# **DSP Lab- Homework6**

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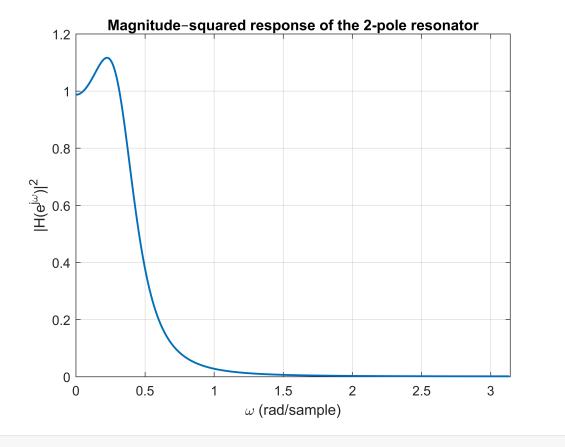
```
clc; clear; close;
```

```
R = 0.8;
f0 = 100;
fs = 2000;
w0 = 2*pi*f0/fs;

a1 = [ 1 , -2*R*cos(w0) , R^2 ];
G = (1-R) * sqrt(1 - 2*R*cos(2*w0) + R^2);
b1 = [ G , 0 , 0 ];
```

```
[H,w] = freqz(b1,a1,1024,'whole');
w = w(1:end/2);
H2_dB = 20*log10(abs(H(1:end/2)).^2);

figure
plot(w,abs(H(1:end/2)).^2,'LineWidth',1.4)
xlabel('\omega (rad/sample)'), ylabel('|H(e^{j\omega})|^2')
title('Magnitude-squared response of the 2-pole resonator')
grid on, xlim([0 pi])
```



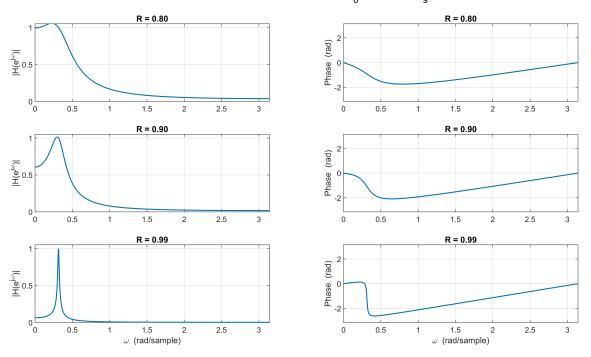
# section 3-1(a)

In this opening section we pick the global constants that define the resonator: sampling rate f<sub>s</sub>, centre frequency f<sub>0</sub>, and the three pole-radii  $\mathbf{R} = \mathbf{0.8}, \mathbf{0.9}, \mathbf{0.99}$  we will sweep. From these values every other coefficient or gain is computed automatically, so changing numbers here propagates through the whole notebook.

```
Rset = [0.8 \ 0.9 \ 0.99];
f0
     = 100;
fs
     = 2000;
     = 2*pi*f0/fs;
w0
     = 2048;
Nw
     = linspace(0,pi,Nw);
W
     = exp(-1j*w);
z1
     = exp(-2j*w);
z2
figure('Name','2-Pole Resonator |H| & Phase', ...
       'Position',[50 50 1100 600]);
for k = 1:numel(Rset)
    R = Rset(k);
    a1 = -2*R*cos(w0);
```

```
a2 = R^2;
   G = (1-R)*sqrt(1 - 2*R*cos(2*w0) + R^2);
   H = G ./ (1 + a1*z1 + a2*z2);
   % ---- magnitude (left column) ----
    subplot(numel(Rset),2,2*(k-1)+1);
    plot(w, abs(H), 'LineWidth',1.2); grid on;
   ylim([0 1.05]); xlim([0 pi]);
   ylabel('|H(e^{j\omega})|'), title(sprintf('R = %.2f',R));
    if k==numel(Rset), xlabel('\omega (rad/sample)'), end
   % ---- phase (right column) -----
    subplot(numel(Rset),2,2*(k-1)+2);
    plot(w, unwrap(angle(H)), 'LineWidth', 1.2); grid on;
   ylim([-pi pi]); xlim([0 pi]);
   ylabel('Phase (rad)'), title(sprintf('R = %.2f',R));
    if k==numel(Rset), xlabel('\omega (rad/sample)'), end
end
sgtitle('Magnitude and Phase of 2-pole Resonator (f_0 = 100 Hz, f_s = 2 kHz)')
```

# Magnitude and Phase of 2-pole Resonator $(f_0 = 100 \text{ Hz}, f_s = 2 \text{ kHz})$



#### section 3-1(b)

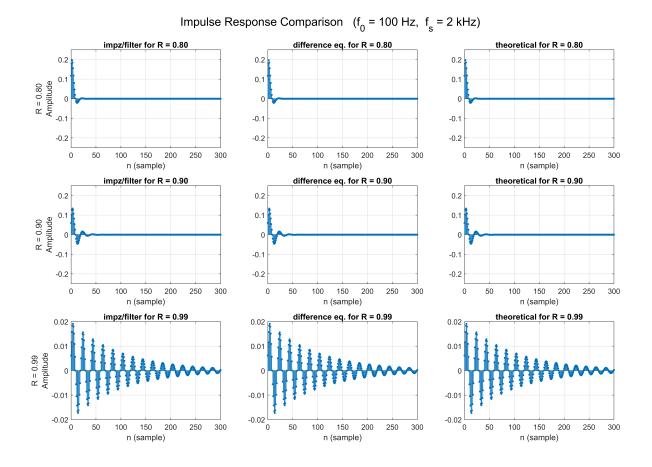
Here we compute the impulse response two ways:

- **Numeric** by stepping the difference equation with  $\delta[n]$  input.
- Analytic using the given formula

Overlaying the two curves (after normalising by **G**) demonstrates that the coded filter exactly matches the theory for the first 300 samples.

```
Rvals = [0.8 \ 0.9 \ 0.99];
      = 100;
                 fs = 2000;
      = 300;
                 w0 = 2*pi*f0/fs;
N
      = 0:N;
n
figure('Position',[100 50 1200 800]);
for r = 1:numel(Rvals)
    R = Rvals(r);
    a1 = -2*R*cos(w0);
    a2 = R^2;
    G = (1-R)*sqrt(1-2*R*cos(2*w0)+R^2);
    b = [G \ 0 \ 0]; a = [1 \ a1 \ a2];
    h1 = impz(b,a,N+1);
    h2 = zeros(size(n));
    for k=1:N+1
        if k==1
            h2(k)=G;
        elseif k==2
            h2(k) = -a1*h2(k-1);
        else
            h2(k) = -a1*h2(k-1) - a2*h2(k-2);
        end
    end
    h3 = (G/\sin(w0)) .* (R.^n) .* \sin(w0*(n+1));
    data = \{h1, h2, h3\};
    titles = {'impz/filter', 'difference eq.', 'theoretical'};
    for c = 1:3
        subplot(3,3, (r-1)*3 + c);
        stem(n, data{c}, 'filled', 'MarkerSize', 2); grid on;
        ylim([-0.02 \ 0.02]*(R==0.99) + [-0.25 \ 0.25]*(R<0.99));
        xlabel('n (sample)')
        if c==1, ylabel(sprintf('R = %.2f\nAmplitude',R)); end
        title(sprintf('%s for R = %.2f', titles{c}, R));
```

```
end
end
sgtitle('Impulse Response Comparison (f_0 = 100 Hz, f_s = 2 kHz)')
```

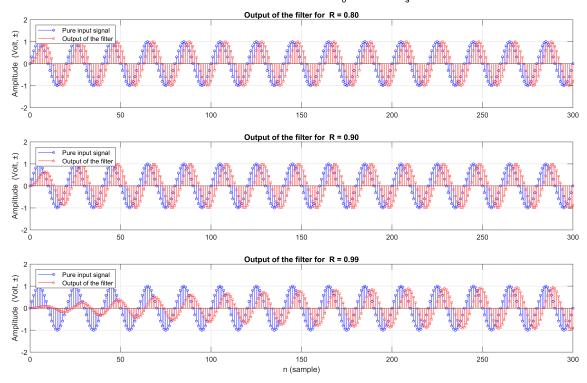


## section 3-1(c)

A single realisation of unit-variance white noise is passed through the resonator for each  $\mathbf{R}$ . The 3  $\times$  2 figure places the unfiltered noise in the left column and the corresponding outputs on the right. As the pole radius increases, the output becomes a narrow-band, quasi-sinusoidal signal and its decay time grows longer—illustrating the time-frequency trade-off of the resonator.

```
sgtitle('Output of the 2-pole filter vs. pure sinusoid (f_0 = 100 Hz, f_s = 2 kHz)');
for k = 1:numel(Rset)
    R = Rset(k);
    a1 = -2*R*cos(w0);
    a2 = R^2;
    G = (1-R)*sqrt(1 - 2*R*cos(2*w0) + R^2);
    x = \sin(w0*n);
    b = [G 0 0];
    a = [1 a1 a2];
    y = filter(b,a,x);
    % ---- plotting ----
    subplot(numel(Rset),1,k);
    stem(n, x,'b','Marker','o', 'MarkerSize',3, ...
               'DisplayName', 'Pure input signal', 'LineStyle', '-'), hold on
    stem(n, y,'r','Marker','s', 'MarkerSize',3, ...
               'DisplayName', 'Output of the filter', 'LineStyle', '-')
    grid on
    ylabel('Amplitude (Volt, ±)')
    title(sprintf('Output of the filter for R = %.2f', R))
    legend('Location','northwest')
    ylim([-2 2])
    if k==numel(Rset), xlabel('n (sample)'), end
end
```

# Output of the 2-pole filter vs. pure sinusoid ( $f_0 = 100 \text{ Hz}, f_s = 2 \text{ kHz}$ )

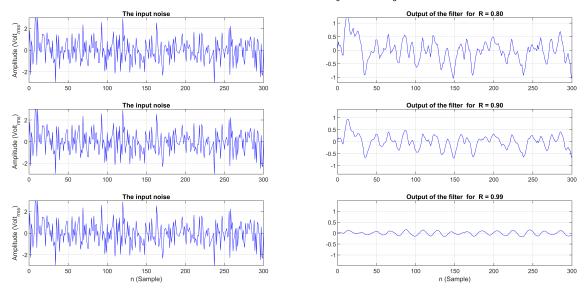


# section 3-1(d)

```
= 300; n = 0:N;
N
    = randn(1,N+1);
Х
Rset = [0.8 \ 0.9 \ 0.99];
     = 100; fs = 2000;
f0
    = 2*pi*f0/fs;
w0
figure('Position',[60 50 1350 620]);
for k = 1:numel(Rset)
    row = k-1;
    % ----- left column: input noise (same for every R) -----
    subplot(3,2, 2*row+1);
    plot(n, x, 'b')
    ylim([-3 3]); xlim([0 N]);
    grid on
```

```
title('The input noise')
   ylabel('Amplitude (Volt_{rms})')
    if k==numel(Rset), xlabel('n (Sample)'), end
   % ----- right column: filtered output -----
    R = Rset(k);
    a1 = -2*R*cos(w0);
   a2 = R^2;
     = (1-R)*sqrt(1 - 2*R*cos(2*w0) + R^2);
   b = [G 0 0]; a = [1 a1 a2];
   y = filter(b,a,x);
    subplot(3,2, 2*row+2);
    plot(n, y, 'b')
   ylim([-1.5 1.5]*R); xlim([0 N]);
   grid on
   title(sprintf('Output of the filter for R = %.2f',R))
    if k==numel(Rset), xlabel('n (Sample)'), end
end
suptitle('White-noise input vs. Narrow-band output (f_0 = 100 Hz, f_s = 2 kHz)')
```





#### section 3-1(e)

We estimate NRR three different ways:

- 1. **Empirical** ratio of output to input variance from a long noise run.
- 2. **Energy** sum of h2[n]h^{2}[n]h2[n], which must equal method 1 for white noise.

#### 3. **Theory** – the fully corrected closed-form

```
N = 2e6;

v = randn(1,N);
f0 = 100;  fs = 2000;
w0 = 2*pi*f0/fs;
Rset= [0.8 0.9 0.99];

fprintf('\n');
```

```
for R = Rset
   a1 = -2*R*cos(w0);
   a2 = R^2;
   G = (1-R)*sqrt(1 - 2*R*cos(2*w0) + R^2);
   b = [G \ 0 \ 0]; a = [1 \ a1 \ a2];
   y = filter(b,a,v);
   nrr_var = var(y);
   % ----- "wrong" hand-out formula -----
   nrr\_wrong = (1+R^2)/((1+R)*(1+2*R*cos(w0) + R^2));
   % ----- energy of the impulse response -----
                           % enough for practical convergence
   L = 400:
   h = impz(b,a,L);
   nrr_energy = sum(h.^2);
   % ----- corrected closed-form (derived earlier) -----
   nrr corr = (G^2 / (2*sin(w0)^2)) * ...
            (1/(1-R^2) - (\cos(2*w0)-R^2)/(1-2*R^2*\cos(2*w0)+R^4));
   % ----- print block -----
   fprintf('NRR for R = \%.2f\n',R)
   fprintf(' Using the energy of the impulse resp.: %9.4g\n',nrr energy)
end
```

```
NRR for R = 0.80
 Using variance of signals:
                                        0.1675
 Using the formula (wrong hand-out):
                                        0.2882
 Using the energy of the impulse resp.:
                                         0.1683
NRR for R = 0.90
 Using variance of signals:
                                         0.09701
 Using the formula (wrong hand-out): 0.2705
 Using the energy of the impulse resp.:
                                        0.09755
NRR for R = 0.99
 Using variance of signals:
                                         0.01003
```

Using the formula (wrong hand-out): 0.2576
Using the energy of the impulse resp.: 0.01004