



Department of Electrical & Computer Engineering
ENEE4113 - Communications Laboratory

Experiment #4 Frequency Modulation

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Section: 4

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Contents

1	Theory	1
1.1	Modulation Scheme	1
1.1.1	Modulation	2
1.1.2	Demodulation - Phase Locked Loop	2
1.2	Carson's Rule	3
1.3	Besel Functions	3
1.4	Zero Crossing Decay	4
2	Procedure and Data Analysis	5
2.1	Modulation	5
2.1.1	Time Domain	5
2.1.2	Frequency Domain	5
2.1.3	Setting Carrier Frequency	6
2.1.4	Characteristics of the FM Modulator	6
2.1.5	FM signal spectrum for different message signals	7
2.1.6	Zero Carrier Crossing	9
2.2	Demodulation	11
2.2.1	Demodulated Signal	11
2.2.2	Loop Filter Effect	12
3	Conclusion	15
	References	16

List of Figures

1	Frequency Modulation Scheme	1
2	Narrowband FM Generation	2
3	Wideband FM Generation	2
4	Phase Locked Loop	3
5	Time Domain	5
6	Frequency Domain	5
7	Carrier Frequency	6
8	Carrier Frequency vs Message Voltage - k_f	7
9	S(F) - 3kHz Message Signal	8
10	S(F) - 200Hz Message Signal	8
11	S(F) - Square Message Signal	9
12	First Amplitude Zero Crossing Decay	10
13	Frequency Zero Crossing Decay	10
14	Demodulated Signal in the time domain	11
15	Demodulated Signal in the frequency domain	11
16	τ_1 vs τ_2 without pre-emphasis	12
17	τ_1 vs τ_2 with pre-emphasis	13
18	τ_1 with pre-emphasis vs τ_1 without pre-emphasis	14
19	τ_2 with pre-emphasis vs τ_2 without pre-emphasis	14

List of Tables

1	Carrier Frequency vs Message Voltage - k_f	7
2	Amplitude zero crossing values	9
3	Amplitude zero crossing values	10
4	Amplitude Values with Pre-emphasis	12
5	Amplitude Values without Pre-emphasis	13

1 Theory

1.1 Modulation Scheme

Frequency modulation is a part of angle modulation, angle modulation is a technique with constant carrier amplitude and varying phase or time derivative of phase. an FM signal can be expressed as:

$$s(t) = A_c \cos(\omega_c t + 2\pi k_f \int_{-\infty}^t m(\tau) d\tau) \quad (1)$$

- A_c is the carrier amplitude
- ω_c is the carrier frequency
- k_f is the modulation sensitivity
- $m(t)$ is the message signal

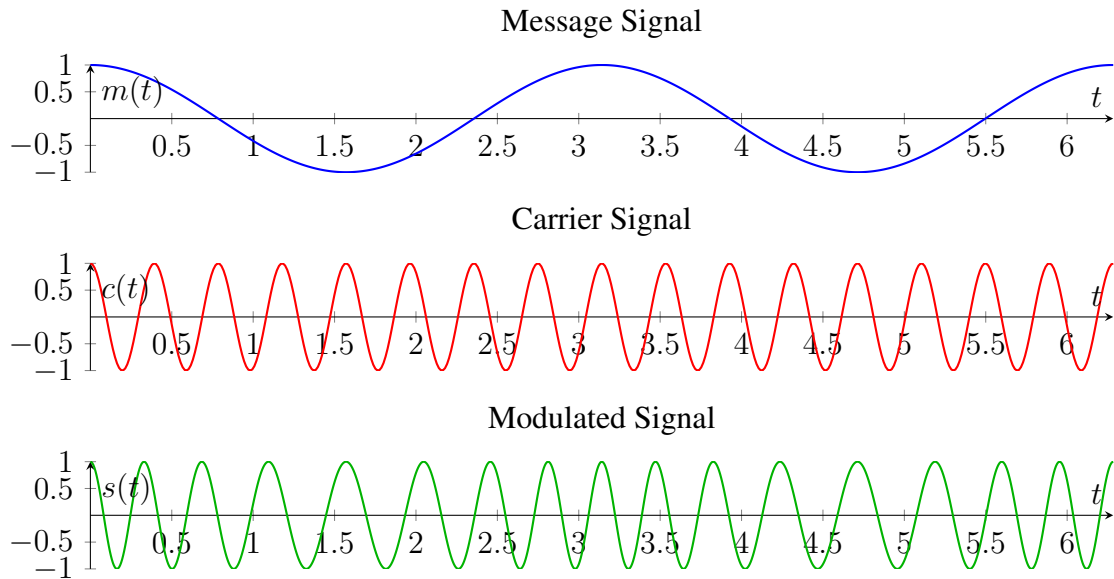


Figure 1: Frequency Modulation Scheme

1.1.1 Modulation

There are two types of modulation, narrowband and wideband. narrowband modulation is when the modulation index is less than 1, and wideband modulation is when the modulation index is greater than 1. The modulation index is defined as:

$$\beta = \frac{\Delta f}{f_m} = \frac{k_f A_m}{f_m} \quad (2)$$

The simplest way of generating an FM signal is using a voltage-controlled oscillator, which is not a stable method. the other method is generating a narrowband FM signal and then using a frequency multiplier to generate a wideband FM signal.

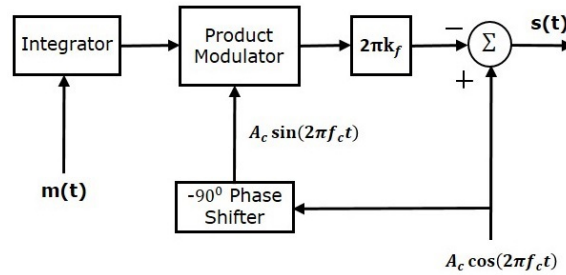


Figure 2: Narrowband FM Generation
[1]

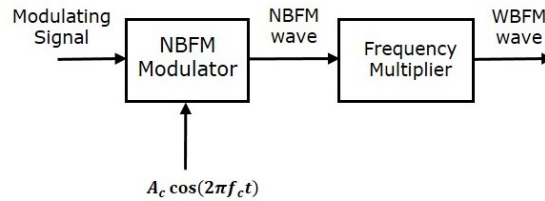


Figure 3: Wideband FM Generation
[1]

1.1.2 Demodulation - Phase Locked Loop

For demodulation, a common method is the phase-locked loop (PLL) a feedback system that synchronizes the phase of the output signal with the phase of the input signal. The

phase detector compares the phase of the input signal with the phase of the output signal and generates an error signal that is used to adjust the phase of the output signal. The output of the phase detector is filtered using a low-pass filter to remove high-frequency components, resulting in a demodulated signal.

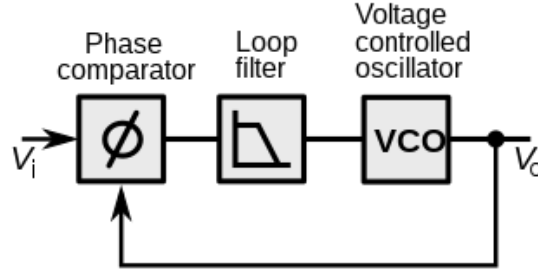


Figure 4: Phase Locked Loop
[2]

1.2 Carson's Rule

Carson's rule is a rule of thumb that gives the approximate bandwidth of an FM signal. It states that A 98% power B.W of an FM signal can be estimated by:

$$B_T \approx 2(\beta + 1)f_m \quad (3)$$

The rule works well when the message signal is continuous. However, it cannot be used when the message contains discontinuities, such as in the case of a square function. [1]

1.3 Bessel Functions

A more accurate expression for the FM signal bandwidth can be obtained using Bessel functions. The FM signal can be expressed as:

$$s(t) = A_c \cos(\omega_c t + \beta \sin(\omega_m t)) \quad (4)$$

where β is the modulation index, and ω_m is the message signal frequency. In terms of Bessel functions, the FM signal can be expressed as:

$$s(t) = A_c \sum_{n=-\infty}^{\infty} J_n(\beta) \cos(\omega_c t + n\omega_m t) \quad (5)$$

In order to find the bandwidth of the FM signal that achieves a certain percentage of the total power, the following equation can be used:

$$J_0^2(\beta) + 2 \sum_{n=1}^m J_n^2(\beta) = p \quad (6)$$

where p is the percentage of the total power, and m is the number of Bessel functions to be summed, the bandwidth of the FM signal can be calculated using:

$$B_T = 2 * m * f_m \quad (7)$$

1.4 Zero Crossing Decay

Zero Crossing is a method to get rid of the carrier spectra, and since the spectral components are proportional to the Bessel functions, the zero crossing of the carrier signal occurs when the Bessel function is zero $J_0(\beta) = 0$.

2 Procedure and Data Analysis

2.1 Modulation

2.1.1 Time Domain

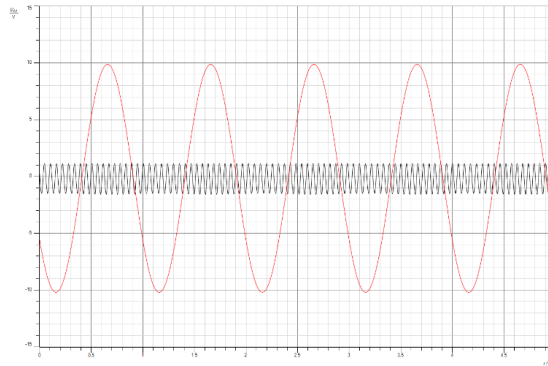


Figure 5: Time Domain

2.1.2 Frequency Domain

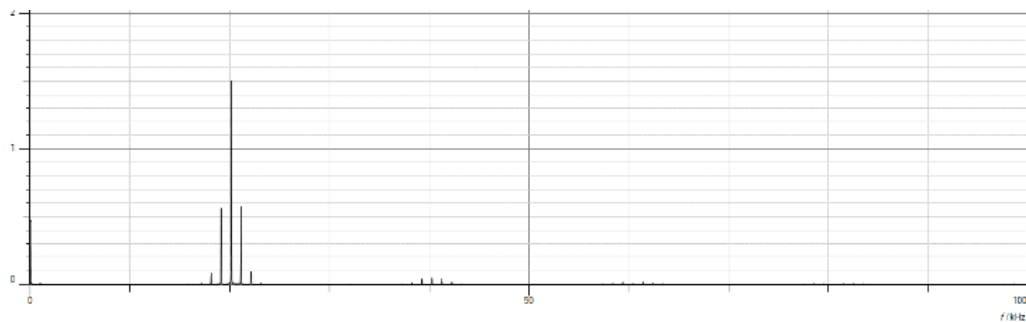


Figure 6: Frequency Domain

The modulated signal isn't affected by the amplitude of the message signal, but the frequency of the message signal affects the frequency of the modulated signal.

2.1.3 Setting Carrier Frequency

In order to set the carrier frequency to 20kHz, we set the message signal to 0, so the modulated signal is just the carrier signal.

$$\begin{aligned} s(t) &= A_c \cos(\omega_c t + \beta \int_0^t m(\tau) d\tau) \\ s(t) &= A_c \cos(\omega_c t + \beta \times 0) \\ s(t) &= A_c \cos(\omega_c t) \end{aligned} \tag{8}$$

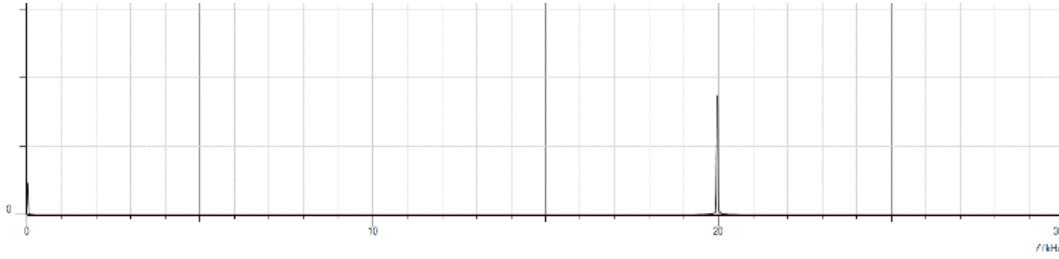


Figure 7: Carrier Frequency

from the classy lab sample, we can see that the carrier frequency is 19.91kHz, which is close to 20kHz.

2.1.4 Characteristics of the FM Modulator

The main Characteristic of the FM modulator is the modulation sensitivity K_f , using the instantaneous frequency equation:

$$f_i = f_c + k_f m(t) \tag{9}$$

using a constant amplitude would result

$$f_i = f_c + k_f \times A_m \tag{10}$$

with f_c being the carrier frequency, k_f being the modulation sensetivity, and A_m being the amplitude of the message signal.

Voltage	-10	-8	-6	-4	-2	0	2	4	6	8	10
Frequency	19.06	19.2	19.32	19.45	19.59	19.73	19.85	20	20.13	20.26	20.39

Table 1: Carrier Frequency vs Message Voltage - k_f

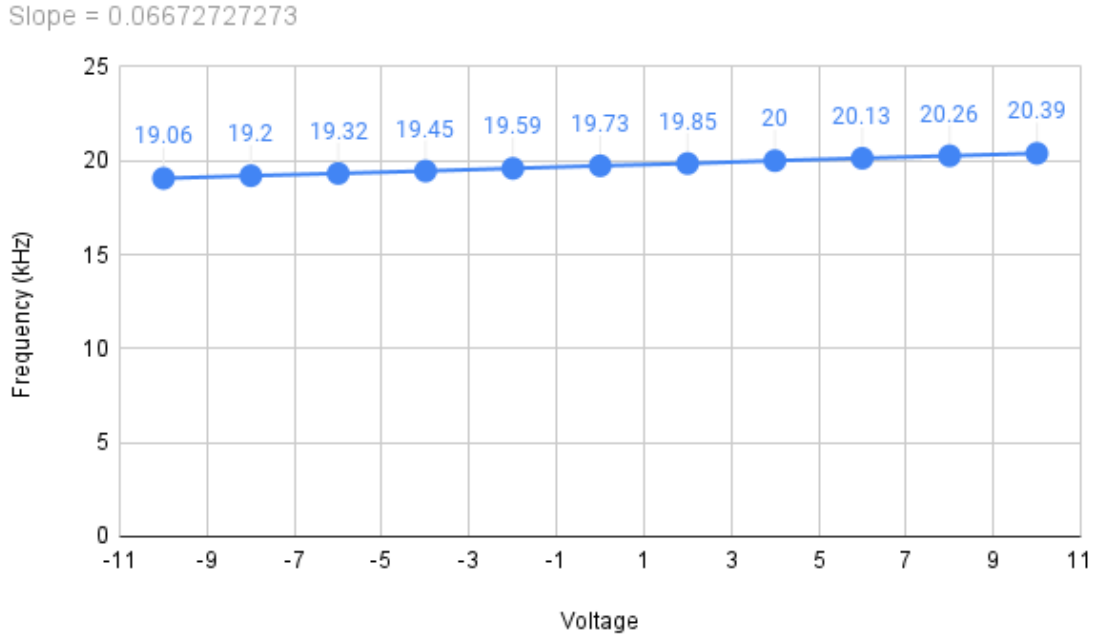


Figure 8: Carrier Frequency vs Message Voltage - k_f

the slope of the graph is the modulation sensitivity k_f , which is 66.6 Hz/V. For a 10V message signal, the maximum frequency deviation is given by:

$$\Delta f = k_f \times A_m = 66.6 \times 10 = 666 \text{ Hz} \quad (11)$$

2.1.5 FM signal spectrum for different message signals

In this part, we used a 10V sinusoidal message signal with two frequencies 3kHz, and 200Hz.

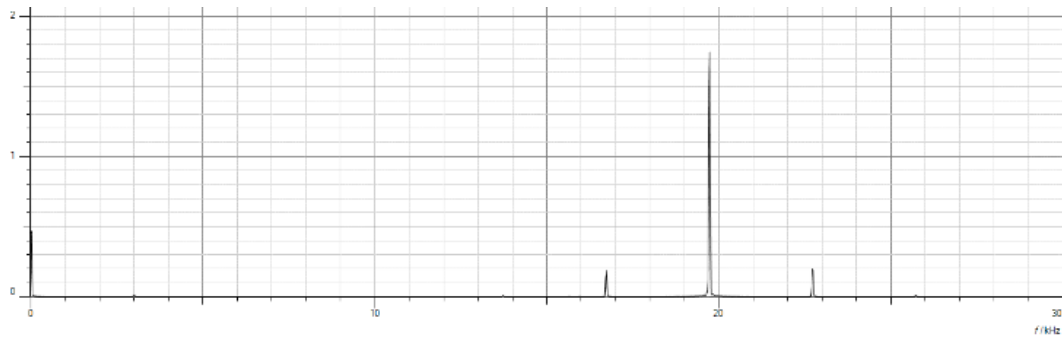


Figure 9: $S(F)$ - 3kHz Message Signal

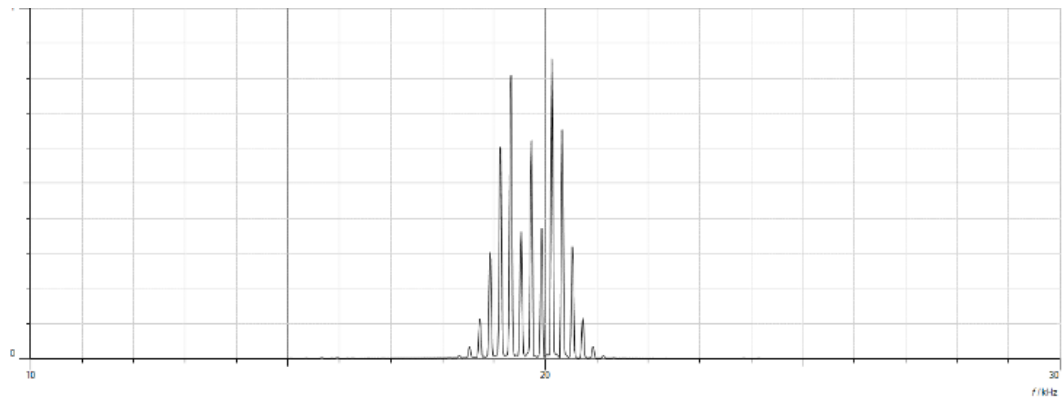


Figure 10: $S(F)$ - 200Hz Message Signal

We notice that the density of the spectrum is higher for the 200Hz message signal, this is because the spectra are 3kHz apart in the first case, and 200Hz apart in the second case. for a square message signal, with 200Hz frequency, 0.5 duty cycle, and 10V the spectrum is as follows:

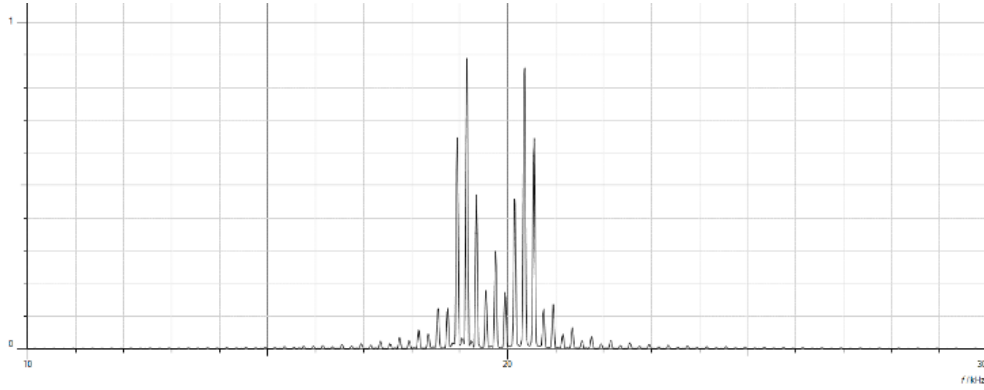


Figure 11: S(F) - Square Message Signal

We noticed that the spectrum is similar to the spectrum of the 200Hz sinusoidal message signal, but the density of the spectrum is higher, this is because the square wave has more harmonics than the sinusoidal wave.

2.1.6 Zero Carrier Crossing

Since the spectral components are proportional to the Bessel functions, the zero crossing of the carrier signal occurs when the Bessel function is zero. The first three values for β that satisfy this condition are 2.4048, 5.5201, and 8.6537. For a constant frequency of 100Hz, the corresponding amplitudes can be calculated as follows:

$$\beta = \frac{K_f \times A_m}{f_m} \quad (12)$$

β	2.4048	5.5201	8.6537
A_m	3.643	8.363	13.111

Table 2: Amplitude zero crossing values

experimentally, we found the first decay at 7.3 VSS, which means $A_M = 3.65$, which is close to the theoretical value.

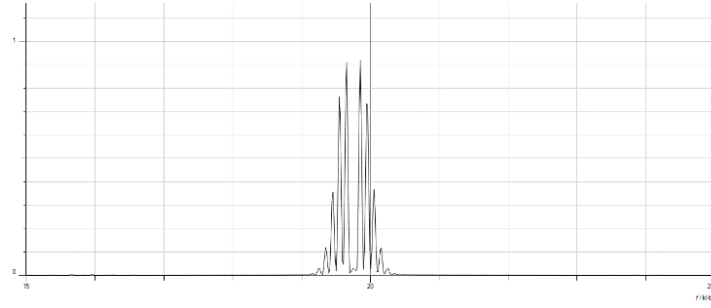


Figure 12: First Amplitude Zero Crossing Decay

Another way of finding the first zero crossing is to use a constant amplitude and vary the frequency. Using (Equation12) and a constant 10V amplitude we can fill the following table:

β	2.4048	5.5201	8.6537
Frequency	276.94	120.65	76.96

Table 3: Amplitude zero crossing values

experimentally, we found that decay at 280kHz is not necessarily the first decay since we didn't check for smaller values.

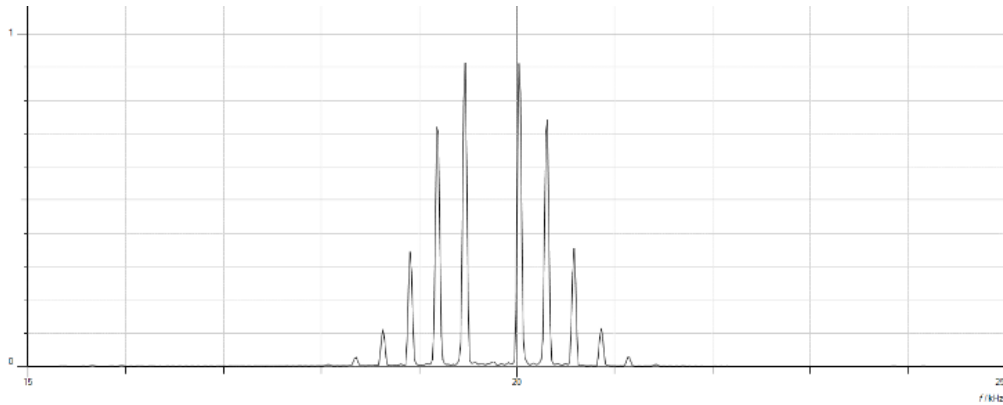


Figure 13: Frequency Zero Crossing Decay

2.2 Demodulation

2.2.1 Demodulated Signal

Setting $V_{ss} = 10 \text{ V}$ and $f_m = 500 \text{ Hz}$ and using demodulator τ_2 filter, we were able to recover the message using PLL demodulation as shown in the following figures:

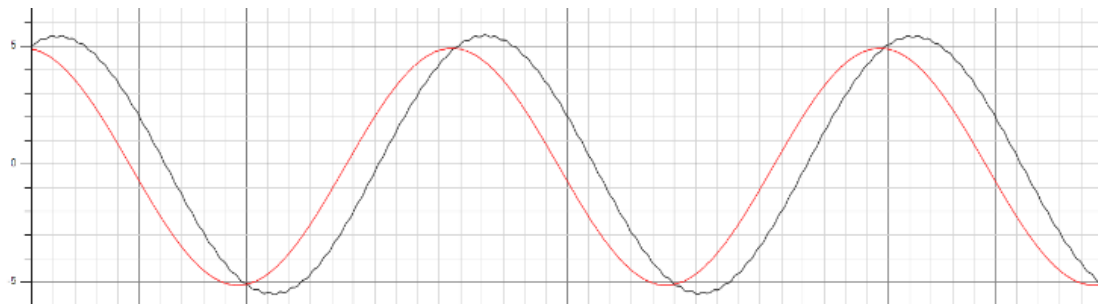


Figure 14: Demodulated Signal in the time domain

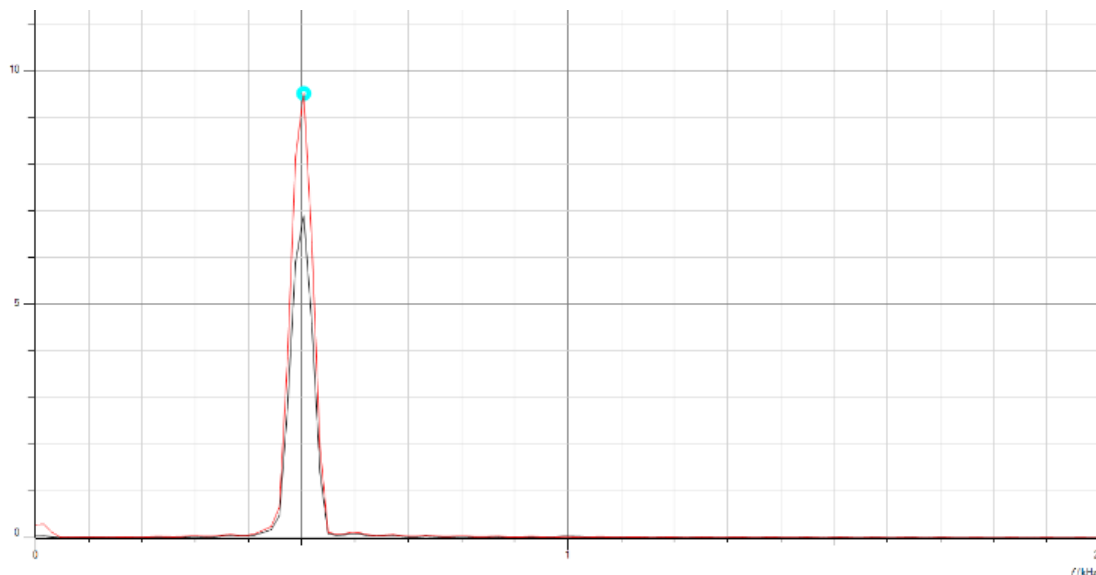


Figure 15: Demodulated Signal in the frequency domain

2.2.2 Loop Filter Effect

for a message signal with 4V amplitude and 500Hz frequency, we used two different loop filters, τ_1 and τ_2 , and we measured the amplitude of the demodulated signal for different message frequencies and the results are as follows:

Frequency	A_D using τ_1	A_D using τ_2
500	8.9	2.75
1000	4.31	2.57
1500	2.93	2.55
2000	2.1	2.25
3000	1.19	1.45
4000	0.81	0.96

Table 4: Amplitude Values with Pre-emphasis

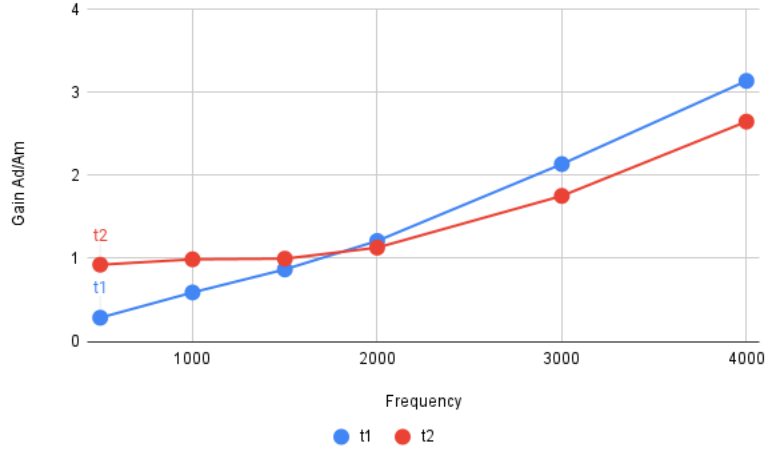


Figure 16: τ_1 vs τ_2 without pre-emphasis

We notice that the loop filter τ_2 is less sensitive to the message frequency than τ_1 , this is because τ_2 has a lower cut-off frequency than τ_1 .

Frequeuncy	A_D using τ_1	A_D using τ_2
500	11.55	3.56
1000	8.87	5.27
1500	7.31	6.36
2000	5.94	6.37
3000	3.61	4.39
4000	2.44	2.9

Table 5: Amplitude Values without Pre-emphasis

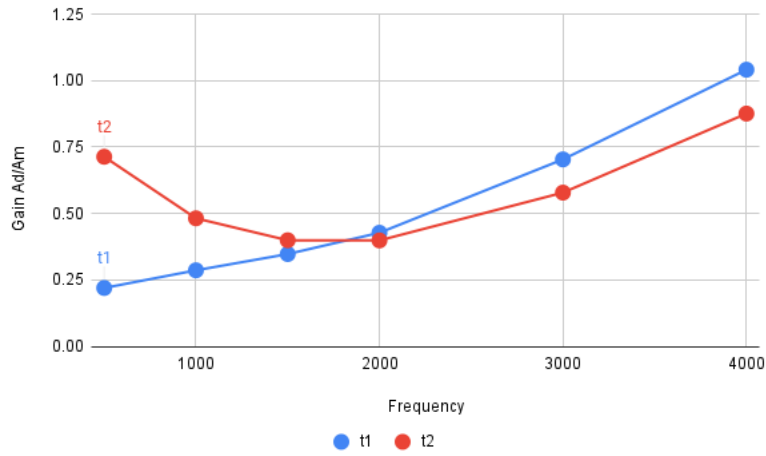


Figure 17: τ_1 vs τ_2 with pre-emphasis

after using the pre-emphasis filter, we notice that the amplitude of the demodulated signal is higher than before, this is because the pre-emphasis filter amplifies the high-frequency portion of the signal, which improves the signal-to-noise ratio of the high-frequency portion of the baseband.

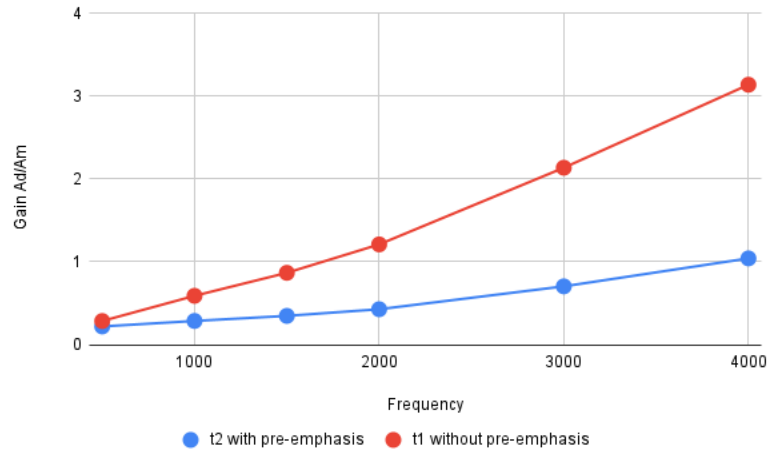


Figure 18: τ_1 with pre-emphasis vs τ_1 without pre-emphasis

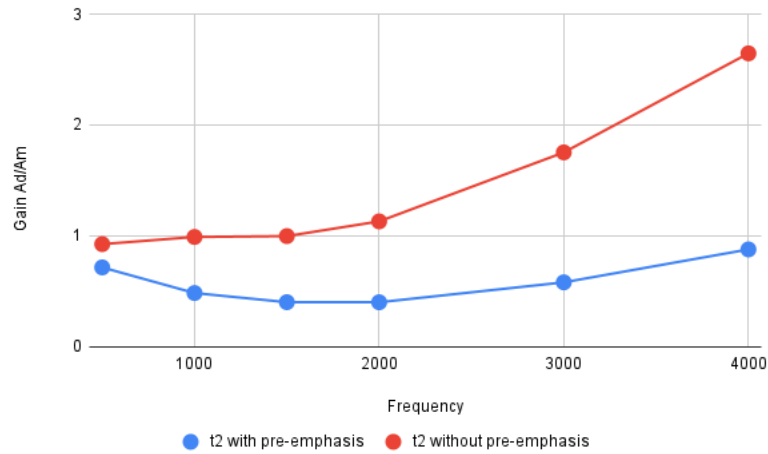


Figure 19: τ_2 with pre-emphasis vs τ_2 without pre-emphasis

The pre-emphasis helps in equalizing the message signal power in terms of deviation ratio. Without pre-emphasis, the received audio would sound unacceptably noisy at high frequencies, especially under conditions of low carrier-to-noise ratio. Pre-emphasis increases the magnitude of the higher signal frequencies, thereby improving the signal-to-noise ratio.

3 Conclusion

In summary, this report illustrates the process of modulating and demodulating FM signals and studied the effect of different parameters such as message frequency, type, amplitude, carrier zeroing, pre-emphasis, and more. The use of pre-emphasis has been shown to improve the signal-to-noise ratio of the high-frequency portion of the baseband.

References

- [1] Tutorialspoint. Analog communication - fm modulators.
https://www.tutorialspoint.com/analog_communication/analog_communication_fm_modulators.htm
- [2] Wikipedia. Phase-locked loop. https://en.wikipedia.org/wiki/Phase-locked_loop