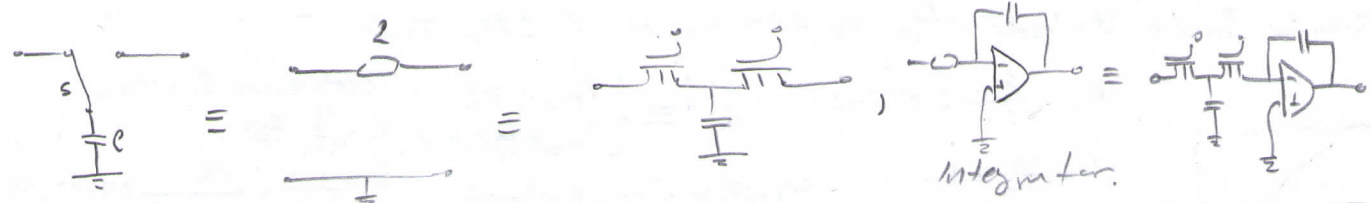


SC filter  
princip rada

$$\omega_0 = \frac{1}{RC} \Rightarrow 2\pi f_0 = \frac{1}{RC}$$

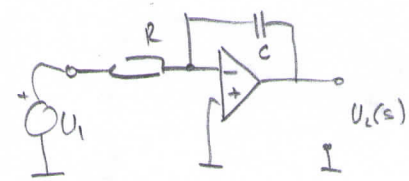
predlopljenje kapaciteta kontinuirnim prolazom naponski (staje)  
te se mogu izvesti otpornici, i ujednačiti



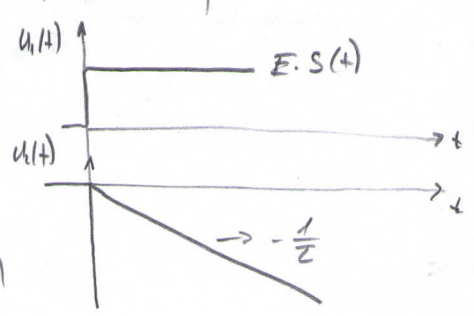
**Zadatak** Design of analog filters. Passive, Active, C, and Switched Capacitor [DESIGN OF ANALOG FILTERS]   
 Prije sklopa, Mohamed S. Chaoui, Kenneth R. Lohr  
 Potrebno osigurati nem. dostavu SC integratora o svoj. opterećenja te realizirati se filter

ponovni integrirajući kruzni HF10.

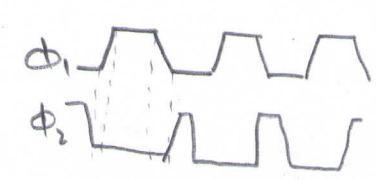
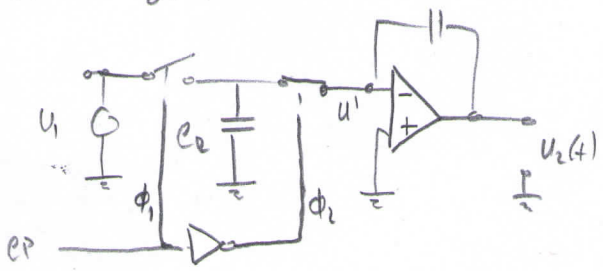
a) RC integrator



- pri. napet. = 0  
 $U_c(0) = U_2(0) = 0$   
 $I_1(s) = \frac{U_1(s)}{R}$   
 $U_2(s) = -I_1(s) \cdot \frac{1}{sC} = -\frac{U_1(s)}{R} \cdot \frac{1}{sC}$   
 $H(s) = \frac{U_2(s)}{U_1(s)} = -\frac{1}{sRC} = -\frac{1}{\tau} \cdot \frac{1}{s}$   
 $\tau = RC$

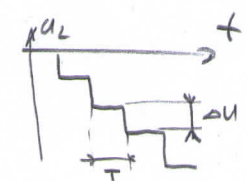


b) SC integrator

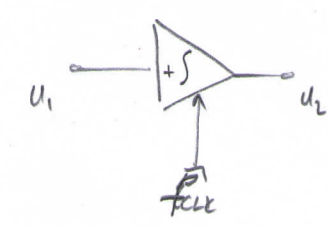
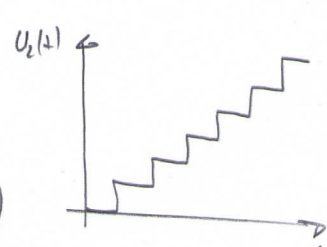
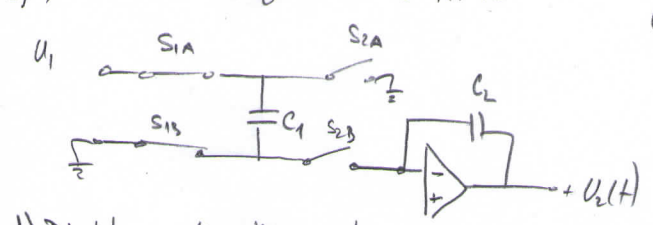


pon. k. des. se ve predlopljuju  
 $\Delta Q = E \cdot C_2$   
 $\Delta U = \frac{\Delta Q}{C_2} = E \frac{\Delta P}{C_2}$

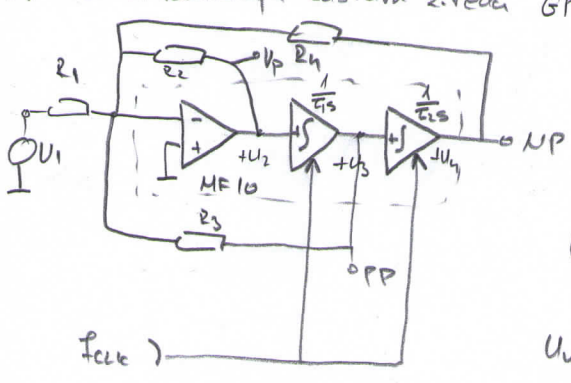
$i = \frac{\Delta Q}{\Delta t} = \frac{E \cdot C_2}{T} = E C_2 \cdot f_{clk}$   
 $I = \frac{E}{T} = \frac{E}{EC_2 f_{clk}} = \frac{1}{C_2 f_{clk}}$   
 $\tau = RC_2 = \frac{C_2}{C_2 f_{clk}}$



c) pozitivni integrator u HF10



d) Direktna realizacija sustava 2. reda GP sekcija (general purpose)



$\tau_1 = \frac{C_2}{f_{clk}}$   
 $\tau_2 = \frac{C_4}{f_{clk}}$   
 $U_3 = \frac{1}{\tau_1 s} U_2$   
 $U_4 = \frac{1}{\tau_2 s} U_3$   
 $U_2 = -R_2 \left[ \frac{U_1}{R_1} + \frac{U_3}{R_3} + \frac{U_4}{R_4} \right]$   
 $\frac{U_4}{U_1} = \frac{-\frac{R_2}{R_1}}{s^2 \left( \frac{R_4}{R_2} \tau_1 \tau_2 \right) + s \left( \frac{R_4}{R_3} \tau_1 \tau_2 \right) + 1} = \frac{K}{\omega_0^2 + \frac{s}{\omega_0 Q} + 1}$   
 $\omega_0 = \sqrt{\frac{R_2}{R_4}} \sqrt{\frac{C_1 C_3}{C_2 C_4}} \cdot f_{clk}$   
 $Q = \sqrt{\frac{R_2}{R_4}} \cdot \frac{R_3}{R_2} \sqrt{\frac{C_2 C_3}{C_1 C_4}}$   
 $\omega_0 = \sqrt{\frac{R_2}{R_4}} \frac{1}{\sqrt{\tau_1 \tau_2}}$   
 $Q = \sqrt{\frac{R_2}{R_4}} \frac{R_3}{R_2} \sqrt{\frac{\tau_1}{\tau_2}}$

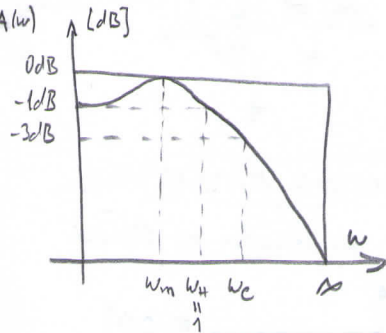
Planarnom tehnologijom na Si moguće je postići vrlo precizno omjer kapaciteta. Proizvođač čipa HF10 je odabrao sljedeći omjer kapaciteta  $C_2:C_1=C_4:C_3=50:1$  i tako mijenja sljedeće formule koje se daju za LP(UP):

$\omega_0 = \sqrt{\frac{R_2}{R_4}} \frac{f_{clk}}{50}$ ,  $K_{LP} = -\frac{R_4}{R_1}$   
 $Q = \sqrt{\frac{R_2}{R_4}} \cdot \frac{R_3}{R_2}$

(10) SC filter

Zadatak: Redizajnirati pasivni integrirani kruzni MF10 čelničevljar NP filter 2. reda sa max. volu'tos'ti u podnožju propu'stanja 1dB i sa graničnom freq. 10kHz ako je freq. otklona f<sub>clk</sub> = 500kHz. U realizaciji ćemo koristiti jednu bičvadratnu sekciju. (za max. cheby, cheby apt, butter---)

Tablice: Tomislav Boudic "Analogna integrirana Elektronika" str. 276., Tablica 5.



$$w_w = \frac{w_m}{w_#} = 0.707$$

$$\left. \begin{matrix} w_0 \\ Q \end{matrix} \right\} \equiv \left\{ \begin{matrix} R_e = 0.5490 \\ I_m = 0.8953206 \end{matrix} \right.$$

Odobrimo  $R_1 = 10k\Omega$

$$w_0 = \sqrt{\frac{R_2}{R_1}} \cdot \frac{f_{clk}}{50}$$

$$w_# = \frac{w_#}{w_#} = 1$$

$$w_# = 2\pi \cdot 10 \cdot 10^3 Hz = 2\pi \cdot 10^4 rad/s$$

$$\sqrt{\frac{R_2}{R_1}} = \frac{50}{f_{clk} \cdot 10^3} = \frac{50}{500 \cdot 10^3} = 6.597344$$

$$w_c = \frac{w_c}{w_#} = 1.308$$

$$w_0 = 1.05 \cdot 2\pi \cdot 10^4 = 65973.44 rad/s$$

$$= 6.597344$$

$$w_0 = \frac{w_0}{w_#} = 1.05000$$

$$Q = 0.95652$$

$$R_2 = \left( \sqrt{\frac{R_2}{R_1}} \right)^2 R_1 = 6.597344^2 \cdot 10^4 = 4352\Omega$$

$$Q = 0.95652$$

$$k_L = 1 - \frac{R_1}{R_2} \Rightarrow R_1 = 10k\Omega$$

$$Q = \sqrt{\frac{R_2}{R_1}} \cdot \frac{1}{R_2} \Rightarrow R_2 = R_1 \cdot \frac{Q}{\sqrt{\frac{R_2}{R_1}}} = 4352 \cdot \frac{0.95652}{6.597344} = 632\Omega$$