-bit of the available 64-bit ACP port while MXP-V16 uses the full 64-bit HP port.

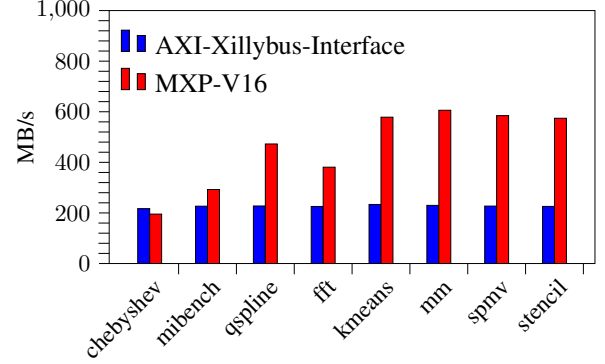
Although the AXI bus-based Xillybus provides an area efficient interface solution for the Xilinx Zynq SoC, the throughput is still far below the theoretical maximum shown in Table I. This is mainly due to not making good use of the available HP/ACP ports (Xillybus uses only one 32-bit port which is operating at a frequency of 66.7% of the theoretical maximum that the AXI bus on Zynq can support).

B. PCIe System

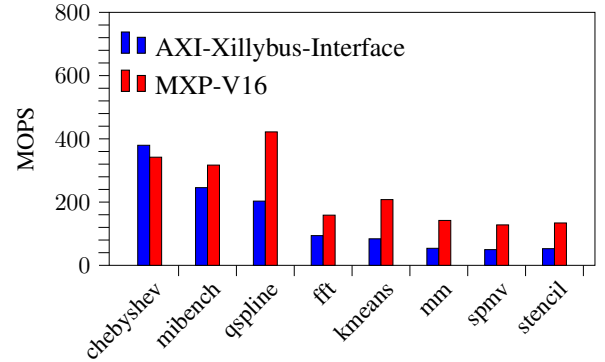
Our PCIe system is based on the Gen2×8 PCIe interface running at a frequency of 250 MHz on the Xilinx VC707 platform, which corresponds to a maximum bandwidth of 8000 MB/s and hence a maximum streaming throughput of 4000 MB/s for a full-duplex system.

1) *Loopback Test*: Similar to the AXI-Xillybus System, we first evaluate the baseline performance of the three PCIe interfaces using a loopback test.

Xillybus has been upgraded to a new version (PCIe-Xillybus) to support high performance PCIe interface communication since 2015. The block diagram of PCIe-Xillybus for data loopback is almost the same as that of AXI-Xillybus, except that the 32-bit FIFO is replaced with a 128-bit FIFO and the Xillybus hardware module *Xillybus.v* (including the Xillybus IP core and a Xilinx 7 series integrated block



(a) Data processing throughput in MB/s



(b) Throughput in MOPS

Fig. 5: Performance comparison for the benchmarks using a transfer size of 1M words.

for PCIe) is connected to an HP Z420 workstation via the PCIe link. Similar to AXI-Xillybus, we developed two host programs to evaluate the round-trip bandwidth in both half-duplex and full-duplex mode. The loopback test design of RIFFA framework is slightly different from that of Xillybus. A first word fall through (FWFT) FIFO is used for data transmission between the RIFFA IP core and the user logic applications. Data transmission is under the control of a finite state machine (FSM), which executes the upstream transfers and downstream transfers in different processing states. Thus, it is not possible to achieve full-duplex data transmission. DyRACK also uses internal FIFOs for temporarily storing data before transmitting/receiving from user logic. Since the back-end interface conforms to AXI4-Stream standard, the transmit and receive interfaces can be tied together to enable loopback. Due to the non-blocking APIs, unlike RIFFA, upstream and downstream transmission can happen in parallel. This provides

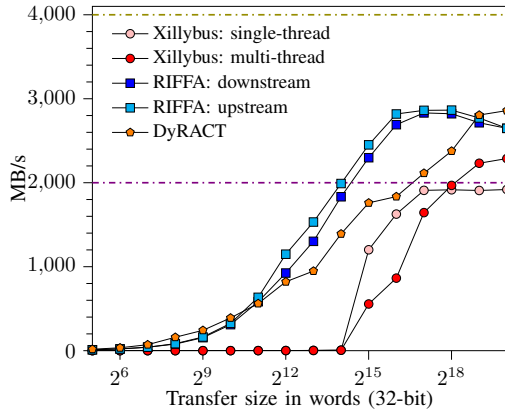
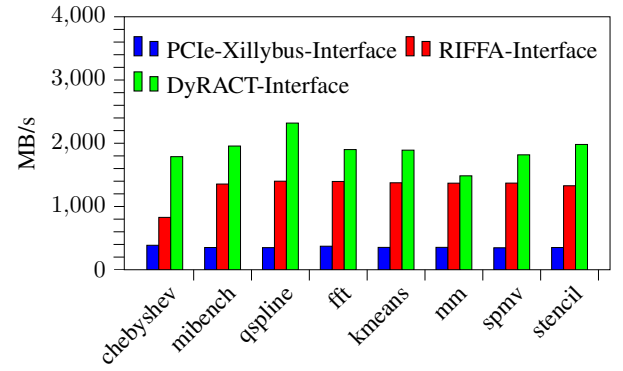


Fig. 6: PCIe system loopback test evaluation.

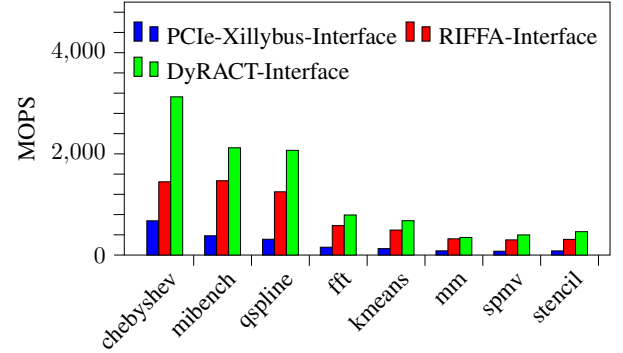
better overall bidirectional performance.

The bandwidth of the loopback test for PCIe-Xillybus, RIFFA and DyRACT using different data block sizes is shown in Figure 6. The two dashed lines indicate the theoretical bandwidth of the Gen2×8 PCIe interface running at 250 MHz. In PCIe-Xillybus, the data transmission is very slow when the transfer size is less than 32K words, due to large system overhead for a data buffer size of less than 128 KBytes [20]. The single-threaded loopback test saturates at a throughput of 1900 MB/s when the transfer size exceeds 128K words, while the multi-thread loopback test saturates (with a maximum throughput of 2300 MB/s) when the transfer size exceeds 500K words. As mentioned previously, the RIFFA hardware was not developed for full-duplex data transmission. Both the upstream transfers (read) and downstream transfers (write) saturate at around 2.85 GB/s when the transfer size is equal to 128K sample words, corresponding a total bandwidth of 2.85 GB/s, as read and write cannot happen simultaneously. The bandwidth degrades slightly when the transfer size exceeds 128k words due to the overutilization of BRAMs. In DyRACT, read and write operations can happen simultaneously and the maximum combined throughput is 5718 MB/s. When tested separately, the write throughput saturates at 3671 MB/s and read performance at 2048 MB/s. The lower read performance is due to the lack of zero memory data copy from kernel to user memory space. Figure 6 shows the average performance of DyRACT saturates at around 2.86 GB/s with 1M words transmission for simultaneous read/write operations.

2) *PCIe-based Overlay Accelerator*: In the PCIe-based overlay accelerator framework, we propose replicating four 32-bit linear TM overlays, interfacing with two 128-bit FIFOs to make full use of the data width. The overlay interfacing with PCIe-Xillybus or RIFFA has to hardcode the instructions to the FUs as it has no access to the FPGA RAMs directly. Thus reconfiguring the overlay is inevitable when the application kernel changes. In comparison, the overlay interfacing with DyRACT can customise three memory arrays to restore the instructions along with their tags to the FUs, and the number of input data or the value of II minus one (II-1).



(a) Data processing throughput in MB/s



(b) Throughput in MOPS

Fig. 7: Performance comparison for the benchmarks.

Figure 7 shows the performance of four 32-bit linear TM overlays integrated with PCIe-Xillybus, RIFFA and DyRACT, respectively. The PCIe-Xillybus based system and DyRACT based system use a block size of 1M words with different parallel programming techniques, while RIFFA based system uses a block size of 128K words. All these overlay accelerators are running on an Ubuntu 14.04.5 workstation (six 3.5 GHz Intel Xeon E5-1650 cores) with a Xilinx VC707 evaluation board plugged into the PCIe Gen2 slot of the motherboard. As can be seen from Figure 7, DyRACT based system achieves an average throughput of 1892 MB/s (47.3% of theoretical maximum), which is around 45% higher than RIFFA based system with an average throughput of 1300 MB/s (32.5% of the theoretical maximum) and 5.3× better than PCIe-Xillybus based system with an average throughput of 358 MB/s (9% of theoretical maximum). The ratio of the throughput in Mega-operations per second (MOPS) is proportional to that of data processing throughput in MB/s. Among the three implementations, DyRACT based system proves to be the best accelerator in terms of throughput. Although all the three overlay accelerators show slight fluctuation of the throughput in MB/s, drastic changes happen when calculating in MOPS. This is mainly due to a large number of I/O ports (especially for the last four benchmarks), which results in larger II values. In a nutshell, the linear TM overlay accelerators are more suitable to the benchmarks with less number of I/O ports.

TABLE VI: Area overhead of PCIe-based systems.

System	Resource Usage			
	LUTs	FFs	BRAMs	DSPs
PCIe-Xillybus-Interface	16,027	12,488	14.5	32
RIFFA-Interface	13,657	18,581	289.5	32
DyRACT-Interface	17,029	16,302	10	32
Available	303,600	607,200	1,030	2,800

Table VI shows the FPGA resource usage of PCIe-Xillybus based system, RIFFA based system and DyRACT based system, respectively. The PCIe-Xillybus based system consumes the least number of FFs (33% fewer than the RIFFA based system), while RIFFA based system costs least number of LUTs (20% fewer than DyRACT based system) and DyRACT based system is the most area efficient in BRAMs consumption (96% fewer than RIFFA based system). While the logic and DSP resources used are comparable among all these implementations, there is a big difference in the BRAM utilization specifically for RIFFA based system. This is due to the half-duplex mode of the RIFFA interface, resulting in a significant BRAM consumption to implement the two large depth FWFT FIFOs. Developing a full-duplex RIFFA based system would minimize the BRAM resource usage. At current stage, DyRACT based system and PCIe-Xillybus based system are much better than RIFFA based system in terms of area efficiency.

VI. DISCUSSION AND CONCLUSIONS

In this paper, we have proposed overlay accelerator systems based on two different memory interfaces, AXI and PCIe. The AXI-Xillybus based system represents a very area efficient implementation compared with VectorBlox MXP-V16, but with about half of the throughput. AXI-Xillybus has only half the theoretical bandwidth of MXP, as Xillybus uses just 32-bits of an ACP port while MXP uses the full 64-bits of an HP port. If Xillybus were modified to use a 64-bit ACP port and two parallel 32-bit overlays were implemented, it would achieve a throughput close to MXP-V16, but with significantly fewer hardware resources (approximately 20% of the LUTs and 8% of the hard macros (BRAM and DSP) required by MXP on Zynq. Thus, the linear TM overlay represents a relatively efficient implementation when FPGA resources are limited, as would be the case when an accelerator was used with other major subsystems in an FPGA based SoC design.

The PCIe-based systems, RIFFA and DyRACT were also proposed for more high-performance accelerator systems. The DyRACT based system has a $5.3\times$ speed improvement when compared to the PCIe-Xillybus based system, and a 45% better throughput than the RIFFA based system, achieving a throughput of 1892 MB/s on average. There is a need to develop a full-duplex system for RIFFA based system, which will not only reduce its BRAM utilization, but also potentially improve the throughput with parallel programming method. The DyRACT based system appears to be the most promising

overlay accelerator, as it provides the best performance and a simple programming model to avoid overlay reconfiguration.

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