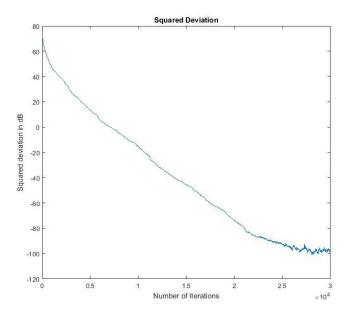
Homework 4

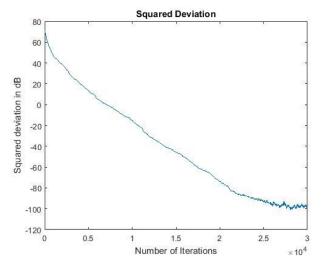
Q1. For P1.mat dataset, NLMS algorithm was applied and the resultant plot can be seen in figure 1.

With the increase in number of iterations, it is converging but the rate of convergence is slow. For NLMS convergence speed is dependant on the eigen value spread. Finally near the convergence, there is oscillations or misadjustments near weiner filter solution which is dependant on the step size.

Similarly the error seems to be fluctuating in it's path towards convergence or minimum error and keeps on fluctuating near the valley which is also dictated by the step size.

Here if stepsize is big, convergence would be fast but misadjustments would be more and vice versa.

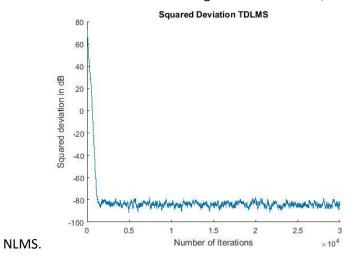


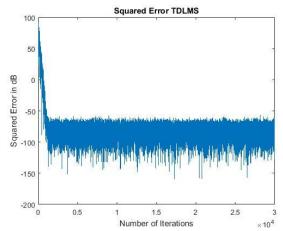


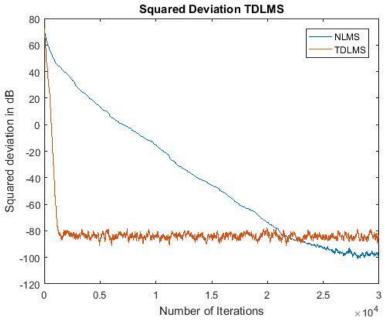
```
clear all;
close all
load('V:\EECS-844\Exam-4\P1.mat');
       %Length of each snapshot
M = 60;
K=length(x); %Length of data
N=K-M+1;
             %Number of Snapshots
X=complex(zeros(M,N));
for k=1:N
 X(:,k)=flipud(x(k:k+M-1)); %Snapshot matrix
R=1/N*(X*X'); %Correlation Matrix
P=zeros(M,1);
                %Cross Correlation Matrix
for i=M:K
 P=P+ flipud(x(i-M+1:i)).*conj(d(i));
end
P=P/N;
                   %Optimum Weiner filter
w opt=inv(R)*P;
%% NLMS
mu = 0.5;
                 %Step Size
del=0.02;
                 %Leakage Factor
w=zeros(M,1);
dev=zeros(K,1);
squared error=zeros(K,1);
for n=M:K
  error=d(n)-w'*flipud(x(n-M+1:n));
                                    %Error
  mu normalized=mu/(del+ctranspose(x(n-M+1:n))*x(n-M+1:n));
  w=w+mu normalized*conj(error)*flipud(x(n-M+1:n));
  dev(n) = (w-w opt) '* (w-w opt);
  squared error(n) = abs(error)^2;
 figure(1);plot(20*log10(dev(M:K)));title('Squared Deviation')
 xlabel('Number of Iterations');
 ylabel('Squared deviation in dB')
 figure(2);plot(20*log10(squared error(M:K)));title('Squared Error')
 xlabel('Number of Iterations');
 ylabel('Squared Error in dB')
```

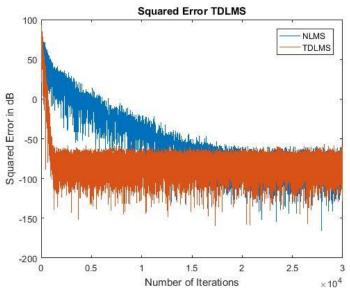
Q2. Here using P1.mat Transform domain LMS was implemented. KLT was applied to input and used for TDLMS algorithm. The squared error and squared deviation is plotted in figure 2.

Compared to NLMS the convergence speed is faster but at the cost of more error than NLMS. Because TDLMS is operated in transform domain where the signal is decorrelated, convergence is faster than









```
clear;
load('V:\EECS-844\Exam-4\P1.mat');
M=60; %Length of each snapshot
K=length(d);
N=K-M+1;
X=complex(zeros(M,N));
D=complex(zeros(M,N));
for k=1:N
 X(:,k) = flipud(x(k:k+M-1)); %Snapshot matrix
  D(:,k)=flipud(d(k:k+M-1)); %Snapshot matrix
R=1/N*(X*X');
                  %Correlation Matrix
P=zeros(M,1); %Cross Correlation Matrix
for i=M:K
 P=P+ flipud(x(i-M+1:i)).*conj(d(i));
end
P=P/N;
w opt=R\P;
              % Optimum Weiner filter
%% TDLMS
mu = 0.5/M;
del=0.02;
[Q,D]=eig(R);
eig vals=eig(R);
D=D+del*eye(size(D));
Z=Q'*X;
                %Transformed input
w optT=Q'*w opt; %Transformed Optimum weight vector
w norm=zeros(M,1);
w td=zeros(M,1);
for n=M:K
 e=d(n)-(w norm'*flipud(x(n-M+1:n))); %Error in normal domain
  z=Q'*flipud(x(n-M+1:n));
  w td=w td+mu*inv(D)*z*conj(e);
 w norm=Q*w td;
                      %Normal domain filter coeff
  squared_error(n) = (abs(e).^2); %Error in normal domain
  dev(n)=(w norm-w opt)'*(w norm-w opt); %Error in normal domain
end
figure(1);hold on;plot(20*log10(dev(M:K)))
title('Squared Deviation TDLMS');
xlabel('Number of Iterations');
ylabel('Squared deviation in dB')
figure(2);hold on; plot(20*log10(squared error(M:K)));
title('Squared Error TDLMS')
xlabel('Number of Iterations');
ylabel('Squared Error in dB')
```

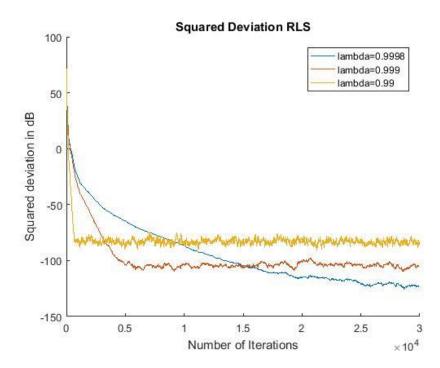
Q3. For P1.mat RLS was implemented and the squared error and squared deviation was plotted. Here RLS with lowest forgetting factor (lambda=0.99) has the fastest convergence speed and highest forgetting factor(lambda=0.9998) has the slowest convergence. But the final squared deviation is maximum for lowest forgetting factor (lambda=0.99) and minimum for highest forgetting factor (lambda=0.9998). Since lowest forgetting factor takes more past inputs into account it reaches to convergence faster.

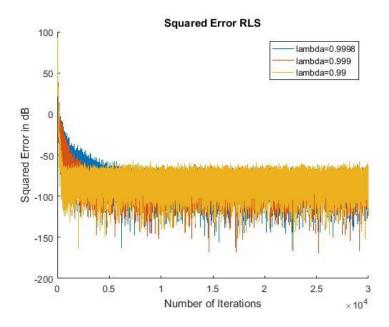
Compared to NLMS convergence is very fast.

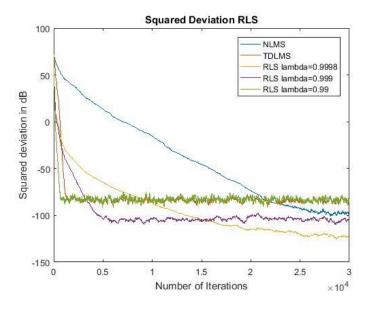
CONVERGENCE (Iterations): NLMS(~26880) vs RLS (~740, lambda=0.99) vs TDLMS(~1566)

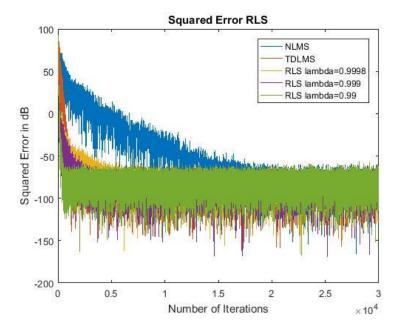
Squared deviation is less for RLS (lambda=0.9998) followed by RLS(lambda=0.999), NLMS, TDLMS, RLS(lambda=0.999).

For squared error after convergence, error remains almost comparable for all algorithms.









```
clear;
%close all
load('V:\EECS-844\Exam-4\P1.mat');
M = 60;
        %Length of each snapshot
K=length(x);
N=K-M+1;
X=complex(zeros(M,N));
for k=1:N
 X(:,k)=flipud(x(k:k+M-1)); %Snapshot matrix
end
R=1/N*X*X';
                  %Correlation Matrix
                  %Cross Correlation Matrix
P=zeros(M,1);
for i=M:K
  P=P+ flipud(x(i-M+1:i)).*conj(d(i));
end
P=P/N;
w opt=inv(R)*P;
                    %Optimum Weiner filter
%% RLS
lambda vect=[0.9998 0.999 0.99];
for lidx=1:length(lambda vect)
 lambda=lambda vect(lidx);
  P=eye(M);
w=zeros(M,1);
for n=M:K
```

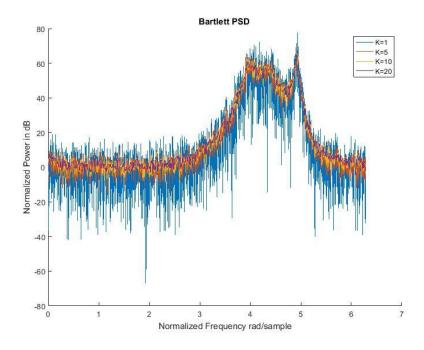
```
PI=P*(flipud(x(n-M+1:n)));
  k=PI/(lambda+(flipud(x(n-M+1:n))'*PI));
  error=d(n)-w'*(flipud(x(n-M+1:n)));
  w=w+k*conj(error);
  P=P/lambda-1/lambda*(k*(flipud(x(n-M+1:n))'*P));
  squared error(n) = (abs(error)^2);
  dev(n) = (w-w opt)'*(w-w opt);
end
figure(1); hold on; plot(20*log10(dev(M:K)));
figure(2);hold on;plot(20*log10(squared error(M:K)));
figure(1);title('Squared Deviation RLS');
xlabel('Number of Iterations');
ylabel('Squared deviation in dB')
figure(2);title('Squared Error RLS');
xlabel('Number of Iterations');
ylabel('Squared Error in dB')
figure (1); legend ('lambda=0.9998', 'lambda=0.999', 'lambda=0.99')
figure (2); legend ('lambda=0.9998', 'lambda=0.999', 'lambda=0.99')
```

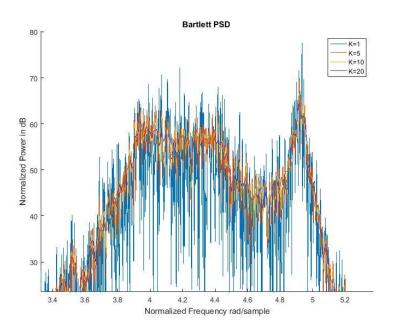
Q 4. Here for P4.mat, Bartlett PSD estimation was implemented for different segment lengths.

For K=1, it can be seen that the resolution is very high but the spectrum variance is the largest.

As we go on increasing the number of segments 'K', the number of samples in a segment decrease which degrades the resolution but it decreases the variance as averaging is done.

For K=20 the variance is least but the resolution is the worst. Also with averaging the peaks become more smooth and distinct at 3.9 rad, 4.3 rad and 4.9 rad.



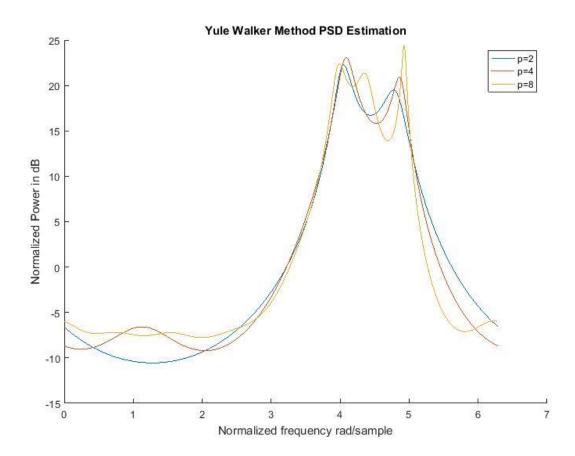


```
%Barttlett window
clear all;
close all;
load('V:\EECS-844\Exam-4\P4.mat');
K=[1 5 10 20];
for iter=1:4
 M=length(x)/K(iter); %Numner of samples for PSD calc
  Xi=reshape(x,M,K(iter)); %Division into chunks
  for div=1:size(Xi,2)
                            %Samples for PSD calculation
    Xii=Xi(:,div);
    % f=-1/2:0.001:1/2;
    f=linspace(0,1,size(x,1)/2); %Frequency
    for idx=1:length(f)
     temp=0;
      for n=0:M-1
        temp=temp+Xii(n+1)*exp((-1i)*2*pi*f(idx)*n);
      P(idx) = (abs(temp)^2)/length(Xii); %Power SPectrum for each chunk
    end
                    %Power Spectrum for all chunks
    P all(:,div)=P;
  end
  P final=1/K(iter) *sum(P all, 2); %Normalizing
  w=2*pi*f;
  figure(1);hold on;plot(w,20*log10(abs(P final)))
figure(1); title('Bartlett PSD ')
xlabel('Normalized Frequency rad/sample')
ylabel('Normalized Power in dB');
legend('K=1','K=5','K=10','K=20')
```

Q5. For P4.mat PSD was calculated by first calculating AR model coefficients and seeing the frequency response for the filter coefficients which is a parametric method. Here SCM was calculated by using snapshot matrix.

The data was modelled using different AR models of order 2, 4 and 8. AR model of 8 showed better frequency resolution as clearly 3 distinct peaks can be visible or differentiated. For p=2 it can only reveal two signals and p=4 has better resolution than p=4.

Compared to Bartlett method, the frequency response is smooth and peaks are distinct. So for properly modelled AR model, it has better PSD estimation than Bartlett. But Bartlett is better than lower order AR model.



```
%Yule Walker Method for PSD Estimation
clear all;
% close all;
load('V:\EECS-844\Exam-4\P4.mat');
pn=[2 4 8]; %AR model order
for idx=1:length(pn)
clearvars -except pn x idx
 p=pn(idx);
 M=p+1;
         %No of AR coefficients
 K=length(x);
 N=K-M+1;
  for k=1:N
   X(:,k) = flipud(x(k:k+M-1)); %Snapshot matrix
  R=1/N*X*X';
               %Sample Correlation matrix from Data
 rxx=R(2:M,1); %Autocorrelation
 Rxx=R(1:p,1:p);
  a=-inv(Rxx)*rxx;
  ar=[1;conj(a)];
                  %AR(p) model filter coefficients
 w=linspace(0,2*pi,1000);
  [H,W]=freqz(1,ar,1000,'whole');
  figure(1);hold on;plot(w,20*log10(abs(H)));
end
title('Yule Walker Method PSD Estimation');
xlabel('Normalized frequency rad/sample');
ylabel('Normalized Power in dB');
legend('p=2','p=4','p=8')
```

Q6. For different data sets X1, X2, X3 and X4, Bayesian Information Criteria was implemented using normal Sample Covariance Matrix (SCM). Which revealed following number of signals for the datasets.

```
For sample X1 no of signals is 1
For sample X2 no of signals is 1
For sample X3 no of signals is 6
For sample X4 no of signals is 6
```

Some of the assumptions made in this are that observed n different measurements are complex Gaussian distributed, the antenna arrays are Uniform Linear Arrays with no calibration errors. The signals are stationary ergodic Gaussian processes with zero mean. And SNR is greater than 15 dB and noise is white noise.

Matlab Code:

```
\mbox{\ensuremath{\$BIC}} and \mbox{\ensuremath{MUSIC}} using normal sample correlation matrix
clear all;
close all
load('V:\EECS-844\Exam-4\P6.mat');
for sample=1:4
if sample==1
  X=X1;
elseif sample==2
  X=X2;
elseif sample==3
  X=X3;
else X=X4;
end
[M,N] = size(X);
p=M; %Number of array elements n=N; %Number of time samples
R=1/n*(X*X');
                  %Correlation Matrix
eig vals=eig(R);
eig vals=sort(real(eig vals),'descend'); %Sort if not sorted
for q=1:p
  term1=sum(log(eig vals(q:p)));
  term2=(p-q+1).*log(sum(eig vals(q:p)/(p-q+1)));
  Loglq=n*(term1-term2);
  BIC(q) = -2*Loglq+((q-1)*(2*p-q+1)+1)*log(n);
% figure;plot([1:p],(BIC)); title(sprintf('BIC for sample %d ',sample))
```

```
[min val,min idx]=min(BIC);
%Number of signals
p=min idx-1;
fprintf('For sample %d no of signals is %d\n',sample,p)
%MUSIC Implementation
[Q,D]=eig(R);
eig values=real(diag(D)); %eigen values
sorted eig=(sort(eig values, 'descend')); %Sorted eigen values
if D(1) ~=sorted eig(1)
                   %Arrange Q based on decreading eigen values
  Q=fliplr(Q);
end
sampling=600; %Number of samples
phi=linspace(-pi/2,pi/2,sampling);
theta=pi*sin(phi);
for idx=1:sampling
 U=0;
  for k=p+1:M
    sv=transpose(exp((-1i)*theta(idx)*[0:M-1])); %Steering vectors
   U=U+(abs(sv'*Q(:,k)))^2;
  end
 P(idx) = 1/U;
end
figure(1); hold on; plot(theta *180/pi,20*log10(P));
figure(1); title('MUSIC Psedospectrum using Bayesian Information Criteria');
xlabel('Theta (deg)');
ylabel('Magnitude in dB')
legend('X1','X2','X3','X4')
```

Q7. Here BIC was applied to the Covariance matrix obtained from Forward Backward Averaging. Here the advantage of FB averaging is that the signal subspace is increased by a factor of 2 for coherent signals. The disadvantage of FB averaging is that It assumes ideal ULA so if there is calibration error it shows false signals.

```
For sample X1 no of signals is 2 (1 for SCM)
```

For sample X2 no of signals is 2 (1 for SCM)

For sample X3 no of signals is 6 (6 for SCM)

For sample X4 no of signals is 12 (6 for SCM)

The signals increased for all by a factor of two except X3.

```
%BIC and MUSIC using FB Averaging
clear all;
close all
load('V:\EECS-844\Exam-4\P6.mat');
for sample=1:4
if sample==1
  X=X1;
elseif sample==2
 X=X2;
elseif sample==3
 X=X3;
else X=X4;
end
[M,N] = size(X);
         %Number of array elements
p=M;
            %Number of time samples
J=flipud(eye(p));
R=(1/(2*n))*(X*X'+J*conj(X)*transpose(X)*J); %Correlation Matrix using FB
averaging
eig vals=eig(R);
eig vals=sort(real(eig vals), 'descend'); %Sort eigen values if not sorted
for q=1:p
  term1=sum(log(eig_vals(q:p)));
 term2 = (p-q+1) *log(sum(eig_vals(q:p)/(p-q+1)));
 Loglq=n*(term1-term2);
 BIC (q) = -2 * Loglq + ((q-1) * (2*p-q+1) + 1) * log(n);
end
[min val, min idx] = min(BIC);
%Number of signals
p=min idx-1;
fprintf('For sample %d no of signals is %d\n',sample,p)
%MUSIC Implementation
[Q,D]=eig(R);
eig values=real(diag(D)); %eigen values
sorted eig=(sort(eig values,'descend')); %Sorted eigen values
if D(1) ~=sorted eig(1)
                %Arrange Q based on decreasing eigen values
  Q=fliplr(Q);
end
sampling=600;
phi=linspace(-pi/2,pi/2,sampling);
theta=pi*sin(phi);
for idx=1:sampling
  U=0;
```

```
for k=p+1:M
    sv=transpose(exp((-1i)*theta(idx)*[0:M-1]));    %Steering vectors
    U=U+(abs(sv'*Q(:,k)))^2;
end
    P(idx)=1/U;
end
figure(1); hold on; plot(theta*180/pi,20*log10(P));
end
figure(1); title('MUSIC Psedospectrum using FB Averaging');
xlabel('Theta (deg)');
ylabel('Magnitude in dB')
legend('X1','X2','X3','X4')
```

Q8 Here BIC was applied to SCM calculated by spatial smoothing which uses sub arrays which increases the rank of the SCM and can reveal more signals that are coherent. The error caused by FB can be removed by Spatial Smoothing as it is less sensitive to array calibrations or mismatch. But the disadvantage is that it degrades the resolution.

	X1	X2	X3	X4
SCM	1	1	6	6
FB Averaging	2	2	6	12
	6	10	6	4
Spatial Smoothing				

Here it has revealed more signals for X1 and X2. Undid the effects of array mismatch for X4 and for X3 no effects which seems to be ideal.

```
%BIC and MUSIC Implementation using Spatial Smoothing
clear all;
close all
load('V:\EECS-844\Exam-4\P6.mat');
for sample=1:4
if sample==1
  X=X1;
elseif sample==2
 X=X2;
elseif sample==3
 X=X3;
else X=X4;
end
[M,N] = size(X);
p=M; %Number of antenna
n=N; %Number of time samples
Mbar=12; %Sub array size
K=M-Mbar+1; %Number of subarrays
Rss=zeros(Mbar, Mbar);
```

```
for subarrayidx=1:K
  Xnew=X(subarrayidx:subarrayidx+Mbar-1,:);
  Rss=Rss+Xnew*Xnew';
                            %Correlation matrix using Spatial Smoothing
end
Rss=Rss/(n*K);
eig vals=eig(Rss);
eig vals=sort(real(eig vals), 'descend'); %Sort eig vals if not sorted
p=Mbar;
for q=1:p
 term1=sum(log(eig vals(q:p)));
  term2=(p-q+1)*log(sum(eig vals(q:p)/(p-q+1)));
  Loglq=n*(term1-term2);
 BIC(q) = -2*Loglq+((q-1)*(2*p-q+1)+1)*log(n);
%figure;plot([1:p],(BIC))
[min_val,min_idx]=min(BIC);
%Number of signals
p=min idx-1;
fprintf('For sample %d no of signals is %d\n', sample, p)
%MUSIC Implementation
[Q,D] = eig(Rss);
eig values=real(diag(D)); %eigen values
sorted eig=(sort(eig values, 'descend')); %Sorted eigen values
if D(1) ~=sorted eig(1)
 Q=fliplr(Q);
                  %Arrange Q based on decreasing eigen values
end
sampling=600;
phi=linspace(-pi/2,pi/2,sampling);
theta=pi*sin(phi);
for idx=1:sampling
  U=0;
  for k=p+1:Mbar
    sv=transpose(exp((-1i)*theta(idx)*linspace(0,Mbar-1,Mbar))); %Steering
vectors
   U=U+(abs(sv'*Q(:,k)))^2;
  end
  P(idx) = 1/U;
figure(1); hold on; plot(theta*180/pi,20*log10(abs(P)));
figure(1); title('MUSIC Psedospectrum using Spatial Smoothing');
xlabel('Theta (deg)');
ylabel('Magnitude in dB')
legend('X1','X2','X3','X4')
```

Q9.

Here as obtained from before the number of signals estimated are:

	X1	X2	X3	X4
SCM	1	1	6	6
FB Averaging	2	2	6	12
	6	10	6	4
Spatial Smoothing				

Here X3 has 6 signals for all so it seems to be the ideal case and it has no calibration errors and the signals are uncorrelated. For X4 there are 6 signals shown by SCM but doubled by FB averaging and SS showed only 4 signals so it must have some calibration error as SS is less prone to calibration error but FB averaging is. So X4 is realistic array but the signals are uncorrelated.

For X1 there are 1, 2 and 6 signals so this seems to be the temporally correlated case and it doesn't seem to have calibration errors. X2 has some calibration errors as it shows more signals than there is which is 6.

Ideal Array & Temporally Uncorrelated	X3
Ideal Array & Temporally Correlated	X1
Realistic Array & Temporally Uncorrelated	X4
Realistic Array & Temporally Correlated	X2

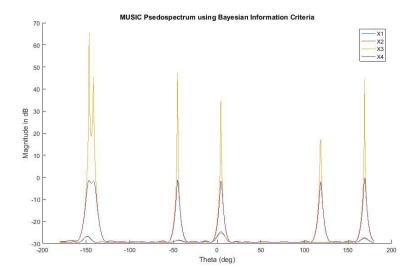
Q10. Using all methods MUSIC was implemented. Here in all methods SCM, FB and SS 6 signals are distint in all cases for X3 which is the ideal array. Also 6 signals can be seen for X4 but for X1 and X2, the signal peaks are not distinct or prominent. X2 has calibration errors so it is showing more peaks in MUSIC.

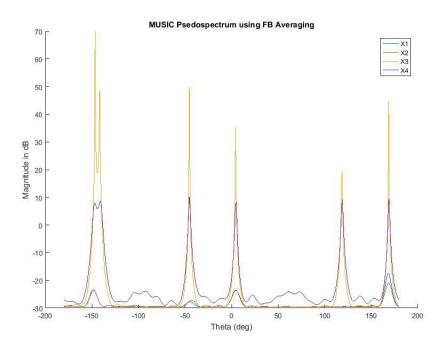
For SCM, X1 shows more 4 peaks same for X2, X3 has 6 and X4 has 6.

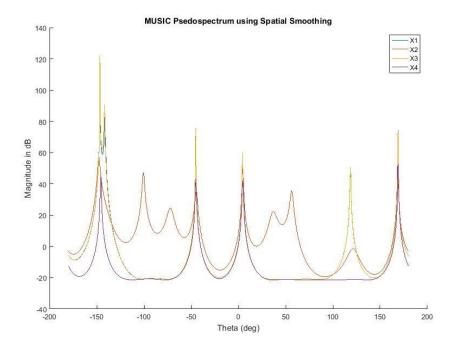
For fb: X1 has 4 peaks, so has X2, X3 and X4 has 6

For SS: X1 has 6, X2 has 9 peaks, X3 has 6 and X4 has 4 peaks.

Here so ideal arrays like X3 has better performance in MUSIC as MUSIC is affected by model mismatch and array calibration and temporal correlation.







MATLAB Code:

Implemented in 6, 7, 8