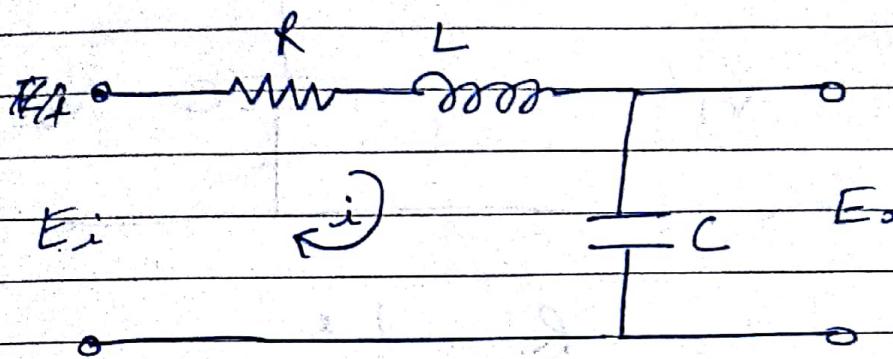


17/03/18

$$\left| \begin{array}{l} R - R \\ L - L_S \\ C - \frac{1}{C_S} \end{array} \right|$$

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## Control Systems

~~Ex. 1.2~~

$$E_i = iR + L\frac{di}{dt} + \frac{1}{C} \int (i + I_{\text{in}}) dt$$

$$E_i(s) = RI(s) + L s I(s) + \frac{I}{Cs}$$

$$E_o(s) = \frac{I}{Cs}$$

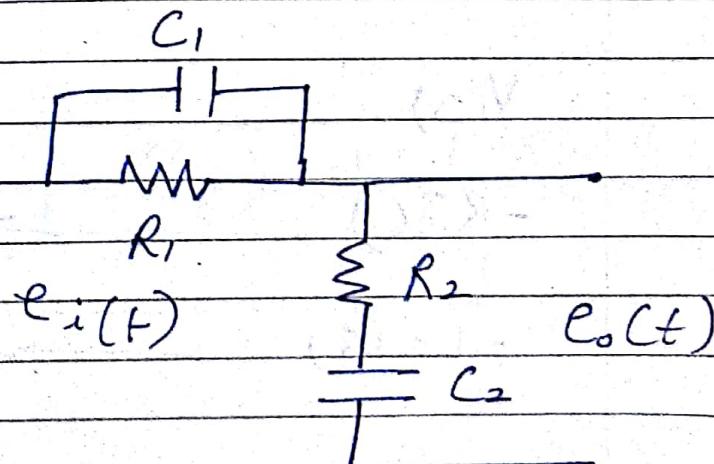
$$\frac{E_o}{E_i} = \frac{I}{Cs} \times \frac{Cs}{CsRI(s) + Cs^2LI(s) + 1}$$

$$= \frac{1}{CsRI(s) + Cs^2LI(s) + 1}$$

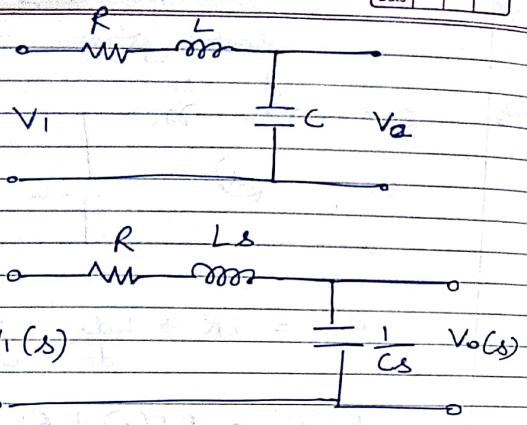
HW  
1.49

HW  $\rightarrow$  Eg. 1-5

Find TF.



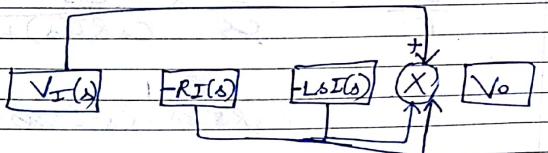
1.29



$$V_1(s) = IR + I(s)s + \frac{I(s)}{Cs}$$

$$V_1(s) = IR + LS I(s) + V_0(s)$$

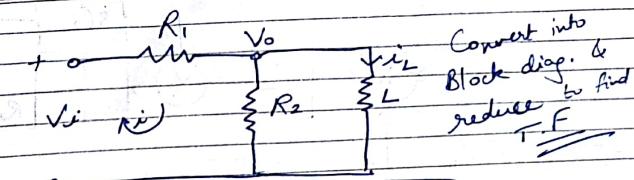
$$V_o = V_t(s) - IR - LS I(s)$$



A circuit diagram for problem 10. It features a dependent voltage source labeled  $V_1(s)$  at the top left. A resistor  $-RI(s)$  is connected in series with it. Below this, another dependent voltage source  $-Ls I(s)$  is shown. To the right of the first dependent source, there is a node with two positive charges (+) and one negative charge (-). From this node, a current  $V_o$  flows to the right. A feedback loop originates from this node, goes up, then right, then down, and finally back to the negative terminal of the first dependent source. This feedback loop is labeled  $\text{L(H)}$ . At the bottom right, there is a box containing the expression  $\frac{G}{1+qH}$ .

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1.31



1.33

$$\begin{array}{c} 1 \\ \hline R_2 \cdot S C_2 \\ + \quad \quad \quad 1 \\ \hline R_2 \cdot S C_2 \end{array}$$

$$\begin{array}{l} \cancel{\text{L}_2\text{S}\text{C}_2} \\ \cancel{\text{R}_2.\text{S}\text{C}_2+1} \\ \cancel{\text{R}_2.\text{S}\text{C}_2} \end{array}$$

$$R_2 \cdot S(G^+)$$

|  |  |  |
|--|--|--|
|  |  |  |
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Convert into  
Block diag. &  
reduce to find  
~~F~~

R2.5

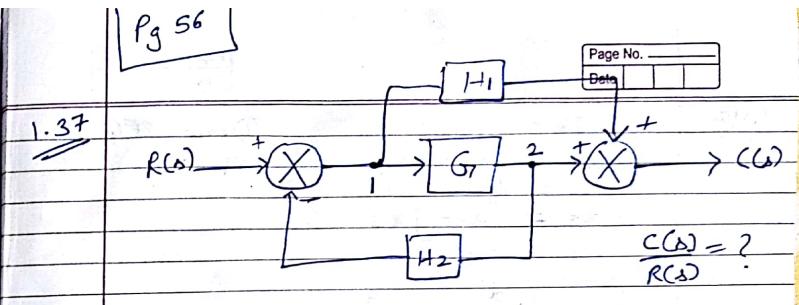
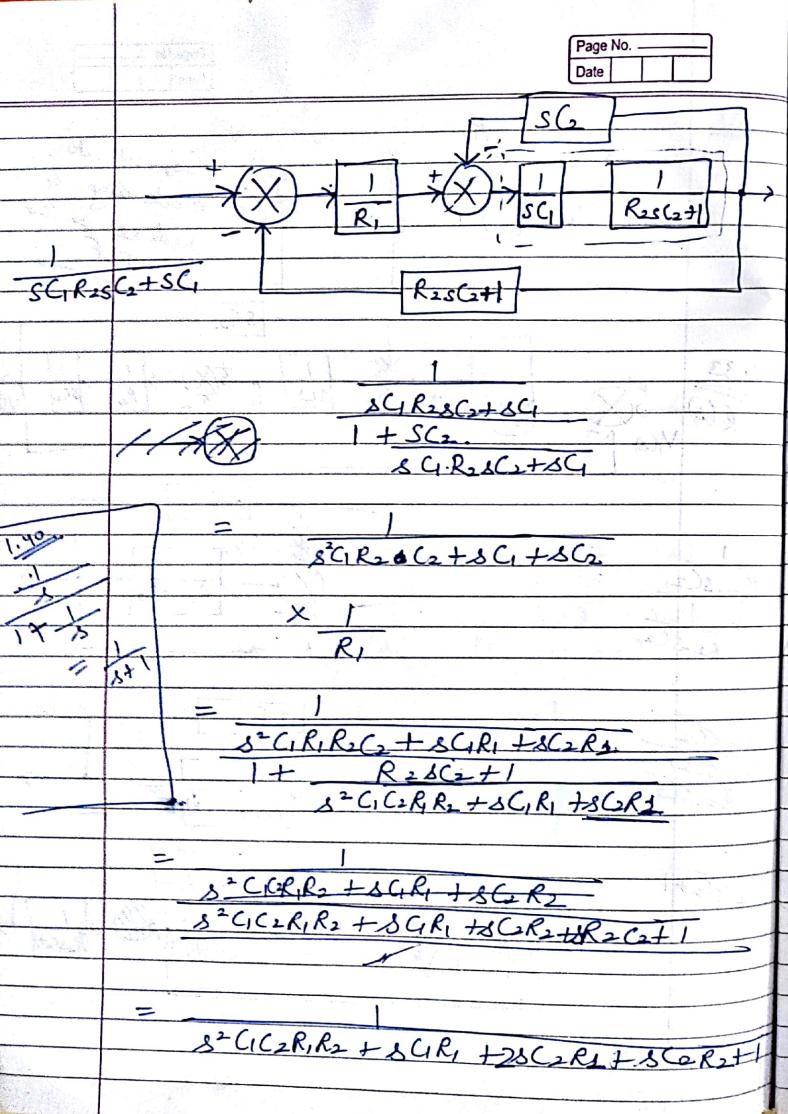
$$|f(z+)|$$

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— 1 —



• Signal Flow Graph

Mason's Formula

T.F. = Transmittance

$$T = \frac{\sum g_k \Delta_k}{\Delta}$$

$g_k \rightarrow$  Gain of fwd. path.

$\Delta_k =$  Part of  $\Delta$  not touching the  $k^{th}$  fwd. path.  
 $1 - (0)$

$\Delta \rightarrow 1 - [\text{sum of all individual loop gains}]$

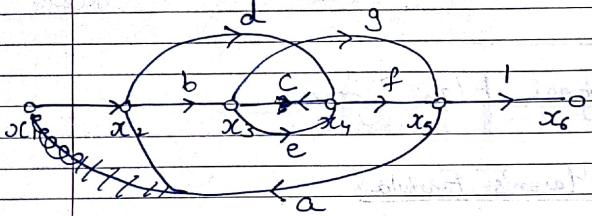
$+ [\text{sum of all product of 2 non-touching loops}]$

$- [\text{sum of all product of 3 non-touching loops}]$

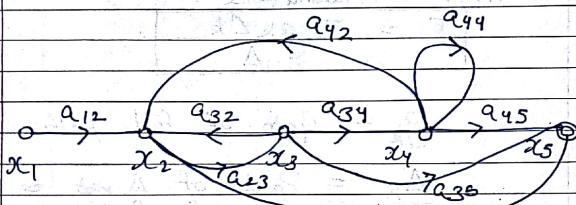
1.42

$$\begin{aligned}x_2 &= x_1 + g x_5 \\x_3 &= b x_2 + (x_4) \\x_4 &= d x_2 + e x_3 \\x_5 &= f x_4 + g x_3 \\x_6 &= x_5\end{aligned}$$

Draw SFG  
TF = ?



1.43



Gain of fwd. paths

$$g_{A_1} = a_{12} a_{23} a_{34} a_{45} \quad A_1 = 1 - 0$$

$$g_2 = a_{12} a_{23} a_{35} \quad A_2 = (1 - a_{44})$$

Gain of loops

$$L_1 = a_{32} a_{23}$$

$$L_2 = a_{23} a_{35} a_{52}$$

$$L_3 = a_{23} a_{34} a_{45} a_{52}$$

$$L_4 = a_{44}$$

$$L_5 = a_{23} a_{34} a_{42}$$

Gain of 2-non touching Loops

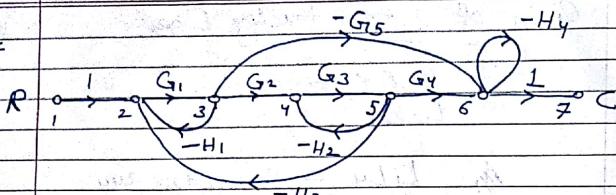
$$\frac{1}{L_1 L_4}, \frac{1}{L_2 L_4}$$

$$\Delta_1 L_1 L_4 = a_{32} a_{23} a_{44} \\ L_2 L_4 = a_{23} a_{35} a_{52} a_{44}$$

$$\Delta T = \frac{g_1 A_1 + g_2 A_2}{\Delta}$$

$$T = \frac{(a_{12} a_{23} a_{34} a_{45}) + [(a_{12} a_{23} a_{35}) - (a_{12} a_{23} a_{33} a_{44})]}{1 - (a_{32} a_{23} + a_{23} a_{35} a_{52} + a_{23} a_{34} a_{45} a_{52} + a_{44} + a_{23} a_{24} a_{45}) + (a_{32} a_{23} a_{44} + a_{23} a_{35} a_{52} a_{44})}$$

1.44



Gain of fwd. Path

$$g_1 = G_1 G_2 G_3 G_4$$

$$g_2 = G_1 G_5$$

Individual Loop Gains

$$L_1 = -G_1 H_1$$

$$L_2 = -G_2 H_2$$

$$L_3 = -G_3 H_3$$

$$L_4 = -H_4$$

Two non-touching Loops

$$L_1 L_2 = G_1 H_1 G_2 H_2$$

$$L_1 L_4 = G_1 H_1 H_4$$

$$L_2 L_4 = G_1 G_2 G_3 H_3 H_4$$

$$L_3 L_4 = G_1 G_2 G_3 H_3 H_4$$

Three Non-touching

$$L_1 L_2 L_4 = -G_1 H_1 G_2 H_2 H_4$$

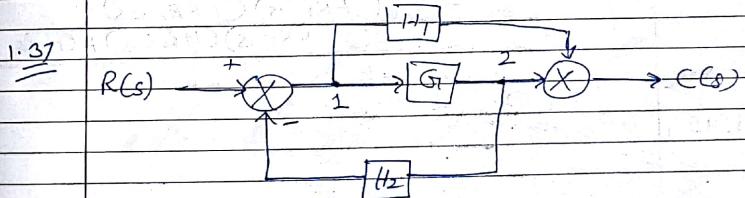
$$\Delta_1 = 1 - (0) = 1$$

$$\Delta_2 = 1 + G_3 H_2$$

$$\frac{C}{R} = \frac{g_1 \Delta_1 + g_2 \Delta_2}{\Delta}$$

$$\Delta = 1 - (L_1 + L_2 + L_3 + L_4) + (L_1 L_2 + L_1 L_4 + L_2 L_4 + L_3 L_4) - (L_1 L_2 L_4)$$

$$\Rightarrow C = \frac{G_1 G_2 G_3 G_4 + G_1 G_5 (1 + G_3 H_2)}{1 + G_1 H_1 + G_2 H_2 + G_1 G_2 G_3 H_3 + H_4 + G_3 G_4 H_1 H_2 + G_1 H_1 H_4 + G_3 H_2 H_4 + G_1 G_2 G_3 H_3 H_4 + G_1 G_3 H_1 H_2 H_4}$$



1.4445

$$Z_1 = \frac{R_1 \cdot 1}{sC_1} = \frac{R_1}{R_1 C_1 s + 1}$$

$$\frac{R_1 + 1}{sC_1}$$

$$Z_2 = \frac{R_2 + 1}{sC_2} = \frac{R_2 s C_2 + 1}{sC_2}$$

$$\frac{E_o}{E_i} = \frac{Z_2(s)}{Z_1(s) + Z_2(s)}$$

$$= \frac{(1 + R_2 s C_2) / s C_2}{\frac{R_1 + 1}{s C_1} + R_2 s C_2 + 1}$$

$$= \frac{(1 + R_1 C_1 s)(1 + R_2 C_2 s)}{(1 + R_1 C_1 s)(1 + R_2 C_2 s) + R_1 C_1 s}$$

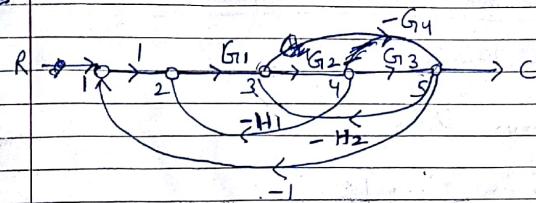
HW

1.45, 1.50, 1.52

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Gain of all feed. paths

$$g_1 = G_1 G_2 G_3$$

$$g_2 = -G_1 G_4$$

Gain of Individual Loops

$$L_1 = -G_1 G_2 G_3 H_2$$

$$L_2 = -G_1 G_3 H_1$$

$$L_3 = -G_2 G_3 H_2$$

$$L_4 = G_4 H_2$$

Gain of all single non-touching loops

$$\Delta_1 = 1 - (0) = 1$$

$$\Delta_2 = 1$$

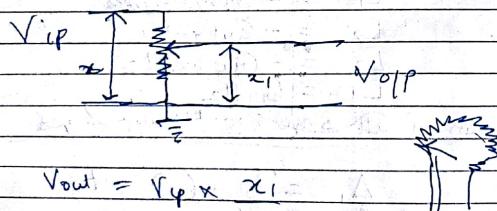
$$T = g_1 \Delta_1 + g_2 \Delta_2$$

$$\Delta = 1 + (G_1 G_2 G_3 + G_1 G_2 H_1 + G_2 G_3 H_2 - G_4 H_2)$$

$$T = \frac{G_1 G_2 G_3 + G_1 G_4}{1 + (G_1 G_2 G_3 + G_1 G_2 H_1 + G_2 G_3 H_2 - G_4 H_2)}$$

### Control System Components

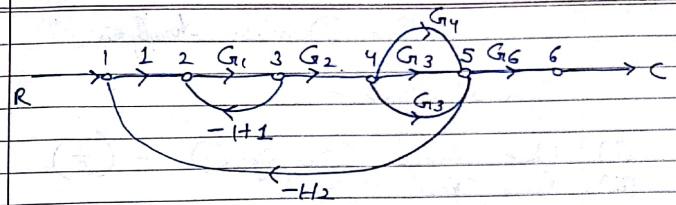
- (i) Potentiometer
- (ii) Servo Motor
- (iii) Tacho Generator
- (iv) Stepper Motor
- (v) Synchro



$$V_{out} = V_{ip} \times x_1$$

$V_{out} \propto x_1$

1.45



$$\begin{aligned} g_1 &= G_1 G_2 G_3 G_6 \\ g_2 &= G_1 G_2 G_4 G_6 \\ g_3 &= G_1 G_2 G_3 G_6 \end{aligned}$$

$$\begin{aligned} \Delta_1 &= 1 - \sigma = 1 \\ \Delta_2 &= 1 - \sigma = 1 \\ \Delta_3 &= 1 - \sigma = 1 \end{aligned}$$

### Individual Loop Gains

$$L_1 = -G_1 H_1$$

$$L_2 = -G_1 G_2 G_3 H_2$$

$$L_3 = -G_1 G_2 G_4 H_2$$

$$L_4 = -G_1 G_2 G_5 H_2$$

### Two non-touching

$$\Delta = 1 - (L_1 + L_2 + L_3 + L_4)$$

$$\frac{C}{R} = \frac{g_1 \Delta_1 + g_2 \Delta_2}{\Delta}$$

$$= \frac{G_1 G_2 G_3 G_6 + G_1 G_2 G_4 G_6 + G_1 G_2 G_5 G_6}{1 + G_1 H_1 + G_1 G_2 G_3 H_2 + G_1 G_2 G_4 H_2 + G_1 G_2 G_5 H_2}$$

25/3/18

|          |  |  |
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### Time Domain Analysis

Types of Signals -

- Unit Signal (Step Signal) -  $(\frac{1}{s})$
- Ramp Signal -  $(\frac{1}{s^2})$
- Impulse Signal -  $(1)$

Types of Response -

- Steady State Response
- Transient Response.

Time Response of 1<sup>st</sup> Order System -



$$T = \frac{C(s)}{R(s)} = \frac{1}{sT}$$

$$\begin{aligned} &= \frac{1 + 1}{sT} \\ &= \frac{\frac{1}{sT}}{\frac{sT + 1}{sT}} \\ &= \frac{1}{1 + sT} \end{aligned}$$

$$= \frac{1}{1 + sT}$$

### Step Functions

$$\frac{C(s)}{1/s} = \frac{1}{sT+1}$$

$$= \frac{1}{s(sT+1)}$$

$$C(s) = \frac{A}{s} + \frac{B}{sT+1}$$

~~$$C(s) = \frac{A(sT+1)+Bs}{s(sT+1)}$$~~

$$1 = A(sT+1) + Bs$$

Put  $s = 0$   
 $A = 1$

$$1 = (T+1) + B$$

$$B = -T$$

$$C(s) = \frac{1}{s} + \frac{(-T)}{sT+1}$$

~~$$C(s) = \frac{1}{s} + \frac{-T}{sT+1}$$~~

$$C(t) = u(t) - \frac{1}{sT+1}$$

$$= u(t) - \frac{1}{T}$$

$$= u(t) - \frac{1}{T} \int_0^t u(\tau) d\tau$$

$$= u(t) - \pi e^{-t/T}$$

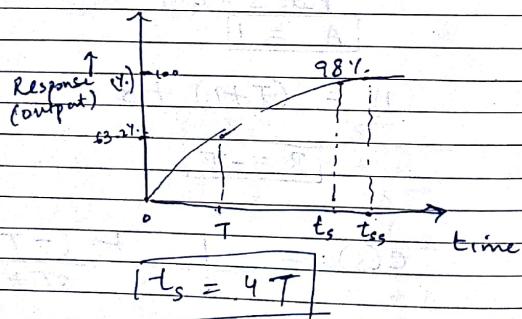
$$[c(t) = 1 - e^{-t/T}]$$

Time constant

$$T = \frac{t}{-e^{-t/T}}$$

$$= 1 - e^{-1} = 0.632$$

= 63.2% of final value



$t_s = T$  = Transient time

$t_{ss}$  onwards = Steady state time

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Ramp

$$R(s) = \frac{1}{s^2}$$

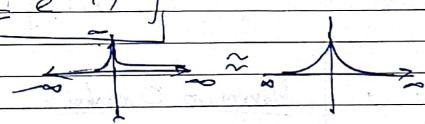
$$[c(t) = t - T + T e^{-t/T}]$$

Impulse

$$R(s) = 1$$

$$[c(t) = \frac{1}{T} e^{-t/T}]$$

$\left. \begin{array}{l} \infty \\ t=0 \end{array} \right\} \quad \left. \begin{array}{l} 0 \\ t \rightarrow \infty \end{array} \right\}$



Time Response for II<sup>nd</sup> Order

$$[s^2 + 2\zeta\omega_n s + \omega_n^2]$$

$\omega \rightarrow$  Angular freq.

$\zeta \rightarrow$  Damping Ratio

P98

$\zeta = 0 \rightarrow$  undamped system

$\zeta = 0-1 \rightarrow$  Underdamped

$\zeta > 1 \rightarrow$  Over damped

$\zeta = 1 \rightarrow$  Critically Damped

→ Definitions on Pg. 91

• Rise Time

$$t_r = \frac{\pi - \tan^{-1}(\sqrt{1-\zeta^2}/\zeta)}{\omega_n \sqrt{1-\zeta^2}}$$

• Peak Time

$$t_p = \frac{\pi}{\omega_n \sqrt{1-\zeta^2}}$$

• Maximum Overshoot

$$\% \text{ } m_p = e^{-\frac{\pi \zeta}{\sqrt{1-\zeta^2}}} \times 100$$

• Settling Time

$$t_s = \frac{4}{\zeta \omega_n}$$

Eg 2.1 Unit Step fn.

$$\zeta = 0.5$$

$$\omega_n = 6 \text{ rad/sec}$$

$$t_r = ? \quad t_p = ? \quad T_s = ? \quad M_p = ?$$

$$t_r = \frac{\pi - \tan^{-1}(\sqrt{1-0.5^2}/0.5)}{6 \sqrt{1-0.5^2}}$$

$$= \frac{\pi - 66.20}{6 \times \frac{\sqrt{3}}{2}} = \frac{79.1}{6 \times \frac{\sqrt{3}}{2}} \text{ sec}$$

$$= \frac{\pi - 1.04}{6 \times \frac{\sqrt{3}}{2}} = \frac{2.101}{3\sqrt{3}}$$

$$= 0.404 \text{ sec}$$

$$t_p = \frac{\pi}{6 \sqrt{1-0.5^2}} = \frac{\pi}{3\sqrt{3}} = 0.60 \text{ sec}$$

$$\% \text{ } M_p = e^{-\frac{\pi \times 0.5}{\sqrt{1-0.5^2}} \times 100}$$

$$= 0.163 \times 100$$

$$= 16.3 \%$$

$$t_d = \frac{4}{0.5 \times 6}$$

$$= 1.33 \text{ sec}$$

$$\underline{g_1} = G_1 G_2 G_3 G_4 G_5$$

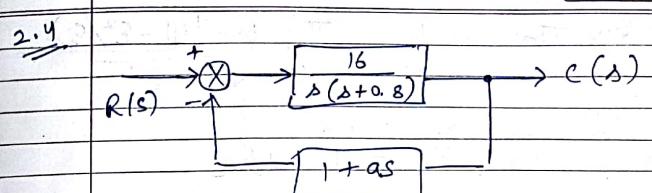
$$G_2 = G_{13} G_{15} G_7$$

$$g_3 = G_2 G_1 G_7$$

$$g_4 = G_{13} G_8 G_6$$

$$g_5 = -G_1 G_4 G_2 G_8 G_6$$

$$g_6 = -G_{13} G_8 G_6 + G_1 G_7$$



$$\frac{C(s)}{R(s)} = \frac{16}{s(s+0.8)}$$

$$= \frac{16}{s(s+0.8) + 16(1+a\delta)}$$

$$= \frac{16}{s^2 + 0.8s + 16 + 16a\delta}$$

$$= \frac{16}{s^2 + s(16a+0.8) + 16}$$

$$\Rightarrow \epsilon \omega_n = 8a + 0.4$$

$$\boxed{\omega_n = 4}$$

$$\Rightarrow \epsilon = 2a + 0.1$$

$$0.5 = 2a + 0.1$$

$$\Rightarrow \boxed{a = 0.2}$$

$$\text{Rise Time} = t_r = \frac{\pi - \tan^{-1}(1 - \zeta^2/\xi)}{\omega_n \sqrt{1 - \xi^2}}$$

$$\xi = 0.5 \\ \omega_n = 4 \text{ rad/sec}$$

$$t_r = \frac{\pi - \tan^{-1}(\sqrt{1 - 0.5^2}/0.5)}{4 \sqrt{1 - 0.5^2}}$$

$$= 1.17 / -20.16862$$

$$\frac{\pi - \pi/3}{4 \sqrt{1 - 0.5^2}}$$

$$t_r = 0.604 \text{ sec.}$$

$$\text{Max. Overshoot} = e^{-\frac{\pi \zeta}{\sqrt{1-\xi^2}}} \times 100 \\ = e^{-\frac{\pi \times 0.5}{\sqrt{1-0.5^2}}} \times 100 \\ = 16.30\%$$

HW → 2.5, 2.6, 2.18

2.12

$$G(s) = \frac{12}{s^2 + 4s + 16}, H(s) = K_s$$

$$\xi = 0.8 \\ V.M.P = ? \\ K = ? \\ C.R = \frac{G(s)}{1 + G(s).H(s)}$$

$$T = \frac{12}{s^2 + 4s + 16} \\ = \frac{12}{s^2 + 4s + 16} \cdot K_s$$

$$= \frac{12}{s^2 + 4s + 16} \\ = \frac{12}{s^2 + 4s + 16 + 12K_s} \\ = \frac{12}{s^2 + 4s + 16}$$

$$1.6 \times 10^{-3} = \frac{12}{s^2 + 4s + 16 + 12K_s}$$

$$2 \times 10^{-3} = 10^{-3} \quad \text{cancel}$$

$$= \frac{12}{s^2 + (4 + 12K_s)s + 16}$$

$$\Rightarrow \xi \omega_n = 2 + 6K \quad \omega_n = 4 \text{ rad/sec}$$

$$\xi = \frac{1}{4} (2 + 6K)$$

$$0.8 \times 4 - 2 = K \\ \frac{6}{6} = K$$

$$\Rightarrow K = \frac{1}{6} = 0.2$$

$$\%MP = e^{\frac{-1}{1-e^2}} \times 100$$

$$= 1.51\%$$

— 1. 51 5.

$$\begin{aligned}
 & \text{2.17} \\
 & \cancel{\boxed{\zeta = 0.5}} \\
 & \frac{16}{s + 0.8} \\
 & \underline{s + 16k} \\
 & s + 0.8 \\
 \\ 
 & = \frac{1}{s} \frac{16}{s + 16k + 0.8} \\
 \\ 
 & = \frac{1}{s^2 + 16sk + 0.8s} \\
 & \underline{1 + \frac{16}{s^2 + 16sk + 0.8s}} \\
 \\ 
 & = \frac{-16}{s^2 + (16k + 0.8)s + 16}
 \end{aligned}$$

$$\sum \omega_n = 8k + 0.4$$

$$\Rightarrow \varepsilon = 28k + 0.41$$

62

$$k = \underline{0.4}$$

82

$$k = \underline{0.0028} \quad \underline{0.2}$$

$$Y.M.P = e^{-\frac{\pi \cdot c}{\sqrt{1-c^2}}} \times 100$$

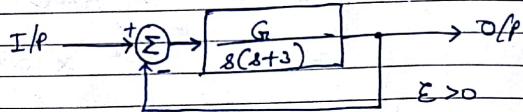
$$= 1.51 \%$$

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$$t_s = \frac{4}{\varepsilon_{wn}} = \frac{4}{0.5 \times 4}$$

$$= \frac{1}{0.5} = 2 \text{ sec}$$

2.5



(a)

$$\frac{C(s)}{R(s)} = \frac{G}{s(s+3)}$$

$$= \frac{G}{s^2 + 3s + G}$$

(b)

$$s^2 + 2\zeta\omega_n s + \omega_n^2 = 0$$

$$\Rightarrow 2\zeta\omega_n = 3 \quad \omega_n = \sqrt{G}$$

$$\zeta = \frac{3}{2} \sqrt{\frac{G}{s}}$$

2.6

$$G(s) = \frac{10}{(s+2)(s+5)}$$

$$1 + G(s)(s) = 0$$

$$\Rightarrow 1 + \frac{10}{(s+2)(s+5)} = 0$$

$$s^2 + 7s + 20 = 0$$

Standard eqn. is  $s^2 + 2\zeta\omega_n s + \omega_n^2 = 0$

$$\Rightarrow 2\zeta\omega_n = 7 \quad \omega_n^2 = 20$$

$$\zeta = \frac{7}{2 \times 4.472} \quad \omega_n = \sqrt{20} = 4.472$$

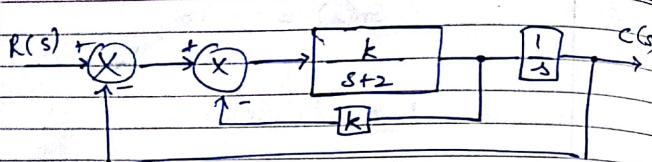
$$\zeta = 0.782$$

$$M_p = e^{-\frac{\pi\zeta}{\sqrt{1-\zeta^2}} \times 100}$$

$$= e^{-\frac{\pi \times 0.782}{\sqrt{1-0.782^2}} \times 100}$$

$$M_p = 1.92\%$$

2.18



$$\frac{C(s)}{R(s)} = \frac{k}{s^2 + (2+k_h)s + k}$$

$$s^2 + (2+k_h)s + k = 0$$

$$\omega^2 + 2\zeta\omega_n s + \omega_n^2 = 0$$

~~$$2\zeta\omega_n s = 2+k_h$$~~

$$\omega^2 = k$$

~~$$2\zeta\omega_n s + 2+k_h = 0$$~~

$$\Rightarrow \omega = \sqrt{k}$$

$$K = 16$$

$$2 \times 0.7 \times 4 = 2 + 16 k$$

$$\Rightarrow \omega = 4$$

$$\Rightarrow k = 0.225$$

2.19

$$c(t) = 1 - e^{-t/T}$$

$$t = 60 \text{ sec}$$

$$T = ?$$

$$c(t) = \frac{98}{100}$$

$$\frac{98}{100} = 1 - e^{-60/T}$$

$$e^{-60/T} = 1 - 0.98$$

$$e^{-60/T} = 0.02$$

$$\frac{-60}{T} \ln(e) = \ln(0.02)$$

$$\frac{-60}{T} = \frac{60}{3.912}$$

$$\Rightarrow T = \frac{60}{3.912}$$

$$\boxed{T = 15.337 \text{ sec}}$$

### Error Analysis

#### Type of Errors

Static Error Coefficients -  
Static Position Error ( $K_p$ )

$$K_p = \lim_{s \rightarrow 0} G(s) \cdot H(s)$$

(i) Steady State Error ( $e_{ss}$ )

$$e_{ss} = \frac{1}{1 + K_p}$$

(ii) Static Velocity Error ( $K_v$ )

$$K_v = \lim_{s \rightarrow 0} s \cdot G(s) \cdot H(s)$$

$$e_{ss} = \frac{1}{K_v}$$

(iii) Static Acceleration Error ( $K_a$ )

$$K_a = \lim_{s \rightarrow 0} s^2 \cdot G(s) \cdot H(s)$$

$$e_{ss} = \frac{1}{K_a}$$

(iv)

$$e_{ss} = \lim_{s \rightarrow 0} \frac{s \cdot R(s)}{1 + G(s)H(s)}$$

[Final Value Theorem]

Eg 3.1

$$G(s) = \frac{50}{(1+0.1s)(s+10)}$$

$$H(s) = 1$$

$$K_p, K_v, K_a = ?$$

$$K_p = \lim_{s \rightarrow 0} \frac{50}{0.1s^2 + 1.1s + 10}$$

$$K_p = \frac{50}{10} = 5$$

$$K_v = \lim_{s \rightarrow 0} s \cdot G(s) H(s)$$

$$= \lim_{s \rightarrow 0} s \cdot \frac{50}{0.1s^2 + 1.1s + 10}$$

$$= \lim_{s \rightarrow 0} \frac{50s}{(0.1s^2 + 1.1s + 10)}$$

$$= 50$$

$$\Rightarrow K_v = 0$$

$$K_a = 0$$

3.2

|          |  |
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$$G(s) = \frac{5(s^2 + 2s + 100)}{s^2 + (s+5)(s^2 + 3s + 10)}$$

$$H(s) = 1$$

$$K_p = \lim_{s \rightarrow 0} \frac{s(100)}{s(s+20)} = 0$$

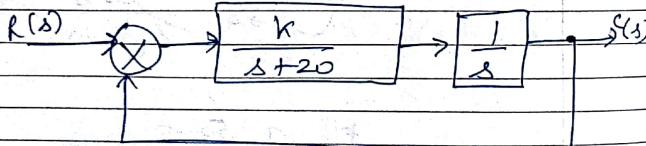
$$K_p = \infty$$

$$K_v = \infty$$

$$K_A = \lim_{s \rightarrow \infty} s \times \frac{5(100)}{5(10)} = \infty$$

$$K_A = 10$$

3.3



$$G(s) = \frac{k}{s(s+20)}$$

$$H(s) = 1$$

$$R(s) = \frac{1}{s^2}$$

$$k = ?$$

(i)  $K = 400$

(ii)  $e_{ss} = 0.02$

$$e_{ss} = \lim_{s \rightarrow 0} \frac{s \cdot R(s)}{1 + G(s) H(s)}$$

$$e_{ss} = \lim_{s \rightarrow 0} \frac{s}{\frac{1}{s+20}} = \lim_{s \rightarrow 0} s^2 \cdot \frac{1}{s+20} = 0$$

$$e_{ss} = \lim_{s \rightarrow 0} s \cdot \frac{1}{s+20} = \lim_{s \rightarrow 0} \frac{s}{s+20} = 0$$

$$e_{ss} = 0$$

$$= \lim_{s \rightarrow 0} \frac{1}{s+20} = \frac{1}{20}$$

$$= \frac{20}{20+400}$$

$$e_{ss} = \frac{20}{420} = \frac{1}{21} = 0.0476$$

(iii)

$$e_{ss} = 0.02$$

$$0.02 = \lim_{s \rightarrow 0} s \cdot \frac{1}{s+20} = \frac{1}{20}$$

$$0.02 = \frac{20}{20+k}$$

$$0.02k + 0.4 = 20$$

$$k = 980$$

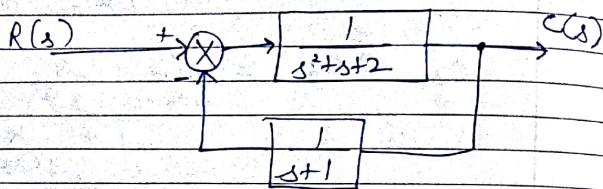
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$H_w \rightarrow 3.5, 3.4$

~~W.B.E.T~~

3.8



$$G(s) = \frac{1}{s^2 + s + 2}$$

$$H(s) = \frac{1}{s+1}$$

$$K_p, e_{ss} = ?$$

$$K_p = \lim_{s \rightarrow 0} G(s)H(s)$$

$$= \lim_{s \rightarrow 0} \frac{1}{(s+1)(s^2 + s + 2)}$$

$$= \frac{1}{2}$$

$$[K_p = 0.5]$$

3.5

$$e_{ss} = \frac{1}{1 + K_p}$$

$$e_{ss} = \frac{1}{1 + 0.5}$$

$$e_{ss} = 0.666$$

$$G(s) = \frac{10s}{s^2(s+4)(s^2 + 3s + 12)}$$

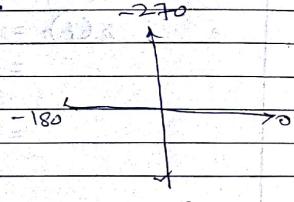
$$\begin{aligned} r(t) &= 2 + 5t + 2t^2 \\ &= \frac{2}{s} + \frac{5}{s^2} + \frac{4}{s^3} \\ &= \frac{2s^2 + 5s + 4}{s^3} \end{aligned}$$

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Unit 3Frequency Domain AnalysisPolar Plots

1. Find the Transfer fn.  $G(s)$ .
2. Substitute  $s = j\omega$
3. At  $\omega = 0$  &  $\omega = \infty$  calculate  $G(j\omega)$ .
4. Calculate phase angle of  $j\omega$  at  $\omega = 0$  &  $\omega = \infty$ .
5. Rationalize  $G(j\omega)$  & separate the real & imaginary parts.



$$\text{Eg. } G(s) = \frac{1}{s}$$

$$G(j\omega) = \frac{1}{j\omega}$$

$$j\omega = \sqrt{\omega}$$

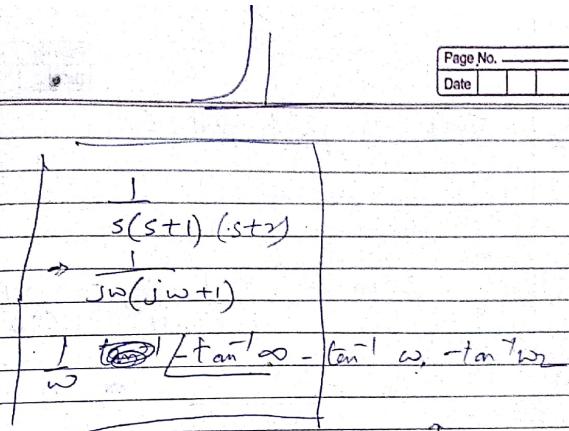
$$= \left(\frac{1}{\sqrt{\omega}}\right)^2 \angle \tan^{-1}\left(\frac{1}{\sqrt{\omega}}\right)$$

$$= \frac{1}{\omega} \angle -90^\circ$$

$$\begin{aligned} \lim_{\omega \rightarrow 0} G(j\omega) &= \frac{1}{\omega} \angle (-90^\circ) = \infty \\ \lim_{\omega \rightarrow \infty} G(j\omega) &= 0 \end{aligned}$$

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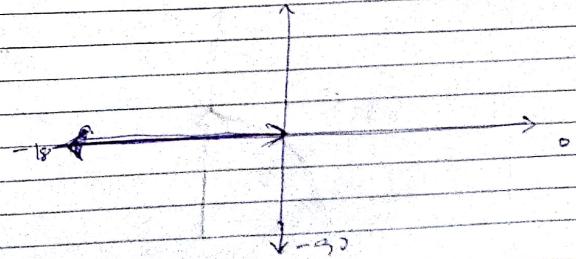
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Ex

$$\begin{aligned} G(j\omega) &= \frac{1}{(j\omega)^2} \\ &= \frac{1}{-\omega^2} \angle -180^\circ \\ &= \frac{1}{-\omega^2} \angle -180^\circ \end{aligned}$$

at  $\omega = 0$ ,  $G(j\omega) = -\infty$   
 $\omega = \infty$ ,  $G(j\omega) = 0$



4.1 Sketch the polar plot for

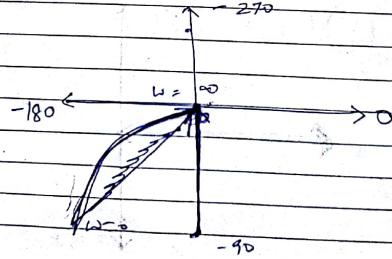
$$G(s) = \frac{1}{s(s+1)}$$

$$\begin{aligned} G(j\omega) &= \frac{1}{j\omega(j\omega+1)} = \frac{1}{(j\omega)^2 + j\omega} \\ &= \frac{1}{-\omega^2 + j\omega} \\ &= \frac{1}{\omega(j\omega+1)} \angle -\tan^{-1}(\infty) - \tan^{-1}\omega \end{aligned}$$

$$\frac{\sqrt{(j\omega)^2}}{\omega \sqrt{j^2 + \omega^2}} \angle -\tan^{-1}\infty - \tan^{-1}\omega$$

$$= \frac{1}{\omega \sqrt{1^2 + \omega^2}} \angle -90 - \tan^{-1}\omega$$

$$\begin{aligned} \text{at } \omega = 0, \quad G(j\omega) &= \infty \angle -90^\circ \\ \omega = \infty, \quad G(j\omega) &= 0 \angle -180^\circ \end{aligned}$$



HW  $\rightarrow$  4.2

$$\underline{4.3} \quad G(s) = \frac{20}{s(s+1)(s+2)}$$

$$G(j\omega) = \frac{20}{j\omega(j\omega+1)(j\omega+2)}$$

$$= \frac{1}{\omega \sqrt{1^2 + \omega^2} \sqrt{2^2 + \omega^2}} \angle -\tan^{-1}\infty - \tan^{-1}\omega - \tan^{-1}\frac{\omega}{2}$$

$$= \frac{1}{\omega \sqrt{2\omega^2 + 5} \sqrt{\omega^2 + 2}} \angle -90 - \tan^{-1}\omega - \tan^{-1}\frac{\omega}{2}$$

$$\begin{aligned} \omega = 0 \rightarrow G(j\omega) &= \infty \angle -90 \\ \omega = \infty \rightarrow G(j\omega) &= 0 \angle -270 \end{aligned}$$

$$\begin{aligned} -270 \quad G(j\omega) &= \frac{20}{j\omega(j\omega+1)(j\omega+2)} \\ &= \frac{20}{[j\omega]^3 + 3j\omega^2 + 2j\omega} \\ &= \frac{20}{-j\omega^3 + 3\omega^2 + 2j\omega} \end{aligned}$$

$$\begin{aligned} &= \frac{20}{-3\omega^2 + 2j\omega} \\ &= \frac{20}{-3\omega^2 + 2j\omega} \end{aligned}$$

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$$= \frac{20}{j(-\omega^2 + 2\omega) - 3\omega^2}$$

$$= \frac{j(20(\omega^3 - 2\omega))}{(\omega^4 + \omega^2)(4 + \omega^2)} = 0$$

$$= 20(\omega^3 - 2\omega) = 0$$

$$\Rightarrow \omega^3 = 2\omega$$

$$\omega = \pm \sqrt[3]{2}$$

## Alternative Method for Polar Plots

4.5

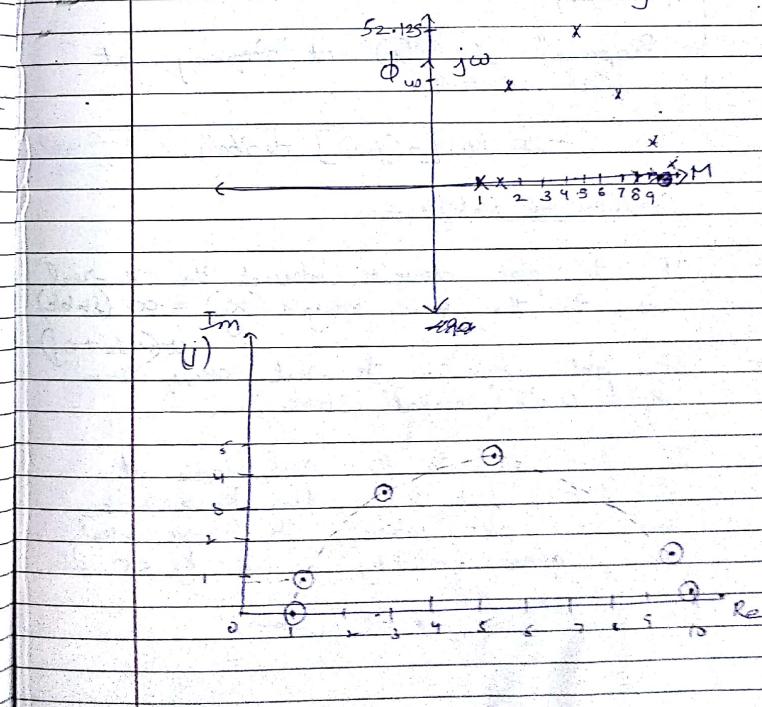
Sketch the polar plot for -

$$G(s) = \frac{10(s+1)}{s+10}$$

$$G(j\omega) = \frac{10(j\omega + 1)}{j\omega + 10}$$

$$= \frac{10 \sqrt{1^2 + \omega^2}}{\sqrt{10^2 + \omega^2}} \sqrt{\tan^{-1}\left(\frac{\omega}{10}\right) + \tan^{-1}\omega}$$

| $\omega$ | M            | $\phi$           | $M \cos \phi + j M \sin \phi$           |
|----------|--------------|------------------|---|
| 0        | 1            | 0                | 1                                       |
| 1        | $10\sqrt{2}$ | $-5.710 + j 4.5$ | $1.407 \cos(39.3) + j 1.407 \sin(39.3)$ |
|          | $\sqrt{101}$ | $= 39.3$         | $= 1.088 + j 0.89$                      |
|          | $= 1.407$    |                  |   |
| 5        | 4.560        | 52.125           | $2.799 + j 3.599$                       |
| 10       | 7.106        | 39.289           | $5.499 + j 4.499$                       |
| 50       | 9.80         | 10.164           | $9.646 + j 1.729$                       |
| 100      | 9.95         | 5.137            | $9.910 + j 0.89$                        |



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Bode Plot

$$\text{Starting Point} = 20 \log k$$

Phase Margin

$$\phi_m = 180^\circ + \phi$$

$\phi \rightarrow$  Angle of TF  $G(s)$  at gain crossover point.

Gain Margin

Reciprocal of  $G(s)$  at frequency at which  $\phi = -180^\circ$ .

$$-20 \log [G(j\omega)] \text{ decibels.}$$

Conditions for Stability -

- If the plot doesn't intersect the -ve real axis, then the gain margin ( $k_g$ ) =  $\infty$  (stable)
- If plot intersects the real axis, then  
 $k_g = 0 \text{ db}$  (marginally stable)
- If plot intersects the real axis at  $< (-1+j0)$  but  $> 0$ , then  $k_g > 0 \text{ db}$ ,  
 and if plot intersects the -ve real axis after  $(-1+j0)$ , then  $k_g < 0 \text{ db}$ .  
 $[- \rightarrow (\text{unstable})]$

4.9

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Bode Plot for TF =  $G(s) = \frac{1000}{(1+0.1s)(1+0.001s)}$

Step 1

$$\text{Initial slope} = 0 \text{ db/dec}$$

$$\begin{aligned} \text{Starting pt.} &= 20 \log(1000) \\ \text{Gain margin} &= \frac{20 \times 3}{60} \\ &= 60 \text{ db.} \end{aligned}$$

$$\omega_1 = \frac{1}{0.1} = 10 \text{ rad/sec}$$

$$\omega_2 = \frac{1}{0.001} = 1000 \text{ rad/sec}$$

Step 2

Q

(Conjugate pair)

|       |         |
|-------|---------|
| 0.1   | 0       |
| 0.2   | 3.98 dB |
| 0.5   | 5.87 dB |
| 1.0   | 7.27 dB |
| 2.0   | 8.61 dB |
| 5.0   | 9.54 dB |
| 10.0  | 9.99 dB |
| 20.0  | 10.2 dB |
| 50.0  | 10.4 dB |
| 100.0 | 10.5 dB |

After every corner freq., slope of magnitude plot reduces by 20 db/decade.

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4.10

Sketch the Bode Plot of the following TF -

$$G(s) = \frac{1000}{s(1+0.1s)(1+0.001s)}$$

Step 1

Type = 1

Initial slope = -20 db

$\therefore$  Initial Starting Point =  $20 \log K$

$$\begin{aligned} &= 20 \log(1000) \\ &= 20 \times 3 \\ &= 60 \text{ db} \end{aligned}$$

$$\omega_1 = \frac{1}{0.1} = 10 \text{ rad/sec}$$

$$\omega_2 = \frac{1}{0.001} = 1000 \text{ rad/sec}$$

$$\Phi_m = -\tan^{-1}(0.1\omega) - \tan^{-1}(0.001\omega) - 90^\circ$$

Step 2-4 Graph

Step 5

| $\omega$ | $-\tan^{-1}(0.1\omega)$ | $-\tan^{-1}(0.001\omega)$ | R        |
|----------|-------------------------|---------------------------|----------|
| 0        | 0                       | 0                         | -90      |
| 5        | -26.56                  | -0.286                    | -116.84  |
| 10       | -45                     | -0.572                    | -135.57  |
| 15       | -56.30                  | -0.859                    | -147.159 |
| 20       | -63.43                  | -1.145                    | -154.58  |
| 50       | -78.69                  | -2.862                    | -171.55  |
| 100      | -84.289                 | -5.710                    | -180.00  |
| 200      | -87.13                  | -11.309                   | -188.439 |