## Problem S11 (Signals and Systems) Solution

1. From the problem statement,

$$\omega_n = \sqrt{2} \frac{9.82 \text{ m/s}^2}{129 \text{ m/s}} = 0.1077 \text{ r/s}$$

$$\zeta = \frac{1}{\sqrt{2}(L_0/D_0)} = \frac{1}{\sqrt{2} \cdot 15} = 0.0471$$

Therefore,

$$\bar{G}(s) = \frac{1}{s(s^2 + 0.01015s + 0.0116)}$$

The roots of the denominator are at s = 0, and

$$s = \frac{-0.01915 \pm \sqrt{0.01015^2 - 4 \cdot 0.0116}}{2}$$
$$= -0.005075 \pm 0.1075 j$$

So

$$\bar{G}(s) = \frac{1}{s\left(s - \left[-0.005075 + 0.1075j\right]\right)\left(s - \left[-0.005075 - 0.1075j\right]\right)}$$

Use the coverup method to obtain the partial fraction expansion

$$\bar{G}(s) = \frac{86.283}{s} + \frac{-43.142 + 2.036j}{s - [-0.005075 + 0.1075j]} + \frac{-43.142 - 2.036j}{s - [-0.005075 - 0.1075j]}$$

Taking the inverse Laplace transform (assuming that  $\bar{q}(t)$  is causal), we have

$$\bar{g}(t) = 86.283\sigma(t)$$

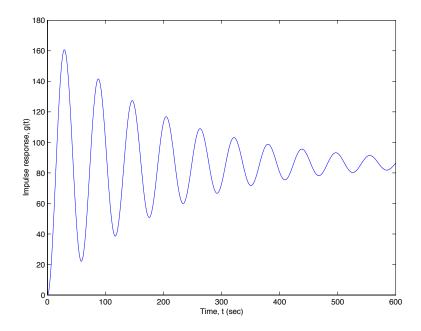
$$+ (-43.142 + 2.036j)e^{(-0.005075 + 0.1075j)t}$$

$$+ (-43.142 - 2.036j)e^{(-0.005075 - 0.1075j)t}$$

Therefore,

$$\bar{g}(t) = \sigma(t) \left[ 86.283 + 2e^{-0.005075t} \left( -43.142 \cos \omega_d t - 2.036 \sin \omega_d t \right) \right]$$
$$= \sigma(t) \left[ 86.283 + \left( -86.284 \cos \omega_d t - 4.072 \sin \omega_d t \right) e^{-0.005075t} \right]$$

where  $\omega_d = 0.1075$  r/s. See below for the impulse response.



## 2. From the problem statement,

$$\frac{H(s)}{R(s)} = \frac{k\bar{G}(s)}{1 + k\bar{G}(s)}$$

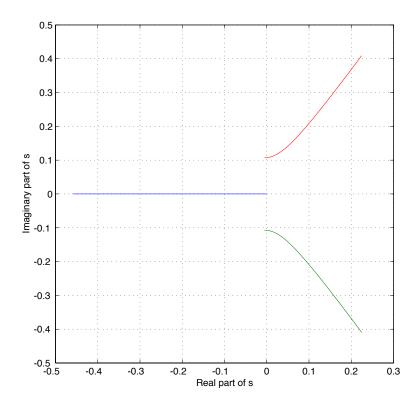
$$= \frac{k \frac{1}{s(s^2 + 2\zeta\omega_n s + \omega_n^2)}}{1 + k \frac{1}{s(s^2 + 2\zeta\omega_n s + \omega_n^2)}}$$

$$= \frac{k}{s^3 + 2\zeta\omega_n s^2 + \omega_n^2 s + k}$$

So the poles of the system are the roots of the denominator polynomial,

$$\phi(s) = s^3 + 2\zeta\omega_n s^2 + \omega_n^2 s + k = 0$$

The roots can be found using Matlab, a programmable calculator, etc. The plot of the roots (the "root locus") is shown below. Note that the oscillatory poles go unstable at a gain of only k = 0.000118.



3. The roots locus for negative gains can be plotted in a similar way, as below. Note that the real pole is unstable for all negative k.

