

# **Report**

## **High Speed Piezo Driver**

### **V6 26/08/2015**

#### **Aim**

Feasibility study of a high speed piezo driver

#### **Requirements**

Driving a piezo brick up to 500 kHz with a full voltage range [0, 150] V.

#### **Design Data**

Capacitive Load (piezo brick) = 170nF

Required Slew rate (sine wave) =  $2\pi \cdot f \cdot V_{\text{peak}} = 2\pi \cdot (500\text{e3}) \cdot 75 \cdot (1\text{e-6})$   
 $\text{V}/\mu\text{s} = \sim 235 \text{ V}/\mu\text{s}$

Required Peak Current =  $C \cdot dV/dt = 170 \text{ nF} \cdot 235 \text{ V}/\mu\text{s} = \sim 40 \text{ A}$

#### **Possible Candidate**

Power Operational Amplifier Apex MP111

#### **FEATURES**

- LOW COST
- HIGH VOLTAGE - 100 VOLTS
- HIGH OUTPUT CURRENT- 50 AMP PULSE  
OUTPUT, 15 AMP CONTINUOUS
- 170 WATT DISSIPATION CAPABILITY
- 130 V/ $\mu$ S SLEW RATE
- 500kHz POWER BANDWIDTH

#### **APPLICATIONS**

- INKJET PRINTER HEAD DRIVE
- PIEZO TRANSDUCER DRIVE
- INDUSTRIAL INSTRUMENTATION
- REFLECTOMETERS
- ULTRA-SOUND TRANSDUCER DRIVE

## Main Issues

- Voltage peak
- Slew Rate

## Possible Solution

From Apex AN 20:

### 5.2 UNIPOLAR OUTPUT

A particularly powerful way of applying the bridge is in the unipolar bridge. By unipolar, we mean that the output can only swing from 0 to one polarity. Figure 5 is used to illustrate this technique.

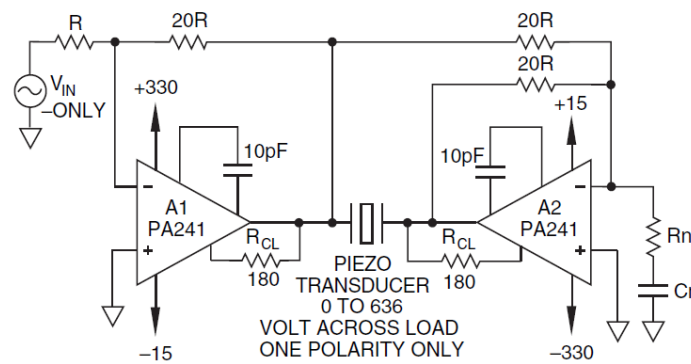


FIGURE 5.

This solution could fulfil all the requirements (using two Apex MP111).

In this case, supposing that the master amplifier (on the left) output works between 0 V and 75 V and the slave one (on the right) instead between 0V and -75V, the floating voltage across the piezo brick ( $V1 - V2$ ) will be in the range [0, 150] V.

In particular when the master reaches 75V the slave reaches -75 V so  $V1 - V2 = 150V$ .

The design data with this solution will change because the voltage swing of the single Apex MP111 is halved.

Capacitive Load (piezo brick) = 170nF

Required Slew rate (sine wave) =  $2 \cdot \pi \cdot f \cdot V_{peak} = 2 \cdot \pi \cdot (500e3) \cdot 37.5 \cdot (1e-6)$   
 $V/\mu s = \sim 117.8 V/\mu s$

Required Peak Current =  $C \cdot dV/dt = 170 \text{ nF} \cdot 117.8 V/\mu s = \sim 20 \text{ A}$

With this solution the previous two issues on the Slew Rate and Voltage Output are solved.

## Supply Rails

From Apex MP111 datasheet, there are two supply rails  $V_s$  and  $V_b$  that can be the same for avoiding that  $V_b$  is less than  $V_s$ .

With  $V_s=V_b=85V$  an output voltage of 75V can be reached ( $V_s - 8.4$  from the datasheet).

With  $-V_s=-V_b=-20V$  an output voltage of 0V can be reached ( $-V_s + 5.8$ ).

Moreover, the Common Mode Voltage Range requirement (15V from the datasheet) is fulfilled.

## Power Dissipation

The power that should be dissipated in the amplifier can be computed as

$$P_{\text{ampl}} = P_{\text{supply}} - P_{\text{load}}$$

Considering the dielectric losses in the piezo brick:

$$P_{\text{load}} = (\pi/4) * \tan(\delta) * f * C * (V_{\text{pp}})^2 = (\pi/4) * 0.3 * (500e3) * (170e-9) * (150)^2 = 450.6 \text{ W}$$

This  $P_{\text{load}}$  should be halved due to the circuit symmetry.

$$\text{For each side of the bridge } P_{\text{supply}} = 2V_s I_{\text{peak}} / \pi = 2 * 52.5 * 20 / \pi = \sim 668.5 \text{ W.}$$

So the power amplifier should dissipate  $P_{\text{ampl}} = \sim 443 \text{ W}$ .

From the datasheet the maximum power that can be dissipated in the Apex MP111 can be computed as  $P_{\text{amplmax}} = (T_j - T_a) / R_{\text{jc}} = (175 - 25) / 0.65 \text{ W} = \sim 231 \text{ W}$ . that is well below the needed one.

- ➔ A possible solution is to use 2 power amplifiers in parallel for each side of the bridge. In this case the power that should be dissipated is halved as in Apex AN 20

#### 5.4 PARALLEL CONNECTION

The bridge circuit can also be combined with the parallel connection of power op amps. Figure 7 shows how substantial audio power outputs can be obtained along with improved reliability since the parallel connection spreads the load among more amplifiers.

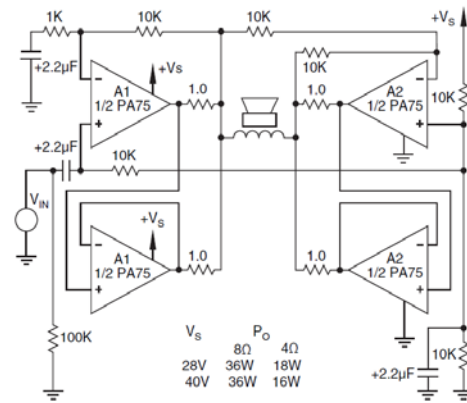


FIGURE 7. SINGLE SUPPLY PARALLEL BRIDGE

In this case the Heat sink thermal resistance can be computed according to

$$R_{HT} = (T_j - T_a)/P_D - R_{jc} = ((175-25)/221.6)-0.65 = 0.0269 \text{ } ^\circ\text{C/W} \text{ (this is a max value).}$$

### Further Considerations

- Heat Dissipation of the Piezo Brick. Its maximum operating temperature is 150 °C. With the considered condition also 100 °C of temperature rise could be expected.
- Considering other piezo bricks with lower capacitance (e. g 3mm\*3mm with 65nF or 2mm\*2mm with 25nF that have all the same nominal stroke and driving voltage). All the critical parameters will be scaled by the ratio between the capacitances.
- Considering the stability issues of the driving circuit with 4 power op amp (2 in parallel for each side of the bridge).
- Considering maybe a mechanical amplification mechanism on the piezo brick -> the driving voltage can be reduced almost by the same factor.

## Evaluation of the use of a lower capacitance brick (e.g. NAC 2011 - 2mm\*2mm)

### - Design Data

Capacitive Load (piezo brick) = 25nF

Required Slew rate (sine wave) =  $2\pi f \cdot V_{\text{peak}} = 2\pi \cdot (500\text{e}3) \cdot 75 \cdot (1\text{e}-6)$   
 $\text{V}/\mu\text{s} = \sim 235 \text{ V}/\mu\text{s}$  (it doesn't change)

Required Peak Current (sine wave) =  $C \text{ d}V/\text{d}t = 25 \text{ nF} \cdot 235 \text{ V}/\mu\text{s} = \sim 6 \text{ A}$

Required Slew rate (triangular wave) =  $4 \cdot f \cdot V_{\text{peak}} = 4 \cdot (500\text{e}3) \cdot 75 \cdot (1\text{e}-6) \text{ V}/\mu\text{s} = 150 \text{ V}/\mu\text{s}$

Required Peak Current (triangular wave) =  $C \text{ d}V/\text{d}t = 25 \text{ nF} \cdot 150 \text{ V}/\mu\text{s} = 3.75 \text{ A}$

### - Solution with 2 MP111 in bridge configuration:

Required Slew rate (sine wave) =  $2\pi f \cdot V_{\text{peak}} = 2\pi \cdot (500\text{e}3) \cdot 37.5 \cdot (1\text{e}-6)$   
 $\text{V}/\mu\text{s} = \sim 117.8 \text{ V}/\mu\text{s}$  (it doesn't change)

Required Peak Current =  $C \text{ d}V/\text{d}t = 25 \text{ nF} \cdot 117.8 \text{ V}/\mu\text{s} = \sim 3 \text{ A}$

Required Slew rate (triangular wave) =  $4 \cdot f \cdot V_{\text{peak}} = 4 \cdot (500\text{e}3) \cdot 37.5 \cdot (1\text{e}-6) \text{ V}/\mu\text{s} = 75 \text{ V}/\mu\text{s}$

Required Peak Current (triangular wave) =  $C \text{ d}V/\text{d}t = 25 \text{ nF} \cdot 75 \text{ V}/\mu\text{s} = \sim 1.88 \text{ A}$

### - Power Dissipation

The power that should be dissipated in the amplifier can be computed as

$$P_{\text{ampl}} = P_{\text{supply}} - P_{\text{load}}$$

Considering the dielectric losses in the piezo brick:

$$P_{\text{load}} = (\pi/4) \cdot \tan(\delta) \cdot f \cdot C \cdot (V_{\text{pp}})^2 = (\pi/4) \cdot 0.3 \cdot (500\text{e}3) \cdot (25\text{e}-9) \cdot (150)^2 = 66.3 \text{ W}$$

This  $P_{\text{load}}$  should be halved due to the circuit symmetry ( $P_{\text{loadhalf}} = 33.15 \text{ W}$ ).

For each side of the bridge  $P_{\text{supply}} = 2V_{\text{s}}I_{\text{peak}}/\pi = 2 \cdot 52.5 \cdot 3 / \pi = \sim 100.3 \text{ W}$ .

So the power amplifier should dissipate  $P_{\text{ampl}} = 67.15 \text{ W}$ .

From the datasheet the maximum power that can be dissipated in the Apex MP111 can be computed as  $P_{\text{amplmax}} = (T_j - T_a)/R_{jc} = (175-25)/0.65 \text{ W} = \sim 231 \text{ W}$ .

For the heat sink

$R_{HT} = (T_j - T_a)/P_D - R_{jc} = ((175-25)/67.15) - 0.65 = 1.58 \text{ }^\circ\text{C/W}$  (this is a max value).

## Circuit (functional) Simulation

### 5.2 UNIPOLAR OUTPUT

A particularly powerful way of applying the bridge is in the unipolar bridge. By unipolar, we mean that the output can only swing from 0 to one polarity. Figure 5 is used to illustrate this technique.

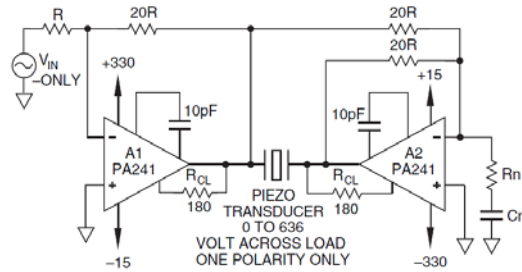
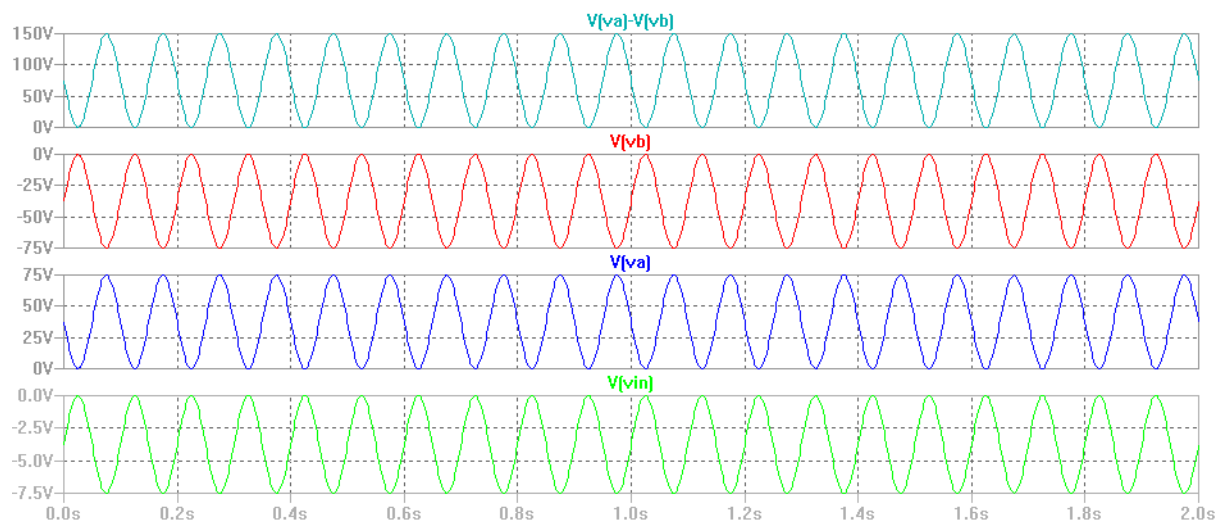
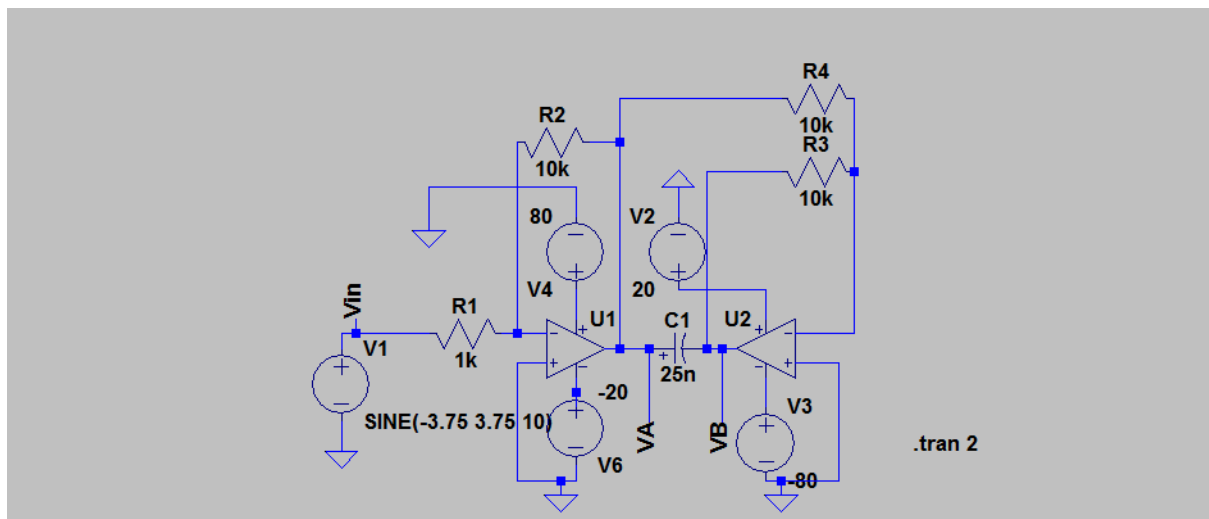
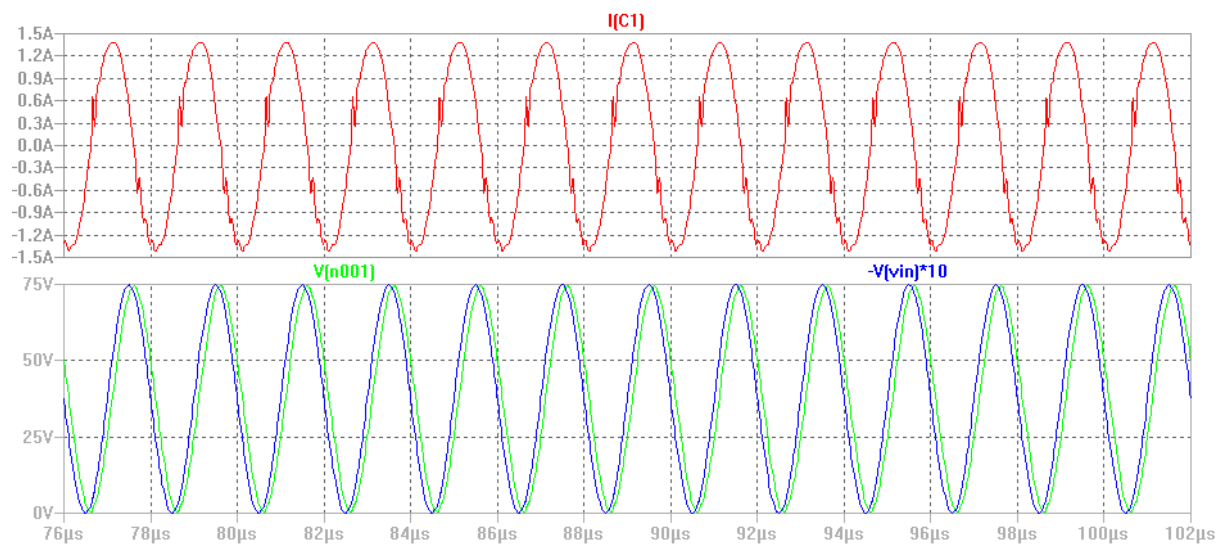
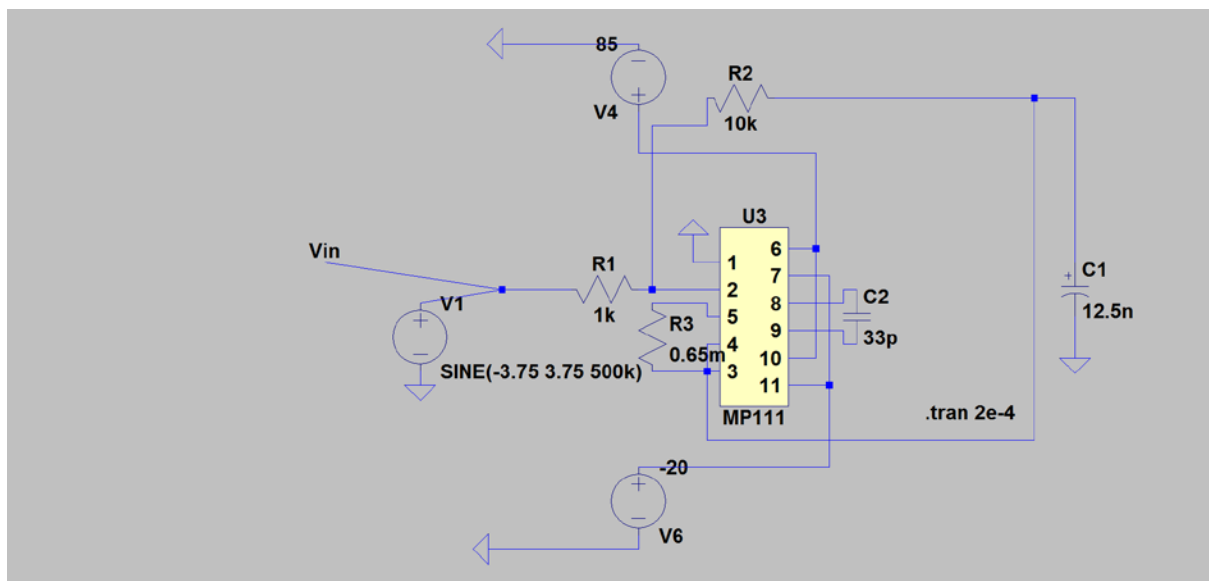


FIGURE 5.

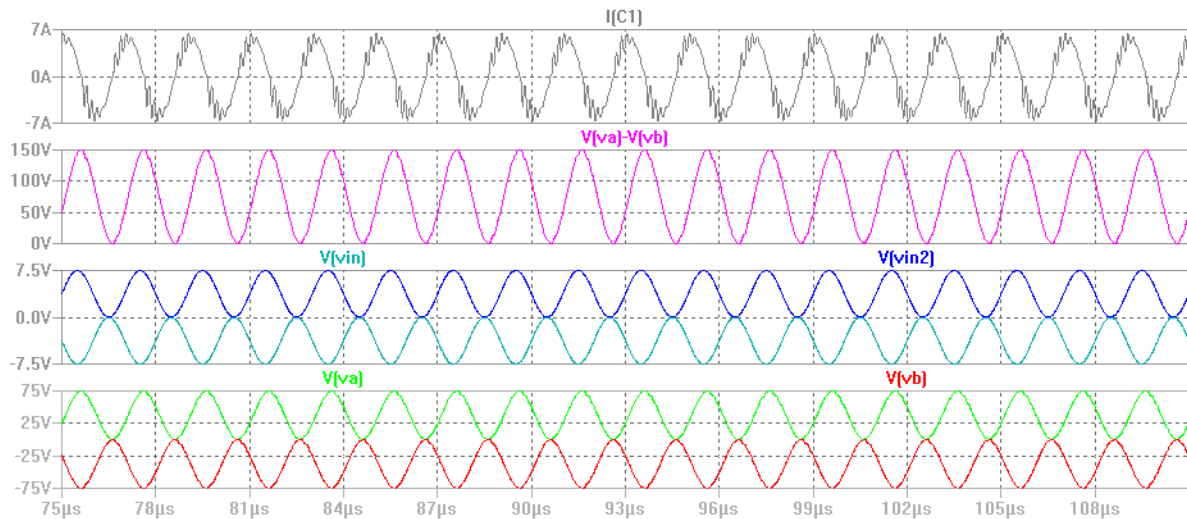
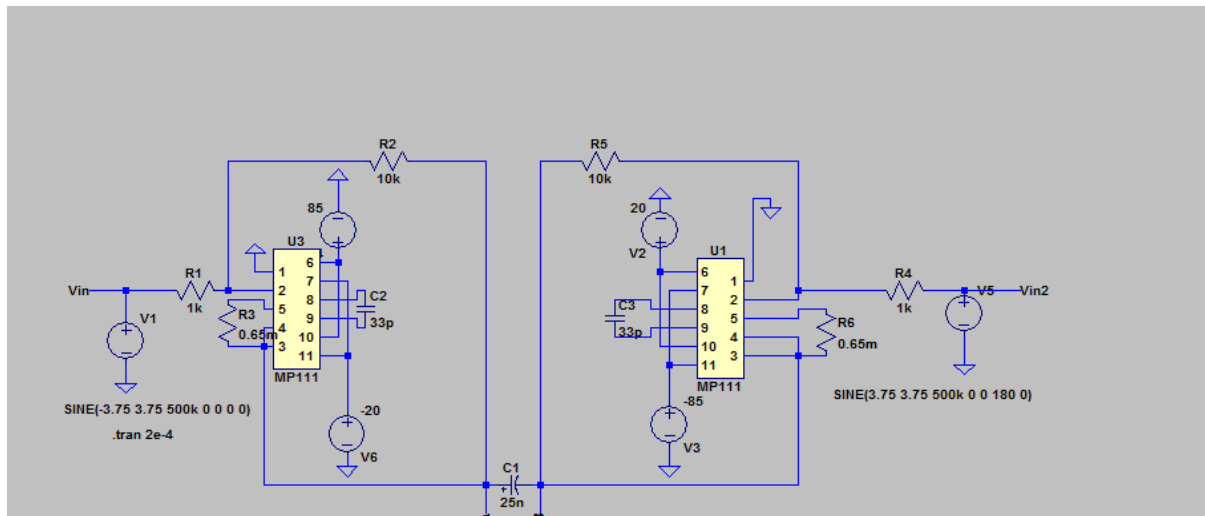


## Circuit Simulation with a single Apex MP111





## Circuit Simulation with a two Apex MP111 in a modified bridge



Note: the bridge has been modified because the slave MP111 in unit gain operation has a slew rate limitation (due to the compensation capacitor that should be bigger).

PHASE COMPENSATION		
$C_c$	GAIN W/O BOOST	TYP. SLEW RATE
100pF	$\geq 1$	55 V/ $\mu$ S
68pF	$\geq 4$	60 V/ $\mu$ S
33pF	$\geq 10$	130 V/ $\mu$ S
$C_c$	GAIN W BOOST	TYP. SLEW RATE
470pF	$\geq 3$	12 V/ $\mu$ S
220pF	$\geq 6$	27 V/ $\mu$ S
100pF	$\geq 10$	55 V/ $\mu$ S