

# Contents

| Chapter 1 | Analogue Control Systems  | Page 2        |
|-----------|---|---------------|
| 1.1       | Mathematical Modelling Signals and Linear Dynamic Systems   | 2             |
|           | 1.1.1 System Models Using Differential Equations  | 2             |
|           | 1.1.2 Signal Representation in the Frequency Domain and Transfer Functions  | 2             |
| 1.2       | Frequency Response Analysis and Design  | 2             |
|           | <ul><li>1.2.1 System Models Using Differential Equations</li><li>1.2.2 Signal Representation in the Frequency Domain and Transfer Functions</li></ul>             | $\frac{2}{2}$ |
|           | 1.2.3 Bode Diagrams   | $\frac{2}{2}$ |
|           | 1.2.4 Design of Compensation  | $\frac{1}{2}$ |
| 1.3       | Complex Frequency Analysis and Design   | 2             |
|           | 1.3.1 Laplace Transforms and Complex Frequency Concepts   | 2             |
|           | 1.3.2 Signal Representation in the Frequency Domain and Transfer Functions  | 2             |
|           | 1.3.3 Root Locus Design Method  | 2             |
|           |   |               |
| Chapter 2 | Digital Control Systems   | Page 3        |
| 2.1       | Design of a Digital Controller Using Continuous System Theory: CNC Controller Case Stud   |               |
| 2.1       | 2.1.1 CNC System Modelling  | 3             |
|           | 2.1.2 CNC Controller Design for Transients, Disturbance Rejection and Multi-Axis Contouring   | 3             |
|           | 2.1.3 Effects of Sampling   | 3             |
| 2.2       | Discrete System Analysis Using Z-Transforms   | 3             |
|           | 2.2.1 Z-Transforms of Sampled Data Signals, Modified Z-Transforms and Fractional Time Delays  | 3             |
|           | 2.2.2 Discrete Transfer Function  2.2.2 Discrete Transfer Function  Approx. Integration MPZ ZOII)   | 3             |
|           | <ul> <li>2.2.3 Digital Equivalent of a Continuous Transfer Function (Approx. Integration, MPZ, ZOH)</li> <li>2.2.4 Root Locus Design in the 'Z' Domain</li> </ul> | 3<br>3        |
|           | 2.2.5 Jury's Stability Test   | 3             |
|           | 2.2.6 Sampling Theorem  | 3             |
|           |   |               |
| Chapter 2 |   |               |
| Chapter 3 | Introduction to State Variable Analysis   | Page 4        |
| 3.1       | State Variable Analysis of Continuous Systems   | 4             |
|           | 3.1.1 State Variable Modelling in Relation to Block Diagrams  | 4             |
|           | 3.1.2 Eigenvalues, Eigenvectors and Characteristic Equation, Stability of State Variable Models   | 4             |
|           | 3.1.3 Conversion Between Transfer Function and State Variable Models 3.1.4 The State Transition Matrix  | $rac{4}{4}$  |
|           | 3.1.5 Closed Loop Systems   | 4             |
|           | 3.1.6 State Variable Feedback   | 4             |
|           | 3.1.7 Design of a Tracking Controller   | 4             |
|           | 3.1.8 Controllability, Observability  | 4             |
| 3.2       | State Variable Representation of Discrete Systems   | 4             |
|           | 3.2.1 Discrete State Variable Model from the Time Response of the Continuous Model  | 4             |
| 2.2       | 3.2.2 Discrete State Variable Model from Discrete Transfer Function G(z)  | 4             |
| 3.3       | Non-examinable Material   | 4             |
|           | 3.3.1 Kalman Filtering 3.3.2 Optimal Control  | 4             |
|           | 3.3.2 Optimal Collitor  | 4             |

## Chapter 1

# Analogue Control Systems

- 1.1 Mathematical Modelling Signals and Linear Dynamic Systems
- 1.1.1 System Models Using Differential Equations
- 1.1.2 Signal Representation in the Frequency Domain and Transfer Functions
- 1.2 Frequency Response Analysis and Design
- 1.2.1 System Models Using Differential Equations
- 1.2.2 Signal Representation in the Frequency Domain and Transfer Functions
- 1.2.3 Bode Diagrams
- 1.2.4 Design of Compensation
- 1.3 Complex Frequency Analysis and Design
- 1.3.1 Laplace Transforms and Complex Frequency Concepts
- 1.3.2 Signal Representation in the Frequency Domain and Transfer Functions
- 1.3.3 Root Locus Design Method

### Chapter 2

# Digital Control Systems

- 2.1 Design of a Digital Controller Using Continuous System Theory: CNC Controller Case Study
- 2.1.1 CNC System Modelling
- 2.1.2 CNC Controller Design for Transients, Disturbance Rejection and Multi-Axis Contouring
- 2.1.3 Effects of Sampling
- 2.2 Discrete System Analysis Using Z-Transforms
- 2.2.1 Z-Transforms of Sampled Data Signals, Modified Z-Transforms and Fractional Time Delays
- 2.2.2 Discrete Transfer Function
- 2.2.3 Digital Equivalent of a Continuous Transfer Function (Approx. Integration, MPZ, ZOH)
- 2.2.4 Root Locus Design in the 'Z' Domain
- 2.2.5 Jury's Stability Test
- 2.2.6 Sampling Theorem

### Chapter 3

# Introduction to State Variable Analysis

- 3.1 State Variable Analysis of Continuous Systems
- 3.1.1 State Variable Modelling in Relation to Block Diagrams
- 3.1.2 Eigenvalues, Eigenvectors and Characteristic Equation, Stability of State Variable Models
- 3.1.3 Conversion Between Transfer Function and State Variable Models
- 3.1.4 The State Transition Matrix
- 3.1.5 Closed Loop Systems
- 3.1.6 State Variable Feedback
- 3.1.7 Design of a Tracking Controller
- 3.1.8 Controllability, Observability
- 3.2 State Variable Representation of Discrete Systems
- 3.2.1 Discrete State Variable Model from the Time Response of the Continuous Model
- 3.2.2 Discrete State Variable Model from Discrete Transfer Function G(z)
- 3.3 Non-examinable Material
- 3.3.1 Kalman Filtering
- 3.3.2 Optimal Control

### Random Examples

#### Definition 3.3.1: Limit of Sequence in $\mathbb{R}$

Let  $\{s_n\}$  be a sequence in  $\mathbb{R}$ . We say

$$\lim_{n\to\infty} s_n = s$$

where  $s \in \mathbb{R}$  if  $\forall$  real numbers  $\epsilon > 0$   $\exists$  natural number N such that for n > N

$$s - \epsilon < s_n < s + \epsilon$$
 i.e.  $|s - s_n| < \epsilon$ 

#### Question 1

Is the set x-axis\{Origin} a closed set

Solution: We have to take its complement and check whether that set is a open set i.e. if it is a union of open

#### Note:-

We will do topology in Normed Linear Space (Mainly  $\mathbb{R}^n$  and occasionally  $\mathbb{C}^n$ ) using the language of Metric Space

#### Claim 3.3.1 Topology

Topology is cool

#### Example 3.3.1 (Open Set and Close Set)

Open Set:  $\bullet \phi$ 

- $\bullet \bigcup_{x \in X} B_r(x) \text{ (Any } r > 0 \text{ will do)}$
- $B_r(x)$  is open

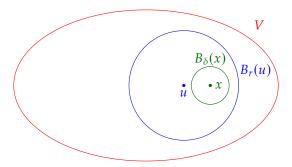
- Closed Set: X,  $\phi$ 
  - $\bullet$   $\overline{B_r(x)}$

x-axis  $\cup y$ -axis

#### Theorem 3.3.1

If  $x \in \text{open set } V \text{ then } \exists \ \delta > 0 \text{ such that } B_{\delta}(x) \subset V$ 

**Proof:** By openness of  $V, x \in B_r(u) \subset V$ 



Given  $x \in B_r(u) \subset V$ , we want  $\delta > 0$  such that  $x \in B_\delta(x) \subset B_r(u) \subset V$ . Let d = d(u, x). Choose  $\delta$  such that  $d + \delta < r$  (e.g.  $\delta < \frac{r-d}{2}$ )

If  $y \in B_{\delta}(x)$  we will be done by showing that d(u, y) < r but

$$d(u, y) \le d(u, x) + d(x, y) < d + \delta < r$$

⊜

#### Corollary 3.3.1

By the result of the proof, we can then show...

#### Lenma 3.3.1

Suppose  $\vec{v_1}, \dots, \vec{v_n} \in \mathbb{R}^n$  is subspace of  $\mathbb{R}^n$ .

#### **Proposition 3.3.1**

1 + 1 = 2.

#### Random

#### Definition 3.3.2: Normed Linear Space and Norm $\|\cdot\|$

Let V be a vector space over  $\mathbb{R}$  (or  $\mathbb{C}$ ). A norm on V is function  $\|\cdot\|$   $V \to \mathbb{R}_{\geq 0}$  satisfying

- $(1) ||x|| = 0 \iff x = 0 \ \forall \ x \in V$
- (2)  $\|\lambda x\| = |\lambda| \|x\| \ \forall \ \lambda \in \mathbb{R}(\text{or } \mathbb{C}), \ x \in V$
- (3)  $||x + y|| \le ||x|| + ||y|| \ \forall \ x, y \in V$  (Triangle Inequality/Subadditivity)

And V is called a normed linear space.

• Same definition works with V a vector space over  $\mathbb{C}$  (again  $\|\cdot\| \to \mathbb{R}_{\geq 0}$ ) where ② becomes  $\|\lambda x\| = |\lambda| \|x\|$  $\forall \lambda \in \mathbb{C}, x \in V$ , where for  $\lambda = a + ib$ ,  $|\lambda| = \sqrt{a^2 + b^2}$ 

#### **Example 3.3.2** (*p*-Norm)

 $V = \mathbb{R}^m, p \in \mathbb{R}_{\geq 0}$ . Define for  $x = (x_1, x_2, \dots, x_m) \in \mathbb{R}^m$ 

$$||x||_p = (|x_1|^p + |x_2|^p + \dots + |x_m|^p)^{\frac{1}{p}}$$

(In school p = 2)

Special Case p = 1:  $||x||_1 = |x_1| + |x_2| + \cdots + |x_m|$  is clearly a norm by usual triangle inequality. Special Case  $p \to \infty$  ( $\mathbb{R}^m$  with  $||\cdot||_{\infty}$ ):  $||x||_{\infty} = \max\{|x_1|, |x_2|, \cdots, |x_m|\}$ 

For m = 1 these p-norms are nothing but |x|. Now exercise

#### Question 2

Prove that triangle inequality is true if  $p \ge 1$  for p-norms. (What goes wrong for p < 1?)

Solution: For Property (3) for norm-2

#### When field is $\mathbb{R}$ :

We have to show

$$\sum_{i} (x_i + y_i)^2 \le \left( \sqrt{\sum_{i} x_i^2} + \sqrt{\sum_{i} y_i^2} \right)^2$$

$$\implies \sum_{i} (x_i^2 + 2x_i y_i + y_i^2) \le \sum_{i} x_i^2 + 2\sqrt{\left[\sum_{i} x_i^2\right] \left[\sum_{i} y_i^2\right]} + \sum_{i} y_i^2$$

$$\implies \left[ \sum_{i} x_i y_i \right]^2 \le \left[ \sum_{i} x_i^2 \right] \left[ \sum_{i} y_i^2 \right]$$

So in other words prove  $\langle x,y\rangle^2 \leq \langle x,x\rangle \langle y,y\rangle$  where

$$\langle x, y \rangle = \sum_{i} x_i y_i$$

#### Note:-

- $\bullet \ \, ||x||^2 = \langle x, x \rangle$
- $\bullet \ \langle x, y \rangle = \langle y, x \rangle$
- $\langle \cdot, \cdot \rangle$  is  $\mathbb{R}$ -linear in each slot i.e.

 $\langle rx + x', y \rangle = r \langle x, y \rangle + \langle x', y \rangle$  and similarly for second slot

Here in  $\langle x, y \rangle$  x is in first slot and y is in second slot.

Now the statement is just the Cauchy-Schwartz Inequality. For proof

$$\langle x, y \rangle^2 \leq \langle x, x \rangle \langle y, y \rangle$$

expand everything of  $\langle x - \lambda y, x - \lambda y \rangle$  which is going to give a quadratic equation in variable  $\lambda$ 

$$\langle x - \lambda y, x - \lambda y \rangle = \langle x, x - \lambda y \rangle - \lambda \langle y, x - \lambda y \rangle$$

$$= \langle x, x \rangle - \lambda \langle x, y \rangle - \lambda \langle y, x \rangle + \lambda^2 \langle y, y \rangle$$

$$= \langle x, x \rangle - 2\lambda \langle x, y \rangle + \lambda^2 \langle y, y \rangle$$

Now unless  $x = \lambda y$  we have  $\langle x - \lambda y, x - \lambda y \rangle > 0$  Hence the quadratic equation has no root therefore the discriminant is greater than zero.

#### When field is $\mathbb{C}$ :

Modify the definition by

$$\langle x, y \rangle = \sum_{i} \overline{x_i} y_i$$

Then we still have  $\langle x, x \rangle \ge 0$ 

### Algorithms

```
Algorithm 1: what
   Input: This is some input
   Output: This is some output
   /* This is a comment */
 1 some code here;
 \mathbf{z} \ x \leftarrow 0;
\mathbf{3} \ \mathbf{y} \leftarrow 0;
4 if x > 5 then
 5 x is greater than 5;
                                                                                          // This is also a comment
 6 else
 7 x is less than or equal to 5;
8 end
9 foreach y in 0..5 do
10 y \leftarrow y + 1;
11 end
12 for y in 0..5 do
13 y \leftarrow y - 1;
14 end
15 while x > 5 do
16 x \leftarrow x - 1;
17 end
18 return Return something here;
```