

## Assignment : 02.

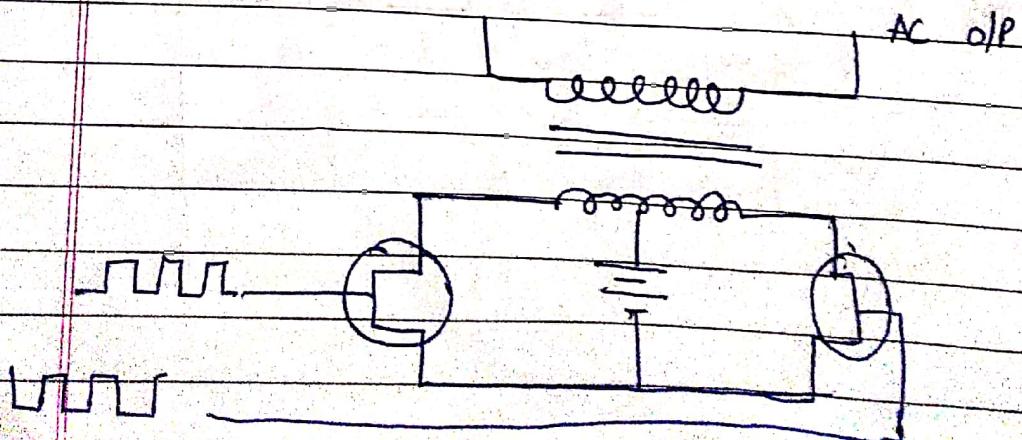
Ques: 01 What is an inverter? List few industrial applications of it.

Ans: An inverter is an electrical device, if it is capable of changing a DC current to an AC current at a given frequency as well as voltage. The AC power can be supplied to homes & industries using the public utility otherwise power grid, the alternating power systems of the batteries can store only DC power.

In some situations like when the DC voltage is low, then we cannot use the low DC voltage in some appliance. So due to this reason, an inverter can be used whenever we utilize solar power panel.

### Types of Inverter:

- Single phase Inverter
- Three phase Inverter



## # Applications:

- used as on UPS (uninterruptible power supplies)
- used as standalone inverters.
- used in Solar power systems
- used in Centrifugal fans, pumps, mixers, extruders, test stands - and web handling equipment.

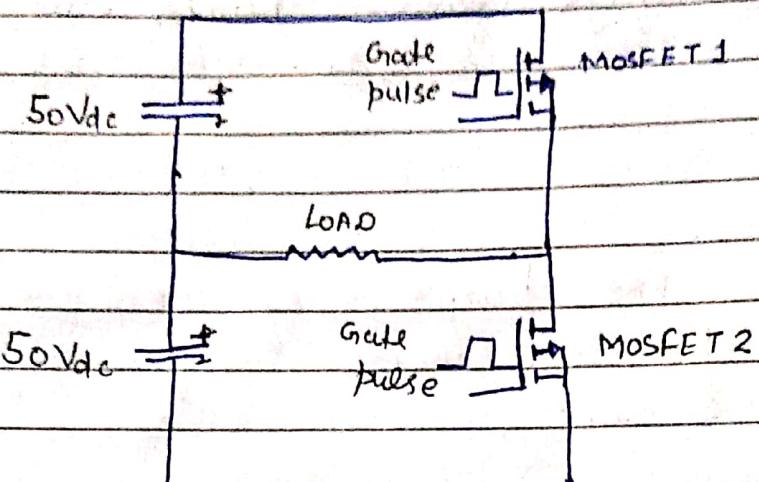
Ques: what are the purpose of connecting diodes in antiparallel with thyristors in inverter circuits?

Ans: when a diode got connected with thyristor or SCR then it became unipolar switch. whenever a diode placed antiparallel to any switching device it became unipolar so when diode connected its voltage drop on forward biased will become reverse voltage drop for SCR during its off-time & voltage drop for power diode is between 0.8 - 1 volt so reverse power loss become less for SCR as initially we may considering some bulky commutation circuit with voltage drop more than 1 volt so power loss more in other then antiparallel diode.

by using antiparallel diode power loss decreases & turn-off time increases.

Ques: Q3 Describe the working of a single phase half bridge inverter, what is its main drawback? Explain how this drawback is overcome.

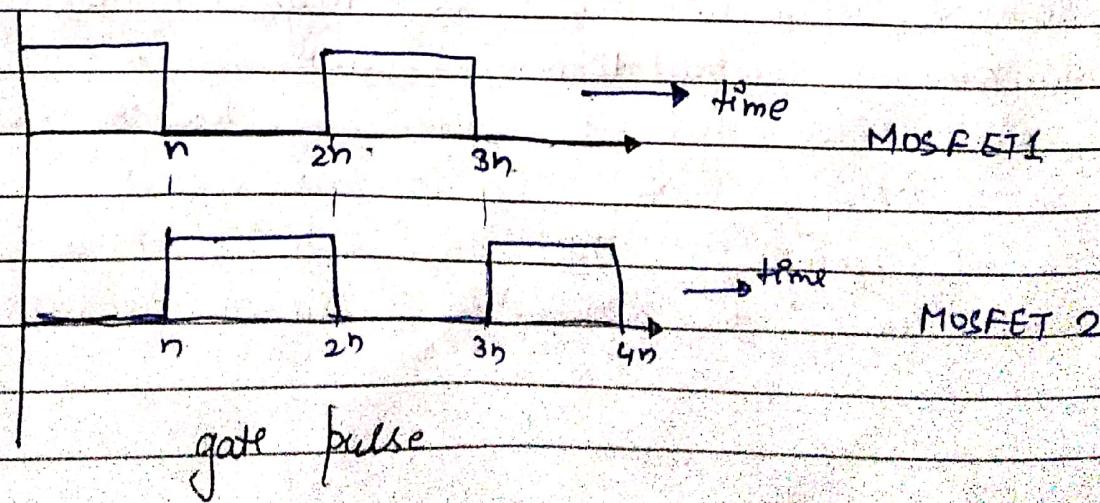
Ans:



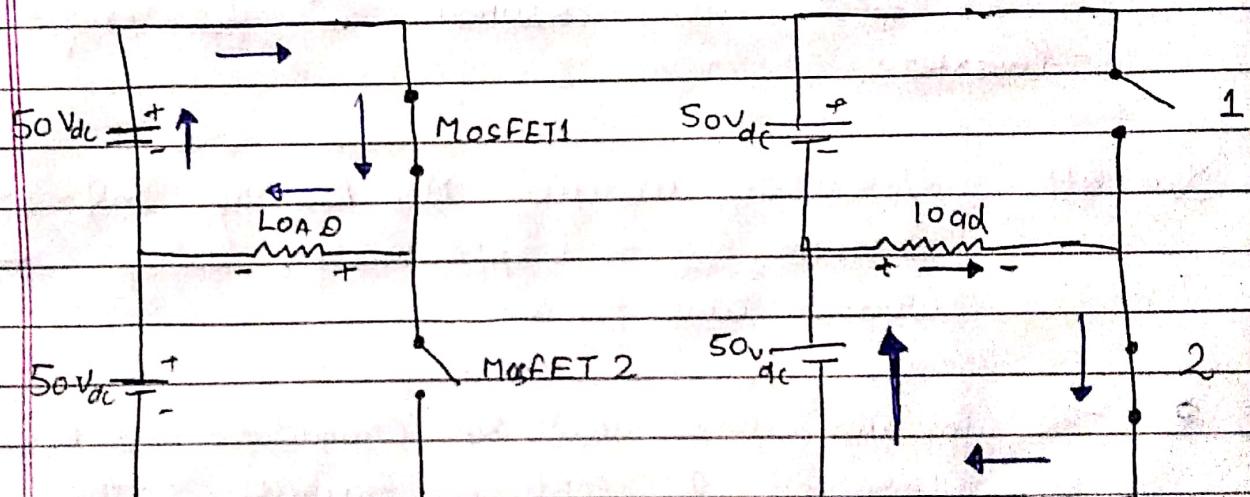
Half Bridge inverter

This type of Inverter requires two power electronics switches (MOSFET). The MOSFET or IGBT is used for switching purpose.

As shown in circuit diagram,  $V_{DC} = 100V$ . This source is divided into two equal parts.



"star" to o/p freq., ON time and OFF time of MOSFET is decided & gate pulse are generated. we need 50Hz AC power, so the time period of one cycle ( $0 < t < 2\pi$ ) is 20 msec, from fig., MOSFET-1 is triggered for first half cycle ( $0 < t < \pi$ ) and during this time period, current will flow in dir<sup>n</sup> of arrow as shown in below figure & half cycle of AC o/p is completed. The current from the load is right to left & load voltage is equal to  $+V_{dc}/2$ .



In second half cycle ( $\pi < t < 2\pi$ ), the MOSFET-2 is triggered and lower voltage source is connected with the load. The current from the load is left to right dir<sup>n</sup> & load voltage is equal to  $-V_{dc}/2$ .

## Ques. 03:

the main drawback is source utilization factor when compared to full bridge inverters.

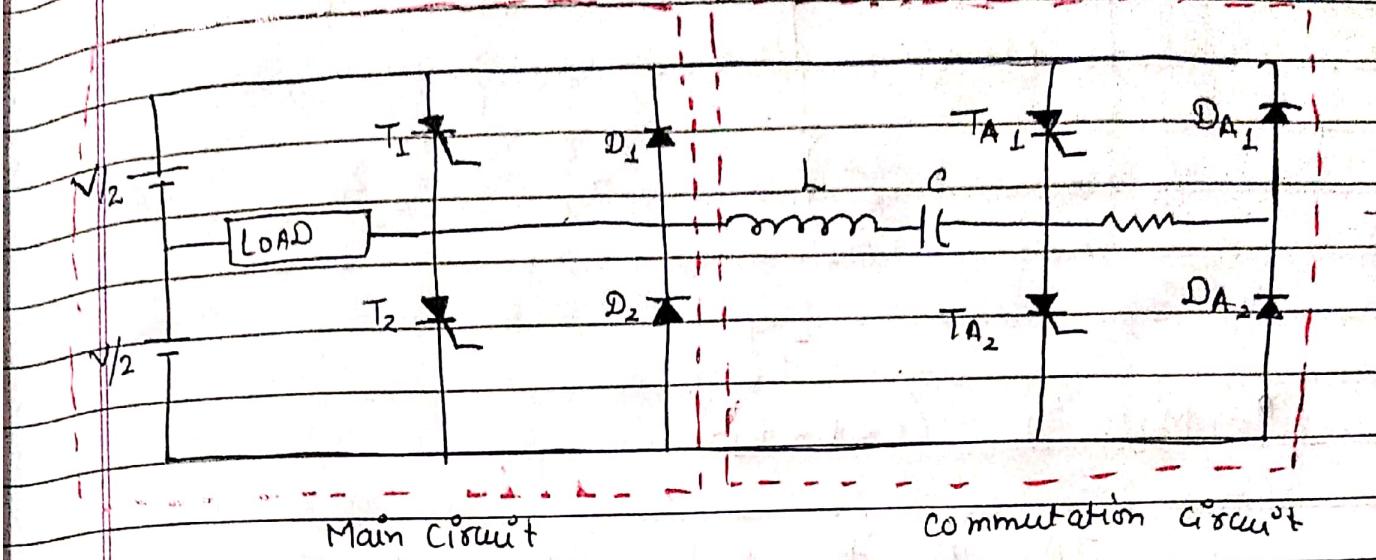
It don't utilise the total source voltage at a time.

In every half cycle, voltage across the load is only half of source voltage i.e.,  $V_s/2$  whereas in full bridge inverter in every half cycle total voltage  $V_s$  appears across the load. So the power consumed by load is half of what is consumed in full bridge inverter.

Ques. 04 Describe modified Mc Murray half-bridge inverter with appropriate voltage & current wave forms.

→ The inverter uses auxiliary commutation scheme to turn off a conducting thyristor. The main inverter circuit is similar to the half-bridge circuit, except that it uses Thyristors  $T_1$  &  $T_2$  in place of commutated switches  $S_1$  &  $S_2$ . The commutation circuit consists of ~~one~~ auxiliary thyristors  $TA_1$  &  $TA_2$  along with anti-parallel diodes  $DA_1$  &  $DA_2$ , Commutating elements L & C, and clamping resistance R. To transfer current from  $T_1$  to  $T_2$ ,  $TA_1$  is triggered and to transfer current from

$T_2$  to  $T_1$   $T_{A21}$  is triggered.



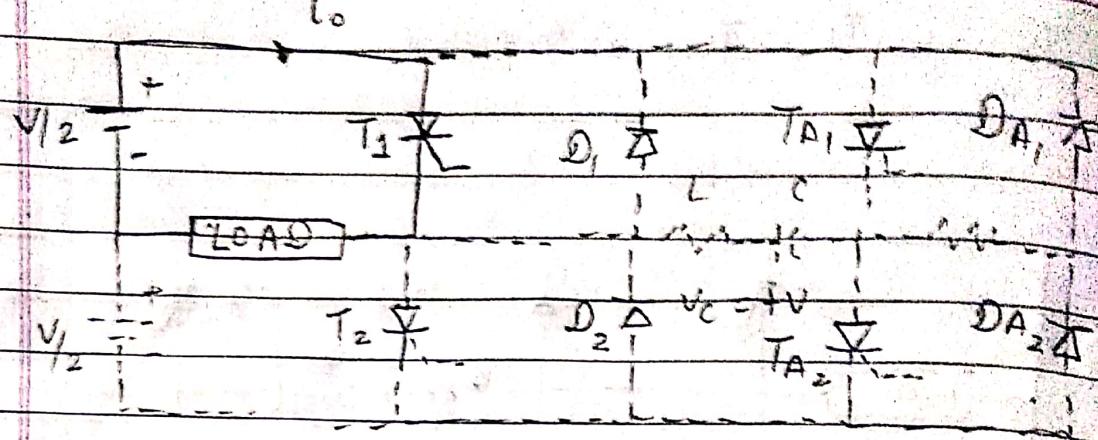
- Circuit of modified Mc Murray half-bridge Inverter

For the sake of simplification, it is assumed that all the devices and circuit elements are ideal. Moreover, it is assumed that the load is highly inductive so that the load current remains constant during commutation interval. The operation is divided into 8-modes.

Thick lines show conducting part & non-conducting parts are shown by dotted lines.

- Mode 1 ( $t < 0$ )

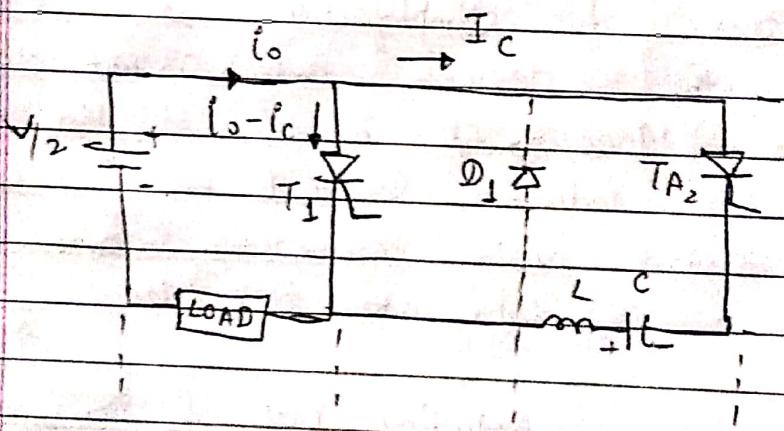
In this mode main thyristor  $T_1$  conducts and a constant positive load current ( $I_0$ ) flows. The load voltage is +ve & the capacitor is charged up to  $+V$ .



• Mode II ( $t_1 < t < t_2$ )

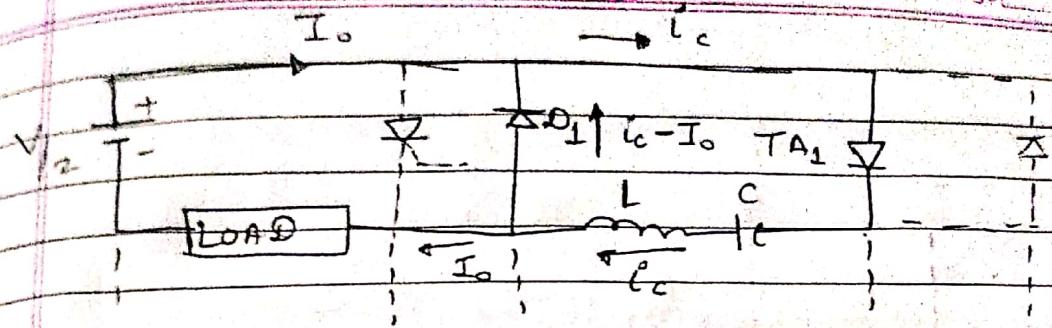
$TA_1$  is triggered at  $t = 0$  to Commutate  $T_1$

$$i_{T_1} = I_o - i_c$$



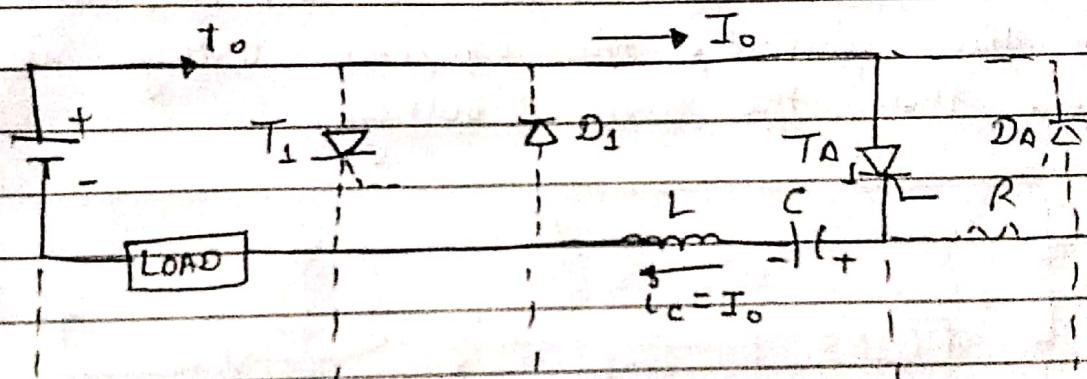
• Mode III ( $t_2 < t < t_3$ )

As the resonating current  $i_c$  becomes higher than the load current ( $I_o$ ), the surplus current ( $i_c - I_o$ ) flows through the diode  $D_2$ .



#### Mode - IV ( $t_2 < t < t_3$ )

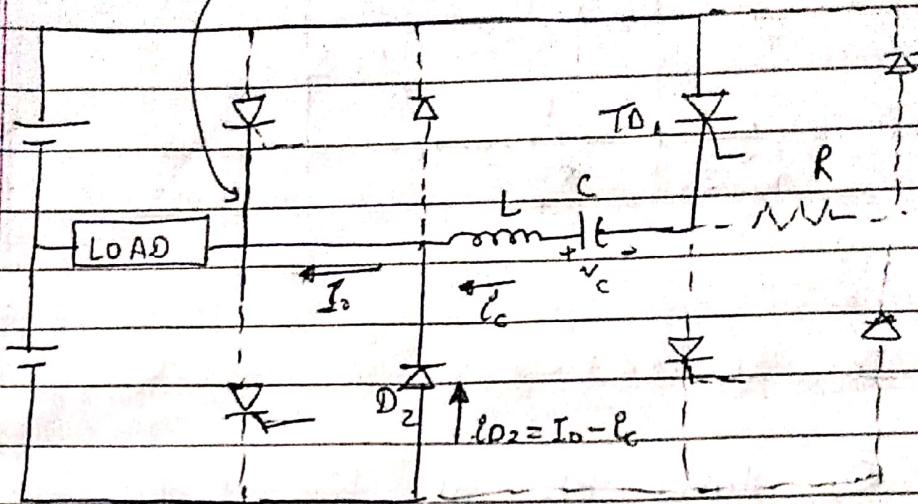
when  $D_1$  turns off at the end of mode - III, the capacitor is charged in reverse direction by the const. load current  $I_o$  during this interval, the voltage across diode  $D_2$ ,  $v_{D_2}$  is equal to  $(v_c - V)$ .



#### Mode V ( $t_3 < t < t_4$ )

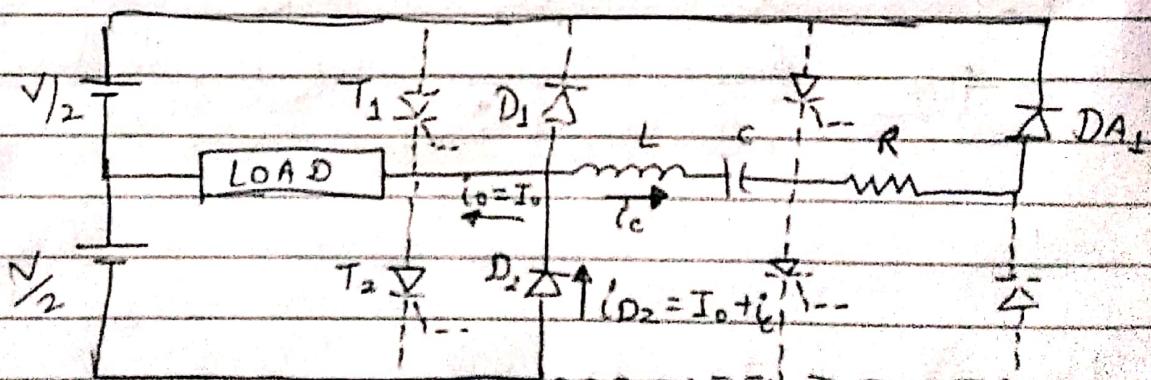
when  $D_2$  turns on,  $I_o$  is shared by  $D_2$  &  $C$  as  $i_c = i_{D_2}$ ,  $i_{D_2} + i_o$  which is equal to  $I_o - i_c$

dotted



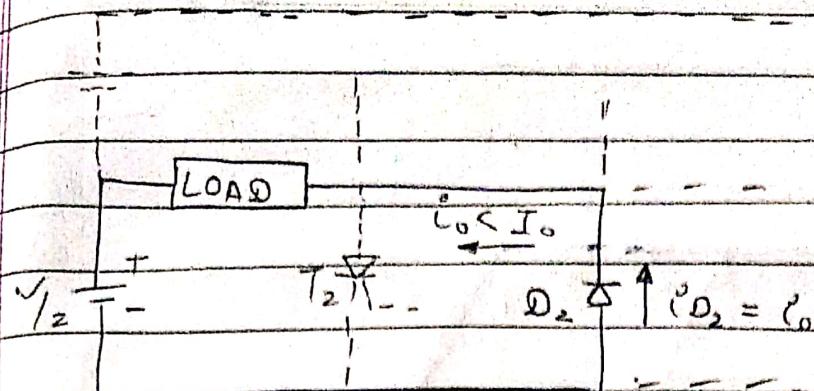
### • Mode: VI ( $t_4 < t < t_5$ )

as  $i_c$  tries to reverse at  $t_4$ ,  $D_{A1}$  turns on. At this instant, the capacitors voltage is more than the source voltage.



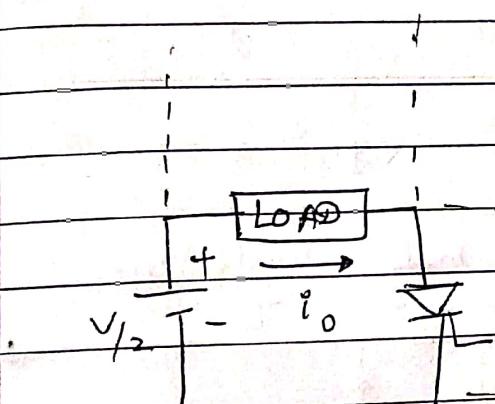
### Mode VII ( $t_5 < t < t_6$ )

During this interval,  $i_o$  flows through  $D_2$  and the lower source.

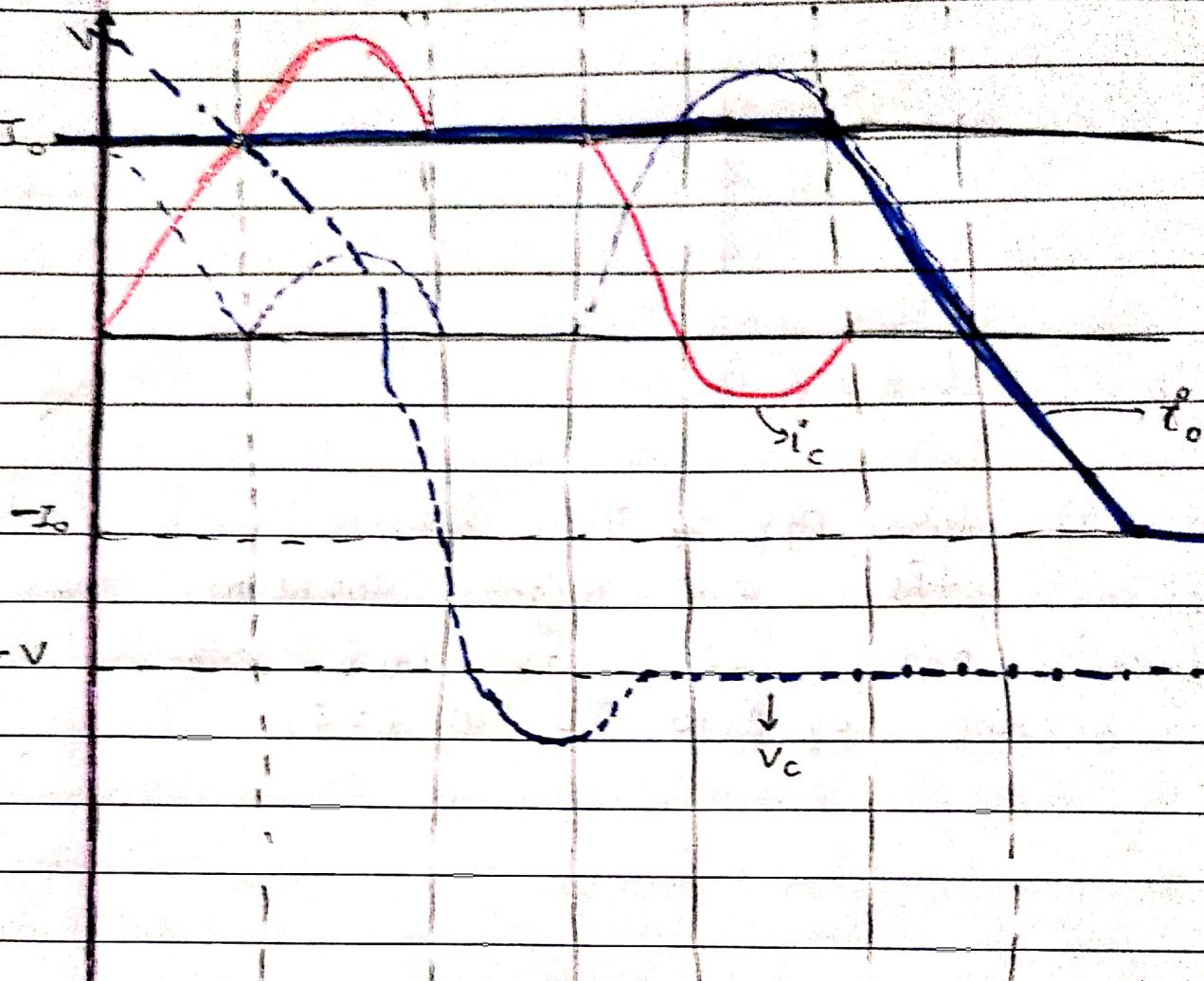


### • Mode VII ( $t > t_6$ )

When D<sub>2</sub> turns off, T<sub>2</sub> which is already receiving the gate signal turns on. It conducts and the -ve load current, raises to become equal to I<sub>o</sub> at  $t = t_7$



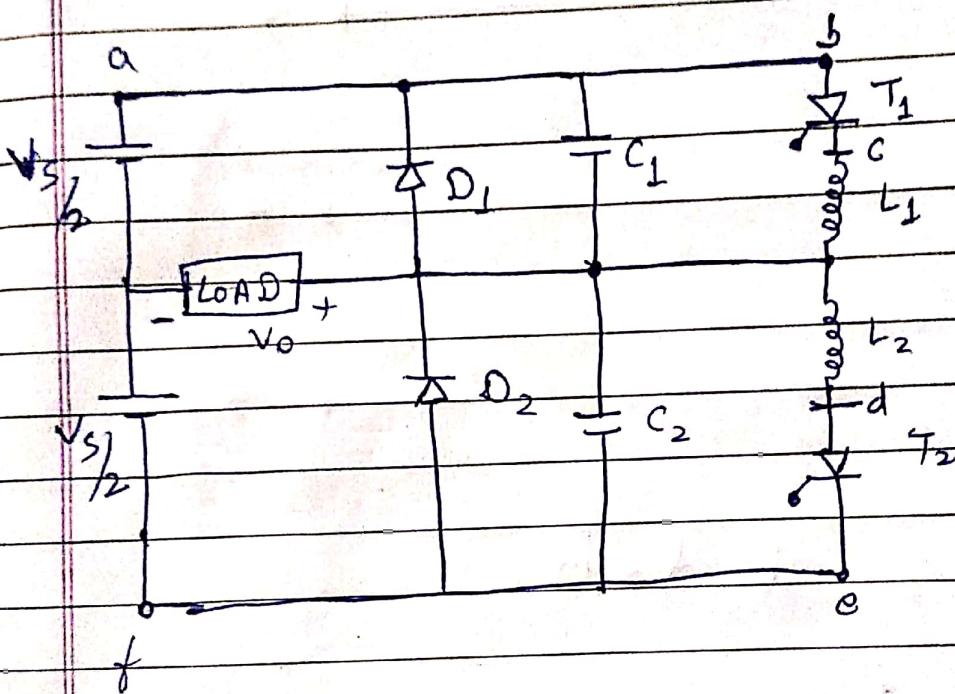
## Waveform



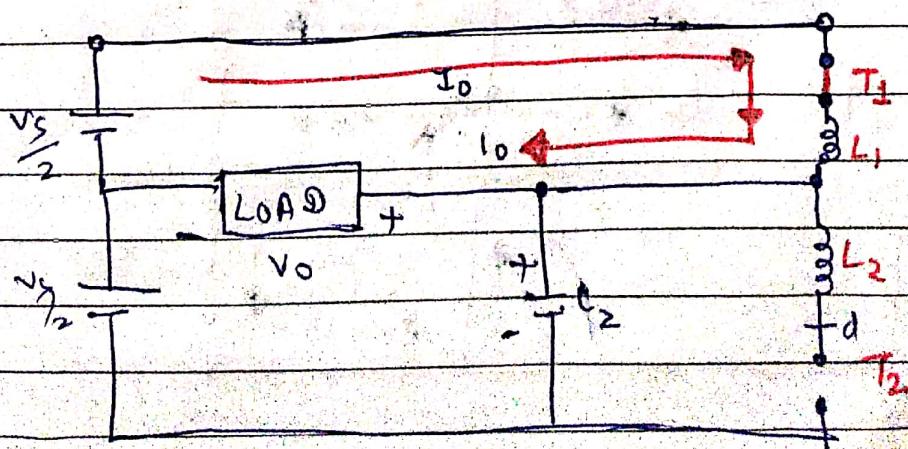
mode I	mode II	mode III	mode IV	mode V	mode VI	mode VII	mode VIII
Cond. $I_1$	$T_1, TA_1$	$D_1, TA_1$	$TA_1$	$TA_2, D_2$	$DA_1, D_2$	$D_2$	$T_2$
devices	$D_1$ , ON						
	$T_1$ off	$D_1$ off	$D_2$ on	$TA_1$ off	$DA_1$ off	$D_2$ off	$T_2$ on

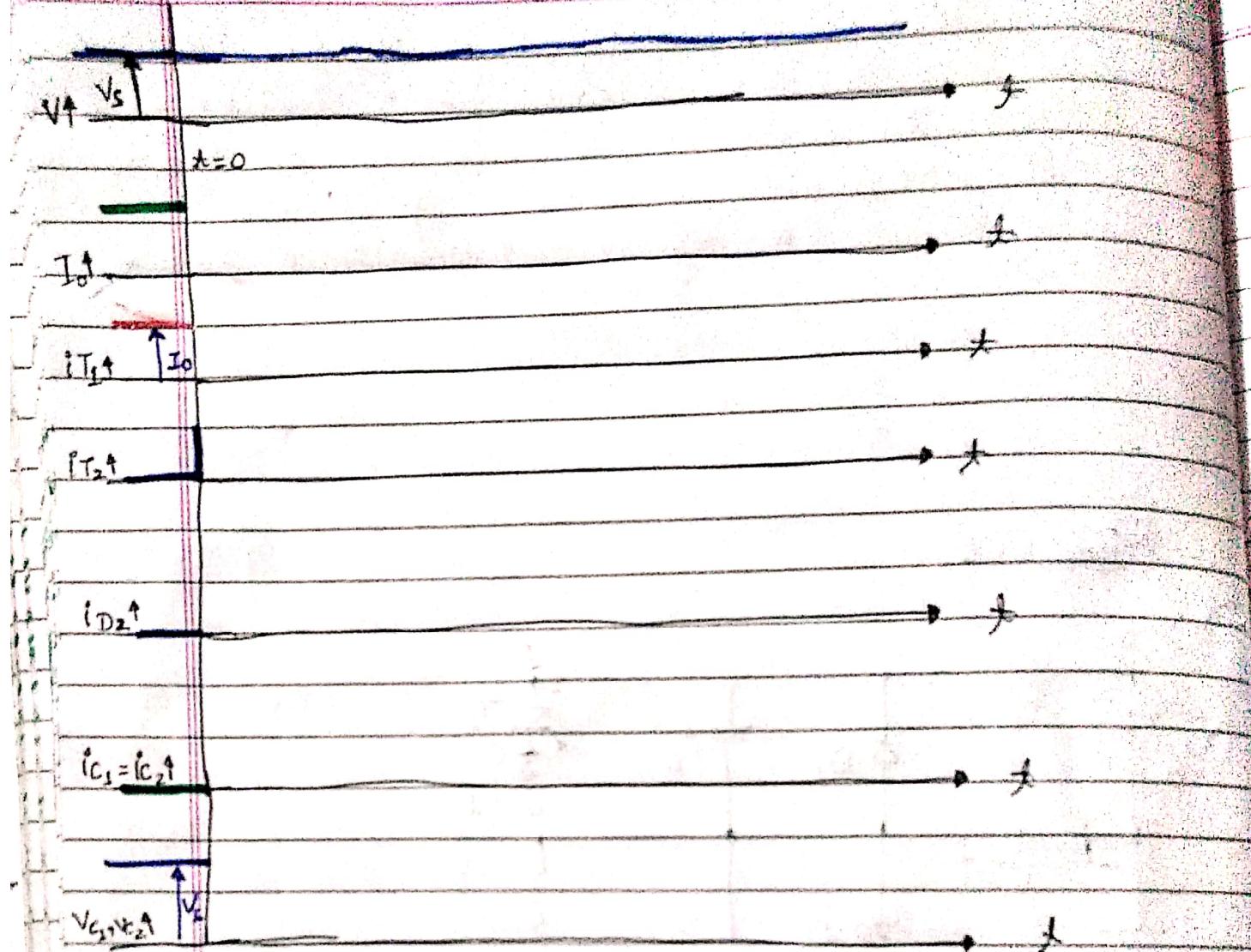
Ques 05 Describe the modified Mc Murray Bedford half bridge single phase inverter with relevant voltage & current waveforms. Also, explain advantages of it over modified Mc Murray half bridge inverter.

Diagram

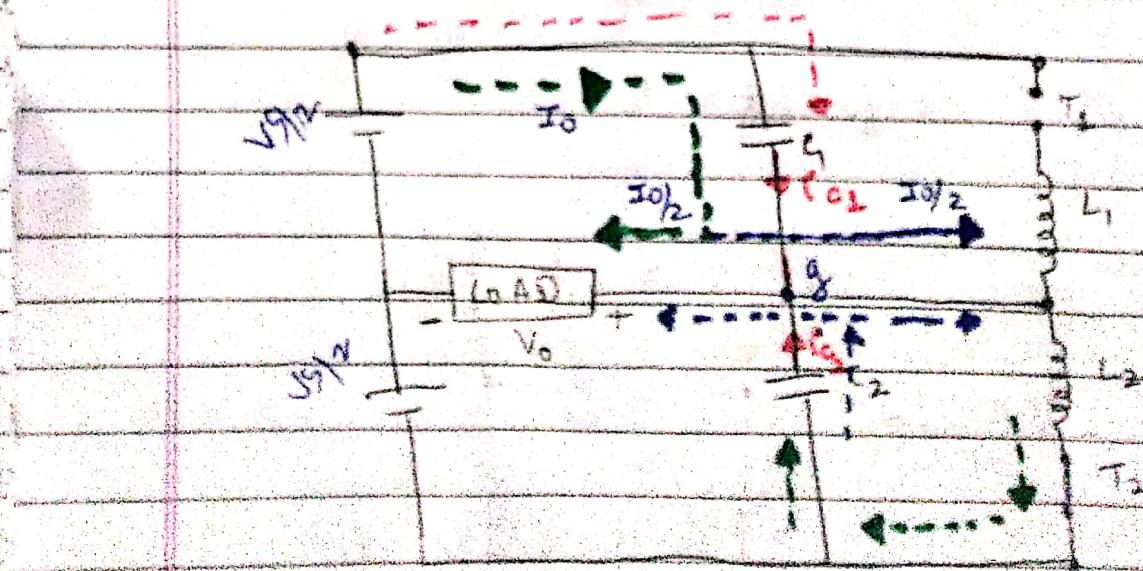


Mode 1 : before  $t=0$





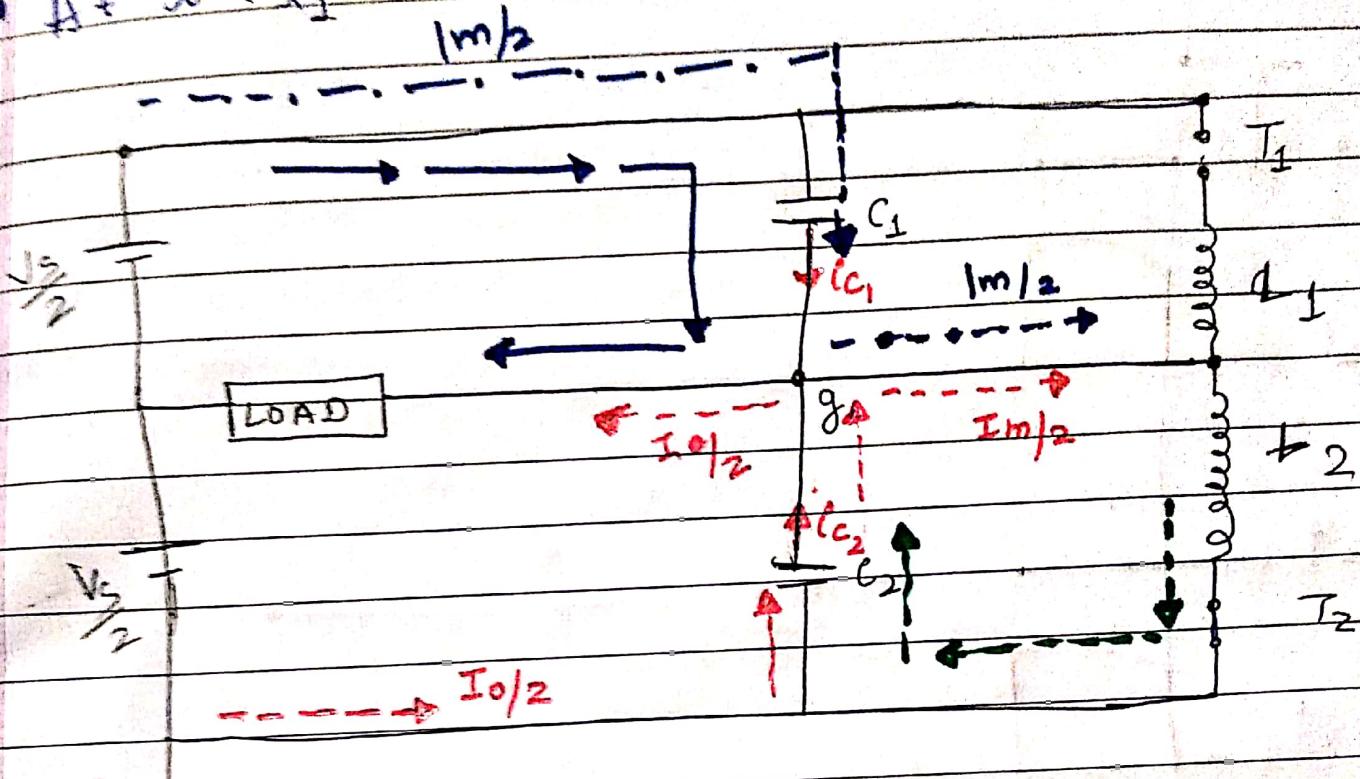
Mode : 02 at  $\lambda = 0$



KVL at node g

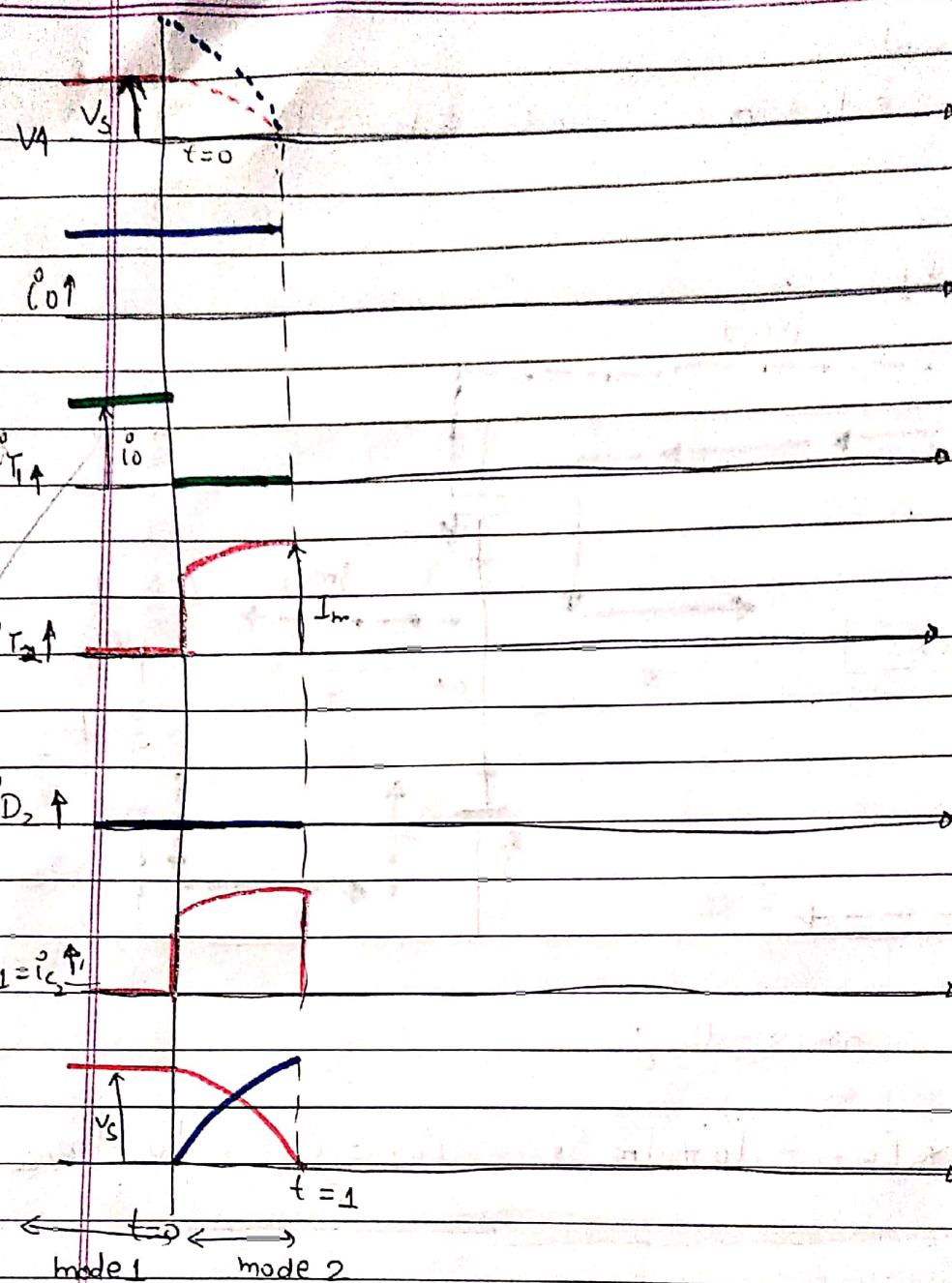
$$i_{C_1} + i_{C_2} = I_0 + I_m \quad , \quad i^{\circ}C_1 = i^{\circ}C_2 = I_0 + \cancel{I_m}$$

$\rightarrow A + \frac{1}{f} < f_1$

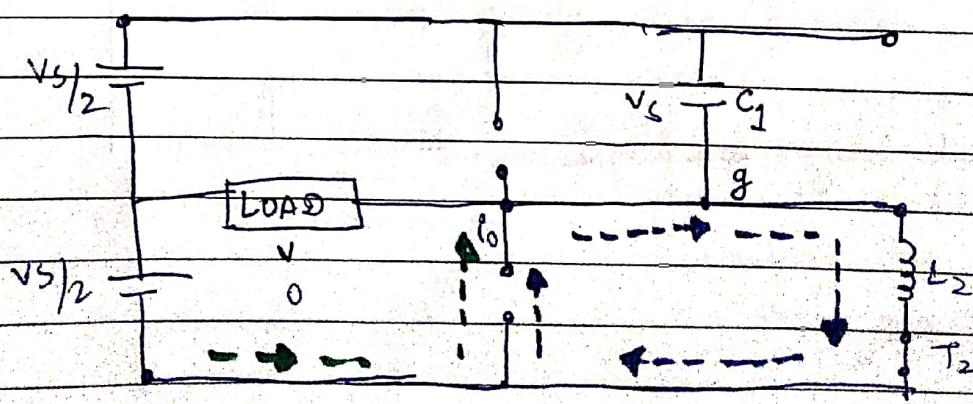


KCL at node g

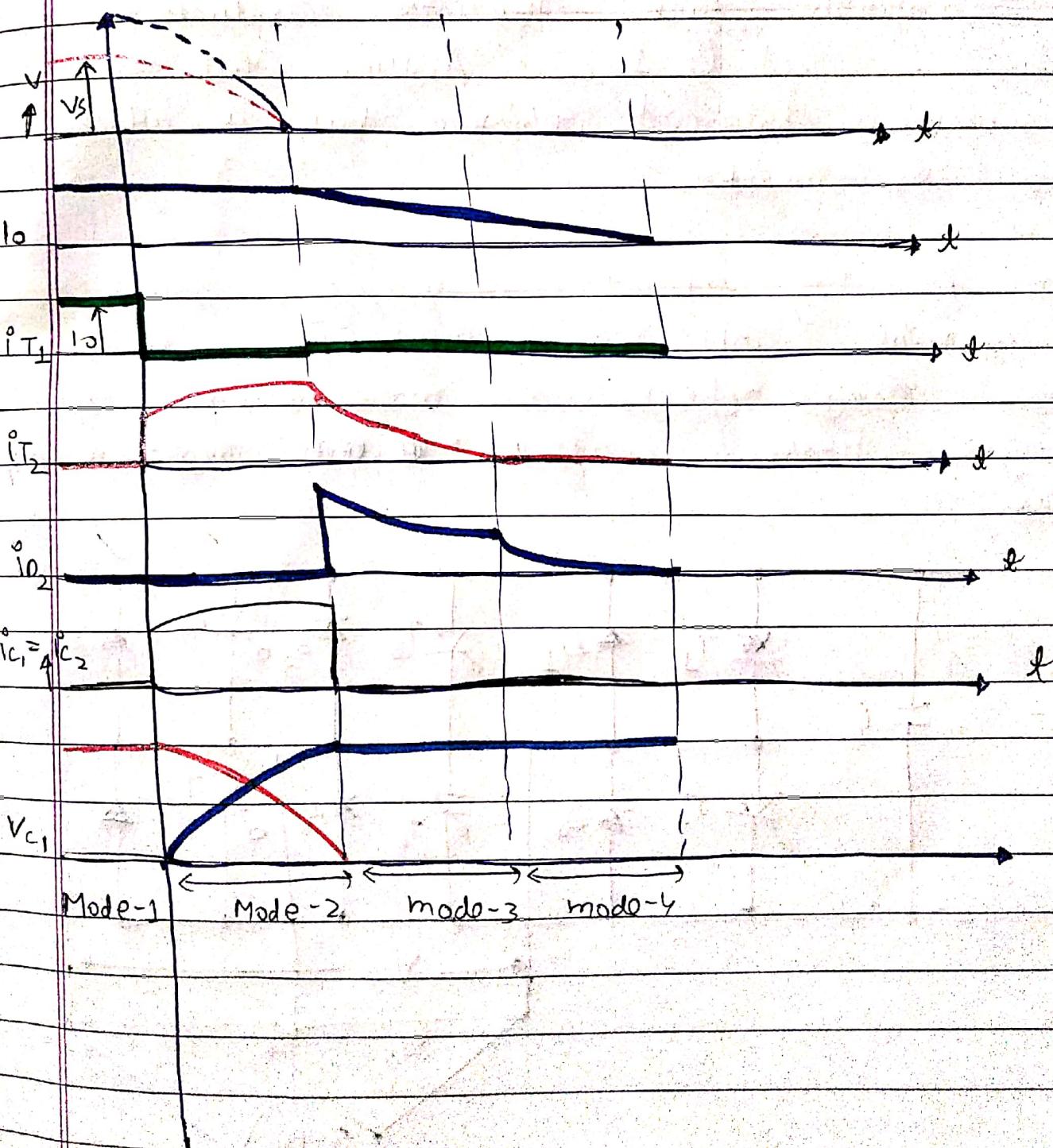
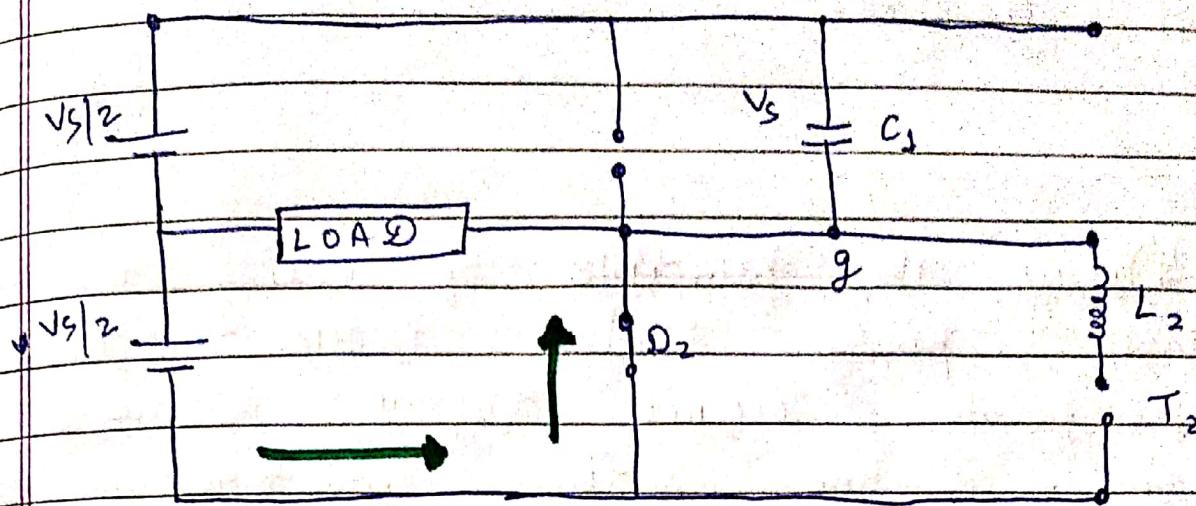
$$i^{\circ}C_1 + i^{\circ}C_2 = I_0 + I_m \quad , \quad i^{\circ}C_1 = i^{\circ}C_2 = I_0 + I_m/2$$



• Mode 3 at  $t_1$



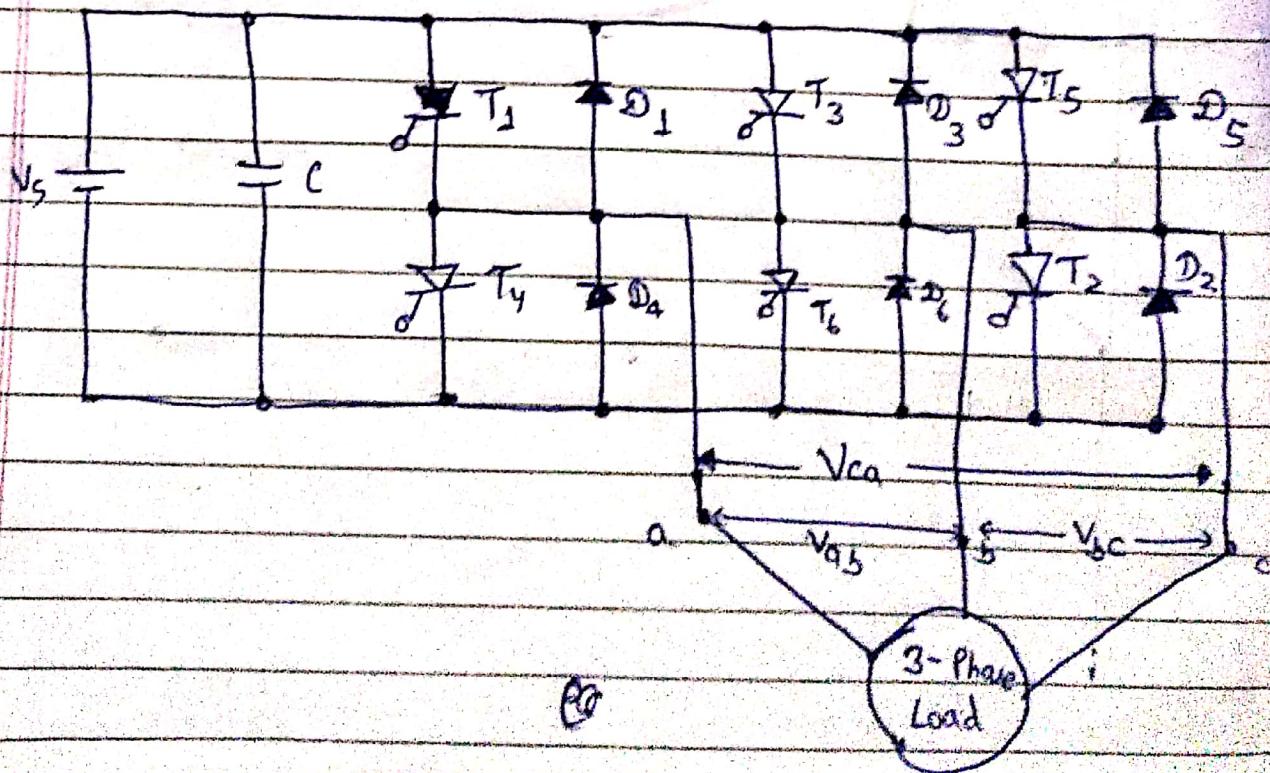
• Mode 3: 04



- It uses less number of Thyristors & diodes compared to modified three phase half bridge inverter.

Ques 06 Discuss the principle of working of three phase bridge inverter with appropriate circuit diagram. Draw phase & line voltage waveforms on the assumption each thyristor conducts for (a)  $180^\circ$  and (b)  $120^\circ$ . Assume that star connected balanced resistive load is attached to the inverter.

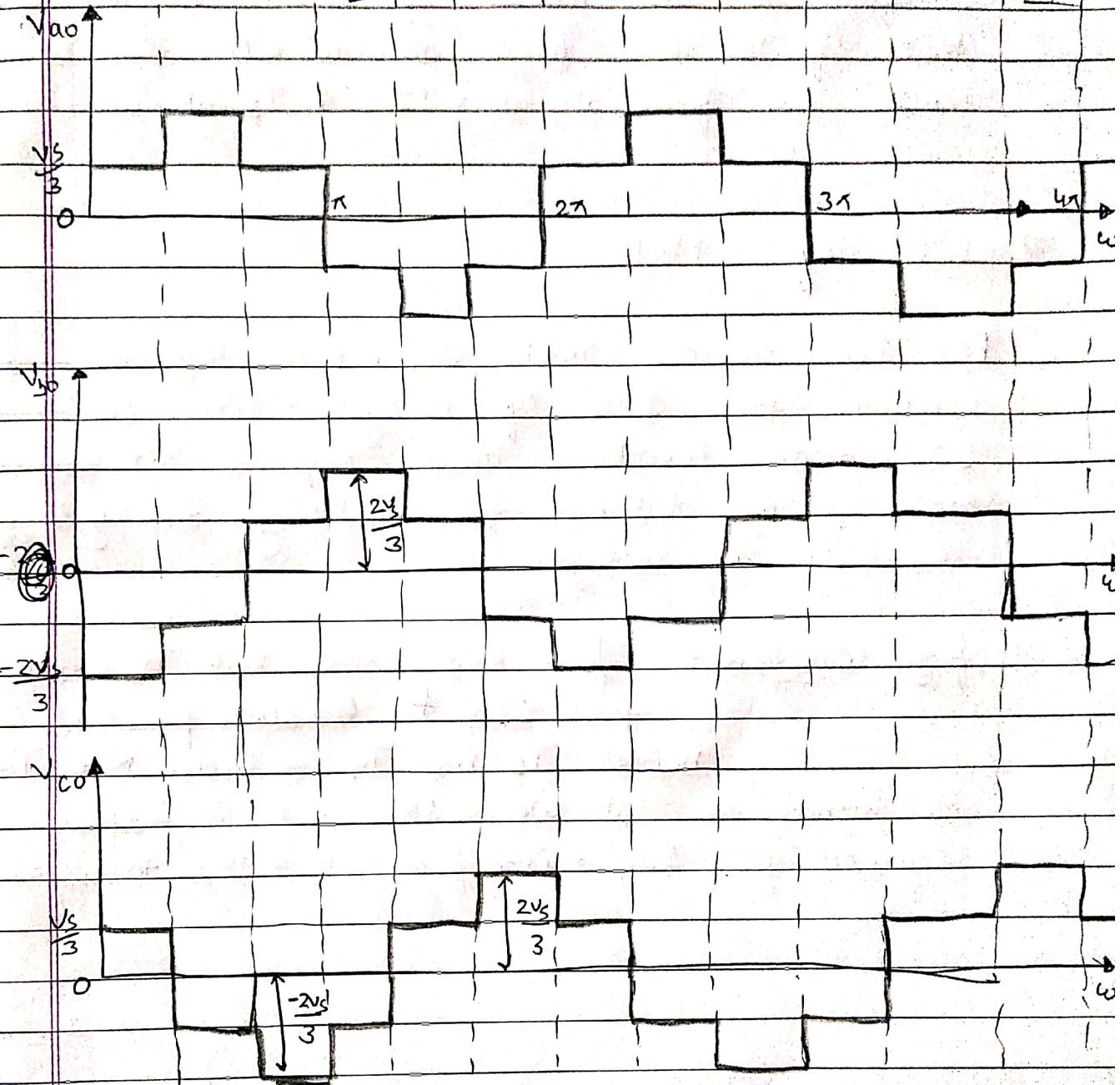
This for providing adjustable - frequency power to industrial applications, three phase inverters are more common than single phase inverters. ~~more efficient~~



$180^\circ$  degree

$0^\circ$	$60^\circ$	$120^\circ$	$180^\circ$	$240^\circ$	$300^\circ$	$360^\circ$	$60^\circ$	$120^\circ$	$180^\circ$	$240^\circ$	$300^\circ$	$360^\circ$
5,6,1,6,1,2	1,2,3,1,2,3,4	3,4,5,4,5,6	15,6,1	6,1,2	1,2,3	12,3,4	3,4,5	4,5,6				

I II III IV V VI



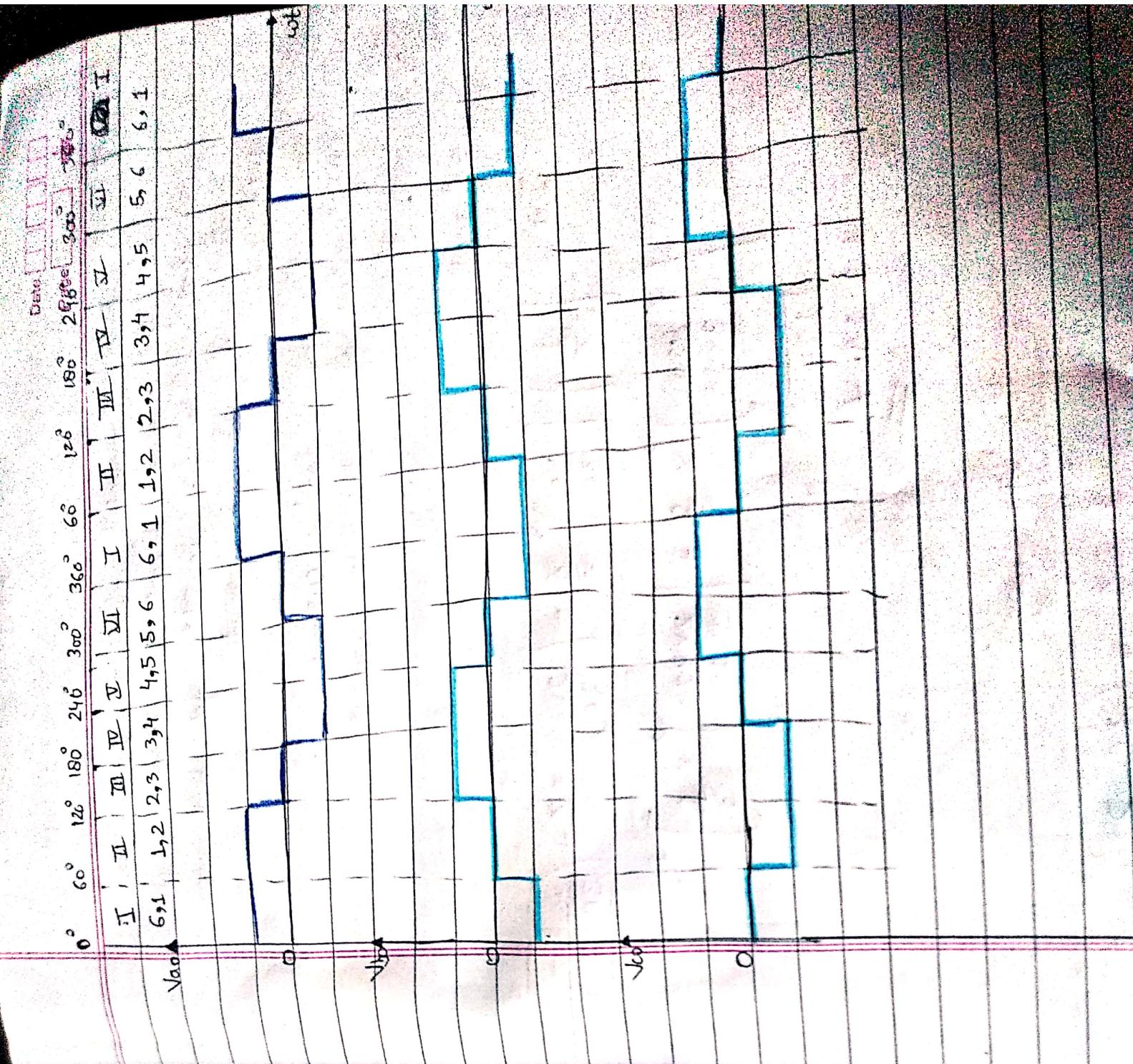
In three phase inverters, each SCR conducts for  $180^\circ$  of a cycle. Thyristor pairs in each arm i.e.,  $T_1, T_4$ ;  $T_3, T_6$  &  $T_5, T_2$  are turned on with a time interval of  $180^\circ$ .

from waveform:  $T_1$  conducts for  $180^\circ$  of  $\omega t$ , for next  $120^\circ$  of a cycle,  
thyristor in the upper group i.e.  $T_2, T_3, T_4$  conducts at an interval of  $120^\circ$ . It implies  
that if  $T_1$  is fired at  $\omega t = 0^\circ$  then  $T_2$   
must be fired at  $\omega t = 120^\circ$  &  $T_3$  at  $\omega t = 240^\circ$ .

$\Rightarrow 120^\circ$  degree Mode

For  $120^\circ$  degree mode VSI, each thyristor  
conducts for  $120^\circ$  of a cycle. Like  $180^\circ$  mode,  
 $120^\circ$  mode inverter also requires six steps,  
each of  $60^\circ$  duration for completing one cycle of  
the o/p ac voltage.

from waveform: first row shows that  $T_1$  conducts  
for  $120^\circ$  & ~~conducts for next 60°~~  
neither  $T_2$  nor  $T_4$  conducts. Now  $T_4$   
is turned on at  $\omega t = 180^\circ$  and it further  
conducts for  $120^\circ$ , i.e. from  $\omega t = 180^\circ$  to  $\omega t = 300^\circ$



$$87.12 \times \frac{2}{3} + 340.4 \times \frac{1}{3}$$

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Ques 07 A star connected load of  $15\Omega$  per phase is fed from  $420V$  dc source through 3-phase bridge inverter for both (a)  $180^\circ$  mode & (b)  $120^\circ$  mode.

determine:

- (i) rms value of load current
- (ii) rms value of thyristor current
- (iii) load power

(a)  $180^\circ$  mode.

$$I_{\text{load}} = \left[ \frac{1}{\pi} \left\{ \left( \frac{v_s}{3R} \right)^2 \frac{\pi}{3} + \left( \frac{2v_s}{3R} \right)^2 \times \frac{\pi}{3} + \left( \frac{v_s}{3R} \right)^2 \frac{\pi}{3} \right\} \right]^{1/2}$$

$$= \left[ \left( \frac{420}{3 \times 15} \right)^2 \times \frac{2}{3} + \left( \frac{2 \times 420}{3 \times 15} \right)^2 \times \frac{1}{3} \right]^{1/2}$$

$$= \sqrt{174}$$

$$= 13.190 A$$

=

Rms value of thyristor current

$$I_{T1} = \left[ \frac{1}{2\pi} \left\{ \left( \frac{420}{3 \times 15} \right)^2 \times \frac{2\pi}{3} + \left( \frac{2 \times 420}{3 \times 15} \right)^2 \times \frac{\pi}{3} \right\} \right]^{1/2}$$

$$= \left[ \frac{1}{2\pi} \cdot \frac{\pi}{3} \left\{ \left( \frac{420}{3 \times 15} \right)^2 \times 2 + \left( \frac{2 \times 420}{3 \times 15} \right)^2 \right\} \right]^{1/2}$$

$$= \left[ \frac{1}{6} \left\{ \left( \frac{420}{45} \right)^2 \times 2 + \left( \frac{840}{45} \right)^2 \right\} \right]^{1/2}$$

$$= \left[ \frac{1}{6} \left\{ 87.12 \times 2 + 340.4 \right\} \right]^{1/2}$$

$$= \sqrt{87}$$

$$= 9.324 \text{ A}$$

Power delivered to Load

$$= 3 I_{\text{av}}^2 R$$

$$= 3 (\sqrt{174})^2 \times 15$$

$$= 7.83 \text{ kW}$$

(b) 120° mode:

$$I_{\text{av}} = \left[ \frac{1}{\pi} \left( \frac{210}{420} \right)^2 \times \frac{2\pi}{3} \right]^{1/2}$$

$$= \sqrt{\frac{(14)^2}{3}} \times \frac{1}{2}$$

$$= \sqrt{130.66}$$

~~$$= \sqrt{9.333}$$~~

~~$$= 11.430 \text{ A}$$~~

Rms value of thyristor current

$$I_{T_1} = \left[ \frac{1}{2\pi} \left( \frac{210}{420} \right)^2 \times \frac{2\pi}{3} \right]^{1/2}$$

$$= \sqrt{65.33}$$

$$= 0.002 \text{ A}$$

### (B) Load power

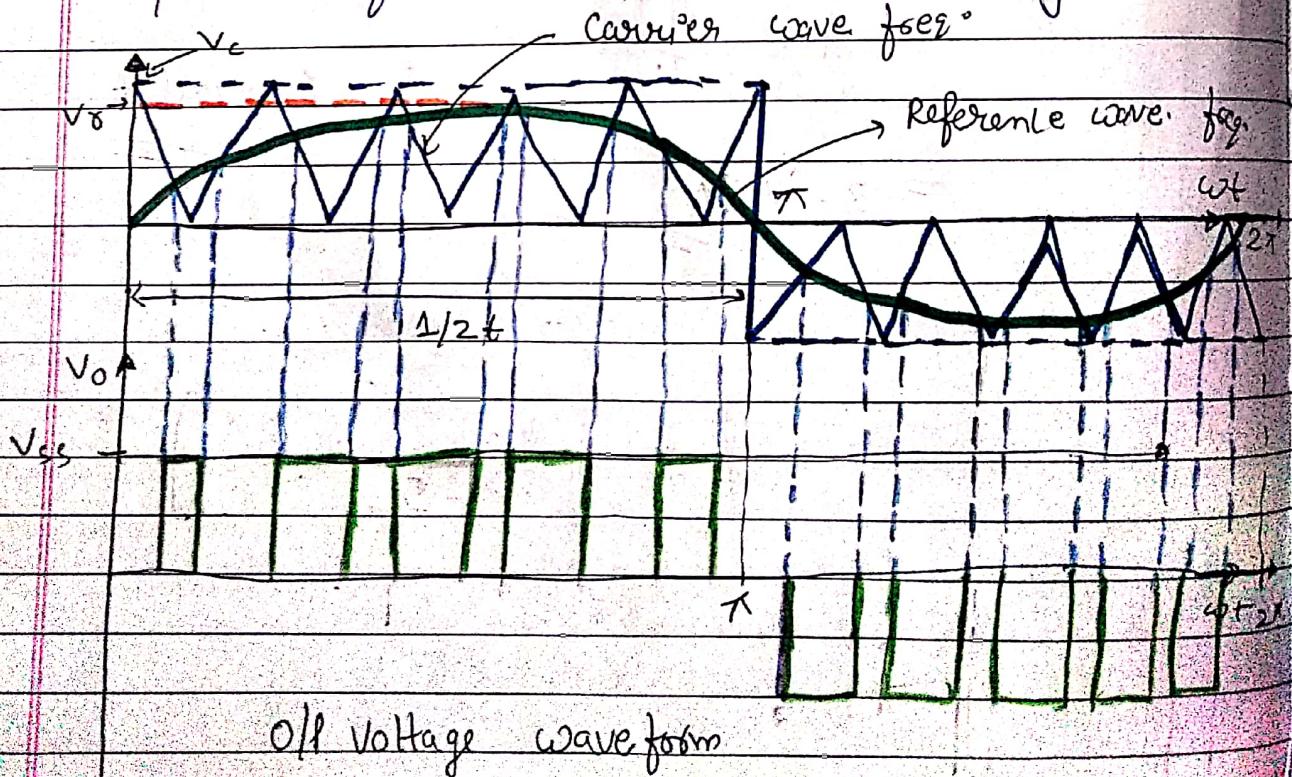
$$= 3 I^2 R$$

$$= 3 (\sqrt{3} \cdot 6)^2 \cdot 15$$

$$= 5.877 \text{ kW}$$

Ques 08 Discuss Sinusoidal pulse modulation as used in PWM Inverters?

In this method of modulation, several pulses per half cycle are used as in the case of multiple-pulse modulation (MPM). In MPM, the pulse width is equal for all the pulses. But in SPM, the pulse width is a sinusoidal func<sup>n</sup> of the angular position of the pulse in a cycle.





for realizing Sime's a high-freq. triangular carrier wave  $V_c$  is compared with a sinusoidal reference wave  $V_r$  of desired freq.

The intersection of  $V_c$  &  $V_r$  waves determines the switching instants and commutation of the modulated pulse.  $V_c$  is the peak value of triangular carrier wave and  $V_r$  that of the reference, or modulating signal.

The carrier & reference waves are mixed in a comparators. When sinusoidal wave has magnitude higher than the triangular wave, the comparator o/p is high, otherwise it is low.