

Linear Dynamical Systems - 2

Malintha Fernando

Department of Intelligent Systems Engineering

Indiana University

*** Examples and codes have been adopted from the book “Data Driven Science & Engineering - Machine Learning, Dynamical Systems, and Control” by Steve Brunton and Nathan Kutz.**

Outline

- Feedback Control Systems
- Control Laws
- Linearization of Nonlinear Dynamical Systems
- Stability of Linear Dynamical Systems
- Inverted Pendulum Example
- Optimal control and Linear Quadratic Regulator

Why Feedback Control?

- To **stabilize** an unstable system.
- To account for **external disturbances**.
- To account for system inherent disturbances such as
 - Unmodeled dynamics,
 - Measurement errors,
 - Model uncertainties.
- To inhibit some **desired behavior** with dynamical systems.
i.e.: trajectory tracking
- By calculating “some gain” parameters (linear control law) we can perform feedback control.

Control Laws

- A control law is a set of parameters that is used to regulate the control input for a system.
- Consider the control law $u = K(r - x)$ for a linear dynamical system. Here, r is the desired state of the system.

$$\dot{\mathbf{x}} = \mathbf{A}\mathbf{x} + \mathbf{B}\mathbf{K}(\mathbf{r} - \mathbf{x})$$

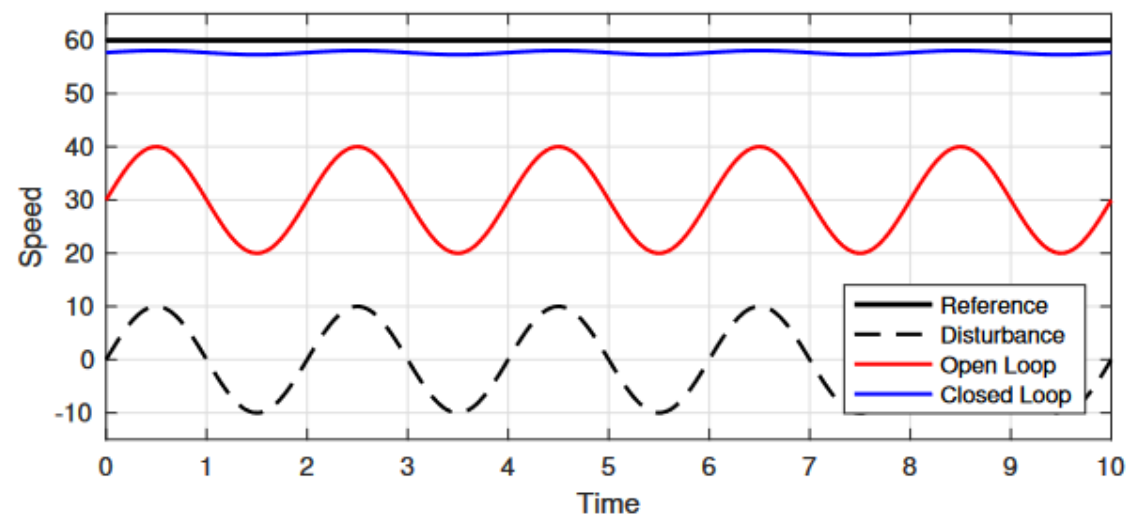
$$\dot{\mathbf{x}} = \mathbf{A}\mathbf{x} + \mathbf{B}\mathbf{K}\mathbf{r} - \mathbf{B}\mathbf{K}\mathbf{x}$$

$$\dot{\mathbf{x}} = (\mathbf{A} - \mathbf{B}\mathbf{K})\mathbf{x} + \mathbf{B}\mathbf{K}\mathbf{r}$$

- Here, A, B are inherent to the system. Thus, by choosing K appropriately, we can bring the system to stabilization, aka to r .

Cruise Control Example

Cruise control with closed (feedback) and open loop control

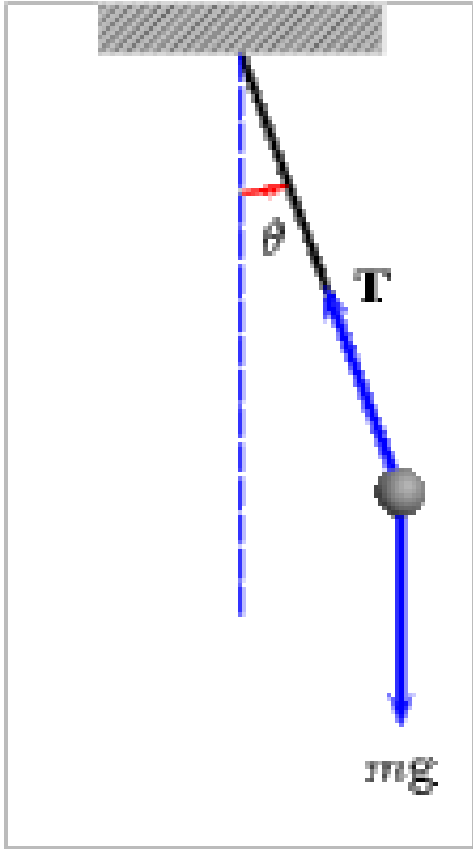


In the open loop control, we set the accelerator to a fixed value. Closed loop control is a result of P control.

$$u = k(r - x), \quad k = 50, r = 60$$

What are Fixed Points?

- A fixed point is any point where our dynamical system is at an equilibrium (or stabilized).
- Formally, \bar{x} is a fixed point if $\dot{x} = f(\bar{x}) = 0$.
- i.e.: A pendulum has two fixed points, $\theta = 0, \theta = \Pi$.
- In trajectory tracking, a fixed point can be any point on the trajectory, that we would like the robot to converge.

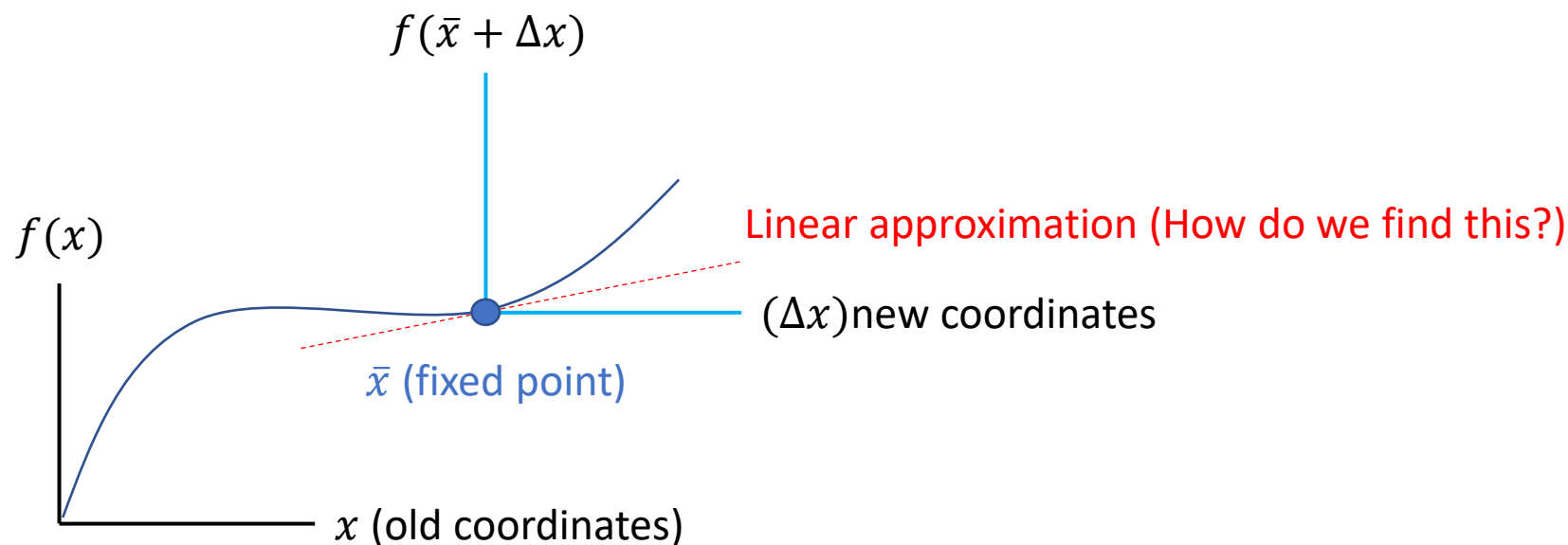


Linearization of Nonlinear Dynamics

- Most dynamical systems in practice are nonlinear.
i.e.: Car, Alfabot, Population dynamic models etc.
- However, linear dynamics have been well studied and established. But nonlinear dynamics are not.
- To analyze nonlinear dynamical systems, we linearize them around their fixed (aka equilibrium/desired) points.
- We assume these linear approximations are valid for a finite neighborhood of points around the fixed points.
- If we look carefully, *most nonlinear dynamical systems are linear for a short duration.*

Linearization

- The key idea of linearization is to **find a linear approximation** of the nonlinear system that is valid for a **close neighborhood around a fixed point**.
- This also defines a new coordinate system whose origin is a **fixed point**. Any measurement on the new coordinate system denotes the deviation of the system from the fixed point.



Linearization of Nonlinear Dynamics

- Consider some nonlinear system of equations in the form $\dot{x} = f(x, u)$.
i.e.: $\dot{x}_1 = f_1(x, u), \dot{x}_2 = f_2(x, u), \dots$ etc.
- Identify the fixed point (\bar{x}, \bar{u}) . Here \bar{u} is the control input at the fixed point \bar{x} .
- Find the deviation from the fixed point.

$$\begin{aligned}\Delta x &= x(t) - \bar{x} \\ \Delta u &= u(t) - \bar{u}\end{aligned}$$

- Now write the Taylor expansion for the system around the fixed point.

$$f(\bar{x} + \Delta x, \bar{u} + \Delta u) = f(\bar{x}, \bar{u}) + \underbrace{\left. \frac{df}{dx} \right|_{(\bar{x}, \bar{u})}}_A \cdot \Delta x + \underbrace{\left. \frac{df}{du} \right|_{(\bar{x}, \bar{u})}}_B \cdot \Delta u + \dots$$

- For small deviations from the fixed points, the higher order terms become negligible.

Gist

- We can obtain the linear dynamical system matrices from a nonlinear differential equations system as below.
- Evaluate the two matrices at the fixed point (\bar{x}, \bar{u}) to get the *numerical* linearized matrices at any point.

$$A = \frac{df}{dx} = \begin{bmatrix} \frac{\partial f_1}{\partial x_1} & \frac{\partial f_1}{\partial x_2} & \cdots \\ \frac{\partial f_2}{\partial x_1} & \frac{\partial f_2}{\partial x_2} & \cdots \\ \vdots & \vdots & \frac{\partial f_n}{\partial x_m} \end{bmatrix} \bigg|_{\bar{x}}$$

$$B = \frac{df}{du} = \begin{bmatrix} \frac{\partial f_1}{\partial u_1} & \frac{\partial f_1}{\partial u_2} & \cdots \\ \frac{\partial f_2}{\partial u_1} & \frac{\partial f_2}{\partial u_2} & \cdots \\ \vdots & \vdots & \frac{\partial f_n}{\partial u_l} \end{bmatrix} \bigg|_{\bar{u}}$$

- In trajectory following, the fixed point could be any “close enough” point in the given **trajectory**.

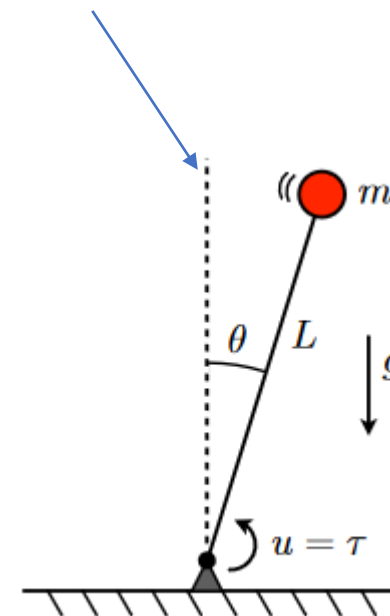
Linearization - Pendulum Example

- Let's consider an inverted pendulum attached to a light rod, angling slightly away from a **fixed point**.
- Write the equations of motion for this system.

$$\ddot{\theta} = -\frac{g}{L} \sin(\theta) + u.$$

- But this doesn't fit our formula. Let's write them as first order differential equations. (Recall the example from last lecture).

$$\mathbf{x} = \begin{bmatrix} x_1 \\ x_2 \end{bmatrix} = \begin{bmatrix} \theta \\ \dot{\theta} \end{bmatrix} \quad \Rightarrow \quad \frac{d}{dt} \begin{bmatrix} x_1 \\ x_2 \end{bmatrix} = \begin{bmatrix} x_2 \\ -\frac{g}{L} \sin(x_1) + u \end{bmatrix}$$



- Call these functions f_1, f_2 .
- Now do the partial differentiation to get the Jacobian matrices.

Linearization - Pendulum Example

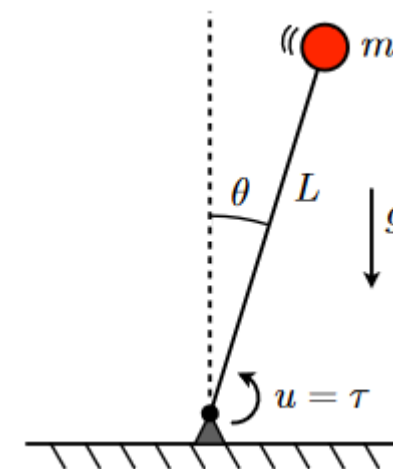
- Let's do the partial differentiation according to our formula.

$$\frac{df}{dx} = \begin{bmatrix} 0 & 1 \\ -\frac{g}{L} \cos(x_1) & 0 \end{bmatrix}, \quad \frac{df}{du} = \begin{bmatrix} 0 \\ 1 \end{bmatrix}.$$

- Now evaluate this system at the **fixed points**. Since we have two fixed points, this will yield two A matrices.

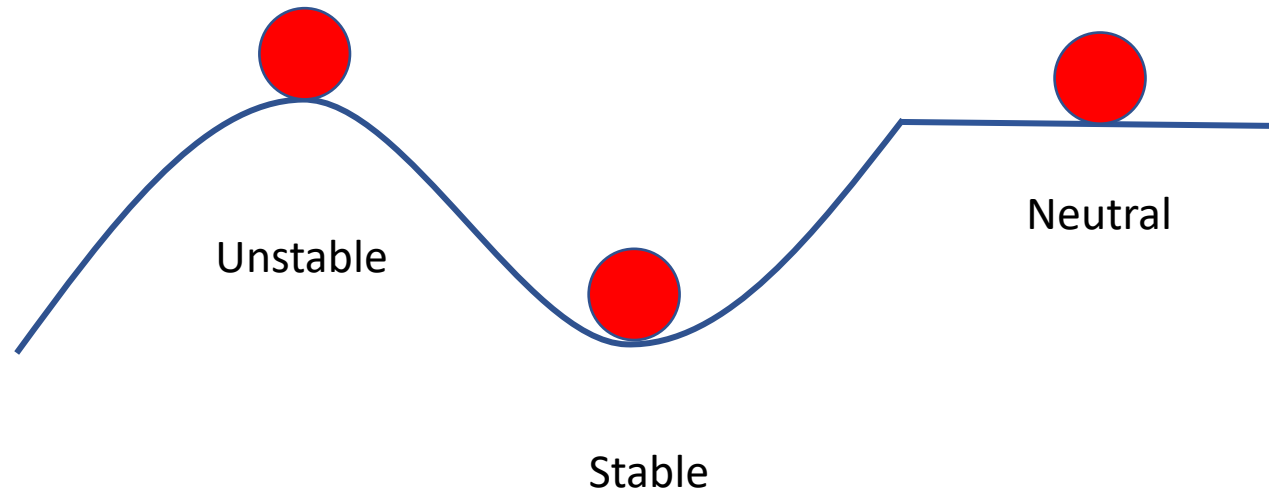
$$A_D = \begin{bmatrix} 0 & 1 \\ -g/L & 0 \end{bmatrix} \Big|_{(\bar{x} = [\Pi, 0])} \quad A_U = \begin{bmatrix} 0 & 1 \\ g/L & 0 \end{bmatrix} \Big|_{(\bar{x} = [0,0])}$$

- Substituting g, L we will get the corresponding linear system near the fixed points.
- Linearization is done!!!!**



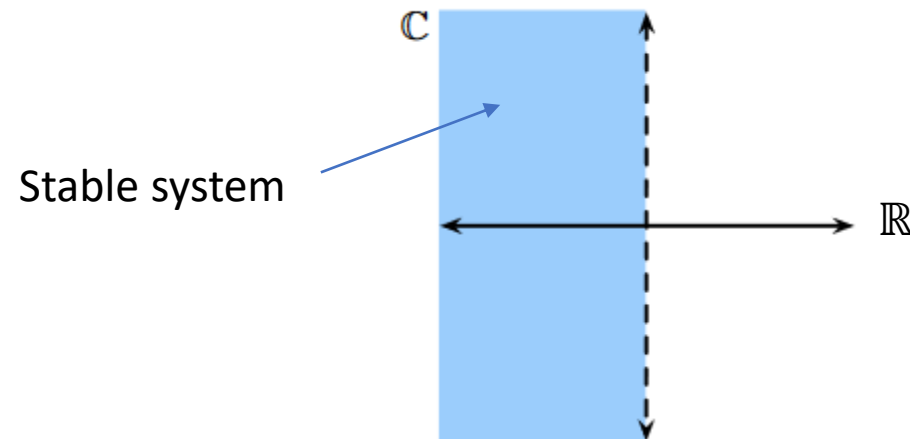
Stability of Linear Systems

- Consider a linear system $\dot{x} = Ax$ with drift ($A \neq 0$) near fixed points.
- Will the system return to its original state after a little perturbation?



Stability of Linear Systems

- We can analyze the A matrix to check if a system is stable at a fixed point.
- For any **continuous time** dynamical system,
 - If the all the **Eigenvalues** of A has **negative real parts** \rightarrow **system is stable** at the fixed point.
 - If any of the **Eigenvalues** has positive real parts \rightarrow unstable at the fixed point.



Stability - Pendulum Example

- Let's consider the linearized A matrices of the pendulum system.
- Add a small **damping** to the system to make sure it has some resistance.

$$A_D = \begin{bmatrix} 0 & 1 \\ -g/L & -0.1 \end{bmatrix} \bigg|_{(\bar{x} = [0,0])}$$

$$A_U = \begin{bmatrix} 0 & 1 \\ g/L & -0.1 \end{bmatrix} \bigg|_{(\bar{x} = [\Pi, 0])}$$

- Set $\frac{g}{L} = 1$ and check for the Eigenvalues.
- The system is unstable near the upper fixed point.

Stability - Pendulum Example

- Let's consider the linearized A matrices of the pendulum system.
- Add a small **damping** to the system to make sure it has some resistance.

$$A_D = \begin{bmatrix} 0 & 1 \\ -g/L & -0.1 \end{bmatrix} \bigg|_{(\bar{x} = [0,0])}$$

$$A_U = \begin{bmatrix} 0 & 1 \\ g/L & -0.1 \end{bmatrix} \bigg|_{(\bar{x} = [\Pi, 0])}$$

- Set $\frac{g}{L} = 1$ and check for the Eigenvalues.
- The system is unstable near the upper fixed point.

How Can We Stabilize It?

- Let's consider the control law $u = K(x)$ again.
- After applying this control, our system becomes,

$$\frac{d}{dt}\mathbf{x} = \mathbf{A}\mathbf{x} + \mathbf{B}u = (\underbrace{\mathbf{A} - \mathbf{B}K}_{\text{New } A \text{ Matrix}})\mathbf{x}.$$

New A Matrix

- Choose $K = [4, 4]$ for the system at the upper fixed point.

$$\dot{\mathbf{x}} = [\mathbf{A} - \mathbf{B}K]\mathbf{x}$$

$$\dot{\mathbf{x}} = \left[\begin{bmatrix} 0 & 1 \\ 1 & 0 \end{bmatrix} - \begin{bmatrix} 0 \\ 1 \end{bmatrix} \begin{bmatrix} 4 & 4 \end{bmatrix} \right] \mathbf{x} = \left[\begin{bmatrix} 0 & 1 \\ -3 & -4 \end{bmatrix} \right] \mathbf{x}$$

How Can We Stabilize It?

- Let's consider the control law $u = K(x)$ again.
- After applying this control, our system becomes,

$$\frac{d}{dt}\mathbf{x} = \mathbf{A}\mathbf{x} + \mathbf{B}u = (\underbrace{\mathbf{A} - \mathbf{B}K}_{\text{New } A \text{ Matrix}})\mathbf{x}.$$

New A Matrix

- Choose $K = [4, 4]$ for the system at the upper fixed point.

$$\dot{\mathbf{x}} = [\mathbf{A} - \mathbf{B}K]\mathbf{x}$$

$$\dot{\mathbf{x}} = \left[\begin{bmatrix} 0 & 1 \\ 1 & 0 \end{bmatrix} - \begin{bmatrix} 0 \\ 1 \end{bmatrix} [4 \quad 4] \right] \mathbf{x} = \left[\begin{bmatrix} 0 & 1 \\ -3 & -4 \end{bmatrix} \right] \mathbf{x}$$

- New Eigenvalues for the matrix becomes $\lambda = -1, \lambda = -3$.
This means **we can stabilize this system** using this control law at **an unstable fixed point!!!**.

How Can We Find K ?

How Can We Find K ?

Linear Quadratic Regulator FTW!

Linear Quadratic Regulator (LQR)

- An optimal control mechanism for **linear dynamical systems** by optimizing a quadratic **cost function**.
- The canonical form for the cost function takes form

$$C = (x - \bar{x})^T Q (x - \bar{x}) + (u - \bar{u})^T R (u - \bar{u})$$

where, $Q \succcurlyeq 0$ and $R \succ 0$ are two weight matrices.

- The weights lets us quantitate the contributions of the state's error and control amount.
- The objective is to find the control input that minimizes this cost and **drives the system to the stability** (or to the desired state).
- The output of the LQR is a gain matrix K that can be used to calculate the optimal control with the control law $u = K(\bar{x} - x)$.

Why Optimal Control?

- To stabilize a system with **minimal control effort**.
- Classical controllers such as PID are much harder to tune when the states are tightly coupled.
- Allows arbitrating between competing goals.
i.e.: convergence time vs control effort.
- Allows finding the optimal gain matrix (control law) accounting for the system model by using **mathematical optimization**.
- Much more **robust** to disturbances.
- *Example: Navigating a drone to a certain location while minimizing the required thrust.*