## Electrical Network and Device Modelling Assignment 2

## David Lynch - 758863, Daniel Landgraf, Zixiang Ren

April 6, 2017

1. (a)

(b)

$$Z_{eq1} = \left(\frac{1}{Z_{R_1}} + \frac{1}{Z_{C_1}}\right)^{-1}$$

$$= \left(\frac{1}{R_1} + \frac{1}{\frac{1}{sC_1}}\right)^{-1}$$

$$= \left(\frac{1}{1.25 \times 10^6} + s \cdot 4 \times 10^{-12}\right)^{-1} \Omega$$

$$= \left(\frac{1 + 1.25 \times 10^6 \cdot (s \cdot 4 \times 10^{-12})}{1.25 \times 10^6}\right)^{-1} \Omega$$

$$= \frac{1.25 \times 10^6}{1 + 1.25 \times 10^6 \cdot (s \cdot 4 \times 10^{-12})} \Omega$$

$$= \frac{2.5 \times 10^{11}}{s + 2 \times 10^5} \Omega$$

(c)

$$\begin{split} Z_{eq2} &= \left(\frac{1}{Z_{R_2}} + \frac{1}{Z_{C_2}}\right)^{-1} \\ &= \left(\frac{1}{R_2} + \frac{1}{\frac{1}{sC_2}}\right)^{-1} \\ &= \left(\frac{1}{5 \times 10^6} + s \cdot 16 \times 10^{-12}\right)^{-1} \Omega \\ &= \left(\frac{1 + 5 \times 10^6 \cdot (s \cdot 16 \times 10^{-12})}{5 \times 10^6}\right)^{-1} \Omega \\ &= \frac{5 \times 10^6}{1 + 5 \times 10^6 \cdot (s \cdot 16 \times 10^{-12})} \Omega \\ &= \frac{6.25 \times 10^{10}}{s + 1.25 \times 10^4} A \end{split}$$

(d) Using voltage division:

$$\begin{split} V_o(s) &= V_{in}(s) \cdot \frac{Z_{eq2}}{Z_{eq1} + Z_{eq2}} \\ &= \frac{10}{s} \cdot \frac{\frac{6.25 \times 10^{10}}{s + 1.25 \times 10^4}}{\frac{6.25 \times 10^{10}}{s + 1.25 \times 10^4}} \cdot \frac{2.5 \times 10^{11}}{s + 2 \times 10^5} V \\ &= \frac{10}{s} \cdot \frac{\frac{6.25 \times 10^{10}}{s + 1.25 \times 10^4} + \frac{2.5 \times 10^{10}}{s + 1.25 \times 10^4}}{\frac{(6.25 \times 10^{10})(s + 2 \times 10^5) + (2.5 \times 10^{11})(s + 1.25 \times 10^4)}{(s + 1.25 \times 10^4) \cdot (s + 2 \times 10^5)}} V \\ &= \frac{10}{s} \cdot \frac{6.25 \times 10^{10}}{s + 1.25 \times 10^4} \cdot \frac{(s + 1.25 \times 10^4) \cdot (s + 2 \times 10^5)}{(6.25 \times 10^{10})(s + 2 \times 10^5) + (2.5 \times 10^{11})(s + 1.25 \times 10^4)}} V \\ &= \frac{6.25 \times 10^{11} \cdot (s + 1.25 \times 10^4) \cdot (s + 2 \times 10^5)}{3.125 \times 10^{11} \cdot s \cdot (s + 1.25 \times 10^4) \cdot (s + 5 \times 10^4)} V \\ &= \frac{2 \cdot (s + 2 \times 10^5)}{s \cdot (s + 5 \times 10^4)} V \\ &= \frac{2s + 4 \times 10^5}{s^2 + 5 \times 10^4 \cdot s} V \end{split}$$

Using Partial fractions to continue.

$$\frac{2s + 4 \times 10^5}{s^2 + 5 \times 10^4 \cdot s} = \frac{A}{s} + \frac{B}{50000}$$

$$A(s + 50000) + Bs = 2(s + 200000)$$

$$(s = 0) : 50000A = 400000$$

$$\implies A = 8$$

$$(s = -50000) : -50000B = 300000$$

$$B = -6$$

$$\implies V_o(s) = \frac{8}{s} - \frac{6}{s + 50000} V$$

Using inverse Laplace transform to find  $v_o(t)$ .

$$v_o(t) = \mathcal{L}^{-1}[V_o(s)]$$

$$= \mathcal{L}^{-1} \left[ \frac{8}{s} - \frac{6}{s + 50000} \right] V$$

$$= (8 - 6e^{-50000t})u(t) V$$

(e) Using Ohm's Law in frequency domain:

$$V_{in}(s) = Z_{eq} \cdot I_o(s)$$

$$= (Z_{eq1} + Z_{eq2}) \cdot I_o(s)$$

$$\implies I_o(s) = \frac{V_{in}(s)}{Z_{eq1} + Z_{eq2}}$$

$$= \frac{\frac{10}{s}}{\frac{2.5 \times 10^{11}}{s + 2 \times 10^5} + \frac{6.25 \times 10^{10}}{s + 1.25 \times 10^4}} A$$

$$= \frac{\frac{10}{s}}{\frac{3.125 \times 10^{11} \cdot (s + 50000)}{(s + 12500) \cdot (s + 200000)}} A$$

$$= \frac{(s + 12500) \cdot (s + 200000)}{3.125 \times 10^{10} \cdot s \cdot (s + 50000)} A$$

After a partial fractions expansion (add working in here)

$$I_o(s) = \frac{9}{2500000 \cdot (s + 50000)} + \frac{1}{625000 \cdot s} + \frac{1}{3.125 \times 10^{10}} A$$

Using inverse Laplace transform to find  $i_o(t)$ .

$$\begin{split} i_o(t) &= \mathcal{L}^{-1} \left[ I_o(s) \right] \\ &= \mathcal{L}^{-1} \left[ \frac{9}{2500000 \cdot (s + 50000)} + \frac{1}{625000 \cdot s} + \frac{1}{3.125 \times 10^{10}} \right] \\ &= 3.6 \times 10^{-6} e^{-50000t} \cdot u(t) + 1.6 \times 10^{-6} \cdot u(t) + 3.2 \times 10^{-11} \delta(t) \; A \end{split}$$

If we assume t > 0 only then:

$$i_o(t) = (3.6 \cdot e^{-50000t} + 1.6) \,\mu A$$

2. (a) Using KCL at the node between capacitor, resistor and inductor:  $\Sigma i_{out} = 0$ 

$$\frac{V_o(s) - \frac{5}{s}}{2.5s} + \frac{V_o(s)}{2} + V_o(s) \cdot s \cdot 0.1 = 0$$

$$\Rightarrow \frac{2 \cdot V_o(s)}{5s} - \frac{2}{s^2} + \frac{V_o(s)}{2} + \frac{s \cdot V_o(s)}{10} = 0$$

$$\Rightarrow V_o(s) \left(\frac{2}{5s} + \frac{s}{10} + \frac{1}{2}\right) = \frac{2}{s^2}$$

$$\Rightarrow V_o(s) \left(\frac{s^2}{10} + \frac{s}{2} + \frac{2}{5}\right) = \frac{2}{s}$$

$$\Rightarrow V_o(s) \cdot (s^2 + 5s + 4) = \frac{20}{s}$$

$$\Rightarrow V_o(s) = \frac{20}{s \cdot (s^2 + 5s + 4)}$$

$$\Rightarrow V_o(s) = \frac{20}{s \cdot (s + 4) \cdot (s + 1)}$$

Now using partial fraction expansion:

$$\frac{20}{s \cdot (s+4) \cdot (s+1)} = \frac{A}{s} + \frac{B}{s+4} + \frac{c}{s+1}$$

$$\Rightarrow A(s+4)(s+1) + B(s)(s+1) + C(s)(s+4) = 20$$

$$(s=0) \Rightarrow A=5$$

$$(s=-1) \Rightarrow 12B = 20$$

$$\Rightarrow B = \frac{5}{3}$$

$$(s=-1) \Rightarrow -3C = 20$$

$$\Rightarrow C = \frac{-20}{3}$$

$$\Rightarrow V_o(s) = \frac{5}{s} + \frac{5}{3(s+4)} - \frac{20}{3(s+1)}$$

(b)

$$v_o(t) = \mathcal{L}^{-1}[V_o(s)]$$

$$= \mathcal{L}^{-1} \left[ \frac{5}{s} + \frac{5}{3(s+4)} - \frac{20}{3(s+1)} \right]$$

$$= (5 + \frac{5}{3}e^{-4t} - \frac{20}{3}e^{-t})u(t)$$

(c)

## 3. (a) NEEDS DIAGRAM!!!!

$$\begin{split} Z_{R||L} &= (\frac{1}{5} + \frac{1}{2s})^{-1} \\ &= (\frac{2s+5}{10s})^{-1} \\ &= \frac{10s}{2s+5} \\ \Longrightarrow H(s) &= \frac{V_o(s)}{V_i(s)} \\ &= \frac{z_{R||L}}{Z_{R||L} + Z_C} \\ &= \frac{\frac{10s}{2s+5}}{\frac{10s}{2s+5} + \frac{10}{s}} \\ &= \frac{\frac{10s}{2s+5}}{\frac{10s^2 + 20s + 50}{2s^2 + 5s}} \\ &= \frac{s^2}{s^2 + 2s + 5} \end{split}$$

## (b) NEEDS DIAGRAM!!!!!

Using phasor analysis:

$$V_o = V_i \cdot \frac{\left(\left(\frac{1}{R} + \frac{1}{j\omega L}\right)^{-1}\right)}{\left(\frac{1}{R} + \frac{1}{j\omega L}\right)^{-1} + \frac{1}{j\omega C}}$$
$$= 10 \angle 0^\circ \cdot \frac{\left(\frac{1}{5} + \frac{1}{40j}\right)^{-1}}{\left(\frac{1}{5} + \frac{1}{40j}\right)^{-1} + \frac{1}{2j}}$$
$$= 10.075 \angle 5.78^\circ V$$

This means that the steady state response for the system when  $v_i(t)=10\cos{(20t)}$  is  $v_o(t)=10.075\cos{(20t+5.78^\circ)}~V$ 

- 4. Q4
- 5. Q5
  - (a)
  - (b)
  - (c)
- 6. Q6