



Faculty of Engineering – Cairo University Electronics and Electrical Communications Engineering Department

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Assignment #1

Space Time Block Codes

Submitted to **Dr. Mai Kafafy**

Submitted by

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Hand Analysis - Decoding the received signal

• STBC A

→ We have 3 antennas send over 4 time slots, and each time slot the Rx receives 1 symbol:

Tx Antenna 1	s_1	$-s_2^*$	0	0
Tx Antenna 2	s_2	s_1^*	0	0
Tx Antenna 3	0	0	s_1	$-s_2^*$
	t_1	t_2	t_3	t_4

→ Since there are 2 symbols sent over 4 time slots:

$$Rate = \frac{\#symbols}{\#time\ slots} = \frac{2}{4} = \frac{1}{2}$$

- \rightarrow Let the channel between each transmitter antenna and the receiver antenna is h_1 , h_2 , h_3 .
- → Now, get the received symbol in terms of transmitted signal and the channel:

$$y_{1} = s_{1}h_{1} + s_{2}h_{2} \rightarrow y_{1} = s_{1}h_{1} + s_{2}h_{2}$$

$$y_{2} = -s_{2}^{*}h_{1} + s_{1}^{*}h_{2} \rightarrow y_{2}^{*} = s_{1}h_{2}^{*} - s_{2}h_{1}^{*}$$

$$y_{3} = s_{1}h_{3} \rightarrow y_{3} = s_{1}h_{3}$$

$$y_{4} = -s_{2}^{*}h_{3} \rightarrow y_{4}^{*} = -s_{2}h_{3}^{*}$$

→ Write the matrix representation of the equations:

$$\begin{bmatrix} y_1 \\ y_2^* \\ y_3 \\ y_4^* \end{bmatrix} = \begin{bmatrix} h_1 & h_2 \\ h_2^* & -h_1^* \\ h_3 & 0 \\ 0 & -h_3^* \end{bmatrix} \begin{bmatrix} s_1 \\ s_2 \end{bmatrix} + \begin{bmatrix} n_1 \\ n_2 \\ n_3 \\ n_4 \end{bmatrix}$$

→ Decode the received signal:

$$s_{1} : \frac{C_{1}^{*T}}{\|C_{1}\|^{2}} * (C_{1} \quad C_{2}) {s_{1} \choose s_{2}} + \underline{n} * \frac{C_{1}^{*T}}{\|C_{1}\|^{2}} = \frac{\|C_{1}\|^{2}}{\|C_{1}\|^{2}} * s_{1} + 0 + \underline{n} * \frac{C_{1}^{*T}}{\|C_{1}\|^{2}} = \boxed{s_{1} + \underline{n} \cdot \frac{C_{1}^{*T}}{\|C_{1}\|^{2}}}$$

$$s_{2} : \frac{C_{2}^{*T}}{\|C_{2}\|^{2}} * (C_{1} \quad C_{2}) {s_{1} \choose s_{2}} + \underline{n} * \frac{C_{2}^{*T}}{\|C_{2}\|^{2}} = \frac{\|C_{2}\|^{2}}{\|C_{2}\|^{2}} * s_{2} + 0 + \underline{n} * \frac{C_{2}^{*T}}{\|C_{2}\|^{2}} = \boxed{s_{2} + \underline{n} \cdot \frac{C_{2}^{*T}}{\|C_{2}\|^{2}}}$$

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Where: $C_1 \& C_2$ are the first and second columns of the channel matrix.

• STBC B

→ Since there are 3 symbols sent over 4 time slots:

$$Rate = \frac{\#symbols}{\#time\ slots} = \frac{3}{4}$$

- ⇒ Rate in this case is higher, leading to higher BER.
- \rightarrow Let the channel between each transmitter antenna and the receiver antenna is h_1 , h_2 , h_3 .
- → Now, get the received symbol in terms of transmitted signal and the channel:

$$y_{1} = s_{1}h_{1} + s_{2}h_{2} + s_{3}h_{3} \rightarrow y_{1} = s_{1}h_{1} + s_{2}h_{2} + s_{3}h_{3}$$

$$y_{2} = -s_{2}^{*}h_{1} + s_{1}^{*}h_{2} \rightarrow y_{2}^{*} = s_{1}h_{2}^{*} + s_{2}(-h_{1}^{*})$$

$$y_{3} = s_{3}^{*}h_{1} - s_{1}^{*}h_{3} \rightarrow y_{3}^{*} = s_{1}(-h_{3}^{*}) + s_{3}h_{1}^{*}$$

$$y_{4} = s_{3}^{*}h_{2} - s_{2}^{*}h_{3} \rightarrow y_{4}^{*} = s_{2}(-h_{3}^{*}) + s_{3}h_{2}^{*}$$

→ Write the matrix representation of the equations:

$$\begin{bmatrix} y_1 \\ y_2^* \\ y_3^* \\ y_4^* \end{bmatrix} = \begin{bmatrix} h_1 & h_2 & h_3 \\ h_2^* & -h_1^* & 0 \\ -h_3^* & 0 & h_1^* \\ 0 & -h_3^* & h_2^* \end{bmatrix} \begin{bmatrix} s_1 \\ s_2 \\ s_3 \end{bmatrix} + \begin{bmatrix} n_1 \\ n_2 \\ n_3 \\ n_4 \end{bmatrix}$$

→ Decode the received signal:

$$s_{1} : \frac{C_{1}^{*T}}{\|C_{1}\|^{2}} * (C_{1} \quad C_{2} \quad C_{3}) \begin{pmatrix} s_{1} \\ s_{2} \\ s_{3} \end{pmatrix} + \underline{n} * \frac{C_{1}^{*T}}{\|C_{1}\|^{2}} = \frac{\|C_{1}\|^{2}}{\|C_{1}\|^{2}} * s_{1} + 0 + \underline{n} * \frac{C_{1}^{*T}}{\|C_{1}\|^{2}} = \boxed{s_{1} + \underline{n} \cdot \frac{C_{1}^{*T}}{\|C_{1}\|^{2}}}$$

$$s_{2} : \frac{C_{2}^{*T}}{\|C_{2}\|^{2}} * (C_{1} \quad C_{2} \quad C_{3}) \begin{pmatrix} s_{1} \\ s_{2} \\ s_{3} \end{pmatrix} + \underline{n} * \frac{C_{2}^{*T}}{\|C_{2}\|^{2}} = \frac{\|C_{2}\|^{2}}{\|C_{2}\|^{2}} * s_{2} + 0 + \underline{n} * \frac{C_{2}^{*T}}{\|C_{2}\|^{2}} = \boxed{s_{2} + \underline{n} \cdot \frac{C_{2}^{*T}}{\|C_{2}\|^{2}}}$$

$$s_{3} : \frac{C_{3}^{*T}}{\|C_{3}\|^{2}} * (C_{1} \quad C_{2} \quad C_{3}) \begin{pmatrix} s_{1} \\ s_{2} \\ s_{3} \end{pmatrix} + \underline{n} * \frac{C_{3}^{*T}}{\|C_{3}\|^{2}} = \frac{\|C_{3}\|^{2}}{\|C_{3}\|^{2}} * s_{3} + 0 + \underline{n} * \frac{C_{3}^{*T}}{\|C_{3}\|^{2}} = \boxed{s_{3} + \underline{n} \cdot \frac{C_{3}^{*T}}{\|C_{3}\|^{2}}}$$

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Where: C_1 , C_2 , and C_3 are the first and second columns of the channel matrix.

Questions Answers

- Which of the two codes could be considered an Orthogonal STBC? And why?
 - → Both of them are orthogonal.
- ⇒ Check orthogonality of STBC (A):

$$\Rightarrow egin{bmatrix} h_1 & h_2 \ h_2^* & -h_1^* \ h_3 & 0 \ 0 & -h_3^* \end{bmatrix}$$

$$C_1^{*T}C_2 = (h_1^* \quad h_2 \quad h_3^* \quad 0) * \begin{pmatrix} h_2 \\ -h_1^* \\ 0 \\ -h_2^* \end{pmatrix} = h_1^*h_2 - h_2h_1^* + 0 + 0 = 0$$

- : STBC (A) is orthogonal.
- ⇒ Check orthogonality of STBC (B):

$$\Rightarrow egin{bmatrix} h_1 & h_2 & h_3 \ h_2^* & -h_1^* & 0 \ -h_3^* & 0 & h_1^* \ 0 & -h_3^* & h_2^* \end{bmatrix}$$

$$C_1^{*T}C_2 = (h_1^* \quad h_2 \quad -h_3 \quad 0) * \begin{pmatrix} h_2 \\ -h_1^* \\ 0 \\ -h_3^* \end{pmatrix} = h_1^*h_2 - h_2h_1^* - 0 - 0 = 0$$

$$C_2^{*T}C_3 = (h_2^* - h_1 \quad 0 \quad -h_3) * \begin{pmatrix} h_3 \\ 0 \\ h_1^* \\ h_2^* \end{pmatrix} = h_2^*h_3 - 0 + 0 - h_3h_2^* = 0$$

$$C_1^{*T}C_3 = (h_1^* \quad h_2 \quad -h_3 \quad 0) * \begin{pmatrix} h_3 \\ 0 \\ h_1^* \\ h_2^* \end{pmatrix} = h_1^*h_3 + 0 - h_3h_1^* + 0 = 0$$

- \therefore STBC (A) is orthogonal.
- How can you decode the received signal of the two codes.
 - \rightarrow Decoding is done in the hand analysis.
- Can the receiver decode the received symbols without knowing the channel? Why?
 - → Yes, Because the STBC first estimates the channel using pilot symbols sent before the data, then it assumes the channel to be static during the total period used to transmit the symbols by making sure that all 4 time slots are sent within the same coherence

time of the channel, hence channel parameters are already assumed to be known to the receiver.

• Performance Metrics Comparison

Metric / Code-word	STBC (A)	STBC (B)	
Divorsity Order	3 (since each symbol hits all 3	2 (less spatial redundancy due	
Diversity Order	antennas across time)	to symbol overlap)	
Average symbol rate	$Rate = \frac{1}{2} \text{ symbols/timeslot}$ (4 time slots for 2 symbols)	$Rate = \frac{3}{4} \text{ symbols/timeslot}$ (4 time slots for 3 symbols)	

Diversity Gain

- → The SNR values are calculated for each code at the same BER, then subtracted from each other to get the diversity gain.
- \rightarrow Diversity gain = SNR (A) SNR (B)

```
Diversity Gain at BER = 1.0e-04: 2.03 dB
Diversity Gain at BER = 1.0e-05: 2.54 dB
```

Figure 1 - Diversity gain calculated with MATLAB.

- → The STBC (A) has higher SNR than the STBC (B) at the same BER, so it has higher diversity gain.
- → Diversity gain is chosen to be calculated at relatively higher SNR values (lower BER) where fading is the dominant error cause and the slope directly reflects the diversity order.

• Which STBC code would you use in a power-limited scenario? And why?

- \rightarrow Use STBC (A).
- → Because in power-limited scenarios, we must prioritize reliability over data rate. This means achieving low BER with low SNR to stay within the power budget. Even though this will cause BW inefficiency due to low rate of the system.
- → STBC (A) gives full diversity and simpler decoding, which improves error performance at low SNR.

Results of system simulation

• BER Comparison

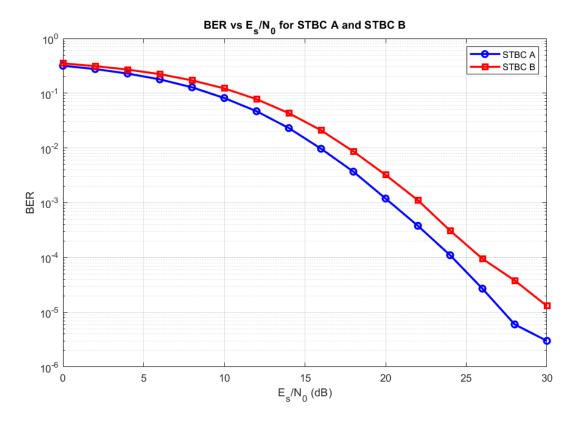


Figure 2 - BER Comparison between STBC (A) & STBC (B).

Comment

- → BER of STBC A is lower (better) than BER of STBC B.
- → We use the same total power for both codes, and since STBC (A) transmits only 2 symbols over 4 timeslots while STBC (B) transmits 3 symbols, this means STBC (A) has higher energy per symbol which makes the symbol more robust to noise and fading, and thus achieving lower BER.
- → Also, STBC (A) has better diversity utilization, because each symbol is sent over all available channel paths at each timeslot.

• Diversity order verification for code A

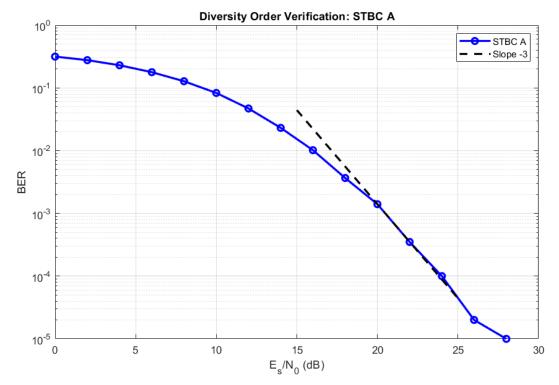


Figure 3 - Diversity order verification for STBC (A) when number of bits is 1e5.

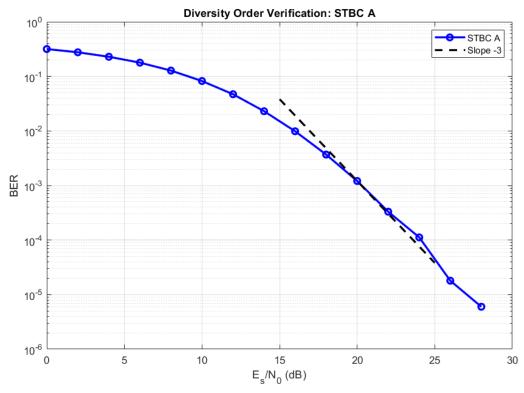


Figure 4 - Diversity order verification for STBC (B) with number of bits (1e6).

Observation:

The slope of the BER curve of STBC A approximately matches the -3 slope at higher SNR, validating the diversity order = 3.

• Diversity order verification for code B

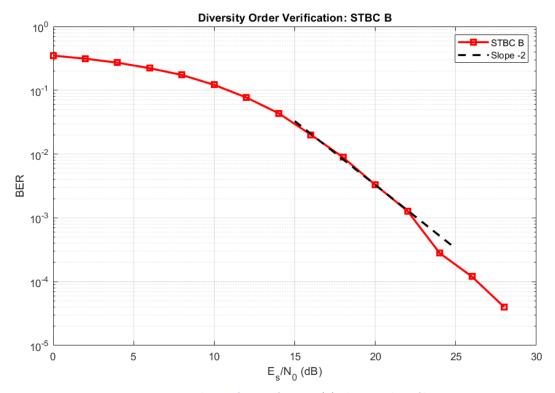


Figure 5 - Diversity order verification for STBC (B) when number of bits is 1e5.

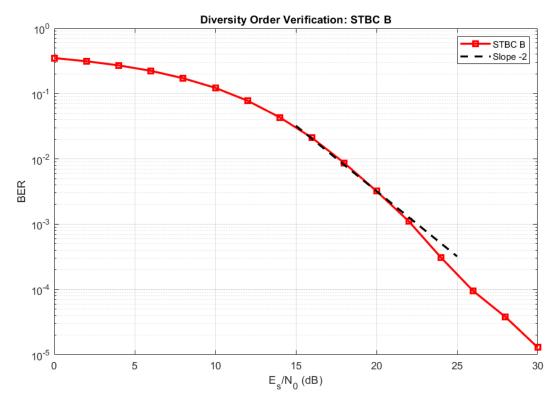


Figure 6 - Diversity order verification for STBC (B) with higher number of bits (1e6).

Observation:

The slope of the BER curve of STBC B approximately matches the -2 slope, validating the diversity order = 2.

The MATLAB Code

```
clear all; close all; clc;
%% Simulation Parameters
N = 1e6;
                           % Number of bits
M = 4;
                          % QPSK modulation
k = log2(M);
                           % Bits per symbol
A = 1;
                           % QPSK Amplitude
Eb = A^2;
Es = 2 * Eb;
                           % QPSK: 2 bits/symbol
numSymbols = N/k;
EsN0 dB = 0:2:30; % Es/N0 in dB
EsN0 lin = 10.^(EsN0 dB/10); % Es/N0 linear
%% QPSK Modulation
qpskMod = comm.QPSKModulator('BitInput', true);
qpskDemod = comm.QPSKDemodulator('BitOutput', true);
%% STBC Encoding Functions
encodeSTBC A = @(s) [
  s(1) s(2) 0;
   -\text{conj}(s(2)) \text{ conj}(s(1)) 0;
          0
                   s(1);
                  -conj(s(2))
   0
           0
];
encodeSTBC_B = @(s) [
  s(1) s(2) s(3);
  -\text{conj}(s(2)) \text{ conj}(s(1)) 0;
   conj(s(3)) 0 -conj(s(1));
       conj(s(3)) - conj(s(2))
];
%% BER Initialization
ber STBC A = zeros(size(EsN0 dB));
ber STBC B = zeros(size(EsN0 dB));
%% Simulation Loop
for idx = 1:length(EsN0 dB)
   errors STBC A = 0;
   errors STBC B = 0;
   totalBits = 0;
   while totalBits < N
       %% Generate random bits and QPSK modulation
       txBits = randi([0 1], N, 1);
       txSym = qpskMod(txBits);
       %% Split into blocks
       numBlocks A = floor(length(txSym)/2);
       numBlocks B = floor(length(txSym)/3);
       %% STBC A Encoding
       txSym A = txSym(1:2*numBlocks A);
```

```
txSym A = reshape(txSym A, 2, numBlocks A);
        encoded A = zeros(4, numAntennas, numBlocks A);
        for n = 1:numBlocks A
            block = encodeSTBC A(txSym A(:,n));
            block = block / sqrt(norm(block, 'fro')^2);
            encoded A(:,:,n) = block;
        end
        %% STBC B Encoding
        txSym B = txSym(1:3*numBlocks B);
        txSym B = reshape(txSym B, 3, numBlocks B);
        encoded B = zeros(4, numAntennas, numBlocks B);
        for n = 1:numBlocks B
            block = encodeSTBC B(txSym B(:,n));
            block = block / sqrt(norm(block, 'fro')^2);
            encoded B(:,:,n) = block;
        end
        %% Improved Block Rayleigh Flat Fading Channel
        H A = (randn(numAntennas, numBlocks A) + 1i*randn(numAntennas,
numBlocks A)) / sqrt(2);
        H B = (randn(numAntennas, numBlocks B) + 1i*randn(numAntennas,
numBlocks B)) / sqrt(2);
        %% AWGN Noise
        N0 = Es ./ EsN0 lin(idx);
        %% Transmission - STBC A
        rx A = zeros(4, 1, numBlocks A);
        for n = 1:numBlocks A
                                              % 4x3
            X = \text{squeeze}(\text{encoded }A(:,:,n));
            h = H A(:,n);
                                                % 3x1
            noise = sqrt(N0/2) * (randn(4,1) + 1i*randn(4,1));
            rx A(:,:,n) = X * h + noise;
                                               % 4x1
        end
        %% Transmission - STBC B
        rx B = zeros(4, 1, numBlocks B);
        for n = 1:numBlocks B
                                              % 4x3
            X = squeeze(encoded B(:,:,n));
            h = H B(:,n);
                                                % 3x1
            noise = sqrt(N0/2) * (randn(4,1) + 1i*randn(4,1));
            rx B(:,:,n) = X * h + noise;
        end
        %% STBC A Decoding
        decSym A = zeros(2, numBlocks A);
        for n = 1:numBlocks_A
            h = H A(:,n);
            y = rx A(:,:,n);
            y \text{ vec} = [y(1); conj(y(2)); y(3); conj(y(4))];
            C1 = [h(1); conj(h(2)); h(3); 0];
            C2 = [h(2); -conj(h(1)); 0; -conj(h(3))];
            s1 hat = (C1' * y_vec) / (norm(C1)^2);
            s2_{hat} = (C2' * y_{vec}) / (norm(C2)^2);
```

```
decSym A(:,n) = [s1 hat; s2 hat];
        decSym A = decSym A(:);
        %% STBC B Decoding
        decSym B = zeros(3, numBlocks B);
        for n = 1:numBlocks B
            h = H_B(:,n);
            y = rx B(:,:,n);
            y_{vec} = [y(1); conj(y(2)); conj(y(3)); conj(y(4))];
            C1 = [h(1); conj(h(2)); -conj(h(3)); 0];
            C2 = [h(2); -conj(h(1)); 0; -conj(h(3))];
            C3 = [h(3); 0; conj(h(1)); conj(h(2))];
            s1 hat = (C1'*y vec) / (norm(C1)^2);
            s2_{hat} = (C2'*y_{vec}) / (norm(C2)^2);
            s3 hat = (C3'*y vec) / (norm(C3)^2);
            decSym B(:,n) = [s1 hat; s2 hat; s3 hat];
        end
        decSym B = decSym B(:);
        %% QPSK Demodulation
        rxBits A = qpskDemod(decSym A);
        rxBits B = qpskDemod(decSym B);
        %% BER Calculation
        numBits A = min(length(txBits), length(rxBits A));
        numBits B = min(length(txBits), length(rxBits B));
        errors STBC A = errors STBC A + sum(txBits(1:numBits A) ~=
rxBits A(1:numBits A));
        errors STBC B = errors STBC B + sum(txBits(1:numBits B) ~=
rxBits_B(1:numBits B));
        totalBits = totalBits + numBits_A;
    end
   ber STBC A(idx) = errors STBC A / N;
   ber STBC B(idx) = errors STBC B / N;
    fprintf('Es/N0 = %2d dB: STBC A BER = %.4e | STBC B BER = %.4e\n', ...
            EsNO dB(idx), ber STBC A(idx), ber STBC B(idx));
end
%% Diversity Gain
% Target BER points
targetBERs = [1e-4, 1e-5];
% Loop over targets
for i = 1:length(targetBERs)
   targetBER = targetBERs(i);
    % Find valid index ranges for interpolation (BER > 0 and BER > target)
    validIdx A = (ber STBC A > 0) & (ber STBC A >= targetBER);
    validIdx B = (ber STBC B > 0) & (ber STBC B >= targetBER);
```

```
if sum(validIdx A) < 2 || sum(validIdx B) < 2</pre>
        fprintf('Not enough data to interpolate at BER = %.1e\n', targetBER);
        continue;
    end
    % Sort valid points in increasing BER order (x-axis for interp1 must be
monotonic)
    [berA sorted, idxA] = sort(ber STBC A(validIdx A), 'descend');
    snrA_sorted = EsN0_dB(validIdx_A);
    snrA sorted = snrA sorted(idxA);
    [berB sorted, idxB] = sort(ber STBC B(validIdx B), 'descend');
    snrB sorted = EsNO dB(validIdx B);
    snrB sorted = snrB sorted(idxB);
    % Interpolation
    try
        SNR A = interp1 (berA sorted, snrA sorted, targetBER, 'linear',
'extrap');
        SNR B = interp1(berB sorted, snrB sorted, targetBER, 'linear',
'extrap');
        diversityGain dB = SNR B - SNR A;
        fprintf('Diversity Gain at BER = %.1e: %.2f dB\n', targetBER,
diversityGain dB);
    catch ME
        fprintf('Interpolation failed at BER = %.1e: %s\n', targetBER,
ME.message);
    end
end
%% Plot
semilogy(EsNO dB, ber STBC A, 'b-o', 'LineWidth', 2); hold on;
semilogy (EsNO dB, ber STBC B, 'r-s', 'LineWidth', 2);
xlabel('E s/N 0 (dB)'); ylabel('BER'); grid on;
title('BER vs E s/N 0 for STBC A and STBC B');
legend('STBC A', 'STBC B');
%% Plot: Slope for Diversity Order Verification - STBC A
figure;
semilogy(EsNO dB, ber STBC A, 'b-o', 'LineWidth', 2); hold on;
snr ref A = 20;
ber ref A = interp1(EsN0 dB, ber STBC A, snr ref A);
ref_snr_A = [snr_ref_A-5, snr_ref_A+5];
ref ber A = ber ref A * 10.^{(-)}(ref snr A - snr ref A) * 3 / 10);
semilogy(ref snr A, ref ber A, '--k', 'LineWidth', 2);
xlabel('E s/N 0 (dB)');
ylabel('BER');
legend('STBC A', 'Slope -3');
title('Diversity Order Verification: STBC A');
grid on;
%% Plot: Slope for Diversity Order Verification - STBC B
semilogy(EsNO dB, ber STBC B, 'r-s', 'LineWidth', 2); hold on;
snr ref B = 20;
```

```
ber ref B = interp1(EsN0 dB, ber STBC B, snr ref B);
ref snr B = [snr ref B-5, snr ref B+5];
ref ber B = ber ref B * 10.^{(-(ref snr B - snr ref B) * 2 / 10)};
semilogy(ref snr B, ref ber B, '--k', 'LineWidth', 2);
xlabel('E s/N 0 (dB)');
ylabel('BER');
legend('STBC B', 'Slope -2');
title('Diversity Order Verification: STBC B');
grid on;
%% Combined Plot: Slope Lines for Diversity Order Verification
figure;
% STBC A data and slope line
semilogy(EsNO dB, ber STBC A, 'b-o', 'LineWidth', 2, 'DisplayName', 'STBC
A');
hold on;
% Reference point for STBC A
snr ref A = 20;
ber ref A = interp1(EsN0 dB, ber STBC A, snr ref A);
ref snr A = [snr ref A-5, snr ref A+5];
ref ber A = ber ref A * 10.^{(-(ref snr A - snr ref A) * 3 / 10)};
semilogy(ref snr A, ref ber A, '--b', 'LineWidth', 2, 'DisplayName', 'Slope -
3 (STBC A)');
% STBC B data and slope line
semilogy(EsNO dB, ber STBC B, 'r-s', 'LineWidth', 2, 'DisplayName', 'STBC
B');
% Reference point for STBC B
snr ref B = 20;
ber ref B = interp1(EsN0 dB, ber STBC B, snr ref B);
ref snr B = [snr ref B-5, snr ref B+5];
ref ber B = ber ref B * 10.^(-(ref snr B - snr ref B) * 2 / 10);
semilogy(ref snr B, ref ber B, '--r', 'LineWidth', 2, 'DisplayName', 'Slope -
2 (STBC B)');
% Plot formatting
xlabel('E s/N 0 (dB)', 'FontSize', 12);
ylabel('BER', 'FontSize', 12);
title('Diversity Order Verification: STBC A vs STBC B', 'FontSize', 12);
legend('Location', 'southwest', 'FontSize', 10);
grid on;
set(gca, 'FontSize', 10);
% Annotation to highlight diversity orders
text(18, 1e-3, 'Diversity Order = 3', 'Color', 'b', 'FontSize', 10);
text(18, 3e-2, 'Diversity Order = 2', 'Color', 'r', 'FontSize', 10);
```