

# A Data-Driven Attack Detection Approach for DC Servo Motor Systems Based on Mixed Optimization Strategy

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**Abstract**—This article is concerned with the data-driven attack detection problem for cyber-physical systems with the actuator attacks and measurement noise. In most of existing data-driven detection methods,  $H_\infty$  index is used to characterize the sensitivity performance. It is well-known that compared with the  $H_\infty$  index,  $H_-$  index can significantly improve the diagnostic performance. However, the detection system design based on the  $H_-/H_\infty$  mixed optimization technique has not been solved within the data-driven framework. In this article, a residual generator is constructed from the available input–output (I/O) data.  $H_\infty$  and  $H_-$  indices are defined from the viewpoint of time-domain to characterize the robustness of residual generator against measurement noise and sensitivity to attack signals, respectively. In particular, a novel weighting system, which is expressed as an I/O model, is designed to transform the  $H_-$  performance into an  $H_\infty$  constraint, and the detection system design problem based on  $H_-/H_\infty$  mixed optimization technique is finally formulated into a constraint-type optimization one, which can be solved by the classical Lagrange multiplier method. Also, the proposed detection method is applied to a networked dc servo motor system to verify its advantages and effectiveness.

**Index Terms**—Actuator attack, cyber-physical systems (CPSs), data-driven, mixed optimization.

## I. INTRODUCTION

**D**UE to innovations in computer technology, embedded technology, and network technology, recent years have

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witnessed a rapid development of cyber-physical systems (CPSs). CPSs integrate computation, communication, and control into one, and are widely used in power generation, energy development, chemical industry control, etc. However, with further progress of networked embedded control technology, cyber attacks have become one of the major threats to CPSs. Therefore, since the discovery of the Stuxnet malware [1], the issue of security in CPSs has become a hot topic of scientific inquiry [2]–[4].

The detection and secure control problems of CPSs with attacks have received increasing attention in the past years [5]–[15]. In [6], the detection problem of two Gaussian signals with uncertain mean and covariance has been studied. It has been proven that the design of a robust linear detector can be cast into a convex optimization problem. In [7], both hybrid-share real-time detector and mitigation mechanism have been proposed for CPSs. In [9], the attack detection problem for cyber-physical direct current (dc) microgrids with false-data injection attacks has been formalized as identifying a change in sets of inferred candidate invariants. In [11], a method has been proposed in order to detect stealthy false data injection attacks based on coding matrix. Besides, from the attackers' standpoint, the attack problem has been modeled as a constrained control problem and that the characterization of the maximum perturbation has been posed as reachable set computation in [12]. A variation of the receding-horizon control law to deal with the replay attacks has been proposed, and the resulting system performance degradation has been analyzed in [13]. Because attackers are constrained by execution resources, attacks tend to be sparse. Among the existing works, sparse attacks for CPSs have been widely studied. In [14], a low-complexity attacking strategy to construct sparse false data injection attack vectors has been designed, and strategic protection schemes have been also proposed based on greedy approaches.

However, the system dynamic matrices in the abovementioned results are all required to be known. As pointed out in [16], it may be strict to assume that all the system dynamic matrices are known in CPSs. Therefore, it is more reasonable to study the attack detection and secure control problems for CPSs with completely unknown system matrices. Up to now, such problems have not been investigated well, which is one of the main motivations of this article.

On the other hand, in the absence of accurate models of the controlled plants, full use of the online or offline data can be

taken to reconfigure system states, diagnose failures, design controllers, or evaluate performance directly [17]–[24]. In view of the approximate dynamic programming approach, the data-driven fault-tolerant control (FTC) of a dc machine system with nonlinearity and load variations was achieved in [20]. Besides, a new recursive total principle component regression based design and implementation approach has been proposed for efficient data-driven fault detection in [21], which significantly increases the online efficiency and releases unnecessary memory occupation. In [22], when the information of the transmission control protocol/Internet protocol (TCP/IP)-based CPSs was eavesdropped, an approach to blocking network communications and injecting false sensor data into the CPSs was explored, and a closed-loop recursive identification strategy for the dynamic characteristic matrix of the CPSs was also designed. In [23], a novel closed-loop numerical algorithm for subspace state space system identification has been proposed, which can deliver effective unbiased pole estimation and show the superior through a practical dc motor system.

Although the detection schemes were given in [17] and [18] within the data-driven framework, the  $H_\infty$  index was used to characterize sensitivity. As is known to all,  $H_-$  index can improve the diagnostic performance significantly compared with the  $H_\infty$  index. In fact, the  $H_-$  index is defined as the lowest level of sensitivity of system outputs to system inputs, which is the worst-case sensitivity measure and fails to be a matrix norm. Moreover, the filter design problem caused by the introducing of  $H_-$  index is essentially nonconvex. Particularly, within the data-driven framework, how to solve the  $H_-/H_\infty$  mixed optimization problem to simultaneously satisfy the robustness of residual generator against measurement noise and sensitivity to attack signals still remains open, which also motivates the current research.

This article is concerned with the data-driven actuator attack detection problem of the networked dc servo motor systems with the completely unknown system matrices and measurement noise. The main contributions are summarized as follows.

- 1) A data-driven residual generator is designed to detect the actuator attacks. The  $H_\infty$  and the  $H_-$  indices are defined to characterize the robustness of residual generator against measurement noise and sensitivity to attack signals, respectively. Then, the attack detection problem is formulated as an  $H_-/H_\infty$  mixed optimization problem. Compared with the  $H_\infty/H_\infty$  detection scheme in [17], the superiority of the proposed  $H_-/H_\infty$  attack detection method is verified in simulation section.
- 2) To solve the nonconvex problem encountered in addressing attack sensitivity performance characterized by  $H_-$  index, a weighting system in the form of I/O model is constructed to transform the attack sensitivity specification into an  $H_\infty$  constraint, which extends the model-based weighting system design approach in [29]. Finally, the  $H_-/H_\infty$  mixed optimization problem is formulated into a constraint-type optimization one, which can be solved directly by using the classical Lagrange multiplier method.

The rest of this article is organized as follows. In Section II, the attack model of CPSs and some basic assumptions are described. The design method of the data-driven residual generator based

on the optimal parity vector via constructing a weighting system is proposed in Section III. In Section IV, the proposed design method is applied to the networked dc servo motor system to illustrate the effectiveness. Finally, Section V concludes this article.

*Notation:* Throughout this article,  $R^n$  denotes the  $n$ -dimensional Euclidean space;  $\Gamma^\perp$  denotes the orthogonal complement of  $\Gamma$ ;  $I$  denotes the identity matrix;  $\|\cdot\|_2$  is used to represent the Euclidian norm of a vector or a matrix; and  $L_2$  denotes the Hilbert space of the functions with the following norm:  $\|\nu(k)\|_2 = \{\sum_{k=0}^{\infty} \nu^T(k)\nu(k)\}^{1/2}$ .

## II. SYSTEM MODELING AND PROBLEM STATEMENT

In this section, the model of a networked dc servo motor system is given, the concepts related to actuator attacks are also described, and the problem of data-driven attack detection is then provided.

### A. Model Description

The networked dc servo motor system has been studied in [22], the state-space model of the dc motor system is established as described below

$$\begin{aligned} \begin{bmatrix} \Delta \dot{I} \\ \Delta \dot{\Omega} \end{bmatrix} &= \begin{bmatrix} -R_m/L_m & -C_{V/\Omega}/L_m \\ C_{T/I}/J & 0 \end{bmatrix} \begin{bmatrix} \Delta I \\ \Delta \Omega \end{bmatrix} \\ &+ \begin{bmatrix} K_u/L_m & 0 \\ 0 & -1/J \end{bmatrix} \begin{bmatrix} u \\ \Delta T \end{bmatrix} \\ y &= \begin{bmatrix} 0 & K_y \end{bmatrix} \begin{bmatrix} \Delta I \\ \Delta \Omega \end{bmatrix} \end{aligned} \quad (1)$$

where  $R_m$ ,  $L_m$ ,  $C_{V/\Omega}$ ,  $C_{T/I}$ ,  $T$ , and  $J$  represent the armature resistance, armature inductance, motor constant, voltage constant, load torque, and total inertia, respectively.  $\Delta I$  denotes the error between the steady-state current and the actual armature current, and  $\Delta \Omega$  represents the error between the desired and actual speeds of the motor. Moreover,  $K_u = U_T/u$ ,  $K_y = y/\Omega$  where  $u$ ,  $y$ ,  $U_T$ , and  $\Omega$  denote the control input, system output, terminal voltage, and rotating speed of the motor, respectively.

The Fig. 1 represents the block diagram structure of the networked dc servo motor system, which shows the connections among the components. More specifically, the client computer sends measurement data  $y$  to the server computer via the Ethernet hub, and after computing a control command by virtue of  $y$ , the server computer sends it back to the client computer through the Ethernet hub again.

It is pointed out in [22] that the TCP/IP-based networked dc servo system is a typical CPS. In addition, it is assumed that the adversary can break into the communication link from the server computer to the client computer, and an attack signal is then designed to devastate the system tracking performance. However, different from [22], this article investigates the attack detection problem of the system (1) from the viewpoint of defense. Due to most of the analytical model based attack detection approaches relying on the system parameters, the main challenge to be solved is how to design the detector only using the system input and output (I/O) data.

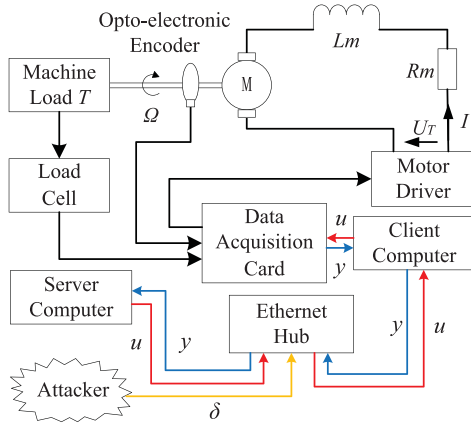


Fig. 1. Block diagram structure of the networked dc servo motor system.

To this end, the dynamics model of the networked dc servo motor system with the actuator attacks is established as

$$\begin{cases} x(k+1) = Ax(k) + B(u(k) + \delta(k)) \\ y(k) = Cx(k) + d(k) \end{cases} \quad (2)$$

where  $x(k) \in R^n$  denotes the unmeasurable system state,  $u(k) \in R^l$  and  $y(k) \in R^m$  are control input and system output, respectively.  $d(k)$  is the measurement noise. Similar to [25],  $\delta(k) = [\delta_1(k), \delta_2(k), \dots, \delta_l(k)]$  is a vector modeling how an attacker changes the control input at time  $k$ . If the  $q$ th,  $q \in \{1, \dots, l\}$  actuator is attacked, the  $q$ th element in the vector  $\delta(k)$  is nonzero, otherwise it is not attacked. Moreover  $A$ ,  $B$ , and  $C$  are unknown matrices with appropriate dimensions.

*Remark 1:* Just as mentioned in [5], both attack detection and fault detection aim to justify whether attacks/faults occurred or not, and for different types of attacks/faults, different detection approaches have been given in the literature. Analogous to the detection problem, the identification problem is to distinguish between distinct attacks/faults [5]. Especially, how to distinguish between actuator fault and actuator attack should also belong to the identification problem, and such an interesting problem will be further investigated in the future work.

### B. Problem Statement

The objective of this article is to design a data-driven attack detector such that the residual signal is sensitive to actuator attacks and also with a certain robustness to system noise.

In order to construct such a residual generator, the system (2) is manipulated into the I/O model (3) in the normal case, i.e.,  $\delta(k) = 0$

$$\mathcal{Y}_f = \Gamma_s \mathcal{X}_k + \mathcal{H}_s^u \mathcal{U}_f + \mathcal{H}_s^d \mathcal{D}_f \quad (3)$$

where

$$\Gamma_s = \begin{bmatrix} C \\ CA \\ \vdots \\ CA^s \end{bmatrix} \in \mathcal{R}^{m_s \times n} \quad (4)$$

is the extended observability matrix.

$$\mathcal{H}_s^u = \begin{bmatrix} 0 & 0 & \cdots & 0 \\ CB & 0 & \cdots & 0 \\ \vdots & \ddots & \ddots & \vdots \\ CA^{s-1}B & \cdots & CB & 0 \end{bmatrix} \in \mathcal{R}^{m_s \times l_s} \quad (5)$$

is a block Toeplitz matrix, and  $\mathcal{H}_s^d$  is an identity matrix. Here,  $m_s = (s+1)m$ ,  $l_s = (s+1)l$ , and  $s \geq n$ . Then, the state sequences and the block Hankel matrices for outputs are defined as follows:

$$\mathcal{X}_k = [x(k), x(k+1), \dots, x(k+N)] \quad (6)$$

$$\mathcal{Y}_p = \begin{bmatrix} y(k-s) & y(k-s+1) & \cdots & y(k-s+N-1) \\ y(k-s+1) & y(k-s+2) & \cdots & y(k-s+N) \\ \vdots & \vdots & \ddots & \vdots \\ y(k) & y(k+1) & \cdots & y(k+N-1) \end{bmatrix} \quad (7)$$

$$\mathcal{Y}_f = \begin{bmatrix} y(k+1) & y(k+2) & \cdots & y(k+N) \\ y(k+2) & y(k+3) & \cdots & y(k+N+1) \\ \vdots & \vdots & \ddots & \vdots \\ y(k+s+1) & y(k+s+2) & \cdots & y(k+s+N) \end{bmatrix} \quad (8)$$

Following the same lines, the matrices  $\mathcal{U}_p$ ,  $\mathcal{U}_f$ , and  $\mathcal{D}_f$  can be defined well, where  $f$  and  $p$  denote the dimensions of future and past windows, respectively, and  $N$  is chosen large enough which satisfies (11).

To investigate the attack detection problem, the underlying basic assumptions are introduced in this article.

*Assumption 1:* The order of the system  $n$  is known.

*Assumption 2:*  $(A, B)$  is controllable and  $(A, C)$  is observable.

Based on the Assumptions 1 and 2, the I/O data will be used to design the residual generator to detect attacks.

### III. DATA-DRIVEN MIXED OPTIMIZATION ATTACK DETECTION APPROACH

In this section, in order to realize the detection of actuator attacks, an observer-based residual generator is first designed by using the I/O data. Then, the  $H_\infty$  and  $H_-$  indices are defined to characterize the robustness of residual generator against system noise and sensitivity to attack signals, respectively. Finally, the observer design problem is transformed into an  $H_-/H_\infty$  mixed optimization problem, which is solved by the classical Lagrange multiplier method, and the parameters can be computed to construct the residual generator.

#### A. Data-Driven Observer-Based Residual Generator Design

Under the Assumption 2, the extended observability matrix  $\Gamma_s$  is full rank of column, and then there exists a left null space of  $\Gamma_s$ ,

denoted by  $\Gamma_s^\perp$ . The construction of an observer-based residual generator can be achieved by identifying  $\Gamma_s^\perp$  and  $\Gamma_s^\perp \mathcal{H}_s^u$ . To this end, let

$$\mathcal{Z}_p = \begin{bmatrix} \mathcal{Y}_p \\ \mathcal{U}_p \end{bmatrix}, \quad \mathcal{Z}_f = \begin{bmatrix} \mathcal{Y}_f \\ \mathcal{U}_f \end{bmatrix}.$$

According to [26], the identification process can be completed by performing singular value decomposition (SVD) on  $\frac{1}{N} \mathcal{Z}_f \mathcal{Z}_p^T$

$$\frac{1}{N} \mathcal{Z}_f \mathcal{Z}_p^T = \begin{bmatrix} \mathcal{U}_{11} & \mathcal{U}_{12} \\ \mathcal{U}_{21} & \mathcal{U}_{22} \end{bmatrix} \begin{bmatrix} \Sigma & 0 \\ 0 & 0 \end{bmatrix} \begin{bmatrix} \mathcal{V}_1^T \\ \mathcal{V}_2^T \end{bmatrix} \quad (9)$$

where

$$\mathcal{U}_{12} \in \mathcal{R}^{m_s \times (m_s - n)}, \quad \mathcal{U}_{22} \in \mathcal{R}^{l_s \times (m_s - n)}.$$

From [27], it is known that

$$\Gamma_s^\perp = \mathcal{U}_{12}^T, \quad \Gamma_s^\perp \mathcal{H}_s^u = -\mathcal{U}_{22}^T. \quad (10)$$

In addition, to make sure  $\Gamma_s^\perp$  and  $\Gamma_s^\perp \mathcal{H}_s^u$  are available, the data employed in constructing  $\frac{1}{N} \mathcal{Z}_f \mathcal{Z}_p^T$  should be collected under the input excitation condition, which satisfies

$$\text{rank} \left( \frac{1}{N} \begin{bmatrix} \mathcal{X}_k \\ \mathcal{U}_f \end{bmatrix} \mathcal{Z}_p^T \right) = n + sl. \quad (11)$$

According to [28], the observer-based residual generator is designed in the following:

$$z(k+1) = A_z z(k) + B_z u(k) + L_z y(k) \quad (12)$$

$$r(k) = g_z y(k) - C_z z(k) \quad (13)$$

where  $A_z, B_z, C_z, L_z$ , and  $g_z$  together with the transformation matrix  $T$  are able to be parameterized by the Luenberger equations

$$\begin{aligned} A_z T &= T A - L_z C \\ B_z &= T B \\ C_z T &= g_z C. \end{aligned} \quad (14)$$

Consider a vector  $v_s = [v_s^0, v_s^1, \dots, v_s^s] \in \mathcal{R}^{m \times s}$ , where

$$v_s \in P_s, \quad P_s = \{v_s \mid v_s \Gamma_s = 0\}. \quad (15)$$

Here,  $v_s$  is the so-called parity vector. Then, the following equations hold:

$$A_z = \begin{bmatrix} 0 & 0 & \cdots & 0 \\ 1 & 0 & \cdots & 0 \\ \vdots & \ddots & \ddots & \vdots \\ 0 & \cdots & 1 & 0 \end{bmatrix}, \quad B_z = \begin{bmatrix} v_s \mathcal{H}_{s,0}^u \\ v_s \mathcal{H}_{s,1}^u \\ \vdots \\ v_s \mathcal{H}_{s,s-1}^u \end{bmatrix}$$

$$C_z = \begin{bmatrix} 0 & 0 & \cdots & 1 \end{bmatrix}, \quad L_z = - \begin{bmatrix} v_s^0 \\ v_s^1 \\ \vdots \\ v_s^{s-1} \end{bmatrix}$$

$$g_z = v_s^s \quad (16)$$

where

$$\mathcal{H}_s^u = \begin{bmatrix} \mathcal{H}_{s,0}^u & \mathcal{H}_{s,1}^u & \cdots & \mathcal{H}_{s,s-1}^u \end{bmatrix}$$

$$\mathcal{H}_{s,i}^u = \begin{bmatrix} 0 \\ \vdots \\ CB \\ \vdots \\ CA^{s-i-1}B \end{bmatrix}$$

for  $i = 0, 1, \dots, s-1$ .

With  $\Gamma_s^\perp$  and  $\Gamma_s^\perp \mathcal{H}_s^u$  being identified by using the I/O data,  $v_s$  can be selected to construct the residual generator (12) and (13). In fact, there exists an optimal parity vector  $v_s$  such that the residual is robust against system noise and sensitive to attack signals simultaneously. The design criterion of the optimal parity vector  $v_s$  depends on the robustness and sensitivity indices defined in the next section.

*Remark 2:* It should be pointed out that the proposed observer design method is also applicable for the multi-input–multi-output (MIMO) systems. In fact, the I/O model (3) and the method of calculating the observer parameters (16) are still valid for the MIMO systems, and the detailed discussions can refer to [17] and [18].

## B. Calculation of the Optimal Parity Vector

Similar to (3), by iterating the equations in (2), we can obtain the I/O model in the presence of an attack

$$y_s(k) = \Gamma_s x(k-s) + \mathcal{H}_s^u u_s(k) + \mathcal{H}_s^d d_s(k) + \mathcal{H}_s^\delta \delta_s(k) \quad (17)$$

where

$$y_s(k) = \begin{bmatrix} y(k-s) \\ y(k-s+1) \\ \vdots \\ y(k) \end{bmatrix}$$

Following the same lines,  $u_s(k)$ ,  $d_s(k)$ , and  $\delta_s(k)$  can be defined well, and  $\mathcal{H}_s^\delta$  is the same as  $\mathcal{H}_s^u$ .

According to [28], the residual signal generated by (12) and (13) can be rewritten as

$$r(k) = v_s(y_s(k) - \mathcal{H}_s^u u_s(k)) = v_s(\mathcal{H}_s^d d_s(k) + \mathcal{H}_s^\delta \delta_s(k)) \quad (18)$$

Thus, on the basis of (18), we define the following robust performance index from a time-domain perspective:

$$R_d := \|v_s \mathcal{H}_s^d\|_\infty = \sup_{d \neq 0} \frac{\|v_s \mathcal{H}_s^d d\|_2}{\|d\|_2} \quad (19)$$

which is a measure of the worst-case influence of the disturbance  $d(k)$  on the residual signal  $r(k)$ . In the existing work [17], [18],



an  $H_\infty$  index is used to characterize the sensitivity from the attack  $\delta(k)$  to  $r(k)$

$$S_{\delta,+} := \|v_s \mathcal{H}_s^\delta\| = \sup_{\delta \neq 0} \frac{\|v_s \mathcal{H}_s^\delta \delta\|_2}{\|\delta\|_2}. \quad (20)$$

According to (19) and (20), the  $H_\infty/H_\infty$  optimization problem formulated as maximizing  $J_{S/R} = \frac{S_{\delta,+}}{R_d}$  is solved to get the best performance of residual generator in [17].

It should be pointed out that (20) is not the worst-case attack sensitivity measure, while  $H_-$  index is a measure of evaluating the minimum sensitivity, which can significantly improve the diagnostic performance according to [10]. Therefore, in this article, we define the sensitivity index as

$$S_{\delta,-} := \|v_s \mathcal{H}_s^\delta\|_- = \inf_{\delta \neq 0} \frac{\|v_s \mathcal{H}_s^\delta \delta\|_2}{\|\delta\|_2}. \quad (21)$$

On the basis of (19) and (21), our problem is cast into an  $H_-/H_\infty$  mixed optimization one described as  $\mathcal{P}_0$

$$\begin{aligned} \mathcal{P}_0 : & \text{maximize } \beta, \text{ subject to } S_{\delta,-} > \beta \\ & \text{with } R_d < \gamma_0, \text{ for a given } \gamma_0 \end{aligned} \quad (22)$$

where  $\beta$  denotes the sensitivity index to be optimized, and  $\gamma_0$  is the prescribed robustness index.

Note that  $S_{\delta,-}$  is an  $H_-$  index, which fails to be a norm. The filter design problem caused by the introducing of the  $H_-$  index is essentially nonconvex. To overcome this problem, we will refer to the method in [29] to transform the sensitivity specification into an  $H_\infty$  constraint.

**Lemma 1:** Consider the attack sensitivity specification  $S_{\delta,-} = \|v_s \mathcal{H}_s^\delta\|_- > \beta$  in (22). Introduce a weighting matrix  $W_f$  such that

$$\|W_f\|_- > \eta > \beta \quad (23)$$

where  $\eta$  is a given positive scalar. If

$$\|v_s \mathcal{H}_s^\delta - W_f\|_\infty < \eta - \beta \quad (24)$$

holds, the specification  $S_{\delta,-} = \|v_s \mathcal{H}_s^\delta\|_- > \beta$  can be ensured.

*Proof:* According to the definition similar to (21), we have

$$\begin{aligned} \|v_s \mathcal{H}_s^\delta\|_- &= \inf_{\delta \neq 0} \frac{\|v_s \mathcal{H}_s^\delta \delta\|_2}{\|\delta\|_2} \\ &\geq \inf_{\delta \neq 0} \left( \left\| \frac{W_f \delta}{\|\delta\|_2} - \frac{(v_s \mathcal{H}_s^\delta - W_f) \delta}{\|\delta\|_2} \right\|_2 \right) \\ &\geq \inf_{\delta \neq 0} \left( \frac{\|W_f \delta\|_2}{\|\delta\|_2} \right) - \inf_{\delta \neq 0} \left( \frac{\|(v_s \mathcal{H}_s^\delta - W_f) \delta\|_2}{\|\delta\|_2} \right) \\ &> \inf_{\delta \neq 0} \left( \frac{\|W_f \delta\|_2}{\|\delta\|_2} \right) - \sup_{\delta \neq 0} \left( \frac{\|(v_s \mathcal{H}_s^\delta - W_f) \delta\|_2}{\|\delta\|_2} \right) \end{aligned} \quad (25)$$

Therefore, we can obtain that

$$\|v_s \mathcal{H}_s^\delta\|_- > \|W_f\|_- - \|v_s \mathcal{H}_s^\delta - W_f\|_\infty. \quad (26)$$

In view of (23), (26) and the sufficient condition (24), the attack sensitivity specification  $S_{\delta,-} = \|v_s \mathcal{H}_s^\delta\|_- > \beta$  can be ensured, which completes the proof. ■

Assuming that the minimal state-space realization of  $W_f$  is

$$\begin{cases} x_w(k+1) = A_w x_w(k) + B_w \delta(k) \\ z_w(k) = C_w x_w(k) \end{cases} \quad (27)$$

where  $x_w(k) \in \mathcal{R}^n$ ,  $A_w$ ,  $B_w$ , and  $C_w$  are the constant matrices to be designed, and the detailed design processes are given in [29].

Within the data-driven framework, we first rewrite the state-space realization (27) into the I/O model

$$z_s^w(k) = \Gamma_s^w x_w(k-s) + \mathcal{H}_{\delta,s}^w \delta_s(k) \quad (28)$$

where  $\Gamma_s^w$  has the same structure as  $\Gamma_s$  in (4) by replacing  $A$  and  $C$  with  $A_w$  and  $C_w$ , and  $\mathcal{H}_{\delta,s}^w$  has the same block Toeplitz structure as  $\mathcal{H}_s^\delta$  in (17), which is derived by replacing  $A$ ,  $B$ , and  $C$  with  $A_w$ ,  $B_w$ , and  $C_w$ . Moreover, the parity vector  $\alpha_s$  similar to  $v_s$  in (15) is also introduced to eliminate the states  $x_w(k-s)$ . Then, we get

$$\alpha_s z_s^w(k) = \alpha_s \mathcal{H}_{\delta,s}^w \delta_s(k). \quad (29)$$

Thus,  $W_f$  is rewritten as

$$W_f = \alpha_s \mathcal{H}_{\delta,s}^w. \quad (30)$$

Note that,  $\alpha_s$  should be given in such a way that the minimum singular value of matrix  $\alpha_s \mathcal{H}_{\delta,s}^w$  is maximized.

By using Lemma 1, the  $H_-$  index used to characterize the attack sensitivity specification is transformed into an  $H_\infty$  constraint, and the problem described in  $\mathcal{P}_0$  can be converted to  $\mathcal{P}_1$

$$\mathcal{P}_1 : \text{minimize } \|v_s \mathcal{H}_s^\delta - \alpha_s \mathcal{H}_{\delta,s}^w\|, \text{ subject to } \|v_s \mathcal{H}_s^d\| < \gamma_0 \quad (31)$$

which is a constraint-type optimization problem and can be solved by the classical Lagrange multiplier method. In order to solve  $\mathcal{P}_1$ , introducing a Lagrange multiplier  $\lambda$ , the optimal parity vector  $v_s$  can be found by minimizing the following objective

$$\begin{aligned} \mathcal{L}(v_s, \lambda) &= \min_{v_s} \|v_s \mathcal{H}_s^\delta - \alpha_s \mathcal{H}_{\delta,s}^w\|^2 + \lambda (\|v_s \mathcal{H}_s^d\|^2 - \gamma_0^2) \\ &= \min_{v_s} [(v_s \mathcal{H}_s^\delta - \alpha_s \mathcal{H}_{\delta,s}^w)(v_s \mathcal{H}_s^\delta - \alpha_s \mathcal{H}_{\delta,s}^w)^T] \\ &\quad + \lambda [(v_s \mathcal{H}_s^d)(v_s \mathcal{H}_s^d)^T - \gamma_0^2] \end{aligned} \quad (32)$$

with

$$\lambda [(v_s \mathcal{H}_s^d)(v_s \mathcal{H}_s^d)^T - \gamma_0^2] = 0. \quad (33)$$

Differentiating  $\mathcal{L}(v_s, \lambda)$  with respect to  $v_s$  leads to

$$v_s [\mathcal{H}_s^\delta (\mathcal{H}_s^\delta)^T + \lambda \mathcal{H}_s^d (\mathcal{H}_s^d)^T] = \alpha_s \mathcal{H}_{\delta,s}^w (\mathcal{H}_s^\delta)^T \quad (34)$$

On the basis of (33) and the constraint in  $\mathcal{P}_1$ , we can obtain

$$v_s [\mathcal{H}_s^\delta (\mathcal{H}_s^\delta)^T] = \alpha_s \mathcal{H}_{\delta,s}^w (\mathcal{H}_s^\delta)^T \quad (35)$$

as  $\lambda = 0$ . Therefore, the solution space containing the optimal parity vectors  $v_s$  can be obtained through (35).

**Algorithm 1:** Data-Driven Mixed Optimization Design of a Residual Generator.

**Step 1:** Collect the data tuples  $(\{y_i, u_i\}, i = k - s, k - s + 1, \dots)$  under an input excitation condition. Subsequently, construct  $\mathcal{Z}_p$  and  $\mathcal{Z}_f$ . Perform SVD on  $\frac{1}{N} \mathcal{Z}_f \mathcal{Z}_p^T$  to identify  $\Gamma_s^\perp$  and  $\Gamma_s^\perp \mathcal{H}_s^u$ .

**Step 2:** Based on the identification results of the original system via the input-output data, calculate  $\mathcal{H}_s^d$  and  $\mathcal{H}_s^\delta$ .

**Step 3:** According to Lemma 1, design a weighting system  $W_f$ , and calculate  $\mathcal{H}_{\delta,s}^w$  in (30).

**Step 4:** Solve the constraint-type optimization problem (32) by the classical Lagrange Multiplier method. The optimal parity vector  $v_s$  can be derived to build the complete residual generator (12) and (13).

**Step 5:** Establish the threshold value  $J_{th}$  (37), if  $J_r(\tau) > J_{th}$ , then an attack is detected.

Finally, the residual generator (12) and (13) can be designed using the optimal parity vector to detect the actuator attack signals.

### C. Detection Threshold Design

In this section, the threshold for detecting actuator attacks is designed and the detection logic unit is based on the result in [30].

Let the root mean square value which means the average energy of residual signal over a time interval  $(k_0, k_\tau)$

$$J_r(\tau) = \sqrt{\frac{1}{\tau} \sum_{k=k_0}^{k_\tau} r^T(k)r(k)} \quad (36)$$

be a residual evaluation function, where  $k_0$  stands for the initial evaluation time instant and  $k_\tau$  denotes the evaluation time.

Let

$$J_{th} = \sup_{d(k) \in L_2, \delta(k)=0} J_r(\tau) \quad (37)$$

as the threshold. According to (37), the occurrence of attacks can be detected by comparing  $J_r(\tau)$  and  $J_{th}$  based on the following test:

$$\begin{cases} J_r(\tau) > J_{th} \Rightarrow \text{alarm} \\ J_r(\tau) \leq J_{th} \Rightarrow \text{no attack} \end{cases} \quad (38)$$

On the basis of above work, the main steps of the designed data-driven attack detection scheme are summarized in Algorithm 1.

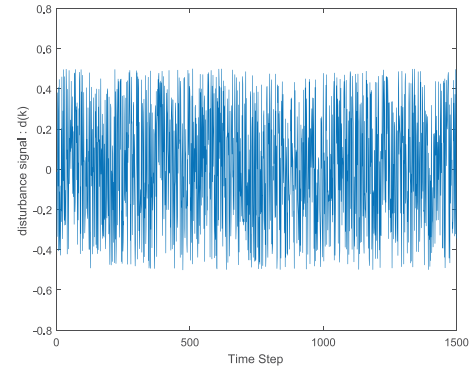
## IV. SIMULATION EXAMPLE

In this section, the networked dc servo motor system is used to verify the effectiveness and advantages of the proposed method.

All the simulation parameters of the dc motor, the digital-to-analog converter, and the optoelectronic encoder are summarized in Table I. The parameters can also be found in [19],

**TABLE I**  
SIMULATION PARAMETERS OF THE MOTOR SYSTEM

Parameter	Symbol	Value	Unit
Motor constant	$C_{T/I}$	0.06	Nm/A
Voltage constant	$C_{V/\Omega}$	$6.27 \times 10^{-3}$	V/(r/min)
Armature inductance	$L_m$	0.003	H
Resistance	$R_m$	3.13	Ohm
Machine load	$T$	0.1	Nm
Total inertia	$J$	$80.45 \times 10^{-6}$	$\text{kg} \times \text{m}^2$
Encoder resolution	$E_R$	60	Pulses/R
D/A resolution	$P_R$	12	Bit



**Fig. 2.** Disturbance signal  $d(k)$ .

and the sampling time is set to be 0.1 s. As the parameters are calculated, the above sampling time and zero-order-hold discretization method are adopted, finally the normal networked dc servo motor system is established as

$$\begin{aligned} x(k+1) &= \begin{bmatrix} -0.0022 & -0.0030 \\ 1.0625 & 0.9808 \end{bmatrix} x(k) \\ &\quad + \begin{bmatrix} 0.9516 & 0.0000 \\ 6.1303 & 4.5728 \end{bmatrix} u(k) \\ y(k) &= \begin{bmatrix} 0 & 0.170667 \end{bmatrix} x(k) + d(k). \end{aligned}$$

The considered time-varying random disturbance signal  $d(k)$  with the magnitude being 0.5 is shown in Fig. 2.

In practice, it may be difficult to obtain the precise system parameters. Therefore, for the case that the parameters in Table I are completely unknown, a data-driven attack detection approach is given in the article for the considered dc servo motor system. In the following, it is shown that the proposed method not only detects the actuator attack effectively, but also improves the related data-driven detection results in terms of enhancing the detection performances.

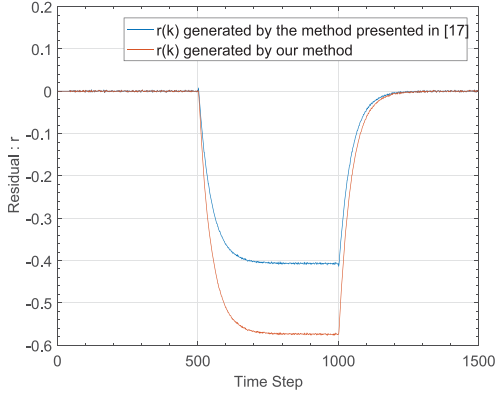


Fig. 3. Residual  $r(k)$  with the magnitude of  $\delta(k)$  being 5.

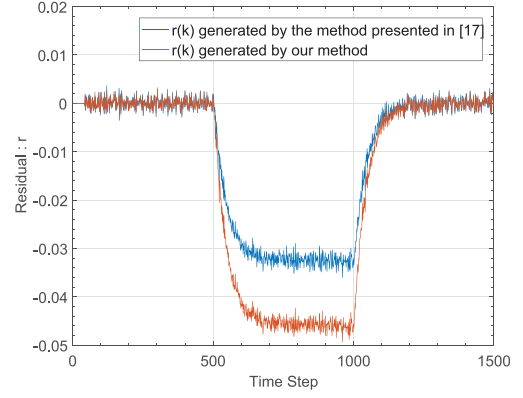


Fig. 5. Residual  $r(k)$  with the magnitude of  $\delta(k)$  being 0.4.

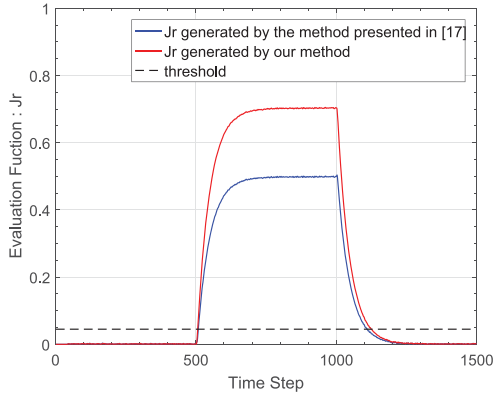


Fig. 4. Residual evaluation function  $J_r(\tau)$  with the magnitude of  $\delta(k)$  being 5.

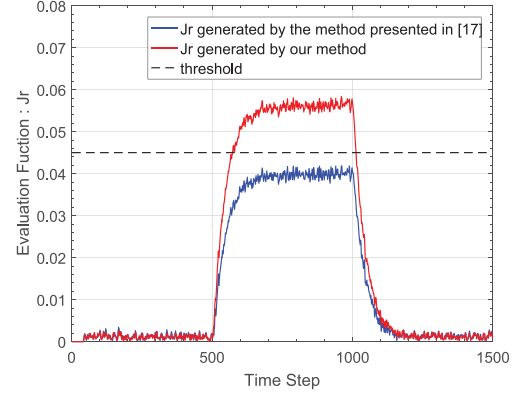


Fig. 6. Residual evaluation function  $J_r(\tau)$  with the magnitude of  $\delta(k)$  being 0.4.

Referring to [29], the following system matrices of the weighting system:

$$A_w = \begin{bmatrix} 0.3781 & -2.2081 \\ 0.0169 & 0.2202 \end{bmatrix}$$

$$B_w = \begin{bmatrix} -4.4658 & 0.1725 \\ -4.8000 & 2.9185 \end{bmatrix}$$

$$C_w = [0.0034 \quad 0.0004]$$

are selected to design  $W_f$  with  $\eta = 0.63$  and  $\alpha_s = [0.0637, -0.1982, -0.0561, 0.9765]$ . Via Lemma 1, one can derive the sensitivity index being  $\beta = 0.59$ , and from Algorithm 1, the optimal parity vector  $v_s$  can be solved as  $v_s = [0, 0.1927, -0.0605, 0]$  with  $\gamma_0 = 0.5$ .

Assuming that at the 500th time step, the actuator attack signal  $\delta(k) = [5, 0]^T$  is injected in the system and removed at the 1000th time step. The residuals  $r(k)$  derived by using the proposed attack detection scheme and the existing one in [17] are then plotted in Fig. 3. The corresponding residual evaluation functions  $J_r(\tau)$  and threshold  $J_{th}$  computed by (36) are depicted in Fig. 4.

From Fig. 3, it can be seen that the residual generated by the  $H_-/H_\infty$  mixed optimization strategy is more sensitive than the one in [17] generated by using  $H_\infty/H_\infty$  detection scheme. It should be pointed out that the magnitude of  $\delta(k)$  is relatively large compared with the one of the system noise  $d(k)$ , then the attack can be detected by using both the proposed detection scheme and the one in [17] (see Fig. 4).

In fact, the attack signals might be weak, because the attacks are artificially launched by the adversary. Therefore, the residual generator should be sensitive enough so that the weak attack signals can still be detected. To further verify the advantages of the proposed detection method, it is assumed that the attack signal is  $\delta(k) = [0.4, 0]^T$ . For the same system noise  $d(k)$  given previously, the residual signals  $r(k)$  and residual evaluation functions  $J_r(\tau)$  generated by the proposed approach and the existing method presented in [17] are reported in Figs. 5 and 6.

As shown in these two figures, the proposed data-driven attack detection approach is still valid when the magnitude of the attack signal is smaller than the one of the system noise, while the existing method in [17] based on the  $H_\infty/H_\infty$  strategy fails to work in this case. Therefore, the comparison shows the superiority of the proposed  $H_-/H_\infty$  detection strategy.

## V. CONCLUSION

This article was concerned with the data-driven attack detection problem for CPSs with the actuator attacks and measurement noise. A residual generator was constructed from the available I/O data. Then, the  $H_\infty$  and  $H_-$  indices characterizing the robustness of residual generator against measurement noise and sensitivity to attack signals, respectively, were defined. Moreover, by introducing a weighting system expressed as the I/O model, the  $H_-$  performance was transformed into an  $H_\infty$  constraint, and the  $H_-/H_\infty$  mixed optimization problem was formulated into a constraint-type optimization one, which can be solved by the classical Lagrange multiplier. Finally, the proposed design method was applied to the networked dc servo motor system to verify its advantages and effectiveness. Note that future research efforts will focus on designing attack detection scheme for CPSs with measurement and process noise, and relevant secure control strategy within the data-driven framework.

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