# TTK4150

# **TTK4150 Nonlinear Systems and Control**

# **Martin Brandt**

# Learning goals:

- \* A thorough knowledge of theory and methods for non-linear dynamical systems.
- \* Know how to find the invariant sets of nonlinear dynamical systems, and know how to analyze the system behavior around these sets.
- \* Know the methods Phase plane analysis, Lyapunov stability analysis, Input-to-state stability analysis, Input-Output stability analysis, Passivity analysis, Lyapunov-based control, Energy-based control, Cascaded control, Passivity-based control, Input-Output linearization, and Backstepping control design.

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# 1 | SECOND-ORDER NONLINEAR TIME-INVARIANT SYSTEMS

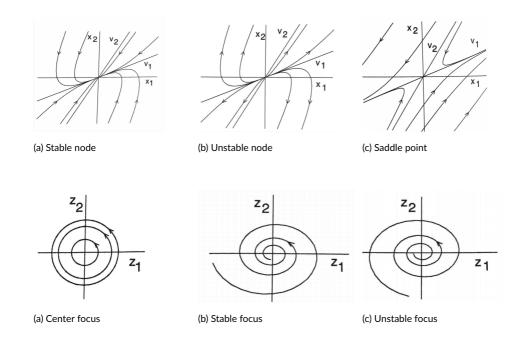
We first consider the system

$$\dot{x}_1 = f_1(x_1, x_2) 
\dot{x}_2 = f_2(x_1, x_2)$$
(1)

**Phase-plane analysis:** Determine the system behavior by constructing a phase portrait, i.e. plotting different IVP solutions in the phase space.

#### Local analysis:

- \* Linearize about x\*.
- \* Find eigenvalues  $\lambda(A)$ .
- \* Classify equilibrium points for  $f(x^*) = 0$ . If  $\lambda$  is real, them we either get a stable node  $(\lambda_2 < \lambda_1 < 0)$ , unstable node  $(0 < \lambda_2 < \lambda_1)$  or a saddle point  $(\lambda_2 < 0 < \lambda_1)$ . In the complex case  $\lambda_{1,2} = \alpha \pm \beta i$ , then we either get a center focus  $(\alpha = 0)$ , a stable focus  $(\alpha < 0)$  or an unstable focus  $(\alpha > 0)$ .



**Topological equivalence:** if the real part of the eigenvalues are nonzero, then the local phase-portrait corresponds to the phase portrait of the linearized system.

## 1.1 | Periodic orbits and limit cycles

**Definition** Periodic orbit:  $\exists T > 0$  s.t.  $x(t + T) = x(t) \quad \forall t \ge 0$ .

**Definition** Limit cycle: non-trivial isolated periodic orbit.

#### Lemma 1 Poincaré-Bendixson criterion:

Let M be a closed bounded subset of the plane s.t.

\* M contains no  $x^*$ , or it contains only one  $x^*$  with the property that the eigenvalues of the Jacobian matrix at  $x^*$  have positive real parts (unstable focus or unstable node).

\* Every trajectory starting in M stays in M  $\forall t > t_0$ .

Then M contains a periodic orbit of the system.

#### Lemma 2 Bendixson negative criterion:

If on a simply connected region D,  $\frac{\partial f_1}{\partial x_1} + \frac{\partial f_2}{\partial x_2}$  is not identically zero and does not change sign, then the system has no periodic orbits lying entirely in D.

**Corollary 3** *C* is a periodic orbit  $\implies \Sigma_i I = 1$  (sum of indeces of equilibrium points in *C*, where saddle points have index -1 and others have index 1)

#### 2 | FUNDAMENTAL PROPERTIES

**Lipschitz:**  $||f(t, x) - f(t, y)|| \le L||x - y||$ 

Either locally Lipschitz on D (L varies), Lipschitz in D or globally Lipschitz.

#### **Theorem 4** Local existence and uniqueness:

lf

- \* f(t, x) is piecewise continuous in t,
- \* f(t, x) is Lipschitz  $\forall x, y \in B = \{x \in \mathbb{R}^n | ||x x_0|| \le r\} \forall t \in [t_0, t_1],$

Then there exists a unique solution of the IVP x(t) on  $t \in [t_0, t_0 + \delta]$ .

#### 3 | LYAPUNOV STABILITY

## 3.1 | Stability of equilibrium points

**Asymptotic stabilization problem:** Find  $\gamma(t, e)$  s.t. e = 0 is an asymptotically stable equilibrium point. Regulation vs. trajectory tracking.

**Definition** Stability: x = 0 is stable iff  $\forall \varepsilon > 0$   $\exists \delta(\varepsilon) > 0$  s.t.  $||x(0)|| < \delta \Rightarrow ||x(t)|| < \varepsilon$   $\forall t \ge 0$ 

**Definition** Asymptotic stability: x = 0 is (locally) asymptotically stable iff it is stable, and  $\exists r > 0$  s.t.  $||x(0)|| < r \Rightarrow \lim_{t \to \infty} x(t) = 0$ 

**Definition** Region of attraction:  $B_r = \{x \in \mathbb{R}^n : ||x|| < r\}$ . We denote  $R_A$  as the union of all the regions of attraction.

**Definition** Global asymptotic stability: x = 0 is GAS iff it is stable, and  $\lim_{t\to\infty} x(t) = 0$   $\forall x(0)$ 

**Definition** Exponential stability: x = 0 is exponentially stable iff

$$\exists r, k, \lambda > 0 \text{ s.t. } ||x(0)|| < r \Rightarrow ||x(t)|| \le k ||x(0)|| e^{-\lambda t} \quad \forall t \ge 0$$

**Definition** Global exponential stability: x = 0 is GES iff  $\exists k, \lambda > 0$  s.t.  $\forall x(0) ||x(t)|| \le k ||x(0)|| e^{-\lambda t}$   $\forall t \ge 0$ 

Remark It is useful to think in terms of stability + convergence to seperate the different stability properties.

#### 3.2 | Lyapunov's indirect method

**Theorem 5** Lyapunov's indirect method:

Let x = 0 be an equilibrium point for

$$\dot{x} = f(x), \quad f: \mathbb{D} \subset \mathbb{R}^n \to \mathbb{R}^n \quad \text{is} \quad C^1$$
 (2)

- **1.** Linearize about x = 0,  $\dot{x} = Ax$ , where  $A = \frac{\partial f}{\partial x}\Big|_{x=0}$ .
- **2.** Find the eigenvalues  $\lambda_1(A), \ldots, \lambda_n(A)$ .
- 3. Categorize the eigenvalues:
- \*  $\forall i \quad \text{Re}(\lambda_i) < 0 \Rightarrow \text{asymptotically(exponentially) stable}$
- \*  $\exists i$  Re  $(\lambda_i) > 0 \Rightarrow unstable$
- \*  $\forall i \quad \text{Re}(\lambda_i) \leq 0 \Rightarrow \text{inconclusive}$

While Lyapunov's indirect method is simple to use, the results are only local and often inconclusive. Let's see if we can do better ey?

#### 3.3 | Lyapunov's direct method

**Definition** Lyapunov function:

V is a Lyapunov function for x = 0 iff

- \* V is  $C^1$
- \* V(0) = 0, V(x) > 0 in  $\mathbb{D} \setminus \{0\}$
- \*  $\dot{V}(0) = 0$ ,  $\dot{V}(x) \le 0$  in  $\mathbb{D} \setminus \{0\}$

If  $\dot{V}(x) < 0$  in  $\mathbb{D} \setminus \{0\}$  then V is a strict Lyapunov function for x = 0.

**Theorem 6** Lyapunov's stability theorem:

- \* If  $\exists V(x)$  for x = 0, then x = 0 is stable.
- \* If  $\exists$  strict V(x) for x = 0, then x = 0 is asymptotically stable.

**Theorem 7** Chetaev's instability theorem:

If  $\dot{V}(x) > 0$  in a set  $U = \{x \in B_r | V(x) > 0\}$ , then x = 0 is unstable.

**Definition** Radially unboundedness:  $||x|| \to \infty \implies V(x) \to \infty$ 

**Theorem 8** If  $\exists$  strict  $V : \mathbb{R}^n \to \mathbb{R}$  for x = 0 and V is radially unbounded, then x = 0 is GAS.

**Theorem 9** If there exist a function  $V: \mathbb{D} \to \mathbb{R}$  and constants  $a, k_1, k_2, k_3 > 0$  s.t.

- \* V is C1
- \*  $k_1 ||x||^a \le V(x) \le k_2 ||x||^a \quad \forall x \in \mathbb{D}$
- \*  $\dot{V}(x) \le -k_3 ||x||^a \quad \forall x \in \mathbb{D}$

then x = 0 is exponentially stable. If these conditions hold for  $\mathbb{D} = \mathbb{R}^n$ , then x = 0 is GES.

Remark  $\lambda_{min}(P) \|x\|^2 \le x^{\top} P x \le \lambda_{max}(P) \|x\|^2$ 

**Remark** How to deal with indeterminate signs in  $\dot{V}$ ?

- \* Completion of squares:  $x_1x_2 \leq \frac{1}{2}(x_1^2 + x_2^2)$
- \* Young's inequality:  $x_1x_2 \le \epsilon x_1^2 + \frac{1}{4\epsilon}x_2^2$
- \* Cauchy-Schwarz' inequality:  $|a_1x_1 + a_2x_2 + \cdots + a_nx_n| \le \sqrt{\left(a_1^2 + a_2^2 + \cdots + a_n^2\right)} ||x||_2$

# 3.4 | The invariance principle

**Definition** Invariant set:  $x(0) \in M \implies x(t) \in M \quad \forall t \in \mathbb{R}$ 

**Definition** Positively invariant set:  $x(0) \in M \implies x(t) \in M \quad \forall t \geq 0$ 

**Definition** Level set:  $\Omega_c = \{x \in \mathbb{R}^n : V(x) \le c\}$ 

**Remark** For V(x) radially unbounded  $\Omega_c$  is an invariant set.

#### Theorem 10 La Salle's theorem:

If  $\exists V : \mathbb{D} \to \mathbb{R}$  s.t.

- \* V is C<sup>1</sup>
- \*  $\exists c > 0$  such that  $\Omega_c = \{x \in \mathbb{R}^n | V(x) \le c\} \subset \mathbb{D}$  is bounded
- \*  $\dot{V}(x) \leq 0 \quad \forall x \in \Omega_c$

Let  $E = \{x \in \Omega_c | \dot{V}(x) = 0\}$ . Let M be the largest invariant set contained in E. Then  $x(0) \in \Omega_c \Rightarrow x(t) \stackrel{t \to \infty}{\longrightarrow} M$ .

#### **Definition** Region of attraction:

Let x=0 be an asymptotically stable equilibrium point of the system  $\dot{x}=f(x)$ , where  $f:\mathbb{D}\to\mathbb{R}^n$  is locally Lipschitz and  $\mathbb{D}\subset\mathbb{R}^n$  contains the origin. Let  $\phi(t,x_0)$  be the solution. Then the region of attraction is

$$R_A = \{x_0 \in \mathbb{D} \mid \phi(t, x_0) \text{ is defined } \forall t \ge 0 \text{ and } \phi(t, x_0) \to 0 \text{ as } t \to \infty\}$$
 (3)

(I.e. all the points with a corresponding solution that converges to the origin).

**Remark** GAS iff  $R_A = \mathbb{R}^n$ .

**Estimate of**  $R_A$ : choose the largest set  $\Omega_c$  in  $\mathbb D$  which is bounded, and only the connected component of  $\Omega_c$  that contains the origin. Then this subset is a subset of  $R_A$ .

# 3.5 | Stability analysis of time-variant systems

We now consider the system  $\dot{x} = f(t, x)$ .

**Definition** Stability:  $\forall \varepsilon > 0$ ,  $\exists \delta(\varepsilon, t_0) > 0$  s.t.  $\|x(t_0)\| < \delta \Rightarrow \|x(t)\| < \varepsilon \quad \forall t \ge t_0 \ge 0$ 

**Definition** Uniform stability: stable with  $\delta(\varepsilon, t_0) = \delta(\varepsilon)$ .

**Definition** Asymptotic stability: stable and  $\exists c(t_0) > 0$  s.t.  $||x(t_0)|| < c \Rightarrow x(t) \stackrel{t \to \infty}{\longrightarrow} 0$ .

**Definition** Uniform asymptotic stability: asymptotically stable with  $\delta(\varepsilon, t_0) = \delta(\varepsilon)$ .

**Definition** Global uniform asymptotic stability: uniform stability with  $\delta(\varepsilon) \xrightarrow{\varepsilon \to \infty} \infty$  and  $\forall c > 0 \quad ||x(t_0)|| < c \Rightarrow x(t) \xrightarrow{t \to \infty} 0$  uniformly in  $t_0$ .

**Definition** Exponential stability:  $\exists c, k, \lambda > 0$  s.t.  $||x(t)|| \le k ||x(t_0)|| e^{-\lambda(t-t_0)}t \ge t_0 \ge 0 ||x(t_0)|| \le c$ . GES if  $\forall c$ .

**Definition** A continuous function  $\alpha:[0,a)\to [0,\infty)$  is a class  $\mathcal{K}$  function iff:  $\alpha(0)=0$  and  $\alpha(r)$  is strictly increasing, i.e.  $\frac{\partial \alpha}{\partial r}>0 \quad \forall r>0$ .

**Definition** If in addition  $a \to \infty$  and  $\alpha(r) \to \infty$  as  $r \to \infty$ , then  $\alpha$  is a class  $\mathcal{K}_{\infty}$  function.

**Definition** A continuous function  $\beta:[0,a)\times[0,\infty)\to[0,\infty)$  is a class  $\mathscr{KL}$  function if for each fixed s

- \*  $\beta(r,s)$  is a class  $\mathscr K$  function w.r.t. r
- and for each fixed r
- \*  $\beta(r, s)$  is decreasing w.r.t. s,
- \*  $\beta(r,s) \to 0$  as  $s \to \infty$ .

We can now define stability in terms of class  ${\mathscr K}$  functions:

 $\textbf{Definition} \quad \text{Uniform stability: } \exists \text{ class } \mathscr{K} \text{ function } \alpha \text{ and } \exists c > 0 \text{ s.t. } \|x(t)\| \leq \alpha \left(\|x(t_0)\|\right) \forall t \geq t_0 \geq 0, \quad \forall \ \|x(t_0)\| < c.$ 

**Definition** Uniform asymptotic stability:  $\exists$  class  $\mathscr{K}\mathscr{L}$  function  $\beta$  and  $\exists c > 0$  s.t.  $\|x(t)\| \le \beta (\|x(t_0)\|, t - t_0) \forall t \ge t_0 \ge 0$ ,  $\forall \|x(t_0)\| < c$ . GUAS if  $\forall c$ .

**Definition** V(t, x) is positive definite iff

- V(t,0) = 0
- \*  $V(t,x) \geq W_1(x)$

 $\forall t \geq 0, W_1(x) > 0$ 

**Definition** V(t, x) is decrescent iff

- V(t,0) = 0
- \*  $V(t,x) \leq W_2(x)$

 $\forall t \geq 0, W_2(x) > 0$ 

We can summarize the stability theorems for time-varying systems like this:

	Stable	Uniformly stable	UAS	GUAS
V	Pos. def.	Pos. def., decrescent	Pos. def., decrescent.	Pos. def., decrescent, radially unbounded
Ÿ	Neg. semidef.	Neg. semidef.	Neg. def.	Neg. def.
	$\forall x \in \mathbb{D}$	$\forall x \in \mathbb{D}$	$\forall x \in \mathbb{D}$	$\forall x \in \mathbb{R}^n$

Estimate of  $R_A$ :  $B_r = \{x \in \mathbb{R}^n : ||x|| \le r\} \subset \mathbb{D}, c < \min_{||x||=r} W_1(x) \implies \{x \in B_r : W_2(x) \le c\}$  is a region of attraction, when the origin is UAS.

#### Lemma 11 Barbalat's lemma:

Let  $\dot{f}: \mathbb{R} \to \mathbb{R}$  be uniformly continuous on  $[0, \infty)$ . If  $\lim_{t \to \infty} f(t)$  exists and is finite, then  $\dot{f} \to 0$  as  $t \to \infty$ . Rephrased: if V is lower bounded,  $\dot{V} \le 0$  and  $\ddot{V}$  is uniformly bounded, then  $\dot{V} \to 0$  as  $t \to \infty$ .

#### 4 | INPUT-TO-STATE STABILITY

# 4.1 | Input-to-state stability

Now we consider the system  $\Sigma : \dot{x} = f(t, x, u)$ , where we consider u(t) to be a disturbance/modelling error.

**Definition** Input-to-state stability (ISS):  $\exists \beta \in \mathcal{KL}, \gamma \in \mathcal{K} \text{ s.t. } \|x(t, x_0, u)\| \leq \max \{\beta(\|x(t_0)\|, t - t_0), \gamma(\|u\|_{\infty})\}$ 

**Remark** This is really just an extension of GUAS that says that x is bounded the input as well. So naturally if  $\Sigma$  is ISS then it is also 0-GUAS.

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Definition V: [0, \infty) \times \mathbb{R}^n \to \mathbb{R} is an ISS-LF for Σ iff *V is C^1. \exists \alpha_1, \alpha_2 \in \mathscr{K}_\infty and \rho \in \mathscr{K} s.t. *\alpha_1(\|x\|) \le V(t, x) \le \alpha_2(\|x\|) *\dot{V}(t, x) = \frac{\partial V}{\partial x}f + \frac{\partial V}{\partial t} \le -W_3(x) \quad \forall \|x\| \ge \rho(\|u\|) > 0 where W_3 > 0. It can be shown that \gamma = \alpha_1^{-1} \circ \alpha_2 \circ \rho. \exists \text{ISS} - \text{LF} for Σ \Rightarrow Σ is ISS
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**Lemma 12** if f is  $C^1$  and globally Lipschitz in (x, u), then  $\Sigma$  is  $0 - GES \Rightarrow \Sigma$  is ISS

**Theorem 13** Consider the cascaded system  $\Sigma_2 \longrightarrow \Sigma_1$ , where  $\Sigma_1 : \dot{x}_1 = f_1(t, x_1, x_2)$  and  $\Sigma_2 : \dot{x}_2 = f_2(t, x_2)$ . If  $\Sigma_2$  is GUAS and  $\Sigma_1$  is ISS, then the cascaded system is GUAS.

#### 4.2 | Input-output stability

We consider the system y = Hu.

**Remark** This makes  $\mathcal{L}_2$  the space of all continuous, square-integrable functions, for instance.

**Definition**  $\mathscr{L}_{pe}^m$  space:  $u \in \mathscr{L}_{pe}^m \Leftrightarrow u_{\tau} \in \mathscr{L}_p^m \quad \forall \tau \in [0, \infty)$ , where  $u_{\tau}$  is the truncated version if u.

**Definition**  $H: \mathcal{L}_{pe}^{m} \to \mathcal{L}_{pe}^{q}$  is  $\mathcal{L}_{p}$  stable iff

\*  $\exists \alpha \text{ class } \mathcal{K} \quad \alpha: [0, \infty) \to [0, \infty)$ \*  $\exists \text{ constant } \beta \geq 0$ s.t.  $\|(Hu)_{\tau}\|_{\mathcal{L}_{p}} \leq \alpha \left(\|u_{\tau}\|_{\mathcal{L}_{p}}\right) + \beta \quad \forall u \in \mathcal{L}_{pe}^{m} \text{ and } \tau \in [0, \infty)$ 

**Definition** Finite-gain  $\mathscr{L}_p$  stable:  $\exists \gamma, \beta \geq 0$  s.t.  $\|(Hu)_{\tau}\|_{\mathscr{L}_p} \leq \gamma \|u_{\tau}\|_{\mathscr{L}_p} + \beta$ 

**Definition** Causal system:  $(Hu)_{\tau} = (Hu_{\tau})_{\tau}$ 

The two definitons above hold for non-truncated signals if the systems are causal.

#### **Theorem 14** *Small-gain theorem:*

The feedback interconnection of  $H_1$  and  $H_2$  are finite-gain  $\mathcal{L}_p$  stable iff  $\gamma_1\gamma_2 < 1$ .

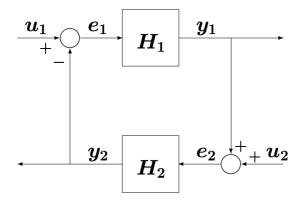


FIGURE 3 Feedback interconnection

## 5 | PASSIVITY

# 5.1 | Passivity for memoryless functions

Consider the memoryless function y = h(t, u)  $h: [0, \infty) \times \mathbb{R}^p \to \mathbb{R}^p$ .

**Definition** The system is passive if  $u^{\top}y \ge 0$  and lossless if  $u^{\top}y = 0$ .

The system is input-strictly passive if  $u^{\top}y \geq u^{\top}\varphi(u)$ , where  $u^{\top}\varphi(u) > 0 \quad \forall u \neq 0$ .

The system is output-strictly passive if  $u^{\top}y \ge y^{\top}\rho(y)$ , where  $y^{\top}\rho(y) > 0 \quad \forall y \ne 0$ .

# 5.2 | Passivity for dynamical systems

Now we extend this property for dynamical systems:  $\Sigma$ :  $\dot{x} = f(x, u), y = h(x, u)$  where  $f : \mathbb{R}^n \times \mathbb{R}^p \to \mathbb{R}^n$  is locally Lipschitz and  $h : \mathbb{R}^n \times \mathbb{R}^p \to \mathbb{R}^p$  is continuous, and f(0, 0) = 0 and h(0, 0) = 0.

**Definition** The system  $\Sigma$  is passive iff  $u^{\top}y \geq \dot{V} \quad \forall (x, u) \in \mathbb{R}^n \times \mathbb{R}^p$ 

The system is lossless if  $u^{T}y = \dot{V}$ .

The system is input-strictly passive if  $u^T y \ge \dot{V} + u^T \varphi(u)$ ,  $u^T \varphi(u) > 0 \quad \forall u \ne 0$ .

The system is output-strictly passive if  $u^T y \ge \dot{V} + y^T \rho(y)$ ,  $y^T \rho(y) > 0 \quad \forall y \ne 0$ .

The system is (state-) strictly passive if  $u^T y \ge \dot{V} + \psi(x)$ ,  $\psi(x) > 0$ .

**Remark** Passivity really just generalizes the idea that the change of stored energy in the system should be less than the energy supplied to the system.

#### 5.3 | Passivity and Lyapunov stability

**Lemma 15** If  $\Sigma$  is passive with a positive definite V(x), then the origin of  $\dot{x} = f(x,0)$  is stable.

**Lemma 16** If  $\Sigma$  is output-strictly passive with  $\rho(y) = \delta y$ ,  $\delta < 0$ , then  $\Sigma$  is finite-gain  $\mathcal{L}_2$  stable with  $\gamma \leq \frac{1}{\kappa}$ .

**Definition** Zero state observability: no solution of  $\dot{x} = f(x,0)$  can stay identically in  $S = \{x \in \mathbb{R}^n | h(x,0) = 0\}$  other than the trivial solution x(t) = 0.

**Lemma 17** The origin of  $\dot{x} = f(x,0)$  is asymptotically stable if  $\Sigma$  is state-strictly passive, or output-strictly passive and zero state observable. If V(x) is radially unbounded  $\dot{x} = f(x,0)$  is GAS.

**Theorem 18** If  $H_1$  and  $H_2$  is passive, then the feedback interconnection of  $H_1$  and  $H_2$  is passive with  $V = V_1 + V_2$ .

**Theorem 19** If  $H_1$  and  $H_2$  satisfies  $e_i^T y_i \ge \dot{V}_i + \varepsilon_i e_i^T e_i + \delta_i y_i^T y_i i = 1, 2$  and  $\varepsilon_1 + \delta_2 > 0$ ,  $\varepsilon_2 + \delta_1 > 0$ , then  $\Sigma$  is finite-gain  $\mathcal{L}_2$  stable.

**Theorem 20** If  $H_1$  and  $H_2$  are state-strictly passive,

or  $H_1$  and  $H_2$  are output-strictly passive and zero state observable,

or  $\mathcal{H}_1$  is state-strictly passive and  $\mathcal{H}_2$  is output-strictly passive and zero state observable or opposite,

then  $\Sigma$  is 0-AS. If  $V_1$  and  $V_2$  are radially unbounded then  $\Sigma$  is 0-GAS.

## 6 | NONLINEAR CONTROL

#### 6.1 | Lyapunov control design

- \* Propose a LFC V(t, x), typically as desired energy in system.
- \* Choose u = g(t, x) s.t.  $\dot{V} < 0$  or  $\dot{V} \le 0$  with La Salle / Barbalat.

## 6.2 | Passivity-based control

**Theorem 21** For the LTI system y(s) = h(s)u(s) with Re  $(p_i) < 0$ ,  $\forall i$  we have:

- \* Passivity  $\Leftrightarrow \text{Re}[h(j\omega)] \ge 0 \forall \omega$  (note that if h(s) has an integrator as well and  $\text{Re}(z_i) < 0$  this still holds)
- \* Input-strict passivity  $\Leftrightarrow \text{Re}[h(j\omega)] \ge \delta > 0 \forall \omega$ , with  $\varphi(u) = \delta u$
- \* Output-strict passivity  $\Leftrightarrow \exists \varepsilon > 0$  s.t.  $\text{Re}[h(j\omega)] \ge \varepsilon |h(j\omega)|^2 \forall \omega$ , with  $\rho(y) = \varepsilon y$

**Remark** This says that with non-negative real part the system is passive, with a strictly positive real part it is input-strictly passive and if it is lower bounded by the square of the magnitude it is output-strictly passive.

Remark Note that the previously discussed passivity theorems can be applied for passivity-based control.

Notably that means that if the controller and plant are passive, then the closed-loop system is passive.

Also if the controller is input- and output strictly passive and the plant is passive, then the system is finite gain  $\mathcal{L}_2$  stable. The final passivity theorem can naturally also be applied.

**Theorem 22** Consider  $H_1$  with  $\dot{x} = f(x, u)$  locally Lipschitz and y = h(x) continuous, both zero in the origin. Also consider  $H_2$ :  $\phi(y)$  locally Lipschitz, memoryless and zero in origin.

If  $H_1$  is passive with V > 0, radially unbounded and zero state observable, and  $H_2$  satisfies  $y^{\top}\phi(y) > 0$ ,  $y \neq 0$  (passive, but not lossness) then the origin is GAS.

Choice of y: for  $\dot{x} = f(x) + G(x)u$ , if  $\exists LF \ V(x)$  radially unbounded, let  $y = \left[\frac{\partial V}{\partial x}G(x)\right]^T$  for the system to be passive. Feedback passivation: choose  $u = \alpha(x) + \beta(x)v$ , y = h(x) s.t.  $\dot{x} = f(x) + G(x)u$  has desired passivity properties  $v \mapsto y$ .

#### 6.3 | Feedback linearization

#### 6.3.1 | Input-state linearization

Consider  $\dot{x} = f(x) + G(x)u$ . Find a state transformation z = T(x) and input transformation  $u = \alpha(x) + \beta(x)v$  s.t. the new system in z coordinates is linear and controllable. This is rearily possible to do, so we consider input-output linearization instead:

### 6.3.2 | Input-output linearization

We now consider the system 
$$\dot{x} = f(x) + g(x)u$$
$$y = h(x)$$

**Definition** Lie derivative:  $L_f h = \frac{\partial h}{\partial x} f$ ,  $L_f^2 h = \frac{\partial L_f h}{\partial x} f$ , ...  $L_f^i h = L_f \left( L_f^{i-1} h \right)$ 

**Definition** The system has relative degree  $\rho$  in a region  $\mathbb{D}_0 \subset \mathbb{D} \subset \mathbb{R}^n$  if

$$L_g L_f^{i-1} h = 0, \quad 1 \le i \le \rho - 1$$

$$L_g L_f^{\rho - 1} h \ne 0$$

$$\forall x \in \mathbb{D}_0$$

$$(4)$$

**Remark** (the number of differentiations of y before u appears)

**Remark** For linear systems we have  $\rho = n - m$ .

**Remark** If the relative degree is well defined in the region of interest  $\mathbb{D}$ , then the system can be input-output linearized.

**Definition** Zero dynamics: internal dynamics when output is kept at zero by the input i.e.  $\psi = 0, \dot{\phi} = f_o(\phi, 0)$ .

**Definition** Minimum-phase system: the zero-dynamics are asymptotically stable.

#### Input-output linearization:

- 1. Find the relative degree  $\rho$
- 2. Write the system in normal form

Let  $\psi_1 = y, \psi_2 = \dot{y}, \dots$  The external dynamics are then  $\dot{\psi_1} = \psi_2, \dots, \dot{\psi}_{\rho} = L_f^{\rho} h + L_g L_f^{\rho-1} h \cdot u$ .

Let  $\phi_1, \ldots, \phi_{n-\rho}$  and  $z = T(x) = [\begin{array}{cc} \varphi^\top & \psi^\top \end{array}]^\top$ .

Choose  $\varphi$  s.t. T is a diffeomorphism,  $L_g \varphi_i = 0$  and  $\varphi_i(0) = 0$ .

If the Jacobian  $\frac{\partial T}{\partial x}\Big|_{x_0}$  is nonsingular, then T is a diffeomorphism. Then the internal dynamics are  $\dot{\psi} = \dots$ . We can then finally write the system in normal form:  $\dot{z} = \dots$ 

3. Choose u to cancel nonlinearities

$$u = \frac{1}{L_g L_f^{\rho - 1} h} \left( -L_f^{\rho} h + v \right) \quad \Longrightarrow \quad \psi_{\rho} = v \tag{5}$$

- 4. Analyze the zero-dynamics
- 5. Choose v to solve the control problem

This is a special case of the system  $\dot{\varphi} = f_0(\varphi, \psi), \quad \dot{\psi} = A\psi + Bv.$ 

**Lemma 23** If the system is minimum phase and  $v = -K\psi$  is chosen s.t. (A - BK) is Hurwitz, then the origin of the system is asymptotically stable.

**Lemma 24** If  $\dot{\phi} = f_0(\phi, \psi)$  is ISS, then the system is GAS.

For tracking we have  $v = -Ke + y_d^{(\rho)}$ , but the same results apply, expect  $\phi$  is now only bounded.

## 6.4 | Adaptive control

# 6.4.1 | MRAC for SISO systems

Consider the SISO system  $\dot{x} = a_p x + c_p f(x) + b_p u$ .

#### SISO MRAC:

- 1. Specify desired closed-loop behaviour by reference model:  $\dot{x}_m = a_m x_m + b_m r(t)$
- **2.** Choose control law s.t. plant output tracks reference model output when parameters are exactly known, by deriving error dynamics. Then replace parameters by estimates.
- **3. Choose adaptation law**  $(\hat{a}_x, \hat{a}_f, \hat{a}_r)$  by first deriving new tracking error dynamics in terms of estimation errors, and then choosing adaptation law from suitable Lyapunov function e.g.

$$V(e, \tilde{a}) = \frac{1}{2}e^2 + \frac{|b_p|}{2\gamma_x}\tilde{a}_x^2 + \frac{|b_p|}{2\gamma_f}\tilde{a}_f^2 + \frac{|b_p|}{2\gamma_r}\tilde{a}_f^2$$
(6)

By letting

$$\dot{\hat{a}}_x = -\gamma_x \operatorname{sgn}(b_p) e x, \quad \dot{\hat{a}}_f = -\gamma_f \operatorname{sgn}(b_p) e f, \quad \dot{\hat{a}}_r = -\gamma_r \operatorname{sgn}(b_p) e r$$
 (7)

the error will go to zero by Barbalat's lemma.

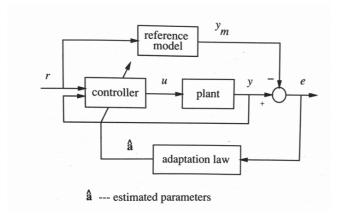


FIGURE 4 Model reference adaptive control loop

## 6.4.2 | Adaptive tracking control for a class of MIMO systems

Consider the system  $M(q)\ddot{q} + C(q, \dot{q})\dot{q} + D(q)\dot{q} + g(q) = u$ , with M > 0,  $\dot{M} - 2C$  is skew symmetric,  $z^T Dz > 0 \quad \forall z \neq 0$ . While the system is nonlinear it is linear in the unknown parameters a by the Regression matrix Y:

$$M(q)\ddot{q} + C(q, \dot{q})\dot{q} + D(q)\dot{q} + g(q) = Y(q, \dot{q}, \ddot{q})a$$
 (8)

#### Adaptive MIMO tracking control:

- **1.** Given the desired trajectory  $q_d(t)$ ,  $\dot{q}_d(t)$ ,  $\ddot{q}_d(t)$  bounded.
- 2. Choose control law:

We introduce the virtual reference velocity  $s = \dot{e} + \lambda e = \dot{q} - \dot{q}_r$ , where  $\dot{q}_r = \dot{q}_d - \Lambda (q - q_d)$ .

$$u = \hat{M}(q)\ddot{q}_r + \hat{C}(q,\dot{q})\dot{q}_r + \hat{D}(q)\dot{q}_r + \hat{g}(q) - K_p(q - q_d) - K_d(\dot{q} - \dot{q}_d) = Y(q,\dot{q},\dot{q}_r,\ddot{q}_r)\hat{a} - K_pe - K_d\dot{e}$$
 (9)

From  $V(s) = \frac{1}{2}s^{T}Ms$  and the cascade stability theorem we get GUAS with perfect tracking.

3. Choose an adaptation law such that tracking is achieved asymptotically. Again done by deriving error dynamics and analyzing Lyapunov function with Barbalat's lemma:

$$V(s, \tilde{a}) = \frac{1}{2} s^{\top} M s + \frac{1}{2} \tilde{a}^{\top} \Gamma^{-1} \tilde{a} \implies \dot{\tilde{a}} = -\Gamma Y^{\top} s$$
 (10)

## 6.5 | Backstepping

We consider the system

$$\begin{bmatrix} \dot{x}_1 \\ \dot{x}_2 \\ \vdots \\ \dot{x}_n \end{bmatrix} = f \begin{pmatrix} \begin{bmatrix} x_1 \\ x_2 \\ \vdots \\ x_n \end{bmatrix} + g \begin{pmatrix} \begin{bmatrix} x_1 \\ x_2 \\ \vdots \\ x_n \end{bmatrix} \\ \cdot x_{n+1}, \quad \dot{x}_{n+1} = u$$

$$(11)$$

Which we can identify as the cascaded system  $\Sigma_1 : \dot{\eta} = f(\eta) + g(\eta)\xi$ ,  $\Sigma_2 : \dot{\xi} = u$ .

We can regard  $\xi$  as the virtual control input of  $\Sigma_1$ .

#### Integrator backstepping:

- 1. Find stabilizing function for  $\Sigma_1$
- $\xi = \varphi(\eta)$ ,  $\varphi(0) = 0$  s.t.  $\xi = 0$  is asymptotically stable by some  $V_1(\eta)$ .
- 2. Design actual control input u
- **a.** Introduce error variable  $z = \xi \varphi(\eta)$ .
- **b.** Derive dynamics in  $(\eta, z)$  coordinates.
- **c.** Choose LFC  $V_2(\eta, z) = V_1(\eta) + \frac{1}{2}x^2$ .
- **d.** Find u that asymptotically stabilizes  $(\eta, z) = (0, 0)$ . Generally we get:

$$u = -\frac{\partial V}{\partial \eta}g(\eta) + \dot{\varphi} - kz \tag{12}$$

If  $V_2$  is radially unbounded in  $\eta$  and  $\mathbb{D} = \mathbb{R}^n$  the results are global.

**Remark** Note that this is a recursive process, and you may therefore do several steps of finding virtual control inputs before finally finding u. Also note that the method works fine even if  $\Sigma_2$  is more exotic.

# A | LINEAR METHODS

**Definition** We define the p-norm as:

$$||x||_p = \left(\sum_{i=1}^n |x_i|^p\right)^{\frac{1}{p}}, \quad p \in [1, \infty]$$
 (13)

$$||f||_{\mathcal{L}_{p}} = \left(\int_{0}^{\infty} |f(t)|^{p} dt\right)^{\frac{1}{p}}, \quad p \in [1, \infty]$$

$$\tag{14}$$

**Theorem 25** *Schwarz' inequality:* 

$$|\langle x, y \rangle| \le ||x|| \cdot ||y||$$
 (15)

**Definition**  $f: \mathbb{R}^n \to \mathbb{R}^m$ , then the Jacobian is defined as:

$$\frac{\partial f}{\partial x} \triangleq \begin{bmatrix} \frac{\partial f_1}{\partial x_1} & \cdots & \frac{\partial f_1}{\partial x_n} \\ \vdots & & \vdots \\ \frac{\partial f_m}{\partial x_1} & \cdots & \frac{\partial f_m}{\partial x_n} \end{bmatrix}$$
(16)

Which in the scalar case m = 1 is the gradient.