# Lecture 16: Friction and Electromechanical systems

- Dynamic friction models
- Electrical motors
- DC motor with constant field
- Some network modeling, passivity, ...

Book: 3.2, 3.3, 5

#### Static friction models

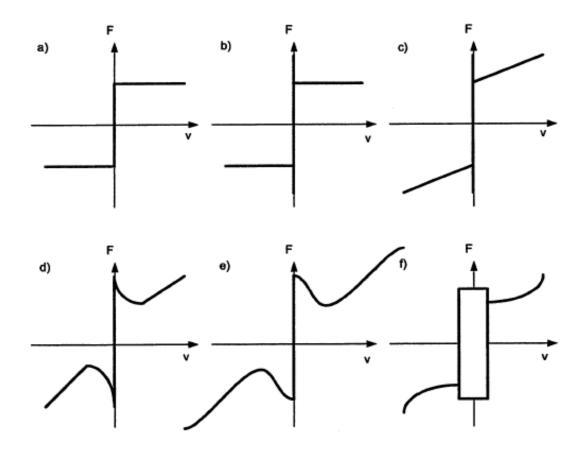
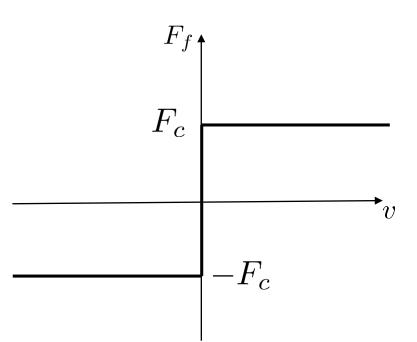
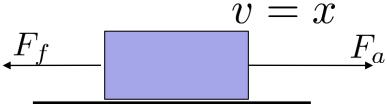


Figure 5.3: Static friction models: a) Colomb friction b) Coulomb+stiction c) Coulomb+stiction+viscous d) Stribeck effect e) Hess and Soom; Armstrong f) Karnopp model

# Problems with the signum terms at

zero velocity I





#### Newton's law:

$$m\dot{v} = F_a - F_f$$

$$F_f = F_c \operatorname{sign}(v) = \begin{cases} -F_c, & v < 0 \\ 0, & v = 0 \\ F_c, & v > 0 \end{cases}$$

# Karnopp's model

### Dynamic friction models

#### The Dahl model

$$\frac{\mathrm{d}F}{\mathrm{d}t} = \sigma \left( v - |v| \frac{F}{F_c} \right)$$

#### Why dynamic friction models?

- Easier to simulate
- Easier to analyze
- They reproduce (to some extent) dynamic friction phenomena
  - Presliding displacement
    - friction force act as a spring in sticking region
  - Frictional lag
    - Dynamic friction force depends on direction of velocity
  - Varying break-away force
    - Break-away force depends on rate-of-change of applied force

#### The LuGre model

$$F = \sigma_0 z + \sigma_1 \frac{\mathrm{d}z}{\mathrm{d}t} + \sigma_2 v$$

$$\frac{\mathrm{d}z}{\mathrm{d}t} = v - \sigma_0 \frac{|v|}{g(v)} z$$

$$g(v) = F_c + (F_s - F_c) e^{-\left(\frac{v}{v_s}\right)^2}$$

Dahl's model 
$$\frac{\mathrm{d}F}{\mathrm{d}t} = \sigma \left( v - |v| \frac{F}{F_c} \right)$$

# Passivity of Dahl's model $\frac{\mathrm{d}F}{\mathrm{d}t} = \sigma \left(v - |v| \frac{F}{F_c}\right)$

#### LuGre model I

$$F = \sigma_0 z + \sigma_z \dot{z} + \sigma_2 v$$

$$\dot{z} = v - \sigma_0 \frac{|v|}{g(v)} z \qquad g(v) = F_c + (F_s - F_c) e^{-\left(\frac{v}{v_s}\right)^2}$$

#### LuGre model II

«time constant»:

$$T = \frac{g(v)}{\sigma_0|v|} \to \infty$$
, if  $|v| \to 0$ 

- Same advantageous/disadvantegous as Dahl's model
- Possible more realistic dynamic behaviour
- LuGre-model is passive from v to F if  $\sigma_1$  is small enough

### Control systems with friction, II

Friction can be used to control motion

Electronic stability control (ESC), "anti-skidding"

Without ESC:







Also ABS systems exploits friction characteristics

### ABS-system – blokkeringsfrie bremser

 Hva er det som gjør at bremsing, gass, styring får bilen til å endre hastighet?

Friksjon mellom hjul og vei

- Hva bestemmer friksjon?
  - Tyngde
  - Underlag og egenskaper ved dekk
    - tørr asfalt, våt asfalt, snø, is
  - Relativ hastighetsforskjell mellom bil og hjul
    - langsgående (longitudinal) slipp, side- (lateral) slipp

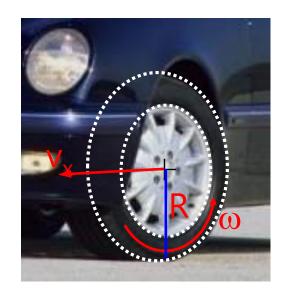
### Slipp – relativ hastighetsforskjell

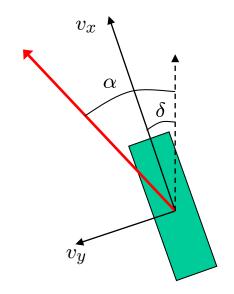
I langsretning:

$$\lambda_x := \frac{v_x - R\omega}{v_x}$$

I sideretning:

$$\lambda_y := \sin \alpha$$
 $\alpha := \arctan \frac{v_y}{v_x}$ 



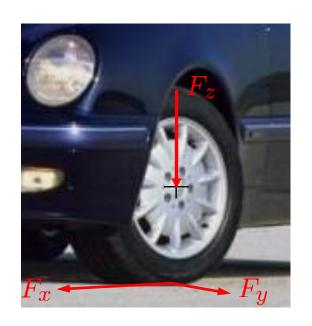


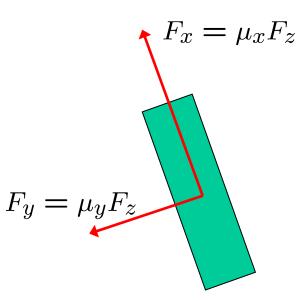
### Friksjonskrefter

#### Coloumbs lov:

- Friksjonskrefter gitt av vertikale krefter og friksjonskoeffisient
- Friksjonskoeffisient gitt av slipp og underlag

$$\mu_x \approx \mu_x(\lambda_x, \lambda_y, \mu_H)$$
 $\mu_y \approx \mu_y(\lambda_y, \lambda_x, \mu_H)$ 



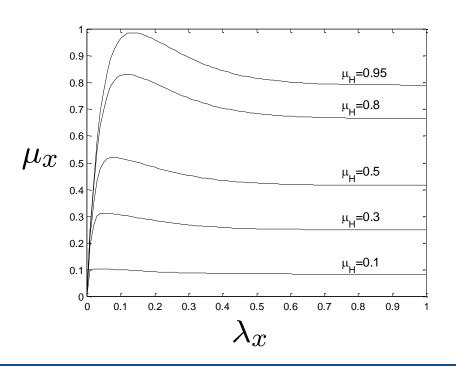


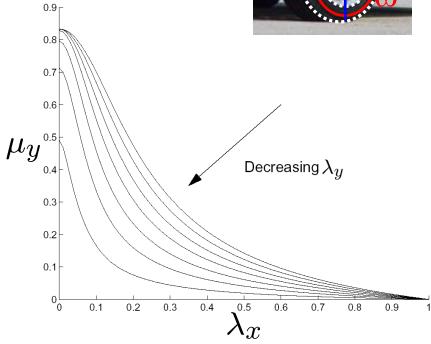
### Friksjonskoeffisienter under bremsing

 Bremsing reduserer hjulhastighet i forhold til bilhastighet

$$\lambda_x := \frac{v_x - R\omega}{v_x}$$

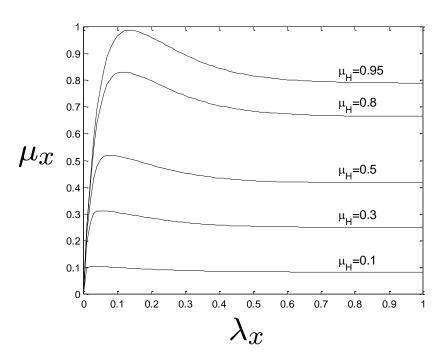


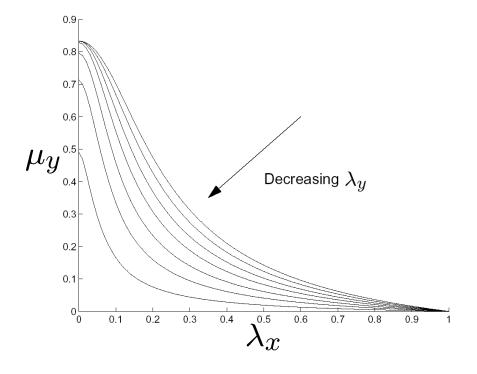




### Blokkeringsfrie bremser – ABS

- Ønsker konstant lav slipp under bremsing fordi
  - Det gjør bremsing mest effektivt
  - Kan styre bilen under bremsing





### ABS i praksis

#### Bremselengde:



#### Unnamanøver:



### Why modeling of electrical motors?

Electrical (and hydraulic) motors are used when something should move

Used everywhere: Process industries, offshore oil&gas production,

electromechanical systems, cars ...

Large and small

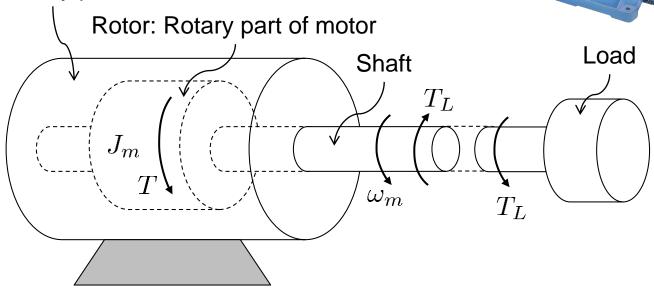


Often actuator (e.g. in a valve, in a compressor, ...)

- Example of modeling across domains (electrical + mechanical), and network modeling
  - Hydraulic motors another example (Ch. 4)
- Example of control-relevant modeling
  - Linear (transfer function) modeling

### **Motors**

Stator: Stationary part of motor



Equation of motion for motor shaft:

$$J_m \dot{\omega}_m = T - T_L$$

where

- T: Motor torque (set up by some device, e.g. DC motor)

-  $T_L$ : Load torque

-  $J_m$ : Moment of inertia for rotor and shaft

-  $\omega_m$ : Angular velocity/motor speed [rad/s, or rev./min]

### **Mechanical Power**

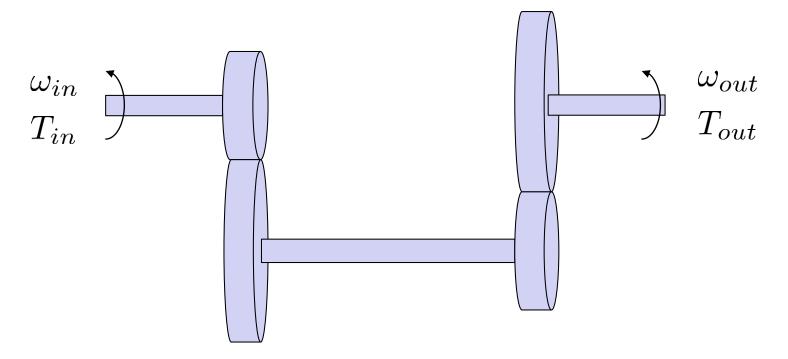
# Gears

# Rotational gear (cogwheel)

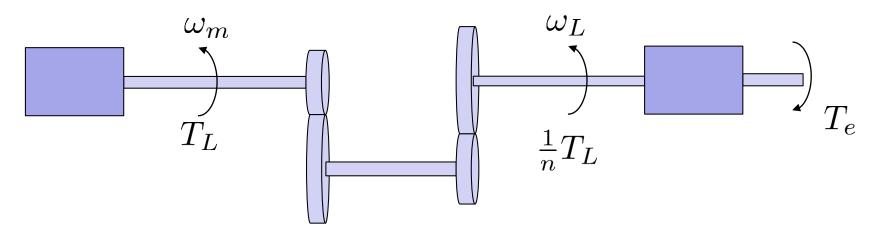


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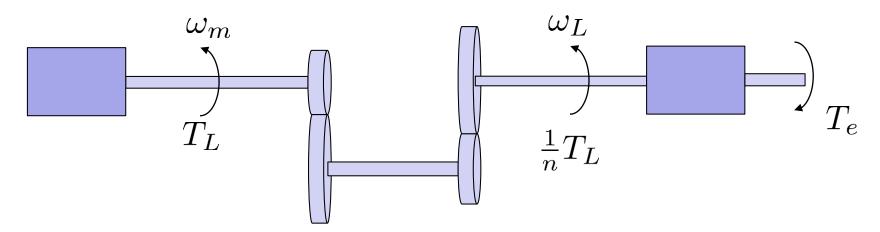
### Gear model



### Motor + Gear I



### Motor + Gear II



## Gears

# Rotational gear (cogwheel)



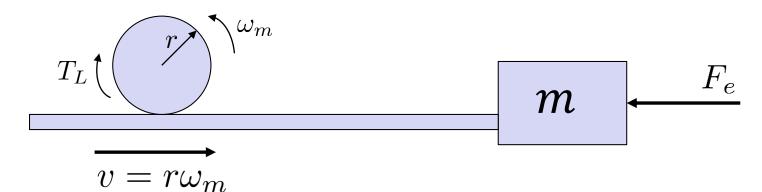
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# Translational gear (rack and pinion)

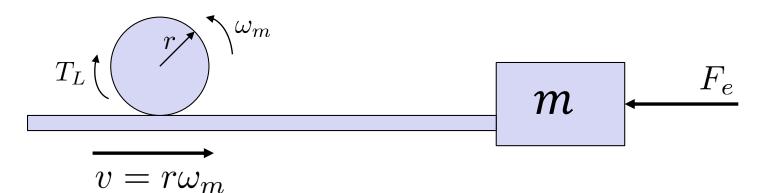


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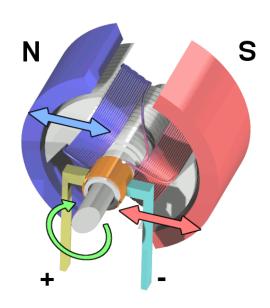
## Translational gear I

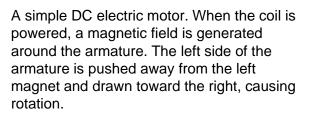


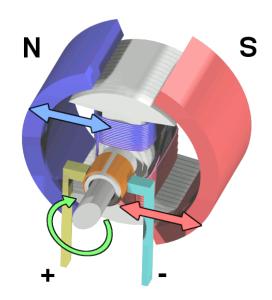
## Translational gear II



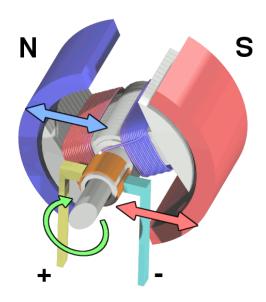
### A simple DC electric motor







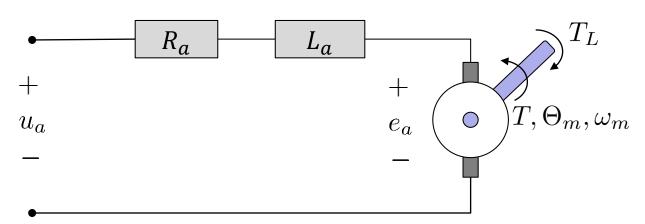
The armature continues to rotate.



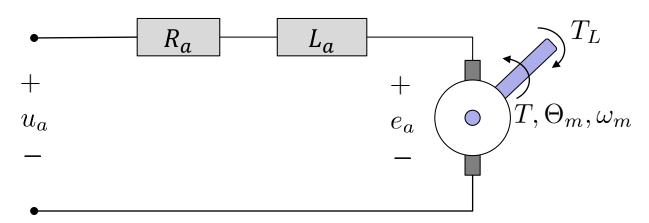
When the armature becomes horizontally aligned, the commutator reverses the direction of current through the coil, reversing the magnetic field. The process then repeats.

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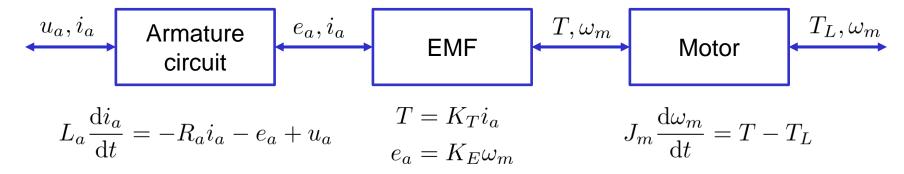
### Armature circuit I



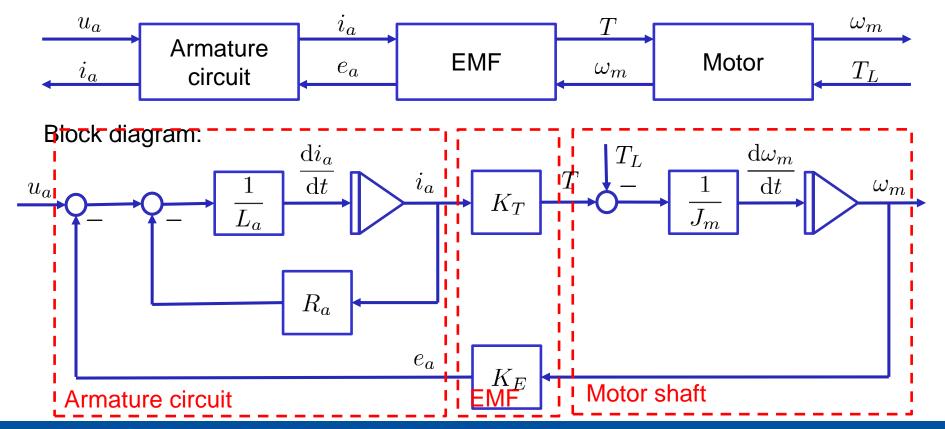
### Armature circuit II



#### Network modeling of DC-motor:



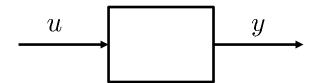
#### Signal flow modeling of DC-motor:



### Dymola Demo: Motor Drive

- File -> Demos -> Motor Drive
- Modelica.Electrical.Machines

### **Passivity**



A system with input u and output y is passive if

$$\int_0^t y(\tau)u(\tau)d\tau \ge -E_0$$

for all  $t \ge 0$ , for all input trajectories.

- If the product yu has power as unit, then if
  - $\int_0^t y(\tau)u(\tau)d\tau \ge 0$ : Energy is absorbed within the system, nothing delivered to the outside
  - $-\int_0^t y(\tau)u(\tau)d\tau \ge -E_0$ : Some energy can be delivered to the outside, limited (typically) by the initial energy in the system.
  - $\int_0^t y(\tau)u(\tau)d\tau \to -\infty$ : There is an inexhaustible energy source in the system. Not passive!

### Storage function

- We can proof passivity via the storage function
- Consider the system:

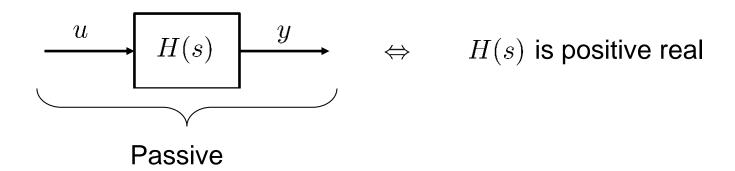
$$\dot{x} = f(x, u)$$
$$y = h(x)$$

- Assume we have a storage function  $V(x) \ge 0$  and a dissipation function  $g(x) \ge 0$
- Such that the time derivative for all control inputs u is:

$$\dot{V} = \frac{\partial V}{\partial x} f(x, u) = u^T y - g(x)$$

 $\rightarrow$  System with input u and output y is passiv

#### Positive real transfer functions



**Definition:** The transfer function H(s) (rational or irrational) is positive real if

- 1. H(s) analytic in Re[s] > 0.
- 2. H(s) is real for all positive and real s.
- 3.  $\operatorname{Re}[H(s)] \ge 0$  for all  $\operatorname{Re}[s] > 0$ .

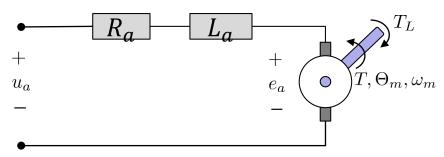
#### Check rational TFs for PRness

**Theorem:** A rational, proper transfer function H(s) is positive real (and hence passive) if and only if

- 1. H(s) has no poles in Re[s] > 0.
- 2. Re[H( $j\omega$ )]  $\geq 0$  for all  $\omega \in [-\infty, \infty]$  such that  $j\omega$  is not a pole of H(s).
- 3. If  $j\omega_0$  is a pole of H(s), then it is a simple pole, and the residual in  $s = j\omega_0$  is real and greater than zero, that is,

$$\operatorname{Res}_{s=j\omega_0} H(s) = \lim_{s \to j\omega_0} (s - j\omega_0) H(j\omega) > 0.$$

### Storage function DC motor



### Example 42 (load = «friction»)

# Transfer function of current controlled DC motor

$$\frac{i_a}{u_a}(s) = \frac{J_m}{K_T K_E} \frac{s}{1 + T_m s + T_m T_a s^2}$$

$$T_m = \frac{J_m R_a}{K_E K_T}$$

$$T_a = \frac{L_a}{R_a}$$

$$\frac{i_a}{u_a}(s) \approx \frac{J_m}{K_T K_E} \frac{s}{(1+T_m s)(1+T_a s)}$$

### Passivity current controlled DC motor

