## UNIVERSITY OF DUBLIN TRINITY COLLEGE

## FACULTY OF ENGINEERING, MATHEMATICS & SCIENCE SCHOOL OF ENGINEERING

**Electronic & Electrical Engineering** 

Senior Sophister Engineering Annual Examinations Trinity Term, 2014

**Digital Signal Processing (4C5)** 

Date: Friday 9<sup>th</sup> May Venue: Luce Upper Time: 14.00 – 16.00

Mr. W. Dowling

Answer THREE questions
All questions carry equal marks

**Permitted Materials:** 

Calculator
Drawing Instruments
Mathematical Tables
Graph Paper

**Q.1** (a) A continuous-time signal,  $x_a(t)$ , has the Fourier transform  $X_a(j\omega)$ .

The discrete-time signal x[n] is derived from  $x_a(t)$  by periodic sampling:

$$x[n] = x_a(nT)$$
,

where T is a positive constant.

Let  $X(e^{j\Omega})$  denote the discrete-time Fourier transform of x[n]. Show that

$$X(e^{j\Omega}) = \frac{1}{T} \sum_{k=-\infty}^{\infty} X_a \left( j \left( \frac{\Omega}{T} - \frac{2\pi k}{T} \right) \right).$$

[11 marks]

**(b)** The continuous-time signal y(t) is given by

$$y(t) = \sum_{n=-\infty}^{\infty} x[n]h(t - nT),$$

where

$$h(t) = \begin{cases} 1, & 0 < t < T \\ 0.5, & t = 0, T \\ 0, & \text{elsewhere.} \end{cases}$$

Let  $Y(j\omega)$  and  $H(j\omega)$  denote the Fourier transforms of y(t) and h(t) respectively.

(i) Obtain an expression for  $H(j\omega)$  and sketch  $|H(j\omega)|$  for  $|\omega| \le \frac{2\pi}{T}$ .

[5 marks]

(ii) Show that  $Y(j\omega) = X(e^{j\Omega})\Big|_{\Omega = \omega T} H(j\omega)$ .

[4 marks]

**Q.2** (a) Show that a linear, time-invariant, discrete-time system is stable in the bounded-input bounded-output sense if, and only if, the unit-sample response of the system, h[n], is absolutely summable, that is

$$\sum_{n=-\infty}^{\infty} |h[n]| < \infty.$$

[9 marks]

- (b) A causal discrete-time system has a unit-sample response, h[n], which is absolutely summable. Let H(z) denote the z-transform of h[n]. Show that the region of convergence of H(z) includes the unit circle and the entire z-plane outside the unit circle. [7 marks]
- (c) The transfer function, H(z), of a causal linear time-invariant discrete-time system is

$$H(z) = \frac{1 + z^{-1}}{1 - oz^{-1}},$$

where  $\alpha$  is a real constant.

- (i) Determine the unit-sample response of the system, h[n]. [1 mark]
- (ii) For what range of values of  $\alpha$  is the system stable in the bounded-input bounded-output sense? [1 mark]
- (iii) The input to the system is

$$x[n] = \begin{cases} (0.5)^n, & n \ge 0, \\ 0, & n < 0. \end{cases}$$

If  $\alpha = 0.8$ , determine the output y[n].

[2 marks]

**Q.3** (a) A discrete-time filter has a unit-sample response h[n] which is zero for n < 0 and n > N - 1. Let  $H(e^{j\Omega})$  denote the frequency response of the filter. If h[n] = h[N - 1 - n] and N is odd, show that

$$H(e^{j\Omega}) = e^{-j\Omega[(N-1)/2]} \left\{ h\left[\frac{N-1}{2}\right] + \sum_{n=0}^{\left(\frac{N-1}{2}\right)-1} 2h[n]\cos\left[\Omega\left(n - \frac{N-1}{2}\right)\right] \right\}$$

[8 marks]

(b) An ideal discrete-time band-pass filter has a frequency response,  $H_{id}(e^{j\Omega})$ , given by

$$H_{id}(e^{j\Omega}) = \begin{cases} 0, & \left|\Omega\right| < \frac{\pi}{4} \\ 1, & \frac{\pi}{4} < \left|\Omega\right| < \frac{3\pi}{4} \\ 0, & \frac{3\pi}{4} < \left|\Omega\right| \le \pi \end{cases}$$

Obtain an expression for the unit-sample response of this filter.

[7 marks]

(c) A nine point Hamming window,  $w_H[n]$ , is given by

$$w_H[n] = \begin{cases} 0.54 + 0.46\cos\left(\frac{\pi}{4}n\right), & -4 \le n \le 4, \\ 0, & \text{otherwise.} \end{cases}$$

Using the Hamming window, design a causal, nine point finite impulse response filter that approximates the magnitude response of the ideal band-pass filter in part (b).

[5 marks]

**Q.4 (a)** Let  $H_c(s)$  denote the transfer function of a continuous-time filter. The transfer function of a discrete-time filter, H(z), is obtained from  $H_c(s)$  by the following transformation:

$$H(z) = H_c(s) |_{s = (1-z^{-1})/(1+z^{-1})}$$

Show that the frequency responses of the discrete-time and continuous-time filters are related by

$$H(e^{j\Omega}) = H_c(j\omega)\Big|_{\omega = \tan(\Omega/2)}$$
.

[8 marks]

**(b)** A discrete-time high-pass filter with frequency response  $H(e^{j\Omega})$  is to be designed to meet the following specifications:

$$0.89 \le \left| H(e^{j\Omega}) \right| \le 1, \quad 0.6\pi \le |\Omega| \le \pi,$$
 and 
$$\left| H(e^{j\Omega}) \right| \le 0.18, \quad \left| \Omega \right| \le 0.2\pi.$$

The filter is to be designed by applying the bilinear transformation to the transfer function of an appropriate Butterworth continuous-time filter. Verify that a second order filter is sufficient to meet the specifications. Determine the transfer function, H(z), of the discrete-time filter. Note that the transfer function of a second order Butterworth low-pass prototype filter is

$$H(s) = \frac{1}{s^2 + \sqrt{2}s + 1}.$$

[12 marks]

**Q.5 (a)** The sequence x[n] is zero outside the interval  $0 \le n \le N-1$ . Assume that  $N=2^{\nu}$ , where  $\nu$  is a positive integer. Let  $x_1[n]=x[2n]$ , and  $x_2[n]=x[2n+1]$ . Show that the N-point discrete Fourier transform (DFT) of the sequence x[n] can be obtained by appropriately combining the N/2-point DFTs of the sequences  $x_1[n]$  and  $x_2[n]$ .

[8 marks]

**(b)** Consider two finite duration signals x[n] and h[n] where both are zero for n < 0 and where

$$x[n] = 0$$
,  $n \ge 32$ 

$$h[n] = 0$$
,  $n \ge 8$ .

The 32-point DFTs of each of the signals are multiplied and the inverse DFT computed. Let r[n] denote this inverse DFT.

The sequence y[n] is obtained by linearly convolving x[n] and h[n]. Specify the values of n for which r[n] is guaranteed to be equal to y[n].

[7 marks]

(c) A 15,000 point sequence is to be linearly convolved with a sequence that is 80 points long. The convolution is to be implemented using DFTs and inverse DFTs of length 512. If the overlap-add method is used, what is the minimum number of 512-point DFTs and the minimum number of 512-point inverse DFTs needed to implement the convolution for the entire 15,000 point sequence?

[5 marks]