# Relatório 6 Outubro

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### 1 ERF Relatorio 6 Out

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#### 1.1 Abstrato

Este trabalho serve como introdução ao software ADS, para desenho e analise de sistemas de radiofrequência e microondas. Para tal, iremos adaptar duas cargas a configurações de linha diferentes, utilizabdo a ferramenta Line Calc para obter as dimensões fisicas da microstrip line implementada. Usamos ainda o diagrama de Smith presente no programa para normalizar a carga da alinea 2. Obtivemos resultados favoraveis, que demonstram um alargamento da banda afectada em torno da frequência de funcionamento, com o acréscimo de secções de linha, sugerindo que estas introduzem sempre alguns pequenos defeitos que se acumulam.

# 1.2 Para uma secção de $\lambda/4$ , com $R_L = 200$

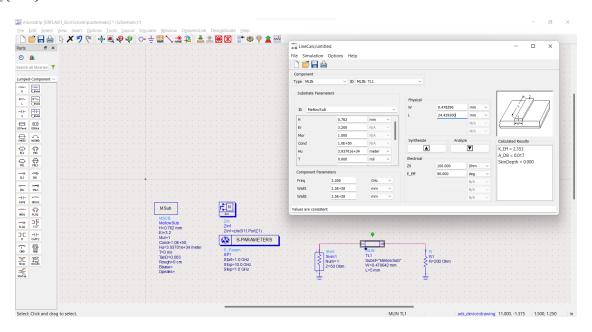
Para este caso, primeiro calculamos a impedancia caracteristica da secção de linha adaptada,  $Z_{line}$ .

```
[1]: import numpy as np
     from IPython.display import Markdown as md
     c = 3e8 \# light speed
     f = 2e9 # 2GHz Operating frequency
     output = ""
     Z_0 = 50 #Characteristic impedance of line
     R_L = 200 \#Load impedance
     Z_line = np.sqrt(Z_0*R_L) #impedance od microstrip line
     e_r = 3.2 \# from \ quide
     wavlen = c/f
     output += f'$$Z_0 = {Z_0} \Omega$$
     output += f'\$R_L = \{R_L\} \backslash m = \{R_L\} \backslash m'
     output += f'$$Z_{{line}} = \sqrt{{Z_{{0}}} \cdot R_{{L}}}} = {Z_line:.
      \rightarrow Of}\Omega$$\n'
     output += f'$$\lambda = {wavlen:.3f}m$$\n'
     md(output)
```

[1]:

$$Z_0 = 50\Omega$$
 
$$R_L = 200\Omega$$
 
$$Z_{line} = \sqrt{Z_0 \cdot R_L} = 100\Omega$$
 
$$\lambda = 0.150m$$

Utilizando estes valores no Line Calc, com  $E_{Eff}=90~(\lambda/4~{\rm corresponde~a}~\frac{360}{4}=90~)$  e \$Z\_0 = Z\_{line} = 100  $\Omega$ \$ obtivemos:



```
[2]: output = ""
#FROM ADS LINECALC
#----
W = 0.478296e-3 #width
L = 24.4293e-3 #Lenght
#----
K_eff = 2.53
A_DB = 0.017
#----
output += f'$$W = {W} m$$\n'
output += f'$$L = {L} m$$\n'
md(output)
```

[2]:

$$W = 0.000478296m$$
  
 $L = 0.0244293m$ 

```
[3]: #Depth of the substrate
H = 0.762 #from the guide
```

```
#Effective dielectric constant
\#l = 1e-3 \# 1mm \ should \ be \ changed \ to \ L \ from \ ads
e_ef = (e_r+1)/2+ (e_r-1)/2*1/np.sqrt(1+12*H/W)
beta = 2*np.pi*f/(c)*np.sqrt(e_ef)
Z_{in} = Z_{line*} (R_L+1j+Z_{line*np.tan(beta*L))/(Z_{line+1j*R_L*np.tan(beta*L))}
Gamma = (Z_{in} - Z_{0})/(Z_{in} + Z_{0})
rho = np.abs(Gamma)
theta = np.angle(Gamma)*180/np.pi
SwR = (1 + rho)/(1 - rho)
output=''
output+=f'$$E_{{ef}} = {e_ef}$$\n'
output+=f'$$B = {beta}$$\n'
output+=f'$$Z_{{in}}= {Z_{in}}$$\n'
output+=f'$\S\backslash Gamma= \{Gamma\}\$\$\backslash n'
output+=f'$\Rho= {rho}$\n'
output+=f'$$\Theta= {theta}$$\n'
output+=f'$$SwR= {SwR}$$\n'
md(output)
```

[3]:

$$E_{ef} = 2.1079553921574252$$

$$B = 60.816205173346454$$

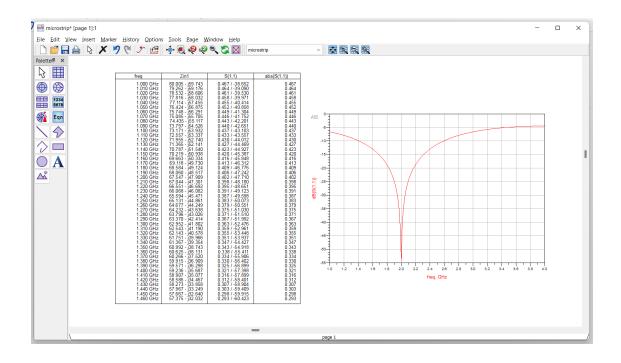
$$Z_{in} = (2.534514689002848 - 58.4223992826539j)$$

$$\Gamma = (0.1489701539673856 - 0.9464102935128225j)$$

$$R = 0.9580629157002651$$

$$\Theta = -81.05472667076315$$

$$SwR = 46.69048762917084$$



# 2 Para 3 Secçoes de Linha de $\lambda/4$ de comprimento

Utilizando a tabela da pag. 254 do livro "Microwave Engineering", vemos que N=3 porque temos 3 secções de linha.

[4]:

$$R_L/Z_0 = 4.0$$

Para N = 3:

$$\frac{Z_1}{Z_0} \qquad \frac{Z_3}{Z_0} \qquad \frac{Z_3}{Z_0}$$

$$\frac{R_L}{Z_0} = 4.0 \quad 1.1907 \quad 2.0000 \quad 3.3594$$

```
[5]: Z_1 = 1.1907 * Z_0
Z_2 = 2 * Z_0
Z_3 = 3.3594 * Z_0

output= ""
output+=f'$$Z_1 = {Z_1}$$\n'
output+=f'$$Z_2 = {Z_2}$$\n'
output+=f'$$Z_3 = {Z_3}$$\n'
md(output)
```

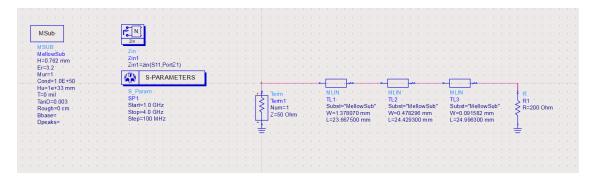
[5]:

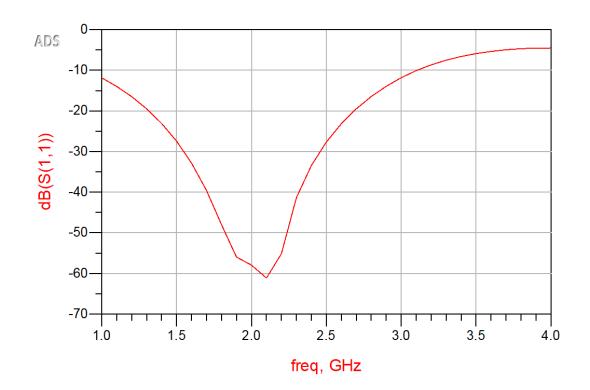
$$Z_1 = 59.53500000000000004$$

$$Z_2 = 100$$

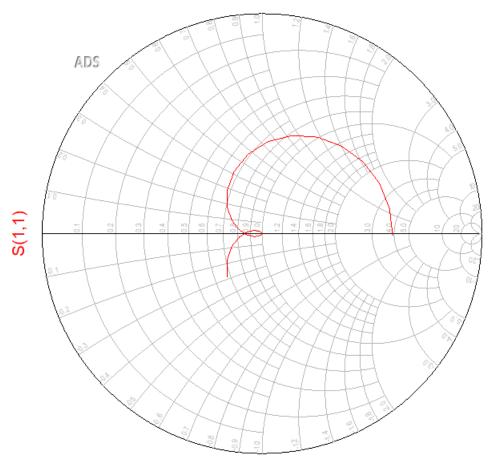
$$Z_3 = 167.97$$

Usando estes valores no ADS temos:





freq	Zin1	S(1,1)	abs(S(1,1))
1.000 GHz 1.100 GHz 1.200 GHz 1.200 GHz 1.300 GHz 1.400 GHz 1.500 GHz 1.500 GHz 1.600 GHz 1.700 GHz 1.800 GHz 2.000 GHz 2.000 GHz 2.000 GHz 2.100 GHz 2.200 GHz 2.300 GHz 2.300 GHz 2.400 GHz 2.500 GHz 2.500 GHz 2.500 GHz 2.500 GHz 2.100 GHz 3.100 GHz 3.100 GHz 3.100 GHz 3.200 GHz 3.200 GHz 3.200 GHz 3.300 GHz 3.300 GHz 3.300 GHz 3.300 GHz 3.300 GHz 3.400 GHz 3.500 GHz 3.500 GHz 3.500 GHz 3.500 GHz 3.500 GHz 3.700 GHz 3.900 GHz 3.900 GHz	36.690 / 36.400 / 37.913 / 40.532 / 43.499 / 46.167 / 48.128 / 49.282 / 49.792 / 49.987 / 50.007 / 49.899 / 49.405 / 48.237 / 46.239 / 40.545 / 37.938 / 36.473 / 36.473 / 36.473 / 36.473 / 39.494 / 44.606 / 52.256 / 62.677 / 76.453 / 94.641 / 118.710 / 149.399 / 181.543 / 196.058 /	0.254 / -1 0.200 / -1 0.150 / -1 0.106 / -1 0.070 / 1 0.042 / 1 0.023 / 1 0.010 / 1 0.004 / 1 0.002 / 1 0.001 / 9 8.785E-4 0.002 / -1 0.009 / -1 0.009 / -1 0.041 / -1 0.106 / 1 0.106 / 1 0.255 / 1 0.255 / 1 0.312 / 1 0.367 / 1 0.367 / 1 0.419 / 8 0.466 / 7 0.506 / 6 0.538 / 4 0.508 / 3 0.581 / 2 0.594 / -0	0.254 0.200 0.150 0.106 0.070 0.042 0.023 0.010 0.004 0.002 0.001 0.002 0.009 0.021 0.041 0.069 0.106 0.150 0.200 0.255 0.312 0.367 0.419 0.466 0.538 0.563 0.591 0.594



freq (1.000GHz to 4.000GHz)

# $2.1 \quad Para \; RL = 100 + 50J$

```
[6]: output = ""

Z_0 = 50 #Characteristic impedance of line

R_L = 100 + 50j #Load impedance

Z_line = np.sqrt(Z_0*R_L) #impedance od microstrip line
e_r = 3.2
wavlen = c/f

output += f'$$Z_0 = {Z_0} \Omega$$\n'
output += f'$$R_L = {R_L} \Omega$$\n'
output += f'$$Z_{{line}} = \sqrt{{Z_{{0}}} \cdot R_{{L}}}} = {Z_line:.

--Of}\Omega$$\n'

output += f'$$\lambda = {\wavlen:.3f}m$$\n'
md(output)
```

[6]:

$$Z_0 = 50\Omega$$
 
$$R_L = (100 + 50j)\Omega$$
 
$$Z_{line} = \sqrt{Z_0 \cdot R_L} = 73 + 17j\Omega$$
 
$$\lambda = 0.150m$$