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A white slide background with a thin black border. In the center, the title 'STOCHASTIC PROCESSES' is written in bold black capital letters. Below it, there is a single bullet point: '• Text'. At the bottom left is the SOICT logo and the text 'VIỆN CÔNG NGHỆ THÔNG TIN VÀ TRUYỀN THÔNG'. A horizontal red line is at the bottom.

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A white slide background with a thin black border. In the center, the title 'IV. Stochastic processes' is written in bold black capital letters. Below it is a bulleted list of topics: '• Definitions', '• Stationary processes', '• Spectral density function', '• Ergodicity of stochastic processes', and '• Stochastic Data Processing Systems'. At the bottom left is the SOICT logo and the text 'VIỆN CÔNG NGHỆ THÔNG TIN VÀ TRUYỀN THÔNG'. A horizontal red line is at the bottom.

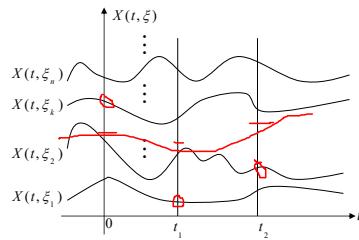
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## IV. Stochastic processes

### 4.1. Definitions

- Stochastic processes

- Let  $\xi$  denote the random outcome of an experiment. To every such outcome suppose a waveform  $X(t, \xi)$  is assigned.
- The collection of such waveforms form a stochastic process.



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## IV. Stochastic processes

### 4.1. Definitions

- Example

- $X(t)=A\cos(\omega_0 t+\varphi)$
- Where  $\varphi$  is a uniformly distributed random variable in  $(0, 2\pi)$  represents a stochastic process.

- Some stochastic processes

- Brownian motion,
- Stock market fluctuations,
- Various queuing systems.



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## IV. Stochastic processes

### 4.1. Definitions

- The set of  $\{\xi_k\}$  and the time index  $t$  can be continuous or discrete (countably infinite or finite) as well.
- For fixed  $\xi_j \in S$  (the set of all experimental outcomes),  $X(t, \xi_j)$  is a specific time function.
- For fixed  $t$ ,  $X_t=X(t, \xi_j)$  is a random variable. The ensemble of all such realizations  $X(t, \xi)$  over time represents the stochastic process  $X(t)$ .



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## IV. Stochastic processes

### 4.1. Definitions

- If  $X(t)$  is a stochastic process, then for fixed  $t$ ,  $X(t)$  represents a random variable.
- Its distribution function is given by

$$F_x(x, t) = P\{X(t) \leq x\}$$

- Notice that  $F_x(x, t)$  depends on  $t$ , since for a different  $t$ , we obtain a different random variable.

$$f_x(x, t) \triangleq \frac{dF_x(x, t)}{dx}$$

- Derivative of  $F_x(x, t)$  represents the first-order probability density function of the process  $X(t)$ .



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## IV. Stochastic processes

### 4.1. Definitions

- For  $t = t_1$  and  $t = t_2$ ,
- $X(t)$  represents two different random variables  $X_1 = X(t_1)$  and  $X_2 = X(t_2)$  respectively.
- Their joint distribution is given by

$$F_x(x_1, x_2, t_1, t_2) = P\{X(t_1) \leq x_1, X(t_2) \leq x_2\}$$

- Their joint density function is:

$$f_x(x_1, x_2, t_1, t_2) \triangleq \frac{\partial^2 F_x(x_1, x_2, t_1, t_2)}{\partial x_1 \partial x_2}$$

- And represents the second-order density function of the process  $X(t)$ .



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## IV. Stochastic processes

### 4.1. Definitions

- Similarly  $f_x(x_1, \dots, x_n, t_1, \dots, t_n)$  represents the  $n^{\text{th}}$  order density function of the process  $X(t)$ .
- Complete specification of the stochastic process  $X(t)$  requires the knowledge of  $f_x(x_1, \dots, x_n, t_1, \dots, t_n)$  for all  $t_i$ ,  $i = 1, \dots, n$  and for all  $n$ . (an almost impossible task in reality).



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## IV. Stochastic processes

### 4.1. Definitions

- Characteristics of stochastic processes
  - Mean of a stochastic process

$$\mu(t) \triangleq E\{X(t)\} = \int_{-\infty}^{+\infty} x f_x(x, t) dx$$

- $\mu(t)$  represents the mean value of a process  $X(t)$ . In general, the mean of a process can depend on the time index  $t$ .

- Autocorrelation function of a process  $X(t)$  is defined as

$$R_{xx}(t_1, t_2) \stackrel{\Delta}{=} E\{X(t_1)X^*(t_2)\} = \iint x_1 x_2^* f_x(x_1, x_2, t_1, t_2) dx_1 dx_2$$

- It represents the interrelationship between the random variables  $X_1 = X(t_1)$  and  $X_2 = X(t_2)$  generated from the process  $X(t)$ .



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## IV. Stochastic processes

### 4.1. Definitions

- Properties of autocorrelation function

$$1. R_{xx}(t_1, t_2) = R_{xx}^*(t_2, t_1) = [E\{X(t_2)X^*(t_1)\}]^*$$

$$2. R_{xx}(t, t) = E\{|X(t)|^2\} \geq 0.$$

(Average instantaneous power)

- 3.  $R_{xx}(t_1, t_2)$  represents a nonnegative definite function, i.e., for any set of constants  $\{a_i\}_{i=1}^n$

$$\sum_{i=1}^n \sum_{j=1}^n a_i a_j^* R_{xx}(t_i, t_j) \geq 0.$$

- This equation follows by noticing that:  $E\{|Y|^2\} \geq 0$  and

$$Y = \sum_{i=1}^n a_i X(t_i).$$



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## IV. Stochastic processes

### 4.1. Definitions

- The **autocovariance** function of the process  $X(t)$ .

$$C_{xx}(t_1, t_2) = R_{xx}(t_1, t_2) - \mu_x(t_1)\mu_x^*(t_2)$$

- Examples

- Given

$$z = \int_{-T}^T X(t)dt.$$

$$\begin{aligned} E[|z|^2] &= \int_{-T}^T \int_{-T}^T E\{X(t_1)X^*(t_2)\} dt_1 dt_2 \\ &= \int_{-T}^T \int_{-T}^T R_{xx}(t_1, t_2) dt_1 dt_2 \end{aligned}$$



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## IV. Stochastic processes

### 4.1. Definitions

- Consider process  $X(t)$

$$X(t) = a \cos(\omega_0 t + \varphi), \quad \varphi \sim U(0, 2\pi).$$

$$\mu_x(t) = E\{X(t)\} = aE\{\cos(\omega_0 t + \varphi)\}$$

$$= a \cos \omega_0 t E\{\cos \varphi\} - a \sin \omega_0 t E\{\sin \varphi\} = 0, \\ \text{since } E\{\cos \varphi\} = \frac{1}{2\pi} \int_0^{2\pi} \cos \varphi d\varphi = 0 = E\{\sin \varphi\}.$$

$$\begin{aligned} R_{xx}(t_1, t_2) &= a^2 E\{\cos(\omega_0 t_1 + \varphi) \cos(\omega_0 t_2 + \varphi)\} \\ &= \frac{a^2}{2} E\{\cos \omega_0(t_1 - t_2) + \cos(\omega_0(t_1 + t_2) + 2\varphi)\} \\ &= \frac{a^2}{2} \cos \omega_0(t_1 - t_2). \end{aligned}$$



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## IV. Stochastic processes

### 4.2. Stationary processes

- Strict sense and wide sense stationarity

- Stationarity

- Stationary processes exhibit statistical properties that are invariant to shift in the time index.
- Thus, for example, second-order stationarity implies that the statistical properties of the pairs  $\{X(t_1), X(t_2)\}$  and  $\{X(t_1+c), X(t_2+c)\}$  are the same for any  $c$ .
- Similarly first-order stationarity implies that the statistical properties of  $X(t_i)$  and  $X(t_i+c)$  are the same for any  $c$ .



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## IV. Stochastic processes

### 4.2. Stationary processes

- Strict sense stationarity

- In strict terms, the statistical properties of a stochastic process are governed by the joint probability density function.
- A process is  $n^{\text{th}}$ -order **Strict-Sense Stationary (S.S.S)** if

$$f_x(x_1, x_2, \dots, x_n, t_1, t_2, \dots, t_n) \equiv f_x(x_1, x_2, \dots, x_n, t_1 + c, t_2 + c, \dots, t_n + c)$$

for any  $c$ , where the left side represents the joint density function of the random variables  $X_1=X(t_1), \dots, X_n=X(t_n)$  and the right side corresponds to the joint density function of the random variables  $X'_1=X(t_1+c), \dots, X'_n=X(t_n+c)$ .

- A process  $X(t)$  is said to be **strict-sense stationary** if equation above is true for all  $t_i, i=1, \dots, n; n=1, 2, \dots$  and any  $c$ .



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## IV. Stochastic processes

### 4.2. Stationary processes

- First order strict sense stationary process

• For any c

$$f_x(x, t) \equiv f_x(x, t + c)$$

• In particular, if c = -t, then

$$f_x(x, t) = f_x(x)$$

• That means, the first-order density of  $X(t)$  is independent of  $t$ . In that case

$$E[X(t)] = \int_{-\infty}^{+\infty} xf(x)dx = \mu, \text{ a constant. (1)}$$



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## IV. Stochastic processes

### 4.2. Stationary processes

- The autocorrelation function is given by

$$\begin{aligned} R_{xx}(t_1, t_2) &\triangleq E\{X(t_1)X^*(t_2)\} \\ &= \iint x_1 x_2^* f_x(x_1, x_2, \tau = t_1 - t_2) dx_1 dx_2 \\ &= R_{xx}(t_1 - t_2) \triangleq R_{xx}(\tau) = R_{xx}^*(-\tau), \quad (2) \end{aligned}$$

- The autocorrelation function of a second order strict-sense stationary process depends only on the difference of the time indices  $t_2 - t_1 = \tau$



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## IV. Stochastic processes

### 4.2. Stationary processes

- Second order strict sense stochastic process

• From definition, we have:

$$f_x(x_1, x_2, t_1, t_2) \equiv f_x(x_1, x_2, t_1 + c, t_2 + c)$$

• for any c. If c is chosen so as c = -t\_2, we get

$$f_x(x_1, x_2, t_1, t_2) \equiv f_x(x_1, x_2, t_1 - t_2)$$

• The second order density function of a strict sense stationary process depends only on the difference of the time indices  $t_1 - t_2 = \tau$ .



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## IV. Stochastic processes

### 4.2. Stationary processes

- Wide sense stationarity

• A process  $X(t)$  is said to be **Wide-Sense Stationary** if:

$$E\{X(t)\} = \mu \text{ and } E\{X(t_1)X^*(t_2)\} = R_{xx}(t_1 - t_2),$$

• Since these equations follow from (1) and (2), strict-sense stationarity always implies wide-sense stationarity.

• In general, the converse is *not true*

- Exception: the Gaussian process (normal process).
- This follows, since if  $X(t)$  is a Gaussian process, then by definition  $X_1 = X(t_1), \dots, X_n = X(t_n)$  are jointly Gaussian random variables for any  $t_1, \dots, t_n$  whose joint characteristic function is given by

$$\phi_x(\omega_1, \omega_2, \dots, \omega_n) = e^{j \sum_{k=1}^n \mu(t_k) \omega_k - \sum_{l,k}^n \sum_{l,k}^n C_{xx}(t_l, t_k) \omega_l \omega_k / 2}$$



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## IV. Stochastic processes

### 4.2. Stationary processes

- Examples

- The process:

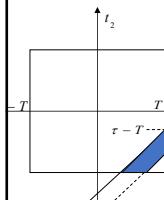
$$X(t) = a \cos(\omega_0 t + \varphi), \quad \varphi \sim U(0, 2\pi).$$

- This process is wide sense stationary but not strict sense stationary.

- If the process  $X(t)$  has zero mean, then  $\sigma_z^2$  is reduced to:

$$\sigma_z^2 = E\{|z|^2\} = \int_{-T}^T \int_{-T}^T R_{xx}(t_1 - t_2) dt_1 dt_2.$$

- As  $t_1$  and  $t_2$  varies from  $-T$  to  $T$ , so  $\tau = t_2 - t_1$  varies from  $-2T$  to  $2T$ .
- $R_{xx}(\tau)$  is a constant over the shaded region in the figure on the left.



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## IV. Stochastic processes

### 4.3. Power spectrum

- Power spectrum

- For a deterministic signal  $x(t)$

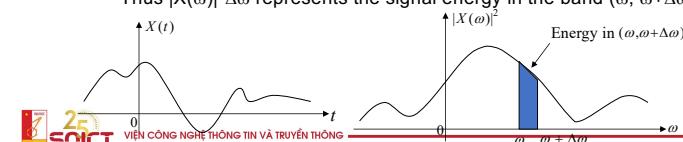
- The spectrum is well defined: If  $X(\omega)$  represents its Fourier transform,

$$X(\omega) = \int_{-\infty}^{+\infty} x(t) e^{-j\omega t} dt,$$

- then  $|X(\omega)|^2$  represents its energy spectrum. This follows from Parseval's theorem since the signal energy is given by

$$\int_{-\infty}^{+\infty} x^2(t) dt = \frac{1}{2\pi} \int_{-\infty}^{+\infty} |X(\omega)|^2 d\omega = E.$$

- Thus  $|X(\omega)|^2 \Delta\omega$  represents the signal energy in the band  $(\omega, \omega + \Delta\omega)$



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## IV. Stochastic processes

### 4.3. Power spectrum

- For stochastic processes,

- A direct application of Fourier transform generates a sequence of random variables for every  $\omega$
- For a stochastic process,  $E\{|X(t)|^2\}$  represents the ensemble average power (instantaneous energy) at the instant  $t$ .
- Partial Fourier transform of a process  $X(t)$  based on  $(-T, T)$  is given by

$$X_T(\omega) = \int_{-T}^T X(t) e^{-j\omega t} dt$$

- The power distribution associated with that realization based on  $(-T, T)$  is represented by

$$\frac{|X_T(\omega)|^2}{2T} = \frac{1}{2T} \left| \int_{-T}^T X(t) e^{-j\omega t} dt \right|^2$$



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## IV. Stochastic processes

### 4.3. Power spectrum

- The average power distribution based on  $(-T, T)$  is ensemble average of power distribution for  $\omega$

$$\begin{aligned} P_r(\omega) &= E \left\{ \frac{|X_T(\omega)|^2}{2T} \right\} = \frac{1}{2T} \int_{-T}^T \int_{-T}^T E\{X(t_1) X^*(t_2)\} e^{-j\omega(t_1 - t_2)} dt_1 dt_2 \\ &= \frac{1}{2T} \int_{-T}^T \int_{-T}^T R_{xx}(t_1, t_2) e^{-j\omega(t_1 - t_2)} dt_1 dt_2 \end{aligned}$$

- We have this represents the power distribution of  $X(t)$  based on  $(-T, T)$ .

- If  $X(t)$  is assumed to be w.s.s, then

$$R_{xx}(t_1, t_2) = R_{xx}(t_1 - t_2)$$

- and we have

$$P_r(\omega) = \frac{1}{2T} \int_{-T}^T \int_{-T}^T R_{xx}(t_1 - t_2) e^{-j\omega(t_1 - t_2)} dt_1 dt_2.$$



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## IV. Stochastic processes

### 4.3. Power spectrum

- Let  $\tau = t_2 - t_1$ , we obtain

$$\begin{aligned} P_r(\omega) &= \frac{1}{2T} \int_{-2T}^{2T} R_{xx}(\tau) e^{-j\omega\tau} (2T - |\tau|) d\tau \\ &= \int_{-2T}^{2T} R_{xx}(\tau) e^{-j\omega\tau} (1 - \frac{|\tau|}{2T}) d\tau \geq 0 \end{aligned}$$

- This is the power distribution of the w.s.s. process  $X(t)$  based on  $(-T, T)$ .

- Letting  $T \rightarrow \infty$ , we obtain

$$S_{xx}(\omega) = \lim_{T \rightarrow \infty} P_r(\omega) = \int_{-\infty}^{+\infty} R_{xx}(\tau) e^{-j\omega\tau} d\tau \geq 0$$

- $S_{xx}(\omega)$  is the *power spectral density* of the w.s.s process  $X(t)$ .



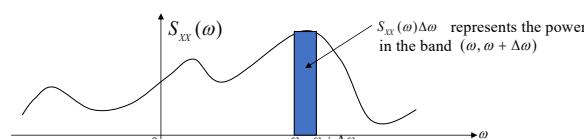
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## IV. Stochastic processes

### 4.3. Power spectrum

- The area under  $S_{xx}(\omega)$  represents the total power of the process  $X(t)$ , and hence  $S_{xx}(\omega)$  truly represents the power spectrum.



- The nonnegative-definiteness property of the auto-correlation function translates into the “nonnegative” property for its Fourier transform (power spectrum).

$$R_{xx}(\tau) \text{ nonnegative - definite} \Leftrightarrow S_{xx}(\omega) \geq 0.$$



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## IV. Stochastic processes

### 4.3. Power spectrum

- Khinchin-Wiener theorem

- The autocorrelation function and the power spectrum of a w.s.s process form a Fourier transform pair.

$$R_{xx}(\omega) \xleftarrow{\text{F.T}} S_{xx}(\omega) \geq 0.$$

$$R_{xx}(\tau) = \frac{1}{2\pi} \int_{-\infty}^{+\infty} S_{xx}(\omega) e^{j\omega\tau} d\omega$$

- For  $\tau = 0$ ,  $S_{xx}(\omega) = \lim_{T \rightarrow \infty} P_r(\omega) = \int_{-\infty}^{+\infty} R_{xx}(\tau) e^{-j\omega\tau} d\tau \geq 0$

$$\frac{1}{2\pi} \int_{-\infty}^{+\infty} S_{xx}(\omega) d\omega = R_{xx}(0) = E\{|X(t)|^2\} = P, \text{ the total power.}$$



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## IV. Stochastic processes

### 4.3. Power spectrum

- If  $X(t)$  is a real w.s.s process, then

$$R_{xx}(\tau) = R_{xx}(-\tau)$$

$$S_{xx}(\omega) = \int_{-\infty}^{+\infty} R_{xx}(\tau) e^{-j\omega\tau} d\tau$$

$$= \int_{-\infty}^{+\infty} R_{xx}(\tau) \cos \omega \tau d\tau$$

$$= 2 \int_0^{\infty} R_{xx}(\tau) \cos \omega \tau d\tau = S_{xx}(-\omega) \geq 0$$

- The power spectrum is an even function, (in addition to being real and nonnegative).



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## IV. Stochastic processes

### 4.4. Ergodicity

- Time averages

- Given wide-sense stationary process  $X(t)$ .

- Time averages

- Mean  $n = \lim_{T \rightarrow \infty} \frac{1}{2T} \int_{-T}^T x(t) dt$

- Autocorrelation

$$R(\tau) = \lim_{T \rightarrow \infty} \frac{1}{2T} \int_{-T}^T x(t + \tau)x(t) dt$$

- These limits are random variables.

- Problems:

$$n = E\{x(t)\} \quad R(\tau) = ?$$



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## IV. Stochastic processes

### 4.4. Ergodicity

- Ergodicity

- $X(t)$  is ergodic if in the most general form if all its statistics can be determined from a single function  $X(t, \zeta)$  of the process.

- $X(t)$  is ergodic if time averages equal ensemble averages (expected values)



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## IV. Stochastic processes

### 4.4. Ergodicity

- Ergodicity of the mean

- Time average of a given process  $X(t)$

$$n_T = \frac{1}{2T} \int_{-T}^T x(t) dt$$

- $n_T$  is a random variable.

- Since  $E\{X(t)\}$  is a constant, we have

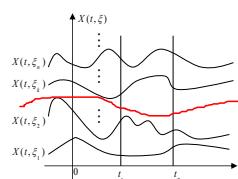
$$E\{n_T\} = E\{X(t)\} = \eta$$

- The variance of  $n_T$  is given by:

$$\sigma_{n_T}^2 = \frac{1}{T} \int_0^{2T} \left(1 - \frac{\tau}{2T}\right) [R(\tau) - \eta^2] d\tau$$

- $R(\tau)$  is the autocorrelation of  $X(t)$ .

- If this variance tends to zero with  $T \rightarrow \infty$ , then  $n_T$  tends to its expected value.



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## IV. Stochastic processes

### 4.4. Ergodicity

- Ergodic theorem for  $E\{X(t)\}$

$$\lim_{T \rightarrow \infty} \frac{1}{2T} \int_{-T}^T x(t) dt = E\{x(t)\} = \eta$$

iff

$$\lim_{T \rightarrow \infty} \frac{1}{T} \int_0^{2T} \left(1 - \frac{\tau}{2T}\right) [R(\tau) - \eta^2] d\tau = 0$$



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## IV. Stochastic processes

### 4.4. Ergodicity

- Ergodicity of autocorrelation
  - We form the average:

$$R_T(\lambda) = \frac{1}{2T} \int_{-T}^T x(t+\lambda)x(t)dt$$

- We have

$$E\{R_T(\lambda)\} = \frac{1}{2T} \int_{-T}^T E\{x(t+\lambda)x(t)\}dt = R(\lambda)$$

- For a given  $\lambda$ ,  $R_T(\lambda)$  is the time average of the process  $\Phi(t)=x(t+\lambda)x(t)$

- The mean of the process  $\Phi(t)$  is given by

$$E\{\Phi(t)\}=E\{x(t+\lambda)x(t)\} = R(\lambda)$$



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## IV. Stochastic processes

### 4.4. Ergodicity

- Its autocorrelation

$$R_{\Phi\Phi}(\tau)=E\{x(t+\lambda+\tau)x(t+\tau)x(t+\lambda)x(t)\}$$

- Hence with  $w(t) = \Phi(t)$ , we have

- Ergodicity theorem for autocorrelation

- For a given  $\lambda$

$$\lim_{T \rightarrow \infty} \frac{1}{2T} \int_{-T}^T x(t+\lambda)x(t)dt = E\{x(t+\lambda)x(t)\} = R(\lambda)$$

iff

$$\lim_{T \rightarrow \infty} \frac{1}{T} \int_0^{2T} \left(1 - \frac{\tau}{2T}\right) [R_{\Phi\Phi}(\tau) - R^2(\lambda)] d\tau = 0$$



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## IV. Stochastic processes

### 4.4. Ergodicity

- Ergodicity of the distribution function

- We determine first order distribution  $F(x) = E\{X(t) \leq x\}$  of a given process  $X(t)$  by a suitable time average.

- Consider the process

$$y(t) = \begin{cases} 1 & \text{if } x(t) \leq x \\ 0 & \text{if } x(t) > x \end{cases}$$

- Its mean is given by

$$E\{y(t)\} = 1.P\{x(t) \leq x\} = F(x)$$

- Where  $F(x, x; \tau)$  is the second order distribution of  $x(t)$

$$E\{y(t+\tau)y(t)\} = 1.P\{x(t+\tau) \leq x, x(t) \leq x\} = F(x, x; \tau)$$



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## IV. Stochastic processes

### 4.4. Ergodicity

- We form the time average

$$y_T = \frac{1}{2T} \int_{-T}^T y(t)dt$$

- We have

$$E\{y_T\} = E\{y(t)\} = F(x)$$

- The variance of  $y_T$  is given by

$$\frac{1}{T} \int_0^{2T} \left(1 - \frac{\tau}{2T}\right) [R(\tau) - \eta^2] d\tau = 0$$

- Where  $R(\tau)$  and  $\eta$  are replaced by  $F(x, x; \tau)$  and  $F(x)$



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## IV. Stochastic processes

### 4.4. Ergodicity

- Ergodic theorem for distribution function
  - For a given  $x$ ,

$$\lim_{T \rightarrow \infty} \frac{1}{2T} \int_{-T}^T y(t) dt = F(x)$$

iff

$$\lim_{T \rightarrow \infty} \frac{1}{T} \int_0^{2T} \left(1 - \frac{\tau}{2T}\right) [F(x, x; \tau) - F^2(x)] d\tau = 0$$



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## Stochastic Data Processing System Models



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## Stochastic Data Processing System Models

- Classification of Linear Systems



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## LTI systems

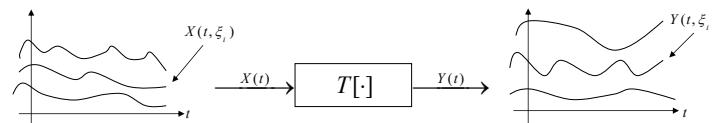
- Systems without memory
- Systems with memory
- Time Invariant systems
- Autocorrelation of output processes
- Power Spectrum and Khinchin-Wiener Theorem



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## LTI systems with stochastic input

- Deterministic system transformation



- $Y(t) = T[X(t)]$ .
- Problems formulation: goal: to study the output process statistics in term of the input process statistics and the system function
  - Is the output process stochastic ?
  - How are statistics of the output process ?
  - What is the relation between input and output processes ?



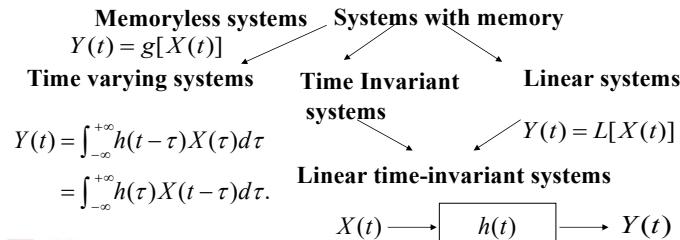
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## LTI systems with stochastic input

- Scope: Response of the deterministics systems under an actions of stochastic processes

### Deterministic systems



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## 4.1 Systems with stochastic inputs

- Memoryless systems

- Response of memoryless systems
  - $Y(t) = g[X(t)]$
  - First order distribution and density:  $F_Y(y; t)$  và  $f_Y(y; t)$  in relation with  $F_X(x; t)$ ,  $f_X(x; t)$ :

$$F_Y(y) = P(Y(\xi) \leq y) = P(g(X(\xi)) \leq y) = P(X(\xi) \leq g^{-1}(-\infty, y]).$$

- Expectation:

$$E\{Y(t)\} = \int_{-\infty}^{\infty} g(x) f_X(x; t) dx$$



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## 4.1 Systems with stochastic inputs

- Second order statistics:

- Second order density  $f_{Y_2}(y_1, y_2; t_1, t_2)$  of  $Y(t)$  can be determined in term of  $f_X(x_1, x_2; t_1, t_2)$
- Autocorrelation:  $E\{Y(t_1)Y(t_2)\}$ :

$$E\{Y(t_1)Y(t_2)\} = \int_{-\infty}^{\infty} \int_{-\infty}^{\infty} g(x_1)g(x_2) f_X(x_1, x_2; t_1, t_2) dx_1 dx_2$$

- n order statistics: n order density function of  $Y(t)$

$$f_{Y_n}(y_1, y_2, \dots, y_n; t_1, t_2, \dots, t_n)$$

$$(t_1) = g(X(t_1)), \dots, Y(t_n) = g(X(t_n)).$$



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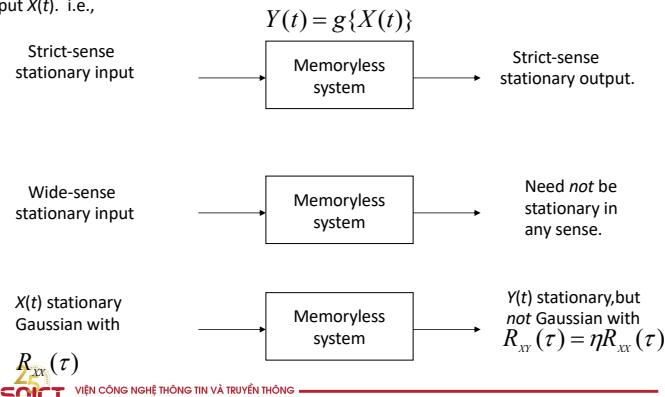
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## 4.1 Systems with stochastic inputs

### Memoryless Systems:

The output  $Y(t)$  in this case depends only on the present value of the input  $X(t)$ . i.e.,



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## 4.1 Hệ thống TTBB dưới tác dụng của quá trình ngẫu nhiên

- Khảo sát tính dừng của tín hiệu đầu ra.

- Khi tín hiệu đầu vào là tín hiệu dừng theo nghĩa hẹp, tín hiệu ra cũng dừng theo nghĩa hẹp;
- Nếu đầu vào  $X(t)$  là dừng theo bậc N, đáp ứng  $Y(t)$  cũng dừng theo bậc N;
- Nếu  $X(t)$  là dừng trong khoảng thì  $Y(t)$  cũng dừng trong khoảng đó;
- Nếu  $X(t)$  là dừng theo nghĩa rộng,  $Y(t)$  có thể không dừng theo mọi nghĩa.

- Ví dụ:

- Bộ thu nhận theo luật bình phương:  $Y(t) = X^2(t)$



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## 4.1 Systems with stochastic inputs

- Consider the memoryless system

$$g(x) = \begin{cases} 1, & x \geq 0 \\ -1, & x < 0 \end{cases}$$



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## 4.1 Systems with stochastic inputs

**Linear Systems:**  $L[\cdot]$  represents a linear system if

$$L\{a_1 X(t_1) + a_2 X(t_2)\} = a_1 L\{X(t_1)\} + a_2 L\{X(t_2)\}.$$

Let

$$Y(t) = L\{X(t)\}$$

represent the output of a linear system.

**Time-Invariant System:**  $L[\cdot]$  represents a time-invariant system if

$$Y(t) = L\{X(t)\} \Rightarrow L\{X(t-t_0)\} = Y(t-t_0)$$

i.e., shift in the input results in the same shift in the output also.

If  $L[\cdot]$  satisfies both equations, then it corresponds to

a linear time-invariant (LTI) system.

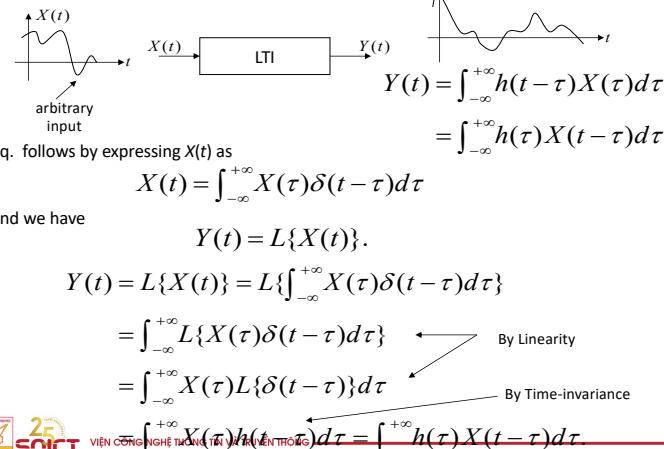
LTI systems can be uniquely represented in terms of their output to a delta function



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## 4.1 Systems with stochastic inputs



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## 4.1 Systems with stochastic inputs

- Output statistics

  - Mean of the output

$$\mu_y(t) = E\{Y(t)\} = \int_{-\infty}^{+\infty} E\{X(\tau)h(t-\tau)\} d\tau = \int_{-\infty}^{+\infty} \mu_x(\tau)h(t-\tau) d\tau = \mu_x(t) * h(t).$$

  - Cross-correlation of input-output

$$\begin{aligned} R_{xy}(t_1, t_2) &= E\{X(t_1)Y^*(t_2)\} \\ &= E\{X(t_1)\} \int_{-\infty}^{+\infty} X^*(t_2 - \alpha)h^*(\alpha) d\alpha \\ &= \int_{-\infty}^{+\infty} E\{X(t_1)X^*(t_2 - \alpha)\}h^*(\alpha) d\alpha \\ &= \int_{-\infty}^{+\infty} R_{xx}(t_1, t_2 - \alpha)h^*(\alpha) d\alpha \end{aligned}$$



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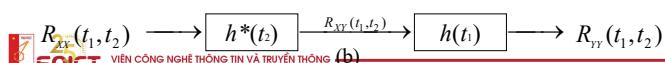
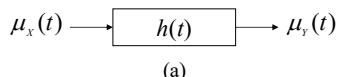
 $R_{xy}(t_1, t_2) = R_{xx}(t_1, t_2 - \alpha) * h^*(\alpha)$ 

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## 4.1 Systems with stochastic inputs

- Autocorrelation of output

$$\begin{aligned} R_{yy}(t_1, t_2) &= E\{Y(t_1)Y^*(t_2)\} \\ &= E\left\{\int_{-\infty}^{+\infty} X(t_1 - \beta)h(\beta) d\beta \int_{-\infty}^{+\infty} X^*(t_2 - \alpha)h^*(\alpha) d\alpha\right\} = \int_{-\infty}^{+\infty} E\{X(t_1 - \beta)Y^*(t_2)\}h(\beta) d\beta \\ &= \int_{-\infty}^{+\infty} R_{xy}(t_1 - \beta, t_2)h(\beta) d\beta = R_{xy}(t_1, t_2) * h(t_1) \\ R_{yy}(t_1, t_2) &= R_{xx}(t_1, t_2) * h^*(t_2) * h(t_1). \end{aligned}$$



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## 4.1 Systems with stochastic inputs

- If  $X(t)$  - wide sense stationary  $\mu_x(t) = \mu_x$

$$\mu_y(t) = \mu_x \int_{-\infty}^{+\infty} h(\tau) d\tau = \mu_x c, \quad a \text{ constant.}$$

$$R_{yy}(t_1, t_2) = R_{xx}(t_1 - t_2)$$

- $X(t), Y(t)$  are jointly w.s.s

$$\begin{aligned} R_{xy}(t_1, t_2) &= \int_{-\infty}^{+\infty} R_{xx}(t_1 - \tau + \alpha, t_2 + \alpha)h^*(\alpha) d\alpha \\ &= R_{xx}(\tau) * h^*(-\tau) = R_{xy}(\tau), \quad \tau = t_1 - t_2. \end{aligned}$$

$$\begin{aligned} R_{yy}(t_1, t_2) &= \int_{-\infty}^{+\infty} R_{xy}(t_1 - \beta - t_2)h(\beta) d\beta, \quad \tau = t_1 - t_2 \\ &= R_{xy}(\tau) * h(\tau) = R_{yy}(\tau). \\ R_{yy}(\tau) &= R_{xy}(\tau) * h^*(-\tau) * h(\tau). \end{aligned}$$



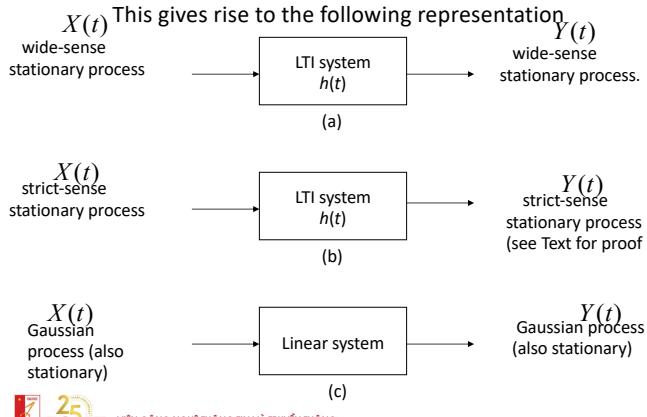
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## 4.1 Systems with stochastic inputs

The output process is also wide-sense stationary.



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## 4.1 Systems with stochastic inputs

- White noise

- $W(t)$  is said to be a white noise process if

$$R_{ww}(t_1, t_2) = q(t_1)\delta(t_1 - t_2),$$

- $E[W(t_1) W^*(t_2)] = 0$  unless  $t_1 = t_2$

- $W(t)$  is said to be wide-sense stationary (w.s.s) white noise if

- $E[W(t)] = \text{constant}$ , and

$$R_{ww}(t_1, t_2) = q\delta(t_1 - t_2) = q\delta(\tau).$$

- If  $W(t)$  is also a Gaussian process (white Gaussian process), then all of its samples are independent random variables



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## 4.1 Systems with stochastic inputs

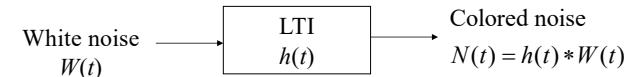
- Theorem:

- For linear systems:

$$E[L[X(t)]] = L[E[X(t)]]$$

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## 4.1 Systems with stochastic inputs



- For w.s.s. white noise input  $W(t)$ , we have

$$E[N(t)] = \mu_w \int_{-\infty}^{+\infty} h(\tau)d\tau, \quad a \text{ constant}$$

$$R_{nn}(\tau) = q\delta(\tau)*h^*(-\tau)*h(\tau) = qh^*(-\tau)*h(\tau) = q\rho(\tau)$$

- where

$$\rho(\tau) = h(\tau)*h^*(-\tau) = \int_{-\infty}^{+\infty} h(\alpha)h^*(\alpha + \tau)d\alpha.$$

- Thus the output of a white noise process through an LTI system represents a (colored) noise process.

- White noise need not be Gaussian.

- "White" and "Gaussian" are two different concepts.



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## 4.2 Discrete Time Stochastic Processes

- Definition

- A discrete time stochastic process  $X_n = X(nT)$  is a sequence of random variables.
- The mean, autocorrelation and auto-covariance functions of a discrete-time process are:

$$\mu_n = E\{X(nT)\}$$

$$R(n_1, n_2) = E\{X(n_1 T) X^*(n_2 T)\}$$

$$C(n_1, n_2) = R(n_1, n_2) - \mu_{n_1} \mu_{n_2}^*$$



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## 4.2 Discrete Time Stochastic Processes

- Tính dừng

- Strict sense stationarity and wide-sense stationarity definitions apply here also.

- $X(nT)$  is wide sense stationary if :

$$E\{X(nT)\} = \mu, \text{ a constant}$$

$$E[X\{(k+n)T\} X^*\{(k)T\}] = R(n) = r_n = r_{-n}^*$$

$$\bullet R(n_1, n_2) = R(n_1 - n_2) = R(n_2 - n_1)$$

- The positive-definite property of the autocorrelation sequence

$$\bullet T_n = \begin{pmatrix} r_0 & r_1 & r_2 & \cdots & r_n \\ r_1^* & r_0 & r_1 & \cdots & r_{n-1} \\ \vdots & & & & \\ r_n^* & r_{n-1}^* & \cdots & r_1^* & r_0 \end{pmatrix} = T_n^*, \quad n = 0, 1, 2, \dots, \infty$$



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## 4.2 Discrete Time Stochastic Processes

- Output of the LTI systems

- $X(nT)$  - W.S.S, LTI system  $h(nT)$ ,  $Y(nT)$  - response.
- Cross-correlation of input-output are:

$$R_{xy}(n) = R_{xx}(n) * h^*(-n)$$

$$R_{yy}(n) = R_{xy}(n) * h(n)$$

$$R_{rr}(n) = R_{xx}(n) * h^*(-n) * h(n).$$

- Thus wide-sense stationarity from input to output is preserved for discrete-time systems also



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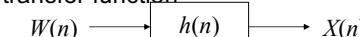
## 4.3. Auto Regressive Moving Average (ARMA) Processes

- Consider an input – output representation of LTIs

$$X(n) = -\sum_{k=1}^p a_k X(n-k) + \sum_{k=0}^q b_k W(n-k),$$

- where  $X(n)$  may be considered as the output of a system  $\{h(n)\}$  driven by the input  $W(n)$

- The transfer function



$$X(z) \sum_{k=0}^p a_k z^{-k} = W(z) \sum_{k=0}^q b_k z^{-k}, \quad a_0 \equiv 1$$

$$H(z) = \sum_{k=0}^{\infty} h(k) z^{-k} = \frac{X(z)}{W(z)} = \frac{b_0 + b_1 z^{-1} + b_2 z^{-2} + \cdots + b_q z^{-q}}{1 + a_1 z^{-1} + a_2 z^{-2} + \cdots + a_p z^{-p}} = \frac{B(z)}{A(z)}$$

$$X(n) = \sum_{k=0}^{\infty} h(n-k) W(k).$$

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### 4.3. Auto Regressive Moving Average (ARMA) Processes

- Transfer function has p poles and q zeros
- The output undergoes regression over p of its previous values and at the same time a moving average based on W(n), W(n-1) ..., W(n-q) of the input over (q + 1) values is added to it (ARMA(p, q))
- The input {W(n)} represents a sequence of uncorrelated random variables of zero mean  $\mu = 0$  and constant variance  $\sigma_w^2$  so that  $R_{ww}(n) = \sigma_w^2 \delta(n)$ .
  - If {W(n)} is normally distributed then the output {X(n)} also represents a strict-sense stationary normal process.
  - If q = 0, AR(p) – all-pole process.
  - If p = 0, represent MA(q) process



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### 4.3. Auto Regressive Moving Average (ARMA) Processes

- Tự tương quan chuẩn hóa của đầu ra:  

$$\rho_x(n) = \frac{R_{xx}(n)}{R_{xx}(0)} = a^{|n|}, \quad |n| \geq 0.$$
  - Trường hợp quá trình X(t) kết hợp với nhiễu V(t):
    - V(t) là chuỗi ngẫu nhiên không tương quan với giá trị trung bình 0 và tự tương quan  $\sigma_v^2$ .
    - $Y(t) = X(t) + V(t)$
- $$R_{yy}(n) = R_{xx}(n) + R_{vv}(n) = R_{xx}(n) + \sigma_v^2 \delta(n) = \sigma_w^2 \frac{a^{|n|}}{1-a^2} + \sigma_v^2 \delta(n)$$
- $$\rho_y(n) = \frac{R_{yy}(n)}{R_{yy}(0)} = \begin{cases} 1 & n=0 \\ c a^{|n|} & n=\pm 1, \pm 2, \dots \end{cases} \quad c = \frac{\sigma_w^2}{\sigma_w^2 + \sigma_v^2(1-a^2)} < 1.$$



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### 4.3. Auto Regressive Moving Average (ARMA) Processes

- Autoregressive process AR(1)

- An AR(1) process has the form:

$$X(n) = aX(n-1) + W(n)$$

- The corresponding system transfer :

$$H(z) = \frac{1}{1 - az^{-1}} = \sum_{n=0}^{\infty} a^n z^{-n}$$

- if  $|a| < 1$  - stable system.

- System impulse response:  $h(n) = a^n, \quad |a| < 1$

- Output autocorrelation:

$$R_{xx}(n) = \sigma_w^2 \delta(n) * \{a^{-n}\} * \{a^n\} = \sigma_w^2 \sum_{k=0}^{\infty} a^{|n|+k} a^k = \sigma_w^2 \frac{a^{|n|}}{1-a^2}$$

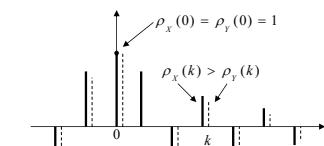


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### 4.3. Auto Regressive Moving Average (ARMA) Processes

- the effect of superimposing an error sequence on an AR(1) model AR(1)



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### 4.3. Auto Regressive Moving Average (ARMA) Processes

- Autoregressive AR(2)

$$X(n) = a_1 X(n-1) + a_2 X(n-2) + W(n)$$

- Transfer function

$$H(z) = \sum_{n=0}^{\infty} h(n)z^{-n} = \frac{1}{1 - a_1 z^{-1} - a_2 z^{-2}} = \frac{b_1}{1 - \lambda_1 z^{-1}} + \frac{b_2}{1 - \lambda_2 z^{-1}}$$

$h(0) = 1, h(1) = a_1, h(n) = a_1 h(n-1) + a_2 h(n-2), n \geq 2$

- with  $\lambda_1, \lambda_2$  - poles of the system, the impulse response:
- Và các hệ thức:  $h(n) = b_1 \lambda_1^n + b_2 \lambda_2^n, n \geq 0$

$$b_1 + b_2 = 1, \quad b_1 \lambda_1 + b_2 \lambda_2 = a_1, \quad \lambda_1 + \lambda_2 = a_1, \quad \lambda_1 \lambda_2 = -a_2,$$

- $H(z)$  stable follows  $|\lambda_1| < 1, |\lambda_2| < 1$ .



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### 4.3. Auto Regressive Moving Average (ARMA) Processes

- Autocorrelation of output process

- Autocorrelation:

$$\begin{aligned} R_{xx}(n) &= E\{X(n+m)X^*(m)\} \\ &= E\{[a_1 X(n+m-1) + a_2 X(n+m-2)]X^*(m)\} \\ &\quad + E\{W(n+m)X^*(m)\} \\ &= a_1 R_{xx}(n-1) + a_2 R_{xx}(n-2) \end{aligned}$$

- Correlation coefficient:

$$\rho_x(n) = \frac{R_{xx}(n)}{R_{xx}(0)} = a_1 \rho_x(n-1) + a_2 \rho_x(n-2).$$



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### 4.3. Auto Regressive Moving Average (ARMA) Processes

- ARIMA processes (Autoregressive Integrated moving average)

- For modeling time series

- Consider ARMA processes:  $W$  – white noise
- $$X(n) = -\sum_{k=1}^p a_k X(n-k) + \sum_{k=0}^q b_k W(n-k),$$

$$(1 + \sum_{i=1}^p D^{i-1}) X(n) = (b_0 + \sum_{k=1}^q D^k) W(n)$$

- $D^i$  – delay operator, or lag operator

- An ARIMA( $p', d, q'$ ) process is a particular case of an ARMA( $p, q$ ) process having the autoregressive polynomial with  $d$  unit roots



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### 4.3. Auto Regressive Moving Average (ARMA) Processes

- Autocorrelation

$$\begin{aligned} R_{xx}(n) &= R_{ww}(n) * h^*(-n) * h(n) = \sigma_w^2 h^*(-n) * h(n) \\ &= \sigma_w^2 \sum_{k=0}^{\infty} h^*(n+k) * h(k) \\ &= \sigma_w^2 \left( \frac{|b_1|^2 (\lambda_1^*)^n}{1 - |\lambda_1|^2} + \frac{b_1^* b_2 (\lambda_1^*)^n}{1 - \lambda_1^* \lambda_2} + \frac{b_1 b_2^* (\lambda_2^*)^n}{1 - \lambda_1 \lambda_2^*} + \frac{|b_2|^2 (\lambda_2^*)^n}{1 - |\lambda_2|^2} \right) \end{aligned}$$

$$\rho_x(n) = \frac{R_{xx}(n)}{R_{xx}(0)} = c_1 \lambda_1^{*n} + c_2 \lambda_2^{*n}$$



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### 4.3. Auto Regressive Moving Average (ARMA) Processes

- Moving average process

$$X(n) = \sum_{k=0}^q b_k W(n-k),$$

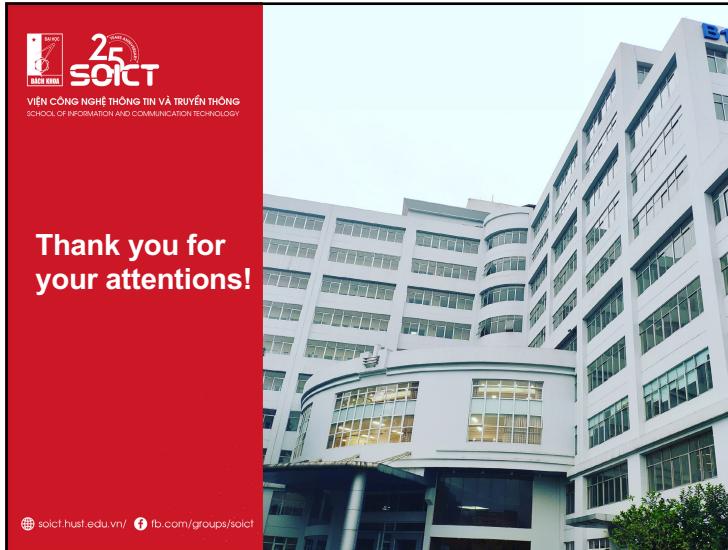
- q zeros; without poles;
- Non-regressive systems;
- Impulse response and transfer function:

$$H(z) = \sum_{k=0}^{\infty} h(k)z^{-k} = b_0 + b_1 z^{-1} + b_2 z^{-2} + \dots + b_q z^{-q}$$



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