



# DW\_fifoctl\_s2\_sf

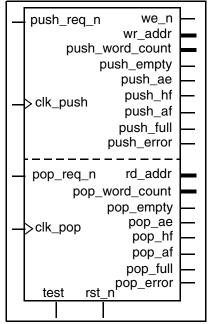
## Synchronous (Dual-Clock) FIFO Controller with Static Flags

Version, STAR and Download Information: IP Directory

### **Features and Benefits**

## **Revision History**

- Fully registered synchronous flag output ports
- Single clock cycle push and pop operations
- Separate status flags for each clock system
- FIFO empty, half full, and full flags
- FIFO push error (overflow) and pop error (underflow) flags
- Parameterized word depth
- Parameterized almost full and almost empty flag thresholds
- Interfaces to common hard macro or compiled ASIC dual-port synchronous RAMs
- Provides minPower benefits with the DesignWare-LP license (Get the minPower version of this datasheet.)



## **Description**

DW\_fifoctl\_s2\_sf is a dual independent clock FIFO RAM controller. It is designed to interface with a dual-port synchronous RAM.

The RAM must have:

- A synchronous write port and an asynchronous read port, or
- A synchronous write port and a synchronous read port (clocks must be independent).

The FIFO controller provides address generation, write-enable logic, flag logic, and operational error detection logic. Parameterizable features include FIFO depth (up to 24 address bits or 16,777,216 locations), almost empty level, almost full level, level of error detection, and type of reset (either asynchronous or synchronous). You specify these parameters when the controller is instantiated in the design.

Table 1-1 Pin Description

Pin Name	Width	Direction	Function
clk_push	1 bit	Input	Input clock for push interface
clk_pop	1 bit	Input	Input clock for pop interface
rst_n	1 bit	Input	Reset input, active low
push_req_n	1 bit	Input	FIFO push request, active low

Table 1-1 Pin Description (Continued)

Pin Name	Width	Direction	Function
pop_req_n	1 bit	Input	FIFO pop request, active low
we_n	1 bit	Output	Write enable output for write port of RAM, active low
push_empty	1 bit	Output	FIFO empty <sup>a</sup> output flag synchronous to clk_push, active high
push_ae	1 bit	Output	FIFO almost empty <sup>a</sup> output flag synchronous to clk_push, active high (determined by <i>push_ae_lvl</i> parameter)
push_hf	1 bit	Output	FIFO half full <sup>a</sup> output flag synchronous to clk_push, active high
push_af	1 bit	Output	FIFO almost full <sup>a</sup> output flag synchronous to clk_push, active high (determined by <i>push_af_lvl</i> parameter)
push_full	1 bit	Output	FIFO full $^{a}$ output flag synchronous to clk_push, active high
push_error	1 bit	Output	FIFO push error (overrun) output flag synchronous to clk_push, active high
pop_empty	1 bit	Output	FIFO empty <sup>b</sup> output flag synchronous to clk_pop, active high
pop_ae	1 bit	Output	FIFO almost empty <sup>b</sup> output flag synchronous to clk_pop, active high (determined by pop_ae_lvl parameter)
pop_hf	1 bit	Output	FIFO half full <sup>b</sup> output flag synchronous to clk_pop, active high
pop_af	1 bit	Output	FIFO almost full <sup>b</sup> output flag synchronous to clk_pop, active high (determined by <i>pop_af_lvl</i> parameter)
pop_full	1 bit	Output	FIFO full <sup>b</sup> output flag synchronous to clk_pop, active high
pop_error	1 bit	Output	FIFO pop error (underrun) output flag synchronous to clk_pop, active high
wr_addr	ceil(log <sub>2</sub> [depth]) bit(s)	Output	Address output to write port of RAM
rd_addr	ceil(log <sub>2</sub> [depth]) bit(s)	Output	Address output to read port of RAM
push_word_count	ceil(log <sub>2</sub> [depth+1]) bit(s)	Output	Words in FIFO (as perceived by the push/pop interface)
pop_word_count	ceil(log <sub>2</sub> [depth+1]) bit(s)	Output	Words in FIFO (as perceived by the push/pop interface)
test	1 bit	Input	Active high, test input control for inserting scan test lock-up latches

a. As perceived by the push interface.

b. As perceived by the pop interface.

## **Table 1-2** Parameter Description

Parameter	Values	Description	
depth	4 to 2 <sup>24</sup> Default: 8	Number of words that can be stored in FIFO	
push_ae_lvl	1 to depth – 1 Default: 2	Almost empty level for the push_ae output port (the number of words in the FIFO at or below which the push_ae flag is active)	
push_af_lvl	1 to <i>depth</i> – 1 Default: 2	Almost full level for the push_af output port (the number of empty memory locations in the FIFO at which the push_af flag is active)	
pop_ae_lvl	1 to <i>depth</i> – 1 Default: 2	Almost empty level for the pop_ae output port (the number of words in the FIFO at or below which the pop_ae flag is active)	
pop_af_lvl	1 to depth – 1 Default: 2	Almost full level for the pop_af output port (the number of empty memory locations in the FIFO at which the pop_af flag is active)	
err_mode	0 or 1 Default: 0	Error mode  0 = stays active until reset [latched],  1 = active only as long as error condition exists [unlatched]	
push_sync	1 to 3 Default: 2	Push flag synchronization mode  1 = single register synchronization from pop pointer,  2 = double register,  3 = triple register)	
pop_sync	1 to 3 Default: 2	Pop flag synchronization mode  1 = single register synchronization from push pointer,  2 = double register,  3 = triple register	
rst_mode	0 or 1 Default: 0	Reset mode 0 = asynchronous reset, 1 = synchronous reset)	
tst_mode	0 or 1 Default: 0	Test Mode 0 = test input not connected 1 = lock-up latches inserted for scan test	

## Table 1-3 Synthesis Implementations

Implementation Name	Function	License Feature Required
str	Synthesis model	DesignWare

Table 1-4 Simulation Models

Model	Function
DW03.DW_FIFOCTL_S2_SF_CFG_SIM	Design unit name for VHDL simulation
dw/dw03/src/DW_fifoctl_s2_sf_sim.vhd	VHDL simulation model source code
dw/sim_ver/DW_fifoctl_s2_sf.v	Verilog simulation model source code

#### Table 1-5 Push Interface Function Table

push_req_n	push_full	Action	push_error
0	0	Push operation	No
0	1	Overrun; incoming data dropped (no action other than error generation)	Yes
1	Х	No action	No

### Table 1-6 Pop Interface Function Table

pop_req_n	pop_empty	Action	pop_error
0	0	Pop operation	No
0	1	Underrun; (no action other than error generation)	Yes
1	Х	No action	No

# **Simulation Methodology**

DW\_fifoctl\_s2\_sf contains synchronization of Gray coded pointers between clock domains for which there are two methods for simulation.

- The first method is to use the simulation models, which emulate the RTL model, with no modeling of metastable behavior. Using this method requires no extra action.
- The second method (only available for Verilog simulation models) is to enable modeling of random skew between bits of the Gray coded pointers that traverse to and from each domain.

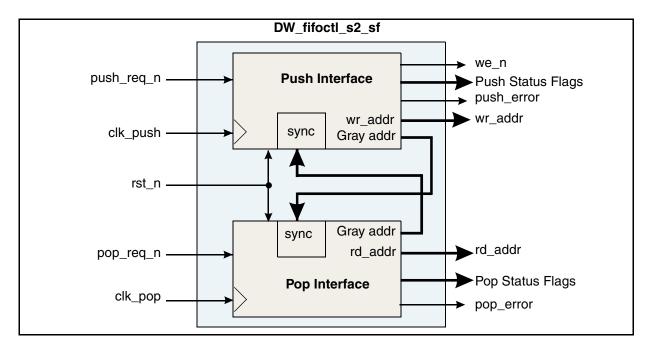
In order to use the second method, a Verilog preprocessing macro named DW\_MODEL\_MISSAMPLES must be defined in one of the following ways:

- Specify the Verilog preprocessing macro in Verilog code:
  - `define DW\_MODEL\_MISSAMPLES
- Or, include a command line option to the simulator, such as
   +define+DW\_MODEL\_MISSAMPLES (which is used for the Synopsys VCS simulator)

## **Block Diagram**

Figure 1-1 shows a simple block diagram of the FIFO controller.

Figure 1-1 DW\_fifoctl\_s2\_sf Block Diagram



# Writing to the FIFO (Push)

The wr\_addr and we\_n output ports of the FIFO controller provide the write address and synchronous write enable, respectively, to the RAM.

A push is executed when:

- The push\_req\_n input is asserted (LOW), and
- The push\_full flag is inactive (LOW)

at the rising edge of clk\_push.

Asserting push\_req\_n when push\_full is inactive causes the following to occur:

- The we\_n is asserted immediately, preparing for a write to the RAM on the next rising clock, and
- On the next rising edge of clk, wr\_addr is incremented (module depth).

Thus, the RAM is written, and wr\_addr (which always points to the address of the next word to be pushed) is incremented on the same rising edge of clk\_push—the first clock after push\_req\_n is asserted. This means that push\_req\_n must be asserted early enough to propagate through the FIFO controller to the RAM before the ensuing clock.

#### Write Errors

An error occurs if a push operation is attempted while the FIFO is full (as perceived by the push interface). That is, the push\_error output goes active if:

- The push\_req\_n input is asserted (LOW), and
- The push\_full flag is active (HIGH)

on the rising edge of clk\_push. When a push error occurs, we\_n stays inactive (HIGH) and the write address, wr\_addr, does not advance. After a push error, although a data word was lost at the time of the error, the FIFO remains in a valid full state and can continue to operate properly with respect to the data that was contained in the FIFO before the push error occurred.

# Reading from the FIFO (Pop)

The read port of the RAM must be asynchronous or be synchronous with its own clock (separate form the write port's clock). The rd\_addr output port of the DW\_fifoctl\_s2\_sf provides the read address to the RAM. rd\_addr always points to, thus prefetches, the next word of RAM read data to be popped.

A pop operation occurs when:

- The pop\_req\_n is asserted (LOW), and
- The pop\_empty flag is not active (LOW) (the FIFO is not empty)

at the rising edge of clk\_pop.

Asserting pop\_req\_n while pop\_empty is not active causes the internal read pointer to be incremented on the next rising edge of clk\_pop. Thus, for asynchronous read port memories, the RAM read data must be captured on the rising edge of clk\_pop following the assertion of pop\_req\_n. For synchronous read port memories, data must be captured on the rising edge of clk\_pop one cycle after the clk\_pop edge that directed the controller to pop.

#### Read Errors

An error occurs if a pop operation is attempted while the FIFO is empty (as perceived by the pop interface). That is, the pop\_error output goes active if:

- The pop\_req\_n input is active (LOW), and
- The pop\_empty flag is active (HIGH)

on the rising edge of clock\_pop. When a pop error occurs, the read address, rd\_addr does not advance. After a pop error the FIFO is still in a valid empty state and can continue to operate properly.

# **Memory Depth**

If the *depth* parameter is an integer power of two (i.e. 4, 8, 16, 32, etc.), then the FIFO controller reads from RAM addresses 0 through depth - 1 requiring a RAM depth of exactly depth.

If the *depth* parameter is an odd value (i.e. 5, 7, 9, 11, etc.), then the RAM depth must be (depth + 1) to allow addresses that range from 0 to depth. If depth is an even value but not an integer power of two, then the RAM depth must be (depth + 2) to allow addresses that range from 0 to depth + 1.

These restrictions are derived from the facts that,

- The memory depth must always be an even number to permit all transitions of the internal Gray coded pointers to be Gray.
- For non-power of two depth, the memory size must be at least one greater than *depth* to allow the pointer arithmetic to unambiguously differentiate between the empty and full states.

### Reset

The  $rst\_mode$  parameter selects whether reset is asynchronous ( $rst\_mode = 0$ ) or synchronous ( $rst\_mode = 1$ ). If asynchronous mode is selected, asserting rst\_n (setting it LOW) immediately causes:

- The internal address pointers to be set to 0, and
- The flags and error outputs to be initialized.

If synchronous mode is selected, after the assertion of rst\_n, at the rising edge of clk\_push the following are initialized:

- Write address pointer,
- Push flags, and
- The push\_error output

At the rising edge of clk\_pop, the following are initialized:

- The read address pointer,
- The pop flags, and
- The pop\_error output

### Metastability Issues Regarding Reset

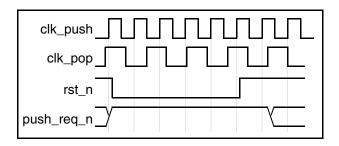
In order to avoid metastability upon reset, the assertion of rst\_n (LOW) should be maintained for at least three cycles of the slower of the two clock inputs, clk\_push and clk\_pop. During the assertion of rst\_n and for at least one cycle of clk\_push after rst\_n goes HIGH, push\_req\_n must be inactive (HIGH). In addition, it is recommended that the pop\_req\_n signal also be held inactive for at least one clock cycle of clk\_pop after the release of rst\_n.

Refer to Figure 1-2.



Since the one input that is critical to proper reset sequencing ( $push\_req\_n$ ) is in the domain of  $clk\_push$ , it is recommended that the reset input,  $rst\_n$ , should be synchronous to  $clk\_push$ .

Figure 1-2 Avoiding Metastability Upon Reset



#### Test

The synthesis parameter, <code>test\_mode</code>, controls the insertion of lock-up latches at the points where signals cross between the clock domains, <code>clk\_push</code> and <code>clk\_pop</code>. Lock-up latches are used to ensure proper cross-domain operation during the capture phase of scan testing in devices with multiple clocks. When <code>test\_mode=1</code>, lock-up latches will be inserted during synthesis and will be controlled by the input, <code>test</code>.

With test\_mode=1, the input, test, controls the bypass of the latches for normal operation where test=0 bypasses latches and test=1 includes latches. In order to assist DFT compiler in the use of the lock-up latches, use the 'set\_test\_hold 1 test\_mode' command before using the insert\_scan command.

When test\_mode=0 (which is its default value when not set in the design) no lock-up latches are inserted and the test input is not connected.



The insertion of lock-up latches requires the availability of an active low enable latch cell. If the target library does not have such a latch or if latches are not allowed (using dont\_use commands for instance), synthesis of this module with *test\_mode*=1 will fail.

# **Error Outputs and Flag Status**

The error outputs and flags are initialized as follows:

- The push\_empty, push\_ae, pop\_empty, and pop\_ae are initialized to 1 (HIGH), and,
- All other flags and the error outputs are initialized to 0 (LOW).

# **Synchronization Between Clock Systems**

Each interface (push and pop) operates synchronous to its own clock: clk\_push and clk\_pop. Each interface is independent, containing its own state machine and flag logic. The pop interface also has the primary read address counter and a synchronized copy of the write address counter. The push interface also has the primary write address counter and a synchronized copy of the read address counter. The two clocks may be asynchronous with respect to each other. The FIFO controller performs inter-clock synchronization in order for each interface to monitor the actions of the other. This enables the number of words in the FIFO at any given point in time to be determined independently by the two interfaces.

The only information that is synchronized across clock domain boundaries is the read or write address generated by the opposite interface. If an address is transitioning while being sampled by the opposite interface (e.g. wr\_addr sampled by clk\_pop), sampling uncertainty can occur. By Gray coding the address

values that are synchronized across clock domains, this sampling uncertainty is limited to a single bit. Single bit sampling uncertainty results in only one of two possible Gray coded addresses being sampled: the previous address or the new address. The uncertainty in the bit that is changing near a sampling clock edge directly corresponds to an uncertainty in whether the new value will be captured by the sampling clock edge or whether the previous value will be captured (and the new value may be captured by a subsequent sampling clock edge). Thus there are no errors in sampling Gray coded pointers, just a matter of whether a change of pointer value occurs in time to be captured by a given sampling clock edge or whether it must wait for the next sampling clock edge to be registered

### push\_sync and pop\_sync

The *push\_sync* and *pop\_sync* parameters determine the number of register stages (1, 2 or 3) used to synchronize the internal Gray code read pointer to <code>clk\_push</code> (for *push\_sync*) and internal Gray code write pointer to <code>clk\_pop</code> (for *pop\_sync*). A value of one (1) indicates single-stage synchronization; a value of two (2) indicates double-stage synchronization; a value of three (3) indicates triple-stage synchronization.

Single-stage synchronization is only adequate when using very slow clock rates (with respect to the target technology). There must be enough timing slack to allow metastable synchronization events to stabilize and propagate to the pointer and flag registers.



Since timing slack and selection of register types is very difficult to control and metastability characteristics of registers are extremely difficult to ascertain, single-stage synchronization is not recommended

Double-stage synchronization is desirable when using relatively high clock rates. It allows an entire clock period for metastable events to settle at the first stage before being cleanly clocked into the second stage of the synchronizer. Double-stage synchronization increases the latency between the two interfaces, resulting in flags that are less up to date with respect to the true state of the FIFO.

Triple-stage synchronization is desirable when using very high clock rates. It allows an entire clock period for metastable events to settle at the first stage before being clocked into the second stage of the synchronizer. Then, in the unlikely event that a metastable event propagates into the second stage, the output of the second stage is allowed to settle for another entire clock period before being clocked into the third stage. Triple-stage synchronization increases the latency between the two interfaces, resulting in flags that are less up to date with respect to the true state of the FIFO.

## **Empty to Not Empty Transitional Operation**

When the FIFO is empty, both push\_empty and pop\_empty are active (HIGH). During the first push (push\_req\_n active [LOW]), the rising edge of clk\_push writes the first word into the FIFO. The push\_empty flag is driven low.

The pop\_empty flag does not go low until one cycle (of clk\_pop) after the new internal Gray code write pointer has been synchronized to clk\_pop. This could be as long as 2 to 4 cycles (depending on the value of the *pop\_sync* parameter). Refer to the timing diagrams for more information. The system design should allow for this latency in the depth budgeting of the FIFO design.

## **Not Empty to Empty Transitional Operation**

When the FIFO is almost empty, both push\_empty and pop\_empty are inactive [LOW] and pop\_ae is active [HIGH]. During the final pop (pop\_req\_n active [LOW]), the rising edge of clk\_pop reads the last word out of the FIFO. The pop\_empty flag is driven high.

The push\_empty flag is not asserted (HIGH) until one cycle (of clk\_push) after the new internal Gray code read pointer has been synchronized to clk\_push. This could be as long as 2 to 4 cycles (depending on the value of the *push\_sync* parameter). Refer to the timing diagrams for more information.

You should be aware of this latency when designing the system data flow protocol.

## **Full to Not Full Transitional Operation**

When the FIFO is full, both push\_full and pop\_full are active [HIGH]. During the first pop (pop\_req\_n active [LOW]), the rising edge of clk\_pop reads the first word out of the FIFO. The pop\_full flag is driven low.

The push\_full flag does not go low until one cycle (of clk\_push) after the new internal Gray code read pointer has been synchronized to clk\_push. This could be as long as 2 to 4 cycles (depending on the value of the push\_sync parameter). Refer to the timing diagrams for more information.

You should be aware of this latency when designing the system data flow protocol.

### **Not Full to Full Transitional Operation**

When the FIFO is almost full, both <code>push\_full</code> and <code>pop\_full</code> are inactive [LOW], and <code>push\_af</code> is active [HIGH]. During the final push (<code>push\_req\_n</code> active [LOW]), the rising edge of <code>clk\_push</code> writes the last word into the FIFO. The <code>push\_full</code> flag is driven high.

The pop\_full flag is not asserted (HIGH) until one cycle (of clk\_pop) after the new internal Gray code write pointer has been synchronized to clk\_pop. This could be as long as 2 to 4 cycles (depending on the value of the pop\_sync parameter). Refer to the timing diagrams for more information.

You should allow for this latency in the depth budgeting of the FIFO design.

#### **Errors**

#### err mode

The *err\_mode* parameter determines whether the push\_error and pop\_error outputs remain active until reset (persistent) or for only the clock cycle in which the error is detected (dynamic).

When the *err\_mode* parameter is set to 0 at design time, persistent error flags are generated. When the *err\_mode* parameter is set to 1 at design time, dynamic error flags are generated.

### push\_error Output

The push\_error output signal indicates that a push request was seen while the push\_full output was active (HIGH) (an overrun error). When an overrun condition occurs, the write address pointer (wr\_addr) cannot advance, and the RAM write enable (we\_n) is not activated.

Therefore, a push request that would overrun the FIFO is, in effect, rejected, and an error is generated. This guarantees that no data already in the FIFO is destroyed (overwritten). Other than the loss of the data accompanying the rejected push request, FIFO operation can continue without reset.

### pop\_error Output

The pop\_error output signal indicates that a pop request was seen while the pop\_empty output signal was active (HIGH) (an underrun error). When an underrun condition occurs, the read address pointer (rd\_addr) cannot decrement, as there is no data in the FIFO to retrieve.

The FIFO timing is such that the logic controlling the pop\_req\_n input would not see the error until 'nonexistent' data had already been registered by the receiving logic. This is easily avoided if this logic can pay close attention to the pop\_empty output and thus avoid an underrun completely.

## **Controller Status Flag Outputs**

The two halves of the FIFO controller each have their own set of status flags indicating their separate view of the state of the FIFO. It is important to note that both the push interface and the pop interface perceives the state of fullness of the FIFO independently based on information from the opposing interface that is delayed up to three clock cycles for proper synchronization between clock domains.

The push interface status flags respond immediately to changes in state caused by push operations but there is delay between pop operations and corresponding changes of state of the push status flags. This delay is due to the latency introduced by the registers used to synchronize the internal Gray coded read pointer to clk\_push. The pop interface status flags respond immediately to changes in state caused by pop operations but there is delay between push operations and corresponding changes of state of the pop status flags. This delay is due to the latency introduced by the registers used to synchronize the internal Gray coded write pointer to clk\_pop.

Most status flags have a property which is potentially useful to the designed operation of the FIFO controller. These properties are described in the following explanations of the flag behaviors.

## push\_empty

The push\_empty output, active HIGH, is synchronous to the clk\_push input. push\_empty indicates to the push interface that the FIFO is empty. During the first push, the rising edge of clk\_push causes the first word to be written into the FIFO, and push\_empty is driven low.

The action of the last word being popped from a nearly empty FIFO is controlled by the pop interface. Thus, the push\_empty output is asserted only after the new internal Gray code read pointer (from the pop interface) is synchronized to clk\_push and processed by the status flag logic.

## Property of push\_empty

If push\_empty is active (HIGH) then the FIFO is truly empty. This property does not apply to pop\_empty.

### push\_ae

The push\_ae output, active HIGH, is synchronous to the clk\_push input. The push\_ae output indicates to the push interface that the FIFO is almost empty when there are no more than <code>push\_ae\_lvl</code> words currently in the FIFO to be popped as perceived at the push interface.

The *push\_ae\_lvl* parameter defines the almost empty threshold of the push interface independent of that of the pop interface. The <code>push\_ae</code> output is useful when it is desirable to push data into the FIFO in bursts (without allowing the FIFO to become empty).

### Property of push\_ae

If push\_ae is active (HIGH) then the FIFO has at least (*depth* – *push\_ae\_lvl*) available locations. Thus such status indicates that the push interface can safely and unconditionally push (*depth* – *push\_ae\_lvl*)

words into the FIFO. This property guarantees that such a 'blind push' operation will not overrun the FIFO.

### push\_hf

The push\_hf output, active HIGH, is synchronous to the clk\_push input, and indicates to the push interface that the FIFO has at least half of its memory locations occupied as perceived by the push interface.

### Property of push hf

If push\_hf is inactive (LOW) then the FIFO has at least half of its locations available. Thus such status indicates that the push interface can safely and unconditionally push (INT(depth/2)+1) words into the FIFO. This property guarantees that such a 'blind push' operation will not overrun the FIFO.

### push\_af

The push\_af output, active HIGH, is synchronous to the clk\_push input. push\_af indicates to the push interface that the FIFO is almost full when there are no more than *push\_af\_lvl* empty locations in the FIFO as perceived by the push interface.

The *push\_af\_lvl* parameter defines the almost full threshold of the push interface independent of the pop interface. The <code>push\_af</code> output is useful when more than one cycle of advance warning is needed to stop the flow of data into the FIFO before it becomes full (to avoid a FIFO overrun).

#### Property of push\_af

If push\_af is inactive (LOW) then the FIFO has at least (*push\_af\_lvl*+1) available locations. Thus such status indicates that the push interface can safely and unconditionally push (*push\_af\_lvl*+1) words into the FIFO. This property guarantees that such a 'blind push' operation will not overrun the FIFO.

### push\_full

The push\_full output, active HIGH, is synchronous to the clk\_push input. The push\_full output indicates to the push interface that the FIFO is full. During the final push, the rising edge of clk\_push causes the last word to be pushed, and push\_full is asserted.

The action of the first word being popped from a full FIFO is controlled by the pop interface. Thus, the <code>push\_full</code> output goes low only after the new internal Gray code read pointer from the pop interface is synchronized to <code>clk\_push</code> and processed by the status flag state logic.

#### pop\_empty

The pop\_empty output, active HIGH, is synchronous to the clk\_pop input. pop\_empty indicates to the pop interface that the FIFO is empty as perceived by the pop interface. The action of the last word being popped from a nearly empty FIFO is controlled by the pop interface. Thus, the pop\_empty output is asserted at the rising edge of clk\_pop that causes the last word to be popped from the FIFO.

The action of pushing the first word into an empty FIFO is controlled by the push interface. That means pop\_empty goes low only after the new internal Gray code write pointer from the push interface is synchronized to clk\_pop and processed by the status flag state logic.

#### pop\_ae

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The pop\_ae output, active HIGH, is synchronous to the clk\_pop input. pop\_ae indicates to the pop interface that the FIFO is almost empty when there are no more than *pop\_ae\_lvl* words currently in the FIFO to be popped as perceived by the pop interface.

The *pop\_ae\_lvl* parameter defines the almost empty threshold of the pop interface independent of the push interface. The pop\_ae output is useful when more than one cycle of advance warning is needed to stop the popping of data from the FIFO before it becomes empty (to avoid a FIFO underrun).

### Property of pop\_ae

If pop\_ae is inactive (LOW) then there are at least (pop\_ae\_lvl +1) words in the FIFO. Thus such status indicates that the pop interface can safely and unconditionally pop (pop\_ae\_lvl +1) words out of the FIFO. This property guarantees that such a 'blind pop' operation will not underrun the FIFO.

### pop\_hf

The pop\_hf output, active HIGH, is synchronous to the clk\_pop input. pop\_hf indicates to the pop interface that the FIFO has at least half of its memory locations occupied as perceived by the pop interface.

### Property of pop\_hf

If pop\_hf is active (HIGH) then at least half of the words in the FIFO are occupied. Thus such status indicates that the pop interface can safely and unconditionally pop INT((*depth*+1)/2) words out of the FIFO. This property guarantees that such a 'blind pop' operation will not underrun the FIFO.

### pop\_af

The pop\_af output, active HIGH, is synchronous to the clk\_pop input. The pop\_af output indicates to the pop interface that the FIFO is almost full when there are no more than *pop\_af\_lvl* empty locations in the FIFO as perceived by the pop interface.

The pop\_af\_lvl parameter defines the almost full threshold of the pop interface independent of that of the pop interface. The pop\_af output is useful when it is desirable to pop data out of the FIFO in bursts (without allowing the FIFO to become empty).

### Property of pop\_af

If pop\_af is active (HIGH) then there are at least ( $depth - pop_af_lvl$ ) words in the FIFO. Thus such status indicates that the pop interface can safely and unconditionally pop ( $depth - pop_af_lvl$ ) words out of the FIFO. This property guarantees that such a 'blind pop' operation will not underrun the FIFO.

## pop\_full

The pop\_full output, active HIGH, is synchronous to the clk\_pop input. pop\_full indicates to the pop interface that the FIFO is full as perceived by the pop interface. The action of popping the first word out of a full FIFO is controlled by the pop interface. Thus, the pop\_full output goes low at the rising edge of the clk\_pop that causes the first word to be popped.

The action of the last word being pushed into a nearly full FIFO is controlled by the push interface. This means the pop\_full output is asserted only after the new write pointer from the pop interface is synchronized to clk\_pop and processed by the status flag state logic.

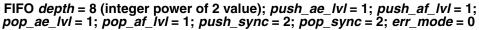
#### Property of pop full

If pop\_full is active (HIGH) then the FIFO is truly full. This property does not apply to push\_full.

# **Timing Waveforms**

This section shows timing diagrams for various conditions of DW\_fifoctl\_s2\_sf.

Figure 1-3 Push Timing Waveforms



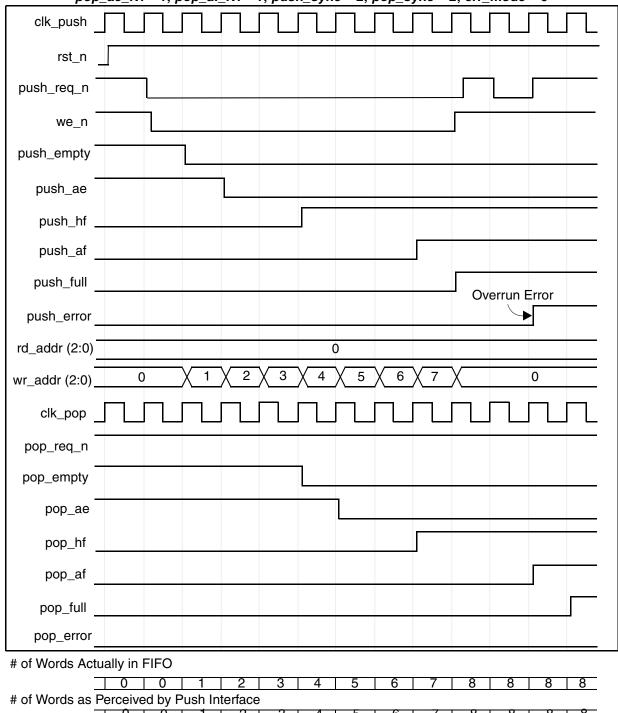
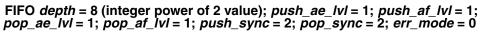


Figure 1-4 Pop Timing Waveforms



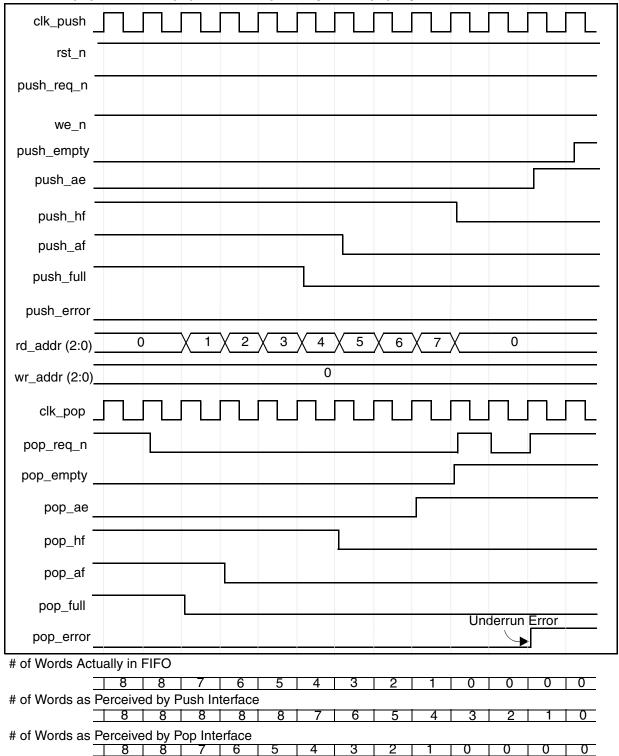
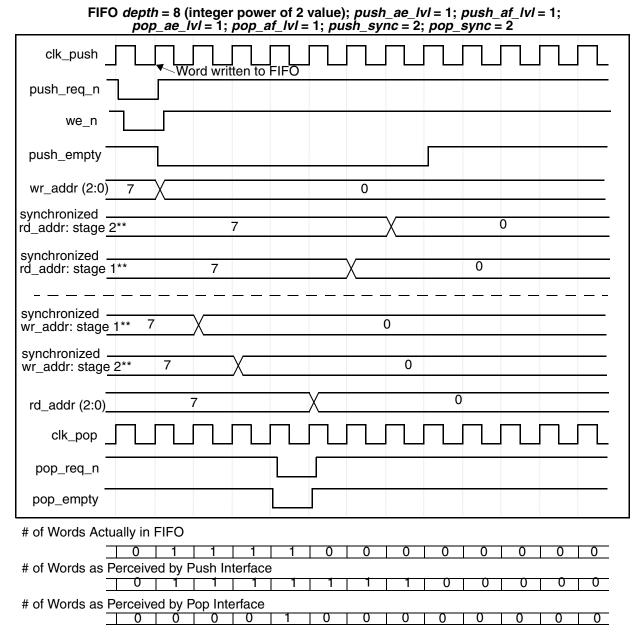
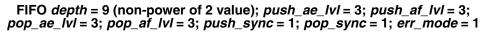


Figure 1-5 FIFO Single Word Timing Waveforms with Double-Stage Synchronization



<sup>\*\*</sup> Note: For clarity in showing the operation, the synchronized addresses inside the controller are not shown as being Gray coded. In the actual synthetic design Gray coded addresses are used in synchronization across clock boundaries.

Figure 1-6 FIFO Non-power of 2 depth Push Timing Waveforms



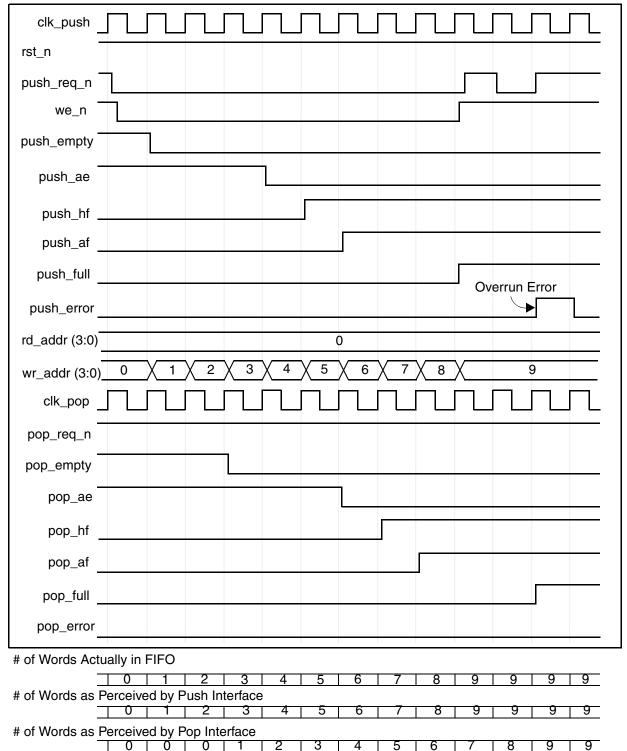


Figure 1-7 FIFO Non-power of 2 depth Pop Timing Waveforms

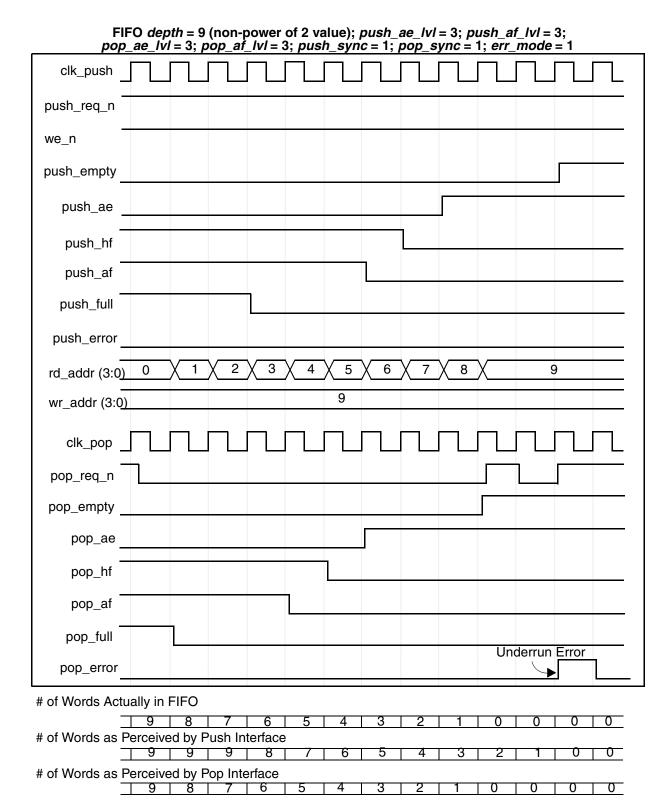
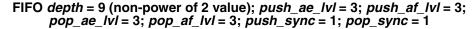
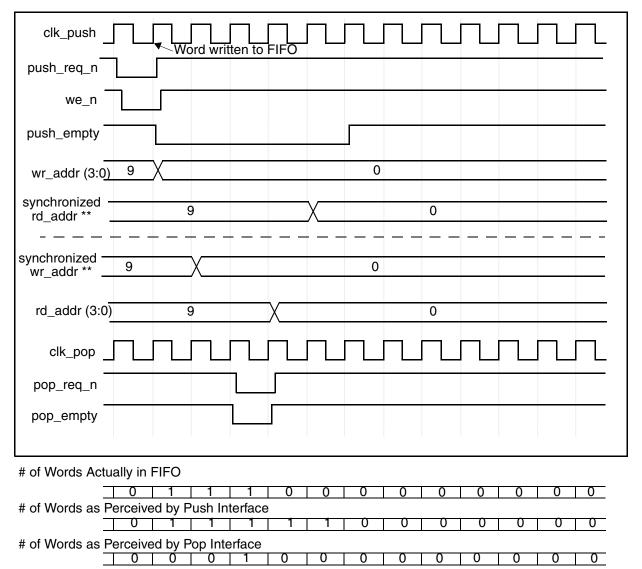


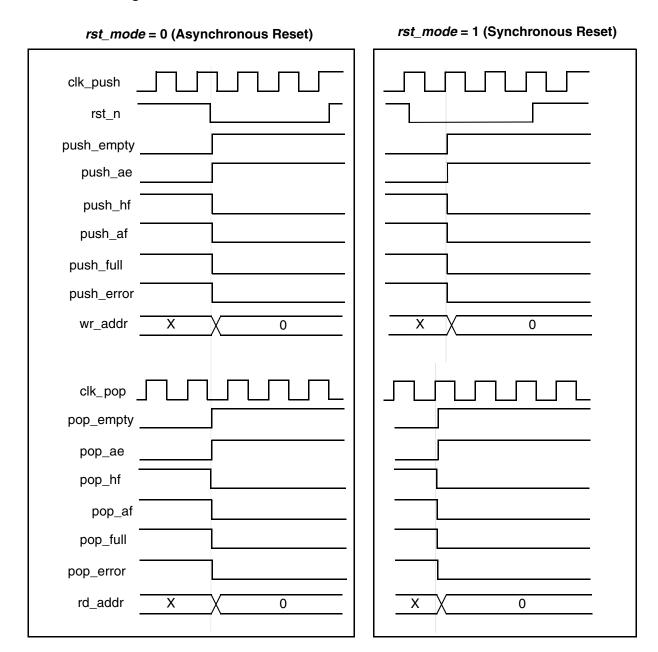
Figure 1-8 FIFO Single Word Timing Waveforms With Single-stage Synchronization





<sup>\*\*</sup> Note: For clarity in illustrating operation, the synchronized addresses inside the controller are not shown as being Gray coded. In the actual synthetic design Gray coded addresses are used in synchronization across clock boundaries.

Figure 1-9 **Reset Timing Waveforms** 



# **Related Topics**

- Memory FIFO Overview
- DesignWare Building Block IP Documentation Overview

# **HDL Usage Through Component Instantiation - VHDL**

```
library IEEE, DWARE, DWARE;
use IEEE.std logic 1164.all;
use DWARE.DWpackages.all;
use DWARE.DW foundation comp.all;
entity DW_fifoctl_s2_sf_inst is
                     : INTEGER := 8; inst push ae lvl : INTEGER := 2;
  generic (inst_depth
          inst_push_af_lvl : INTEGER := 2;
          inst_pop_ae_lv1 : INTEGER := 2;
          inst_pop_af_lvl : INTEGER := 2;
          inst_err_mode : INTEGER := 0;
          inst_push_sync : INTEGER := 2;
          inst_pop_sync : INTEGER := 2;
          inst_rst_mode
                        : INTEGER := 0;
          inst tst mode : INTEGER := 0);
 port (inst_clk_push : in std_logic;
                                       inst clk pop : in std logic;
       inst_rst_n : in std_logic;
                                       inst_push_req_n : in std_logic;
       inst_pop_req_n : in std_logic;    we_n_inst : out std_logic;
       push_empty_inst : out std_logic;  push_ae_inst : out std_logic;
       push_hf_inst : out std_logic; push_af_inst : out std_logic;
       push_full_inst : out std_logic; push_error_inst : out std_logic;
       pop_empty_inst : out std_logic; pop_ae_inst
                                                   : out std_logic;
       pop hf inst
                     : out std_logic; pop_af_inst
                                                      : out std logic;
       pop_full_inst : out std_logic;
                                       pop_error_inst : out std_logic;
       wr_addr_inst : out std_logic_vector(bit_width(inst_depth)-1
                                                          downto 0);
       rd_addr_inst : out std_logic_vector(bit_width(inst_depth)-1
       push word count inst: out std logic vector(bit width(inst depth+1)-1
                                                          downto 0);
       pop_word_count_inst : out std_logic_vector(bit_width(inst_depth+1)-1
                                                          downto 0);
                          : in std_logic
       inst_test
                                           );
end DW fifoctl s2 sf inst;
architecture inst of DW_fifoctl_s2_sf_inst is
begin
  -- Instance of DW fifoctl s2 sf
 U1 : DW_fifoctl_s2_sf
   generic map (depth => inst depth,
                                  push_ae_lvl => inst_push_ae_lvl,
               push_af_lvl => inst_push_af_lvl,
               pop_ae_lvl => inst_pop_ae_lvl,
               pop af lvl => inst pop af lvl,
                                             err mode => inst err mode,
               rst_mode => inst_rst_mode, tst_mode => inst_tst_mode )
   rst_n => inst_rst_n, push_req_n => inst_push_req_n,
```

```
pop_req_n => inst_pop_req_n, we_n => we_n_inst,
              push_empty => push_empty_inst,    push_ae => push_ae_inst,
              push_hf => push_hf_inst,    push_af => push_af_inst,
              push_full => push_full_inst,    push_error => push_error_inst,
              pop_empty => pop_empty_inst,
                                           pop_ae => pop_ae_inst,
             pop_hf => pop_hf_inst,     pop_af => pop_af_inst,
              pop_full => pop_full_inst,     pop_error => pop_error_inst,
              wr_addr => wr_addr_inst, rd_addr => rd_addr_inst,
              push_word_count => push_word_count_inst,
              pop_word_count => pop_word_count_inst, test => inst_test);
end inst:
-- pragma translate off
configuration DW_fifoctl_s2_sf_inst_cfg_inst of DW_fifoctl_s2_sf_inst is
  for inst
 end for; -- inst
end DW_fifoctl_s2_sf_inst_cfg_inst;
-- pragma translate_on
```

# **HDL Usage Through Component Instantiation - Verilog**

```
module DW fifoctl s2 sf inst(inst clk push, inst clk pop, inst rst n,
                             inst push req n, inst pop req n, we n inst,
                             push_empty_inst, push_ae_inst, push_hf_inst,
                             push af inst, push full inst, push error inst,
                             pop_empty_inst, pop_ae_inst, pop_hf_inst,
                             pop_af_inst, pop_full_inst, pop_error_inst,
                             wr addr inst, rd addr inst,
                             push_word_count_inst, pop_word_count_inst,
                             inst_test );
  parameter depth = 8;
  parameter push_ae_lvl = 2;
  parameter push_af_lv1 = 2;
  parameter pop ae lvl = 2;
  parameter pop_af_lvl = 2;
  parameter err mode = 0;
  parameter push sync = 1;
  parameter pop_sync = 1;
  parameter rst_mode = 0;
  parameter tst mode = 0;
  `define addr_width 3 // ceil(log2(depth))
  `define count_width 4 // ceil(log2(depth+1))
  input inst clk push;
  input inst_clk_pop;
  input inst_rst_n;
  input inst push req n;
  input inst_pop_req_n;
  output we n inst;
  output push empty inst;
  output push_ae_inst;
  output push_hf_inst;
  output push_af_inst;
  output push_full_inst;
  output push error inst;
  output pop_empty_inst;
  output pop_ae_inst;
  output pop_hf_inst;
  output pop_af_inst;
  output pop_full_inst;
  output pop_error_inst;
  output [`addr width-1: 0] wr addr inst;
  output [`addr_width-1 : 0] rd_addr_inst;
  output ['count width-1: 0] push word count inst;
  output [`count_width-1 : 0] pop_word_count_inst;
  input inst_test;
  // Instance of DW_fifoctl_s2_sf
```

```
DW_fifoctl_s2_sf #(depth, push_ae_lvl, push_af_lvl, pop_ae_lvl, pop_af_lvl,
                    err_mode, push_sync, pop_sync, rst_mode, tst_mode)
   U1 (.clk_push(inst_clk_push), .clk_pop(inst_clk_pop),
        .rst_n(inst_rst_n),
                             .push_req_n(inst_push_req_n),
        .pop_req_n(inst_pop_req_n), .we_n(we_n_inst),
        .push_empty(push_empty_inst),
                                       .push_ae(push_ae_inst),
        .push_hf(push_hf_inst), .push_af(push_af_inst),
        .push_full(push_full_inst), .push_error(push_error_inst),
        .pop_empty(pop_empty_inst),
                                     .pop_ae(pop_ae_inst),
        .pop_hf(pop_hf_inst), .pop_af(pop_af_inst),
        .pop_full(pop_full_inst), .pop_error(pop_error_inst),
        .wr_addr(wr_addr_inst), .rd_addr(rd_addr_inst),
        .push word count (push word count inst),
        .pop_word_count(pop_word_count_inst), .test(inst_test) );
endmodule
```

## **Revision History**

For notes about this release, see the *DesignWare Building Block IP Release Notes*.

For lists of both known and fixed issues for this component, refer to the STAR report.

For a version of this datasheet with visible change bars, click here.

Date	Release	Updates
December 2017	N-2017.09-SP2	<ul> <li>Added "Simulation Methodology" on page 4 to explain how to simulate synchronization of Gray coded pointers between clock domains</li> </ul>
October 2017	N-2017.09-SP1	■ Replaced the synthesis implementations in Table 1-3 on page 3 with the str implementation
		■ Added this Revision History table and the document links on this page

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