#### Home Work 10

63. end Recursive Binary Search;

The problems in this problem set cover lectures C11 and C12

1. Define a recursive binary search algorithm. a. If lb > ubReturn -1 else Mid := (lb+ub)/2If Array(Mid) = elementReturn Mid Elsif Array(Mid) < Element Return Binary Search(Array, mid+1, ub, Element) Else Return Binary Search(Array, lb, mid-1, Element) End if End if Implement your algorithm as an Ada95 program. 46. function Binary\_Search (My\_Search\_Array : My\_Array; Lb : Integer; Ub: Integer; Element : Integer) return Integer is 47. mid: integer; 48. begin 49. if (Lb> Ub) then 50. return -1; 51. else 52. Mid := (Ub+Lb)/2;53. if My\_Search\_Array(Mid) = Element then return(Mid); 54. 55. elsif My Search Array(Mid) < Element then return (Binary Search(My Search Array, Mid+1, Ub, Element)); 57. else 58. return (Binary\_Search(My\_Search\_Array, Lb, Mid-1, Element)); 59. end if; 60. end if; 62. end Binary Search;

c. What is the recurrence equation that represents the computation time of your algorithm?

```
Recursive Binary Search
                                                                                            Cost
if (Lb> Ub) then
                                                                                            c1
    return -1;
                                                                                            c2
else
                                                                                            c3
    Mid := (Ub+Lb)/2;
                                                                                            c4
    if My Search Array(Mid) = Element then
                                                                                            c5
        return(Mid);
                                                                                            c6
    elsif My Search Array(Mid) < Element then
                                                                                            c7
        return (Binary Search (My Search Array, Mid+1, Ub, Element));
                                                                                            T(n/2)
                                                                                    c8
        return (Binary Search (My Search Array, Lb, Mid-1, Element));
                                                                                            T(n/2)
                                                                                            c9
end if;
                                                                                            c10
```

In this case, only one of the recursive calls is made, hence only one of the T(n/2) terms is included in the final cost computation.

Therefore 
$$T(n)$$
 =  $(c1+c2+c3+c4+c5+c6+c7+c8+c9+c10) + T(n/2)$   
=  $T(n/2) + C$ 

d. What is the Big-O complexity of your algorithm? Show all the steps in the computation based on your algorithm.

- 2. What is the Big-O complexity of:
- a. Heapify function

T(n) =

A heap is an array that satisfies the heap properties i.e.,  $A(i) \le A(2i)$  and  $A(i) \le A(2i+1)$ .

The heapify function at 'i' makes A(i .. n) satisfy the heap property, under the assumption that the subtrees at A(2i) and A(2i+1) already satisfy the heap property.

Heapify function	Cost
Lchild := Left(I);	c1
Rchild := Right(I);	c2
if (Lchild <= Heap_Size and Heap_Array(Lchild) > Heap_Array(I))	<b>c</b> 3
Largest:= Lchild;	c4
else	c5
Largest := I;	c6
if (Rchild <= Heap Size)	c7
if Heap_Array(Rchild) > Heap_Array(Largest)	c8
Largest := Rchild;	c9
if (Largest /= I) then	c10
Swap(Heap_Array, I, Largest);	c11
Heapify(Heap_Array, Largest);	T(2n/3)

$$T(n) = T(2n/3) + C'$$
  
=  $T(2n/3) + O(1)$ 

a = 1, b = 3/2, f(n) = 1, therefore by master theorem,

$$T(n) = O\left(n^{\log_b a} \log n\right)$$

$$= O\left(n^{\log_{3/2} 1} \log n\right)$$

$$= O(1 * \log n)$$

$$= O(\log n)$$

The important point to note here is the T(2n/3) term, which arises in the worst case, when the heap is asymmetric, i.e., the right subtree has one level less than the left subtree (or vice-versa).

## b. Build\_Heap function

Code	Cost t(n)
Heap_Size := Size;	c1
for I in reverse 1 (Size/2) loop	n/2+1
Heapify(Heap_Array, I);	$(n/2) \log n$
end loop;	n/2

Therefore 
$$T(n) = c1 + n/2 + 1 + (n/2)\log n + n/2$$
  
=  $(n\log(n))/2 + n + (c1+1)$ 

```
Simplifying
=> T(n) = O(n log(n))

c. Heap_Sort

Heap Sort

Build_Heap(Heap_Array, Size);
for I in reverse 2.. size loop
   Swap(Heap_Array, 1, I);
```

Heap\_Size:= Heap\_Size -1;

Heapify(Heap Array, 1);

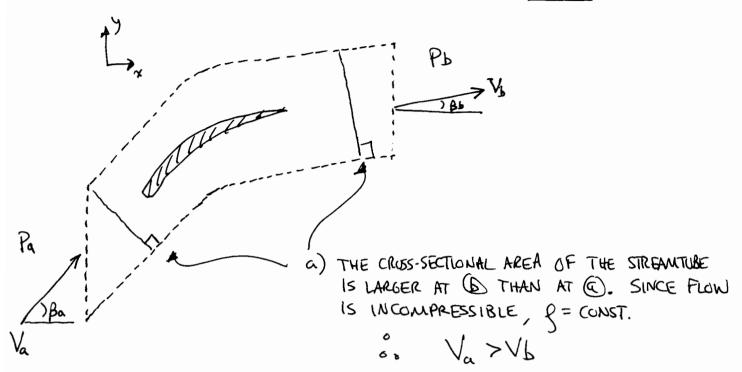
$$T(n) = 2 O(n\log n) + (c1+c2+1)n - O(\log n) +$$
= 2 O(nlog n) - O(log n) + c'n

Simplifying,

$$\Rightarrow$$
  $T(n) = O(nlogn)$ 

# Cost t(n)

O(nlogn))
n
c1(n-1)
c2(n-1)
O(log n)(n-1)



b) STEADY FLOW, NO ACCEL OF C.V.

NOTES: a) BY SYMMETRY PRESSURE FORCES ON UPPER AND LUWER STREAMSURFACES WILL BALLINGE .. ONLY NEED TO CONSIDER PRESSURE FORCES ON LEFT AND RIGHT SURFACES OF C.V.

> b) SINCE UPPER AND LOWER SURFACES ARE STREAMLINES (EVERYWHERE PARALLEL TO FLOW) THERE U NO FLUX ACROSS THEM. NEED ONLY CONSIDER FLUX TERMS ON LEFT AND RIGHT SURFACES OF C.V.

MASS FLOW THROUGH
SURFACE AT (a) = g Va COSBaS

MUST BE EQUAL
SURFACE AT (b) = g Vb COSBbS

AUST BE EQUAL
SURFACE AT (b) = g Vb COSBbS

FLUXES IN OR
OUT OF C.V.

No y-force on left ; right surfaces & bottom surfaces on top & bottom

.. FORCE ON BLADE IS IN + Y-DIRECTION

UNIFIED PROPULSION P4 SOLUTIONS

a) 
$$V_{\text{max range}} = \left[4\left(\frac{V}{S}\right)^2 \frac{1}{9^2} \frac{1}{C_{Do}} \left(\frac{1}{\text{TRR}}\right)\right]^{1/4}$$

$$\frac{W}{S} \approx \frac{4807}{54^{3}} - 144\frac{N}{m^{2}}$$

$$C_{Do} \approx 0.3$$

$$(per Gleman)$$

$$Q \approx 0.96$$

$$PER$$

$$Cou Man$$

$$V \approx 1507 = 42N$$

Vmax runge ≈ 3.3 m/s

b) 
$$V_{\text{max endurance}} = 3^{-1/4} (V_{\text{max range}}) = 2.5 \text{m/s}$$
  
(= min power)

C) ASSUME THE AIRPLANE IS BEING FLOWN AT MAX ENDURANCE CONDITIONS

DMINPOWER = 
$$W \left[ \frac{16}{3} \frac{Co}{\pi e R} \right]^{1/2} = 1.34 N$$

: ENERGY = TIME (POWER REQD) 
$$\times 1 = 30.6 \text{ kJ}$$

NOTE: PROF. COLEMAN CALCULATED E= 18.1 KT 50 PERHAPS NO IS BETTER THAN O.1!

$$\begin{pmatrix}
600 \text{ mA-hr} = 2150 \text{ A-s} & 8.4 \text{ Volts} = \frac{W}{A} \\
E = 8.4.2150 \text{ W·s} = \frac{1}{5}.5 = J = 18144
\end{pmatrix}$$

# Problem S10 (Signals and Systems) Solution

1. Because the numerator is the same order as the denominator, the partial fraction expansion will have a constant term:

$$G(s) = \frac{3s^2 + 3s - 10}{s^2 - 4}$$
$$= \frac{3s^2 + 3s - 10}{(s - 2)(s + 2)}$$
$$= a + \frac{b}{s - 2} + \frac{c}{s + 2}$$

To find a, b, and c, use coverup method:

$$a = G(s)|_{s=\infty} = 3$$

$$b = \frac{3s^2 + 3s - 10}{s + 2}\Big|_{s=2} = 2$$

$$c = \frac{3s^2 + 3s - 10}{s - 2}\Big|_{s=-2} = 1$$

So

$$G(s) = 3 + \frac{2}{s-2} + \frac{1}{s+2}, \quad \text{Re}[s] > 2$$

We can take the inverse LT by simple pattern matching. The result is that

$$g(t) = 3\delta(t) + \left(2e^{2t} + e^{-2t}\right)\sigma(t)$$

2.

$$G(s) = \frac{6s^2 + 26s + 26}{(s+1)(s+2)(s+3)}$$
$$= \frac{a}{s+1} + \frac{b}{s+2} + \frac{c}{s+3}$$

Using partial fraction expansions,

$$a = \frac{6s^2 + 26s + 26}{(s+2)(s+3)}\Big|_{s=-1} = 3$$

$$b = \frac{6s^2 + 26s + 26}{(s+1)(s+3)}\Big|_{s=-2} = 2$$

$$c = \frac{6s^2 + 26s + 26}{(s+1)(s+2)}\Big|_{s=-3} = 1$$

$$G(s) = \frac{3}{s+1} + \frac{2}{s+2} + \frac{1}{s+3}, \quad \text{Re}[s] > -1$$

The inverse LT is given by

$$(3e^{-t} + 2e^{-2t} + e^{-3t}) \sigma(t)$$

3. This one is a little tricky — there is a second order pole at s = -1. So the partial fraction expansion is

$$G(s) = \frac{4s^2 + 11s + 9}{(s+1)^2(s+2)} = \frac{a}{s+1} + \frac{b}{(s+1)^2} + \frac{c}{s+2}$$

We can find b and c by the coverup method:

$$b = \frac{4s^2 + 11s + 9}{s + 2} \Big|_{s = -1} = 2$$

$$c = \frac{4s^2 + 11s + 9}{(s + 1)^2} \Big|_{s = -2} = 3$$

So

$$G(s) = \frac{a}{s+1} + \frac{2}{(s+1)^2} + \frac{3}{s+2}$$

To find a, pick a value of s, and plug into the equation above. The easiest value to pick is s=0. Then

$$G(0) = \frac{a}{1} + \frac{2}{(1)^2} + \frac{3}{2} = \frac{9}{2}$$

Solving, we have

$$a = 1$$

Therefore,

$$G(s) = \frac{1}{s+1} + \frac{2}{(s+1)^2} + \frac{3}{s+2}, \quad \text{Re}[s] > -1$$

The inverse LT is then

$$g(t) = (e^{-t} + 2te^{-t} + 3e^{-2t}) \sigma(t)$$

4. This problem is similar to above. The partial fraction expansion is

$$G(s) = \frac{4s^3 + 11s^2 + 5s + 2}{s^2(s+1)^2} = \frac{a}{s} + \frac{b}{s^2} + \frac{c}{s+1} + \frac{d}{(s+1)^2}$$

We can find b and d by the coverup method

$$b = \frac{4s^3 + 11s^2 + 5s + 2}{(s+1)^2} \Big|_{s=0} = 2$$

$$d = \frac{4s^3 + 11s^2 + 5s + 2}{s^2} \Big|_{s=-1} = 4$$

$$G(s) = \frac{4s^3 + 11s^2 + 5s + 2}{s^2(s+1)^2} = \frac{a}{s} + \frac{2}{s^2} + \frac{c}{s+1} + \frac{4}{(s+1)^2}$$

To find a and c, pick two values of s, say, s = 1 and s = 2. Then

$$G(1) = \frac{4+11+5+2}{1^2(1+1)^2} = \frac{a}{1} + \frac{2}{1^2} + \frac{c}{1+1} + \frac{4}{(1+1)^2}$$

$$G(2) = \frac{4 \cdot 2^3 + 11 \cdot 2^2 + 5 \cdot 2 + 2}{2^2(2+1)^2} = \frac{a}{2} + \frac{2}{2^2} + \frac{c}{2+1} + \frac{4}{(2+1)^2}$$

Simplifying, we have that

$$a + \frac{c}{2} = \frac{5}{2}$$
$$\frac{a}{2} + \frac{c}{3} = \frac{3}{2}$$

Solving for a and c, we have that

$$a = 1$$
 $c = 3$ 

$$G(s) = \frac{1}{s} + \frac{2}{s^2} + \frac{3}{s+1} + \frac{4}{(s+1)^2}$$

and

$$g(t) = (1 + 2t + 3e^{-t} + 4te^{-t}) \sigma(t)$$

### 5. G(s) can be expanded as

$$G(s) = \frac{s^3 + 3s^2 + 9s + 12}{(s^2 + 4)(s^2 + 9)}$$

$$= \frac{s^3 + 3s^2 + 9s + 12}{(s + 2j)(s - 2j)(s + 3j)(s - 3j)}$$

$$= \frac{a}{s + 2j} + \frac{b}{s - 2j} + \frac{c}{s + 3j} + \frac{d}{s - 3j}$$

The coefficients can be found by the coverup method:

$$a = \frac{s^3 + 3s^2 + 9s + 12}{(s - 2j)(s + 3j)(s - 3j)} \Big|_{s = -2j} = 0.5$$

$$b = \frac{s^3 + 3s^2 + 9s + 12}{(s + 2j)(s + 3j)(s - 3j)} \Big|_{s = +2j} = 0.5$$

$$c = \frac{s^3 + 3s^2 + 9s + 12}{(s + 2j)(s - 2j)(s - 3j)} \Big|_{s = -3j} = 0.5j$$

$$d = \frac{s^3 + 3s^2 + 9s + 12}{(s + 2j)(s - 2j)(s + 3j)} \Big|_{s = -3j} = -0.5j$$

Therefore

$$G(s) = \frac{0.5}{s+2j} + \frac{0.5}{s-2j} + \frac{0.5j}{s+3j} + \frac{-0.5j}{s-3j}, \qquad \text{Re}[s] > 0$$

and the inverse LT is

$$g(t) = 0.5 \left( e^{-2jt} + e^{2jt} + je^{-3jt} - je^{3jt} \right) \sigma(t)$$

This can be expanded using Euler's formula, which states that

$$e^{ajt} = \cos at + j\sin at$$

Applying Euler's formula yields

$$g(t) = (\cos 2t + \sin 2t) \, \sigma(t)$$

### Problem S11 (Signals and Systems) Solution

1. From the problem statement,

$$\omega_n = \sqrt{2} \frac{9.82 \text{ m/s}^2}{129 \text{ m/s}} = 0.1077 \text{ r/s}$$

$$\zeta = \frac{1}{\sqrt{2}(L_0/D_0)} = \frac{1}{\sqrt{2} \cdot 15} = 0.0471$$

Therefore,

$$\bar{G}(s) = \frac{1}{s(s^2 + 0.01015s + 0.0116)}$$

The roots of the denominator are at s = 0, and

$$s = \frac{-0.01915 \pm \sqrt{0.01015^2 - 4 \cdot 0.0116}}{2}$$
$$= -0.005075 \pm 0.1075j$$

So

$$\bar{G}(s) = \frac{1}{s\left(s - \left[-0.005075 + 0.1075j\right]\right)\left(s - \left[-0.005075 - 0.1075j\right]\right)}$$

Use the coverup method to obtain the partial fraction expansion

$$\bar{G}(s) = \frac{86.283}{s} + \frac{-43.142 + 2.036j}{s - [-0.005075 + 0.1075j]} + \frac{-43.142 - 2.036j}{s - [-0.005075 - 0.1075j]}$$

Taking the inverse Laplace transform (assuming that  $\bar{q}(t)$  is causal), we have

$$\bar{g}(t) = 86.283\sigma(t)$$

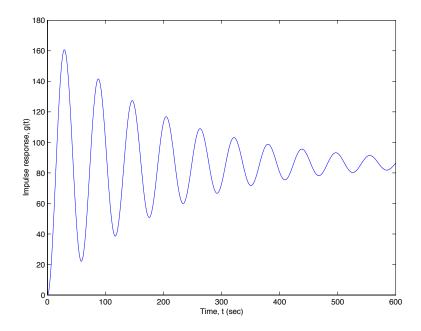
$$+ (-43.142 + 2.036j)e^{(-0.005075 + 0.1075j)t}$$

$$+ (-43.142 - 2.036j)e^{(-0.005075 - 0.1075j)t}$$

Therefore,

$$\bar{g}(t) = \sigma(t) \left[ 86.283 + 2e^{-0.005075t} \left( -43.142 \cos \omega_d t - 2.036 \sin \omega_d t \right) \right]$$
$$= \sigma(t) \left[ 86.283 + \left( -86.284 \cos \omega_d t - 4.072 \sin \omega_d t \right) e^{-0.005075t} \right]$$

where  $\omega_d = 0.1075$  r/s. See below for the impulse response.



#### 2. From the problem statement,

$$\frac{H(s)}{R(s)} = \frac{k\bar{G}(s)}{1 + k\bar{G}(s)}$$

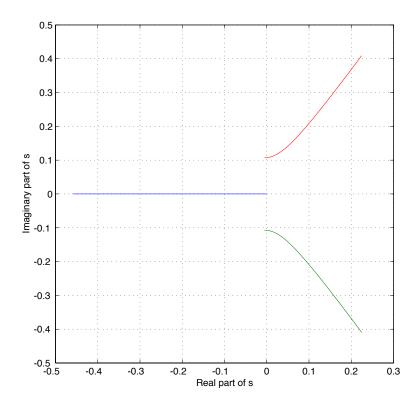
$$= \frac{k \frac{1}{s(s^2 + 2\zeta\omega_n s + \omega_n^2)}}{1 + k \frac{1}{s(s^2 + 2\zeta\omega_n s + \omega_n^2)}}$$

$$= \frac{k}{s^3 + 2\zeta\omega_n s^2 + \omega^2 s + k}$$

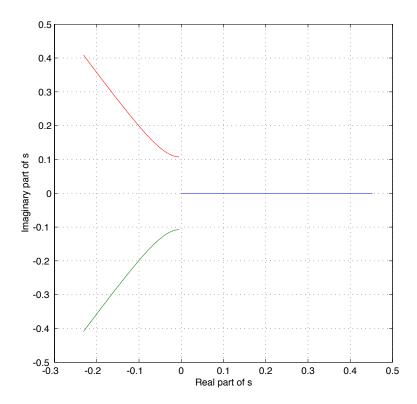
So the poles of the system are the roots of the denominator polynomial,

$$\phi(s) = s^3 + 2\zeta\omega_n s^2 + \omega_n^2 s + k = 0$$

The roots can be found using Matlab, a programmable calculator, etc. The plot of the roots (the "root locus") is shown below. Note that the oscillatory poles go unstable at a gain of only k = 0.000118.



3. The roots locus for negative gains can be plotted in a similar way, as below. Note that the real pole is unstable for all negative k.



### Problem S12 (Signals and Systems) Solution

For each signal below, find the bilateral Laplace transform (including the region of convergence) by directly evaluating the Laplace transform integral. If the signal does not have a transform, say so.

1.

$$g(t) = \sin(at)\sigma(-t)$$

To do this problem, expand the sinusoid as complex exponentials, so that

$$g(t) = \left[\frac{e^{ajt} - e^{-ajt}}{2j}\right] \sigma(-t)$$

Therefore, the LT is given by

$$G(s) = \int_{-\infty}^{0} \left[ \frac{e^{ajt} - e^{-ajt}}{2j} \right] e^{-st} dt$$

For the LT to converge, the integrand must go to zero as t goes to  $-\infty$ . Therefore, the integral converges only for Re[s] < 0. The integral is then

$$G(s) \int_{-\infty}^{0} \left[ \frac{e^{ajt} - e^{-ajt}}{2j} \right] e^{-st} dt$$

$$= \frac{1}{2j} \left[ \frac{1}{-s + aj} e^{(aj - s)t} \Big|_{-\infty}^{0} - \frac{1}{-s - aj} e^{(-aj - s)t} \Big|_{-\infty}^{0} \right]$$

$$= \frac{1}{2j} \left[ \frac{1}{-s + aj} - \frac{1}{-s - aj} \right]$$

$$= \frac{-a}{s^2 + a^2}, \quad \text{Re}[s] < 0$$

2.

$$g(t) = te^{at}\sigma(-t)$$

The LT is given by

$$G(s) = \int_{-\infty}^{0} t e^{at} e^{-st} dt = \int_{-\infty}^{0} t e^{(a-s)t} dt$$

For the LT to converge, the integrand must go to zero as t goes to  $-\infty$ . Therefore,

the integral converges only for Re[s] < a. To find the integral, integrate by parts:

$$G(s) = \int_{-\infty}^{0} t e^{(a-s)t} dt$$

$$= t \frac{1}{a-s} e^{(a-s)t} \Big|_{-\infty}^{0} - \frac{1}{a-s} \int_{-\infty}^{0} e^{(a-s)t} dt$$

$$= 0 - \frac{1}{a-s} \int_{-\infty}^{0} e^{(a-s)t} dt$$

$$= -\frac{1}{(a-s)^{2}} e^{(a-s)t} \Big|_{-\infty}^{0}$$

$$= -\frac{1}{(s-a)^{2}}, \quad \text{Re}[s] < a$$

3.

$$g(t) = \cos(\omega_0 t) e^{-a|t|}, \quad \text{for all } t$$

The LT is given by

$$G(s) = \int_{-\infty}^{\infty} \cos(\omega_0 t) e^{-a|t|} e^{-st} dt$$

For the LT to converge, the integrand must go to zero as t goes to  $-\infty$  and  $\infty$ . Therefore, the integral converges only for -a < Re[s] < a. The integral is given by

$$G(s) = \int_{-\infty}^{\infty} \cos(\omega_0 t) e^{-a|t|} e^{-st} dt$$
$$= \int_{-\infty}^{0} \cos(\omega_0 t) e^{at} e^{-st} dt + \int_{0}^{\infty} \cos(\omega_0 t) e^{-at} e^{-st} dt$$

Expanding the cosine term as

$$\cos(\omega_0 t) = \frac{e^{j\omega_0 t} + e^{-j\omega_0 t}}{2}$$

yields

$$G(s) = \int_{-\infty}^{0} \frac{e^{j\omega_0 t} + e^{-j\omega_0 t}}{2} e^{at} e^{-st} dt + \int_{0}^{\infty} \frac{e^{j\omega_0 t} + e^{-j\omega_0 t}}{2} e^{-at} e^{-st} dt$$

$$= \int_{-\infty}^{0} \frac{e^{(j\omega_0 + a - s)t} + e^{(-j\omega_0 + a - s)t}}{2} dt + \int_{0}^{\infty} \frac{e^{(j\omega_0 - a - s)t} + e^{(-j\omega_0 - a - s)t}}{2} dt$$

$$= \frac{1}{2} \left[ \frac{1}{j\omega_0 + a - s} + \frac{1}{-j\omega_0 + a - s} - \frac{1}{j\omega_0 - a - s} - \frac{1}{-j\omega_0 - a - s} \right]$$

$$= \frac{s + a}{s^2 + 2as + a^2 + \omega_0^2} - \frac{s - a}{s^2 - 2as + a^2 + \omega_0^2}, \quad -a < \text{Re}[s] < a$$