#### **Course Outline - 1st Half**



- Review Classical Feedback Control
- Review Vector/Matrix Theory
- State-Space Representations
- LTI Response, Matrix Exponential
- Transfer Functions & Eigenvalues
- Frequency-Domain Analysis
- Harmonic & Impulse Responses
- Pole Placement
- Controllability

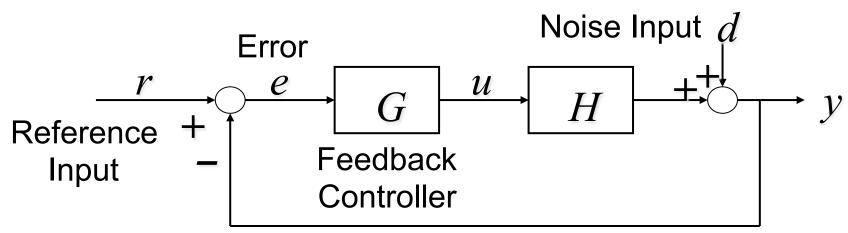
### **Course Summary - I**



# 1. For the SISO open-loop transfer function *GH*, the closed-loop is characterized by:

$$T(s) = \left(\frac{G(s)H(s)}{1+G(s)H(s)}\right) \Rightarrow \begin{array}{l} \text{Complementary Sensitivity} \\ \text{(Tracking Performance)} \end{array}$$

$$S(s) = \left(\frac{1}{1+G(s)H(s)}\right) \Rightarrow \begin{array}{l} \text{Sensitivity Function} \\ \text{(Disturbance Rejection)} \end{array}$$



# **Course Summary - II**



The SISO Characteristic Equation (CE) is

$$1 + G(s)H(s) = 0$$

3. Linear time invariant (LTI) systems can be modeled with first order differential equations of the form

$$\dot{\mathbf{x}}_{[N\times 1]} = \mathbf{A}_{[N\times N][N\times 1]} \mathbf{x} + \mathbf{B}_{[N\times M][M\times 1]}$$
 State Equations

$$\mathbf{y} = \mathbf{C}_{[P \times N][N \times 1]} + \mathbf{D}_{[P \times M][M \times 1]}$$
Output Equations

# **Course Summary - III**



4. State-space <u>realizations</u> of a dynamic system are <u>not unique</u>, but the <u>input-output relationships are invariant</u>. The input-output transfer functions are expressed by the matrix

$$\mathbf{H}(s) = \mathbf{C}[s\mathbf{I} - \mathbf{A}]^{-1}\mathbf{B} + \mathbf{D}$$

# **Course Summary - IV**



 Nonlinear state-space systems may be linearized about some operating point using the Jacobian matrices

$$\dot{\mathbf{x}} = \mathbf{f}(\mathbf{x}, \mathbf{u}) \Rightarrow \dot{\mathbf{x}} = \mathbf{A}(\mathbf{x}_0, \mathbf{u}_0)\mathbf{x} + \mathbf{B}(\mathbf{x}_0, \mathbf{u}_0)\mathbf{u}$$

$$\mathbf{y} = \mathbf{g}(\mathbf{x}, \mathbf{u}) \Rightarrow \mathbf{y} = \mathbf{C}(\mathbf{x}_0, \mathbf{u}_0)\mathbf{x} + \mathbf{D}(\mathbf{x}_0, \mathbf{u}_0)\mathbf{u}$$

• The N Eigenvalues  $(\lambda)$  and the corresponding Eigenvectors  $(\mathbf{v})$  are solutions of  $\mathbf{A} \quad \mathbf{v} = \lambda \quad \mathbf{v}$   $[N \times N][N \times 1] \quad [1 \times 1][N \times 1]$ 

$$(\mathbf{A} - \lambda_n \mathbf{I}) \mathbf{v}_n = 0 \qquad n = 1, \dots, N$$

# **Course Summary - V**



7. The time-domain solution of the State-Space differential equation is

$$\mathbf{x}(t) = e^{\mathbf{A}(t-t_0)} \mathbf{x}(t_0) + \int_{t_0}^{t} \left[ e^{\mathbf{A}(t-\tau)} \mathbf{B} \right] \mathbf{u}(\tau) d\tau$$
Response due to Initial Conditions

Forced Response

8. A key part of this solution is the <u>state</u> <u>transition matrix</u>, or <u>matrix exponential</u> which is related to the matrix **A** through the inverse Laplace transform

$$\Phi(t,t_0) = e^{\mathbf{A}(t-t_0)} = L^{-1}\left\{ \left[ s\mathbf{I} - \mathbf{A} \right]^{-1} \right\}$$

# **Course Summary - VI**



9. Transfer functions can be transformed into a state-space representation through one of the <u>canonical forms</u>.

$$\begin{bmatrix} \dot{x}_1(t) \\ \dot{x}_2(t) \\ \vdots \\ \dot{x}_{n-1}(t) \\ \dot{x}_n(t) \end{bmatrix} = \begin{bmatrix} 0 & 1 & 0 & \cdots & 0 \\ 0 & 0 & 1 & \cdots & 0 \\ \vdots & \vdots & \vdots & \ddots & \vdots \\ 0 & 0 & 0 & \cdots & 1 \\ -a_0 & -a_1 & -a_2 & \cdots & -a_{n-1} \end{bmatrix} \begin{bmatrix} x_1(t) \\ x_2(t) \\ \vdots \\ x_{n-1}(t) \\ x_n(t) \end{bmatrix} + \begin{bmatrix} 0 \\ 0 \\ \vdots \\ 0 \\ 1 \end{bmatrix} u(t)$$

$$[y(t)] = [b_0 \quad b_1 \quad \cdots \quad b_m \quad 0] \begin{bmatrix} x_1(t) \\ x_2(t) \\ \vdots \\ x_{n-1}(t) \\ x_n(t) \end{bmatrix} + [0]u(t)$$

# **Course Summary - VII**



10. The <u>poles</u> of the system are the <u>eigenvalues</u> of the characteristic equation

$$|\mathbf{A} - \lambda \mathbf{I}| = 0$$

which are also the N roots of

$$|s\mathbf{I} - \mathbf{A}| = s^N + a_{N-1}s^{N-1} + \dots + a_1s + a_0 = 0$$

$$(s-s_1)(s-s_2)\cdots(s-s_N) = \prod_{i=1}^{N} (s-s_i) = 0$$

# **Course Summary - VIII**



11. The pole locations in the complex splane provide information about the stability and time response of the open-loop system

Asymptotically Stable  $\Rightarrow$  RE $(\lambda_n) < 0 \ \forall n$ Marginally Stable  $\Rightarrow$  RE $(\lambda_n) = 0$  for at least one nUnstable  $\Rightarrow$  RE $(\lambda_n) > 0$  for at least one nOscillatory  $\Rightarrow$  Complex poles near IM- axis

# **Course Summary - IX**



- 12. The steady-state response of an asymptotically-stable LTI system is related to the magnitude and phase of  $\mathbf{H}(j\omega)$ .
  - The steady-state response amplitude is the amplitude of the input multiplied by  $|\mathbf{H}(j\omega)|$ .
  - The phase shift between the input and the output is equal to the phase of  $\mathbf{H}(j\omega)$ .

### **Course Summary - X**



# 13. The impulse response function is related to the transfer function

$$h_{pm}(t) = L^{-1} \{ H_{pm}(s) \} \qquad \Leftrightarrow \qquad H_{pm}(s) = L \{ h_{pm}(t) \}$$

14. Once the impulse response is known for an LTI system, the response to <u>any arbitrary input</u> can be determined from the convolution integral:  $y_p(t) = \sum_{n=0}^{M} \left( \int_{0}^{t} h_{pm}(t-\tau) u_m(\tau) d\tau \right)$ 

ME 5554 / AOE 5754 / ECE 5754

# **Course Summary - XI**



15. For an LTI state-space system, the closed-loop poles can be "placed" using full-state feedback:

$$\mathbf{u}(t) = -\mathbf{G}_{[M \times N]} \mathbf{x}(t)$$

$$[M \times 1]$$

and the resulting closed-loop system does not have any inputs:

$$\dot{\mathbf{x}}(t) = \underbrace{\left[\mathbf{A} - \mathbf{B}\mathbf{G}\right]}_{\text{Closed-Loop}} \mathbf{x}(t)$$
State Matrix

# **Course Summary - XII**



16. For a <u>Single-Input</u> state-space system, we can use the "brute force" method of pole placement which equates coefficients of the desired and actual closed-loop CE's:

$$|s\mathbf{I} - \mathbf{A}_c| = \underbrace{s^N + a_{N-1}s^{N-1} + \dots + a_1s + a_0}_{\text{Desired CL characteristic equation}}$$

$$|s\mathbf{I} - \mathbf{A} + \mathbf{b}\mathbf{g}| = \underline{s^N + \overline{a}_{N-1}(\mathbf{A}, \mathbf{b}, \mathbf{g})s^{N-1} + \dots + \overline{a}_1(\mathbf{A}, \mathbf{b}, \mathbf{g})s + \overline{a}_0(\mathbf{A}, \mathbf{b}, \mathbf{g})}$$
Actual CL characteristic equation

#### **Course Goals**



# Remember the original course goals from the first meeting (L1):

- Represent multi-input-multi-output (MIMO) dynamic systems using <u>state-space</u> models
- Solve for the time response of a linear dynamic system and relate the response to the <u>state-space</u> system description
- Design linear feedback control systems using frequency domain, state estimation, and pole placement techniques

#### **Midterm Review**



#### Midterm Rules

- Thursday October 15, 11:00-12:15
- You may NOT use: laptop computers, notes, books
- You may bring 1 sheet of 8.5"x11" paper with anything you want printed or written on it (both sides)
- Any required Laplace transforms will be provided
- Exam will cover material up to and including full state feedback for SISO

# Fair game for the Midterm Exam



- ALGEBRA!!!!!
- Transfer functions
- Basic matrix operations
- Eigenvalues & eigenvectors
- State-space representations
- State transition matrix
- Impulse response
- Time domain response
- Harmonic response
- Poles, zeros, and stability

#### **Transfer Functions**



- Convert higher-order differential equations into Laplace-domain transfer functions (L1)
- Convert a state-space model into a Laplace domain transfer function matrix (L6)





- Add and multiply matrices & vectors (L2)
- Compute the inverse of a matrix (up to 3x3) by hand (L2)
- Compute the determinant of a matrix (up to 3x3) by hand (L2)
- Determine whether a matrix is singular or non-singular (L2)





- Define the eigenvalue problem for any square matrix (L2)
- Solve the eigenvalue problem (up to 3x3) by hand for the eigenvalues and corresponding eigenvectors (L2)





- Write one or more higher-order diffeq's in first-order form (L3)
- Place first-order diffeq's into statespace matrix form (L3)
- Write the output equations in statespace matrix form (L3)
- Place a SISO transfer function into state-space canonical form (L6)





- Understand how state-space representations are not unique, but the input/output relationships are invariant (L3)
- Transform a state-space model with a non-singular transformation (L3)
- Understand the difference between LTI, LTV, and NL systems (L3)

#### **State Transition Matrix**



- Use the series definition of the matrix exponential (L4)
- Use basic properties of the matrix exponential (state transition matrix) (L4)
- Compute the state transition matrix using Laplace transforms (L5)
- Use eigenvectors to compute the state transition matrix (L5)

# Impulse Response



- Understand the relationship between an impulse response and a transfer function (L7)
- Understand how to use an impulse response to compute the time response to an arbitrary excitation

# **Time Domain Response**



- Solve for a forced response and/or the IC response using the state transition matrix (L4)
- Solve for a forced response using the convolution integral (L7)
- Understand that the total response is a summation of multiple modes (L9)

# **Harmonic Response**



- Compute the magnitude and phase of a Laplace domain transfer function at a given frequency (L7)
- Compute the steady-state time response to a harmonic excitation (L7)
- Understand superposition



# Poles, Zeros, and Stability

- Obtain the SISO closed-loop characteristic equation (L1)
- Solve the quadratic equation
- Determine the poles of a statespace model (L6, L7)
- Understand the difference between asymptotic stability, marginal stability, and instability (L7)





- Understand the concepts of bandwidth and time constant and how they are related (L7)
- Understand how the time response is related to pole locations (L8)

#### **Pole Placement**



You should know how to:

 Place the poles for a single-input state-space system using the "brute force" method. (L8)