

电磁波传输类型



电流连续性方程

非限定形式

特性方程

限定形式

波动方程

$$\left\{ \begin{array}{l} \nabla \times \vec{H} = \vec{J}_e + \frac{\partial \vec{D}}{\partial t} \\ \nabla \times \vec{E} = -\frac{\partial \vec{B}}{\partial t} \\ \nabla \cdot \vec{B} = 0 \\ \nabla \cdot \vec{D} = \rho \end{array} \right.$$

$$\begin{aligned}\vec{D} &= \epsilon \vec{E} \\ \vec{B} &= \mu \vec{H} \\ \vec{J} &= \sigma \vec{E}\end{aligned}$$

$$\begin{cases} \nabla \times \vec{H} = \sigma \vec{E} + \varepsilon \partial \vec{E} / \partial t \\ \nabla \times \vec{E} = -\mu \partial \vec{H} / \partial t \\ \nabla \cdot \vec{H} = 0 \\ \nabla \cdot \vec{E} = \rho / \varepsilon \end{cases}$$

无源

$\rho = 0$

$$\vec{J} = 0$$

$$\begin{aligned}\nabla^2 \bar{E} - \mu\epsilon \frac{\partial^2 \bar{E}}{\partial t^2} &= 0 \\ \nabla^2 \bar{H} - \mu\epsilon \frac{\partial^2 \bar{H}}{\partial t^2} &= 0\end{aligned}$$

$$\nabla \times (\vec{E} \times \vec{H}) = -\frac{\partial}{\partial t} \left(\frac{1}{2} \epsilon E^2 + \frac{1}{2} \mu H^2 \right) - \sigma E^2$$

坡印廷定律

$$\vec{S} = \vec{E} \times \vec{H} \quad \vec{S} \text{ 坡印廷矢量}$$

\vec{S} 坡印廷矢量

积分形式

$$\left\{ \begin{array}{l} \oint_C \vec{H} \cdot d\vec{l} = \int_S (\vec{J}_e + \frac{\partial \vec{D}}{\partial t}) \cdot d\vec{S} \\ \oint_C \vec{E} \cdot d\vec{l} = - \int_S \frac{\partial \vec{B}}{\partial t} \cdot d\vec{S} \\ \oint_S \vec{B} \cdot d\vec{S} = 0 \\ \oint_S \vec{D} \cdot d\vec{S} = \int_V \rho dV = Q \end{array} \right.$$

$$\begin{aligned}\hat{n} \times (\vec{H}_1 - \vec{H}_2) &= \vec{J}_s \\ \hat{n} \times (\vec{E}_1 - \vec{E}_2) &= 0 \\ \hat{n} \cdot (\vec{B}_1 - \vec{B}_2) &= 0 \\ \hat{n} \cdot (\vec{D}_1 - \vec{D}_2) &= \rho_s\end{aligned}$$

边界条件

$$\vec{H}_2 = 0$$

$$\vec{E}_2 = 0$$

$$\begin{aligned}\hat{n} \times \vec{H} &= \vec{J}_s \\ \hat{n} \times \vec{E} &= 0 \\ \hat{n} \cdot \vec{B} &= 0 \\ \hat{n} \cdot \vec{D} &= \rho_s\end{aligned}$$

理想导体边界条件

$$\nabla \square \vec{B} = 0$$

$$\vec{B} = \nabla \times \vec{A}$$

\vec{A} 矢量位

$$\nabla \times \vec{E} = -\partial \vec{B} / \partial t$$

$$\nabla \times \left(\bar{E} + \frac{\partial \bar{A}}{\partial t} \right) = 0$$

$$\vec{E} + \frac{\partial \vec{A}}{\partial t} = -\nabla \varphi \quad \varphi \text{ 标量位}$$

$$\vec{E} = -\frac{\partial \vec{A}}{\partial t} - \nabla \varphi$$

$$\nabla \cdot \vec{A} + \mu \epsilon \frac{\partial \varphi}{\partial t} = 0 \quad \text{罗伦兹条件}$$

$$\begin{aligned}\nabla^2 \bar{A} - \mu \varepsilon \frac{\partial^2 \bar{A}}{\partial t^2} &= -\mu \bar{J} \\ \nabla^2 \varphi - \mu \varepsilon \frac{\partial^2 \varphi}{\partial t^2} &= -\frac{\rho}{\varepsilon}\end{aligned}$$

达朗贝尔方程

麦克斯韦方程组

波动方程

时变电磁场

$e^{j\omega t}$

时谐电磁场

$e^{j\omega t}$

麦克斯韦方程组

$\vec{E}(\vec{r}, t)$
瞬时

$$\begin{cases} \nabla \times \vec{H} = \vec{J}_e + \frac{\partial \vec{D}}{\partial t} \\ \nabla \times \vec{E} = -\frac{\partial \vec{B}}{\partial t} \\ \nabla \cdot \vec{B} = 0 \\ \nabla \cdot \vec{D} = \rho \end{cases}$$

$e^{j\omega t} \rightarrow \frac{\partial}{\partial t} \rightarrow j\omega$

时谐场

麦克斯韦方程组 (时谐场)

$\vec{E}(\vec{r})$
复数

$$\begin{cases} \nabla \times \vec{H} = \vec{J} + j\omega \vec{D} \\ \nabla \times \vec{E} = -j\omega \vec{B} \\ \nabla \cdot \vec{B} = 0 \\ \nabla \cdot \vec{D} = \rho \end{cases}$$

无源

$$\begin{cases} \rho = 0 \\ \vec{J} = 0 \end{cases}$$

$$\begin{cases} \nabla^2 \vec{E} - \mu\epsilon \frac{\partial^2 \vec{E}}{\partial t^2} = 0 \\ \nabla^2 \vec{H} - \mu\epsilon \frac{\partial^2 \vec{H}}{\partial t^2} = 0 \end{cases}$$

波动方程

$e^{j\omega t} \rightarrow \frac{\partial^2}{\partial t^2} \rightarrow -\omega^2$

$$k = \omega \sqrt{\mu\epsilon}$$

波数

无源

$$\begin{cases} \rho = 0 \\ \vec{J} = 0 \end{cases}$$

$$\begin{cases} \nabla^2 \vec{E} + k^2 \vec{E} = 0 \\ \nabla^2 \vec{H} + k^2 \vec{H} = 0 \end{cases}$$

亥姆霍兹方程

电磁波在无损介质中传播

$$k = \omega \sqrt{\mu \epsilon}$$

$$\nabla^2 \vec{E} + k^2 \vec{E} = 0$$

$$\nabla^2 \vec{H} + k^2 \vec{H} = 0$$

?

$$\frac{d^2 E_x}{dz^2} + k^2 E_x = 0$$

$$E_x = E_m e^{-jkz}$$

$$E_x = E_m \cos(\omega t - kz)$$

$$T, f, \lambda$$

$$v = \omega/k = 1/\sqrt{\mu \epsilon}$$

$$v_g = d\omega/dk$$

$$\vec{E} = \hat{e}_x E_m e^{-jkz}$$

电场取任意方向

$$\vec{E} = \vec{E}_0 e^{-jkz}$$

沿任意方向传播

$$\vec{E} = \vec{E}_0 e^{-j\vec{k} \cdot \vec{r}}$$

$$\nabla \times \vec{E} = -j\omega \mu \vec{H}$$

$$\vec{H} = \frac{1}{\eta} \hat{e}_n \times \vec{E}$$

$$\eta = \sqrt{\frac{\mu}{\epsilon}}$$

本征阻抗

$$\vec{E}(z) = \hat{e}_x E_x(z)$$

电场取沿x轴

$$\vec{E}(z)$$

均匀平面波
(沿z方向传播)

$$\vec{E}(x, y, z) = ??$$

极化

旋转

线圆极化

$$\varphi_x - \varphi_y = 0, \pm \pi$$

椭圆极化

$$\varphi_x - \varphi_y = \text{任意}$$

$$\varphi_x - \varphi_y = \pm \pi/2$$

$$E_{mx} = E_{my}$$

圆极化

左旋

右旋

判断法则

■ 大拇指指向传播方向;

■ 四指指向相位滞后分量;

■ 符合左手定律则为左旋

■ 符合右手定律则为右旋;

电磁波在有损介质中传播

$$\nabla \times \vec{H} = \sigma \vec{E} + j\omega \epsilon \vec{E} = j\omega \epsilon_c \vec{E}$$

$$\epsilon_c = \epsilon - j \frac{\sigma}{\omega}$$

$$\nabla^2 \vec{E} + k^2 \vec{E} = 0$$

$$k_c = \omega \sqrt{\mu \epsilon_c}$$

$$\nabla^2 \vec{E} + k_c^2 \vec{E} = 0$$

$$\Gamma = jk_c = \alpha + j\beta$$

传播系数

$$\nabla^2 \vec{E} - \Gamma^2 \vec{E} = 0$$

$$\vec{E} = \hat{e}_x E_x(z)$$

均匀平面波；
沿z传播

$$\frac{d^2 E_x}{dz^2} - \Gamma^2 E_x = 0$$

$$E_x = E_m e^{-\Gamma z} = E_m e^{-\alpha z} e^{-j\beta z}$$

$$E_x = E_m e^{-\alpha z} \cos(\omega t - kz)$$

$$\nabla \times \vec{E} = -j\omega \mu \vec{H}$$

$$\vec{H} = \hat{e}_y \frac{E_m}{\eta_c} e^{-\alpha z} e^{-j\beta z} = \hat{e}_y \frac{E_m}{|\eta_c|} e^{-\alpha z} e^{-j\beta z} e^{-j\phi}$$

$$\eta_c = |\eta| e^{j\phi}$$

$$\alpha = \omega \sqrt{\frac{\mu \epsilon}{2} \left[\sqrt{1 + \left(\frac{\sigma}{\omega \mu} \right)^2} - 1 \right]}$$

$$\beta = \omega \sqrt{\frac{\mu \epsilon}{2} \left[\sqrt{1 + \left(\frac{\sigma}{\omega \mu} \right)^2} + 1 \right]}$$

$$\frac{\gamma}{\omega \epsilon}$$

$$\frac{\sigma}{\omega \epsilon} \ll 1$$

弱导电媒质

$$\alpha \approx \frac{1}{2} \sqrt{\frac{\mu}{\epsilon}}$$

$$\beta \approx \omega \sqrt{\mu \epsilon}$$

$$\eta_c \approx \sqrt{\frac{\mu}{\epsilon}} \left(1 + j \frac{\sigma}{2\omega \epsilon} \right)$$

$$\frac{\sigma}{\omega \epsilon} \gg 1$$

强导电媒质

$$\alpha \approx \sqrt{\pi f \mu \sigma}$$

$$\beta \approx \sqrt{\pi f \mu \sigma}$$

$$\eta_c \approx \sqrt{\frac{\pi f \mu}{\sigma}} (1 + j)$$

电磁波垂直入射 - 理想导体

入射波

$$E_x^+ = E_m^+ e^{-j\beta z}$$

$$H_y^+ = \frac{E_m^+}{\eta} e^{-j\beta z}$$

反射波

$$E_x^- = E_m^- e^{j\beta z}$$

$$H_y^- = \frac{E_m^-}{\eta} e^{j\beta z}$$

合波

$$E_x = E_x^+ + E_x^-$$

$$= E_m^+ e^{-j\beta z} + E_m^- e^{j\beta z}$$

边界条件

$$\bar{E}_x(0) = 0$$

$$E_m^+ = -E_m^-$$

$$E_x = -2jE_m^+ \sin(\beta z)$$

$$E_x = 2E_m^+ \sin(\beta z) \sin(\omega t)$$

$$H_y^- = -\frac{E_x^-}{\eta}$$

$$H_y = H_y^+ + H_y^- = \frac{2E_m^+}{\eta} \cos \beta z$$

$$H_y = \frac{2E_m^+}{\eta} \cos(\beta z) \cos(\omega t)$$

$$\vec{J}_s = n \times \vec{H} \Big|_{z=0} = \hat{e}_x 2E_m^+ / \eta$$

$$z = -n \frac{\lambda}{2}$$

$$E_x = 0$$

驻波

$$z = -(2n+1) \frac{\lambda}{4}$$

$$H_y = 0$$

电磁波垂直入射 - 导电媒质

入射波

$$\begin{aligned}\vec{E}_1^+ &= \hat{e}_x E_{m1}^+ e^{-\Gamma_1 z} \\ \vec{H}_1^+ &= -\hat{e}_y \frac{E_{m1}^+}{\eta} e^{-\Gamma_1 z}\end{aligned}$$

合波

(媒质1)

$$\begin{aligned}\vec{E}_1 &= \vec{E}_1^+ + \vec{E}_1^- \\ \vec{H}_1 &= \vec{H}_1^+ + \vec{H}_1^-\end{aligned}$$

反射波

$$\begin{aligned}E_1^- &= E_{m1}^- e^{\Gamma_1 z} \\ H_1^- &= -\frac{E_{m1}^-}{\eta} e^{\Gamma_1 z}\end{aligned}$$

边界条件:

场量切向连续

$$\begin{aligned}E_{1t}(0) &= E_{2t}(0) \\ H_{1t}(0) &= H_{2t}(0)\end{aligned}$$

透射波

(媒质2波)

$$\begin{aligned}E_2^+ &= \hat{e}_x E_{m2}^+ e^{-\Gamma_2 z} \\ H_2^+ &= -\hat{e}_y \frac{E_{m2}^+}{\eta} e^{-\Gamma_2 z}\end{aligned}$$

$$E(x, t) = \hat{e}_x E_m e^{-\alpha z} \cos(\omega t - kz)$$

$$\delta = 1/\alpha$$

$$\begin{aligned}\Gamma &= \frac{E_{m1}^-}{E_{m1}^+} = \frac{\eta_2 - \eta_1}{\eta_2 + \eta_1} \\ \tau &= \frac{E_{m2}^+}{E_{m1}^+} = \frac{2\eta_2}{\eta_2 + \eta_1}\end{aligned}$$

$$\begin{aligned}E_{m1}^+ + E_{m1}^- &= E_{m2}^+ \\ \frac{E_{m1}^+}{\eta_1} - \frac{E_{m1}^-}{\eta_1} &= \frac{E_{m2}^+}{\eta_2}\end{aligned}$$

$$\Gamma + 1 = \tau$$

媒质1

媒质2

$$\begin{cases} \nabla \times \vec{H} = \vec{J} + j\omega \vec{D} & \nabla \cdot \vec{B} = 0 \\ \nabla \times \vec{E} = -j\omega \vec{B} & \nabla \cdot \vec{D} = \rho \end{cases}$$

导行电磁波

无源

$$\begin{cases} \nabla \times \vec{H} = j\omega \vec{D} \\ \nabla \times \vec{E} = -j\omega \vec{B} \\ \nabla \cdot \vec{H} = 0 \\ \nabla \cdot \vec{E} = 0 \end{cases}$$

$$\vec{E}(x, y, z)$$

$$\begin{cases} \nabla^2 \vec{E} + k^2 \vec{E} = 0 \\ \nabla^2 \vec{H} + k^2 \vec{H} = 0 \end{cases}$$

$$\nabla^2 = \frac{\partial^2}{\partial x^2} + \frac{\partial^2}{\partial y^2} + \frac{\partial^2}{\partial z^2} = \nabla_{xy}^2 + \Gamma^2$$

$$\vec{E}(x, y)$$

$$\begin{cases} \nabla_{xy}^2 \vec{E} + (\Gamma^2 + k^2) \vec{E} = 0 \\ \nabla_{xy}^2 \vec{H} + (\Gamma^2 + k^2) \vec{H} = 0 \end{cases}$$

沿z轴传播

$$\vec{E}(x, y, z)$$

$$= \vec{E}_m(x, y) e^{-\Gamma z}$$

$$\frac{\partial}{\partial z} = -\Gamma$$

纵横关系

$$\begin{aligned} E_x &= -\frac{1}{\Gamma^2 + k^2} \left(\Gamma \frac{\partial E_z}{\partial x} + j\omega\mu \frac{\partial H_z}{\partial y} \right) \\ E_y &= -\frac{1}{\Gamma^2 + k^2} \left(\Gamma \frac{\partial E_z}{\partial y} - j\omega\mu \frac{\partial H_z}{\partial x} \right) \\ H_x &= -\frac{1}{\Gamma^2 + k^2} \left(\Gamma \frac{\partial H_z}{\partial x} - j\omega\varepsilon \frac{\partial E_z}{\partial y} \right) \\ H_y &= -\frac{1}{\Gamma^2 + k^2} \left(\Gamma \frac{\partial H_z}{\partial y} + j\omega\varepsilon \frac{\partial E_z}{\partial x} \right) \end{aligned}$$

$$\vec{E}(x, y)$$

$$E_z, H_z$$

$$\frac{\partial^2}{\partial z^2} = \Gamma^2$$

$$\begin{cases} H_z = 0 \\ E_z \neq 0 \end{cases}$$

TM波

$$\begin{aligned} E_x &= -\frac{\Gamma}{\Gamma^2 + k^2} \frac{\partial E_z}{\partial x} & H_x &= \frac{j\omega\varepsilon}{\Gamma^2 + k^2} \frac{\partial E_z}{\partial y} \\ E_y &= -\frac{\Gamma}{\Gamma^2 + k^2} \frac{\partial E_z}{\partial y} & H_y &= -\frac{j\omega\varepsilon}{\Gamma^2 + k^2} \frac{\partial E_z}{\partial x} \end{aligned}$$

$$\begin{cases} E_z = 0 \\ H_z = 0 \end{cases}$$

TEM波

$$\Gamma^2 + k^2 = 0 \rightarrow \Gamma = jk = j\omega\sqrt{\mu\varepsilon} \rightarrow v_p = \omega/k = 1/\sqrt{\mu\varepsilon}$$

$$Z_{TEM} = E_x/H_y = j\omega\mu/\Gamma = \Gamma/j\omega\varepsilon = \eta$$

$$\vec{H} = \frac{1}{Z_{TEM}} (\hat{e}_z \times \vec{E})$$

$$\begin{cases} E_z = 0 \\ H_z \neq 0 \end{cases}$$

TE波

$$\begin{aligned} E_x &= -\frac{j\omega\mu}{\Gamma^2 + k^2} \frac{\partial H_z}{\partial y} & H_x &= -\frac{\Gamma}{\Gamma^2 + k^2} \frac{\partial H_z}{\partial x} \\ E_y &= +\frac{j\omega\mu}{\Gamma^2 + k^2} \frac{\partial H_z}{\partial x} & H_y &= -\frac{\Gamma}{\Gamma^2 + k^2} \frac{\partial H_z}{\partial y} \end{aligned}$$

矩形波导

$$\begin{aligned}\nabla_{xy}^2 \vec{H} + (\Gamma^2 + k^2) \vec{H} &= 0 \\ \nabla_{xy}^2 \vec{E} + (\Gamma^2 + k^2) \vec{E} &= 0\end{aligned}$$

$$\vec{E}(E_x, E_y, E_z)$$

$$\nabla_{xy}^2 E_z + (\Gamma^2 + k^2) E_z = 0$$

分离变量

边界条件

$$E_z = X(x)Y(y)$$

$$E_z(0, y) = E_z(a, y) = 0$$

$$E_z(x, 0) = E_z(x, b) = 0$$

$$E_z = E_0 \sin\left(\frac{m\pi}{a}x\right) \sin\left(\frac{n\pi}{b}y\right)$$

TM_{nm}

$$\begin{cases} E_x = -\frac{\Gamma}{h^2} \frac{m\pi}{a} E_0 \cos\left(\frac{m\pi}{a}x\right) \sin\left(\frac{n\pi}{b}y\right) e^{-\Gamma z} \\ E_y = -\frac{\Gamma}{h^2} \frac{n\pi}{b} E_0 \sin\left(\frac{m\pi}{a}x\right) \cos\left(\frac{n\pi}{b}y\right) e^{-\Gamma z} \\ H_x = j \frac{\omega\epsilon}{h^2} \frac{n\pi}{b} E_0 \sin\left(\frac{m\pi}{a}x\right) \cos\left(\frac{n\pi}{b}y\right) e^{-\Gamma z} \\ H_y = -j \frac{\omega\epsilon}{h^2} \frac{m\pi}{a} E_0 \cos\left(\frac{m\pi}{a}x\right) \sin\left(\frac{n\pi}{b}y\right) e^{-\Gamma z} \end{cases}$$

$$H_z = H_0 \cos\left(\frac{m\pi}{a}x\right) \cos\left(\frac{n\pi}{b}y\right)$$

TE_{nm}

$$\begin{aligned} E_x &= j \frac{\omega\mu}{h^2} \frac{n\pi}{b} H_0 \cos\left(\frac{m\pi}{a}x\right) \sin\left(\frac{n\pi}{b}y\right) e^{-\Gamma z} \\ E_y &= -j \frac{\omega\mu}{h^2} \frac{m\pi}{a} H_0 \sin\left(\frac{m\pi}{a}x\right) \cos\left(\frac{n\pi}{b}y\right) e^{-\Gamma z} \\ H_x &= \frac{\Gamma}{h^2} \frac{m\pi}{a} H_0 \sin\left(\frac{m\pi}{a}x\right) \cos\left(\frac{n\pi}{b}y\right) e^{-\Gamma z} \\ H_y &= \frac{\Gamma}{h^2} \frac{n\pi}{b} H_0 \cos\left(\frac{m\pi}{a}x\right) \sin\left(\frac{n\pi}{b}y\right) e^{-\Gamma z} \end{aligned}$$

$$\begin{cases} \text{截止区} & 2a \sim \infty \\ \text{单模区} & a \sim 2a \\ \text{多模区} & < a \end{cases}$$

基模— TE_{10}

$$\begin{cases} E_x = 0 \\ E_y = -j\omega\mu \frac{a}{\pi} H_0 \sin\left(\frac{\pi}{a}x\right) \\ E_z = 0 \\ H_x = jk_z \frac{a}{\pi} H_0 \sin\left(\frac{\pi}{a}x\right) \\ H_y = 0 \\ H_z = H_0 \cos\left(\frac{\pi}{a}x\right) \end{cases}$$

$$h^2 = k_x^2 + k_y^2 = \left(\frac{m\pi}{a}\right)^2 + \left(\frac{n\pi}{b}\right)^2$$

$$\Gamma = \sqrt{h^2 - k^2} = \sqrt{\left(\frac{m\pi}{a}\right)^2 + \left(\frac{n\pi}{b}\right)^2 - \omega^2 \mu\epsilon}$$

$$h^2 = \Gamma^2 + k^2$$

$$\Gamma = 0$$

$$\lambda_c = \frac{1}{2\pi\sqrt{\mu\epsilon}} \sqrt{\left(\frac{m\pi}{a}\right)^2 + \left(\frac{n\pi}{b}\right)^2}$$