

(120 min.) DUS - M1 - 2016/17

157RE:

- [5] **1** NAPIŠATI IZRAZE MATEM. MODELA SHUPM U dg SUSTAVU, U RELATIVNIM JEDINICAMA (PER UNIT) ZA ELEKTRIČ. VELIČINE ($\frac{di_d}{dt}$ i $\frac{di_g}{dt}$). BAZNE VRIJEDNOSTI SU:
- $U_a = \sqrt{2} \underline{U}_{fn}$, $I_a = \sqrt{2} \underline{I}_{fn}$, $\omega_e = 2\pi f_n = \omega_{eb}$
- $\underline{L}_{eff. \text{ vrij. } \text{lozuvaj. napona}}$, $\underline{L}_{eff. \text{ vrij. } \text{lozne struje}}$
- \sum NAZIVNA VRIJEDNOST ELEKTRIČ. BAZNE VRTNE

R_{do}: (1) $\frac{di_d}{dt} = \frac{1}{L_d} (u_d - R_{sid} i_d + \omega_e L_g i_g)$

(2) $\frac{di_g}{dt} = \frac{1}{L_g} (u_g - R_{sig} i_g - \omega_e L_d i_d - \omega_e \phi_{mg})$

$i_{dpu} = \frac{i_d}{I_B}$ | $L_{dpu} = \frac{L_d}{L_B}$; $L_B = \frac{Z_B}{\omega_B}$ | $R_{spu} = \frac{R_s}{Z_B}$

$\omega_{epu} = \frac{\omega_e}{\omega_B}$ | $\phi_{mgpu} = \frac{\phi_{mg}}{\phi_B} = \frac{\phi_B}{\phi_B} = \frac{U_B}{\omega_B}$ | $u_{dpu} = \frac{u_d}{U_B}$

(1) $I_B \cdot \frac{di_{dpu}}{dt} = \frac{\omega_B}{Z_B \cdot L_{dpu}} \left[u_{dpu} \cdot U_B - R_{spu} Z_B \cdot i_{dpu} \cdot I_B + \omega_{epu} \omega_B \cdot L_{gpu} \cdot \frac{Z_B}{\omega_B} \cdot i_{gpu} \cdot I_B \right]$

$\frac{di_{dpu}}{dt} = \frac{\omega_B}{Z_B L_{dpu}} \left[u_{dpu} \cdot Z_B - i_{dpu} R_{spu} \cdot Z_B + \omega_{epu} L_{gpu} i_{gpu} \cdot Z_B \right]$

$\frac{di_{dpu}}{dt} = \frac{\omega_B}{L_{dpu}} \left[u_{dpu} - i_{dpu} R_{spu} + \omega_{epu} L_{gpu} i_{gpu} \right]$

$$(2) \frac{di_g}{dt} = \frac{1}{L_g} (u_g - R_s i_g - \omega_e L_d i_d - \omega_e \phi_{mg})$$

$$I_B \cdot \frac{di_{gpv}}{dt} = \frac{\omega_B}{Z_B L_{gpv}} \left[U_B u_{gpv} - Z_B R_{spv} i_{gpv} I_B - \right. \\ \left. - \omega_B \omega_{epv} \frac{Z_B L_{dpv}}{\omega_B} I_B i_{dpv} - \omega_B \omega_{epv} \frac{U_B \phi_{mgpv}}{\omega_B} \right]$$

$$\frac{di_{gpv}}{dt} = \frac{\omega_B}{L_{gpv}} \left[u_{gpv} - R_{spv} i_{gpv} - \omega_{epv} L_{dpv} i_{dpv} - \omega_{epv} \phi_{mgpv} \right]$$

VERZJA HOJA BI SE MOGLA POJAVITI NA ISPITU?

ODREDI $\frac{d\omega_m}{dt}$ U P.V., SMUPM :

$$\frac{P_{J.}}{(3)} J_m \frac{d\omega_m}{dt} = M_{em} - D_{wm} - M_t \\ = \frac{3}{2} p \left[\phi_{mg} i_g - (L_d - L_g) i_d i_g \right] \quad \begin{matrix} \text{ZA} \\ \text{SMUPM} \end{matrix} \\ = P \text{ ZA SMUPM}$$

$$J_m \omega_B \frac{d\omega_{mpv}}{dt} = M_B M_{empv} - D_B D_{pv} \omega_{mB} \omega_{mpv} - M_B M_{tpv}$$

$$J_m \omega_{mB} \frac{d\omega_{mpv}}{dt} = M_B M_{empv} - \frac{M_B}{\omega_B} \cdot \frac{\omega_B}{p} \cdot D_{pv} \omega_{mpv} - M_B M_{tpv}$$

$$\frac{d\omega_{mpv}}{dt} = \frac{M_{ts}}{J_m \omega_{mB}} \left[M_{empv} - \frac{D_{pv} \omega_{mpv}}{p} - M_{tpv} \right]$$

$\frac{d\omega_{mpv}}{dt} \rightarrow$ SMO SE POMNOŽI DEJNA S 'p'

② [5] SINHRONI STROJ S PERMAN. MAGN. 1 MA SLJ. PARAMETRE:
 $p=2$ (BROJ PARI POLOVA), $L_d = 7 \cdot 10^{-3} \text{ H}$ (INDUKTIVITET ϕ D OSI),

$L_q = 7 \cdot 10^{-3} \text{ H}$ (IND. UZ OSI), $R_s = 2.98 \Omega$ (OTPOR STATORA),

$\Phi_{mg} = 0.125 \text{ Wb}$ (MAGNETOG PERMAN. MAGN.), $J_m = 0.47 \cdot 10^{-4} \text{ kgm}^2$ (MOMENT INERCIJE),

$D = 1.1 \cdot 10^{-4} \text{ Nms}$ (FAKTOR PRIGUŠENJA).

ODREDITI POJAČANJE REGULATORA STRUJE i_g AKO JE STAC.

POJAČANJE ZATVORENOG KRUGA: (a) $\alpha = 0.4$

P tip!

(b) $\alpha = 0.9$

Rj: $L_d = L_q \Rightarrow \text{SMVPM!}$

$$\frac{di_g}{dt} = \frac{1}{L_g} (u_g - R_s i_g - \omega_e L_d i_d - \omega_e \Phi_{mg})$$

$$\text{SVPST: } \hat{u}_g = u_g - R_e L_d \hat{i}_d - R_e \Phi_{mg}$$

$$s I_g = \frac{1}{L_g} \hat{u}_g - \frac{R_s}{L_g} I_g$$

P tip =

$$\frac{I_g}{\hat{u}_g} = \frac{1}{s + R_s/L_g} ; \quad \hat{u}_g = K_c^2 (I_g^* - I_g)$$

$$s I_g = \frac{K_c^2}{L_g} (I_g^* - I_g) - \frac{R_s}{L_g} I_g$$

$$\frac{I_g}{I_g^*} = \frac{K_c^2 / L_g}{s + \frac{K_c^2 + R_s}{L_g}}$$

$$\frac{I_g}{I_g^*} = \frac{k_c^2 M}{sLg + k_c^2 + R_s}$$

$$\alpha = ? \Rightarrow \alpha = \lim_{s \rightarrow 0} s I_g \Big|_{\substack{I_g^* = 1 \\ \frac{1}{s} = 5}} = \frac{k_c^2}{k_c^2 + R_s}$$

$$k_c^2 = k_c^2 \alpha + R_s \alpha$$

$$k_c^2 = \frac{R_s \alpha}{(1 - \alpha)}$$

$$(a) \alpha = 0.4 \Rightarrow k_c^2 = 1.99$$

$$(b) \alpha = 0.9 \Rightarrow k_c^2 = 26.82 //$$

3. [6]

ODREDI IZRAZE ZA PARAMETRE PID REG. POZICIJE (k_c, τ_i, τ_d) SMVPM. PRIJ. FJA PID REG. JE $C(s) = k_c (1 + \frac{1}{\tau_i s} + \tau_d s)$.
REGULATOR STRUJE i_g JE P TIPIA (KAO U ZAD. 2.)

RJ: $\sigma_e = \frac{1}{s} \Sigma e$ SMVPM ($L_d = L_g$)

$$s \Sigma e = \frac{1}{J_m} \left\{ \frac{3}{2} n \phi_{mg} \bar{i}_g + \underbrace{(L_d - L_g) \frac{d \bar{i}_g}{dt}}_{SMVPM, = 0} - \frac{D}{n} \underbrace{\bar{\omega}_e - \bar{\omega}_k}_{= 0} \right\}$$

$$s \Sigma e = \frac{1}{J_m} \cdot \frac{3}{2} n \phi_{mg} \bar{i}_g - \frac{D}{J_m} \Sigma e$$

$$\frac{\Omega_e}{I_g} =$$

$$\frac{\frac{3n^2}{2Jm} \phi_{mg}}{s + \frac{D}{Jm}}$$

$$k_c \frac{D}{1-\alpha}$$

$$\frac{I_g}{I_g^*}$$

$$2.2AD \approx \frac{k_c^2}{s l_2 + k_c^2 + R_s}$$

$$\frac{\Omega_e}{I_g^*} = X \cdot \frac{\frac{R_{sd}}{1-\alpha}}{s l_2 (1-\alpha) + R_{sd} + R_s (1-\alpha)}$$

$$\frac{\Omega_e}{I_g^*} = X \cdot \frac{R_{sd}}{s l_2 (1-\alpha) + R_s} = \frac{\frac{\frac{3}{2} \frac{n^2}{D} \phi_{mg}}{Jm s + 1} \cdot R_{sd}}{s l_2 (1-\alpha) + R_s}$$

$$\frac{\Omega_e}{I_g^*} = \frac{\frac{\frac{3}{2} \frac{n^2}{D} \phi_{mg} d}{Jm}}{s \left(s + \frac{D}{Jm} \right)} \cdot \frac{l}{s+a}$$

$$C(s) = k_c \left(1 + \frac{1}{T_i s} + \frac{T_D s}{1} \right) ; G(s) = \frac{l}{(s+a)s}$$

$$\text{ZATV. KPNUG (PID): } \frac{C(s)G(s)}{1 + C(s)G(s)}$$

Jeru
2.2AD

$$\frac{k_c s + k_c / T_i + k_c T_D s^2}{s^2 (s+a)}$$

$$\frac{s^3 + s^2 a + k_c s l + \left(\frac{k_c}{T_i} \right) l + s^2 k_c T_D l}{s^2 (s+a)}$$

$$\frac{Y(s)}{R(s)} = \frac{A_0}{\Theta_e} = \frac{k_c s + k_c / T_i + s^2 k_c T_D}{s^3 + s^2 (a + k_c T_D l) + s (k_c l) + \frac{k_c l}{T_i}}$$

$$\frac{(s^2 + 2\xi\omega_n s + \omega_n^2)(s + n\omega_n)}{y} = \frac{s^3 + s^2(a + k_c^\theta \tau_0^\theta b) + s(k_c^\theta b) + \frac{k_c^\theta b}{\tau_1^\theta}}{x}$$

$$s^3 - s^2(n\omega_n + 2\xi\omega_n) + s(2\xi n\omega_n^2 + \omega_n^2) - n\omega_n^3 = y \quad (3)$$

$$y = x$$

$$(2) \quad k_c^\theta = \frac{2\xi n\omega_n^2 + \omega_n^2}{a}$$

$$a = D/J_m$$

IVE

$$b = \frac{3 \cdot 10^{-2} \cdot 0.002}{2 \cdot J_m}$$

$$(3) \quad \tau_1^\theta = \frac{2\xi n + 1}{n\omega_n}$$

$$(1) \quad a + k_c^\theta \tau_0^\theta b = n\omega_n + 2\xi\omega_n \Rightarrow \tau_0^\theta = \frac{\omega_n(2\xi + n) - a}{2\xi n\omega_n^2 + \omega_n^2}$$

(4) [4] SHPM (PARAMETRI IZ ZAD. 2) RADU U REGULACIJI MOMENTA. REG. STRUJE JE (P) TIPA S POJACANJEM $k_c^L = 10 R_s$. ODREĐITI STATIČKU GREŠKU (ODSTUPANJE STVARNE OD ŽELJENE VRIJ. U STAC. STANJU) STRUJE i_L NA SKOKOVITU PROMJENU POSTAVJENE VELIČINE MOMENTA, M_{om}^* S VRIJ. 1 Nm!

SHVPM! $L_d = L_g$

$$\frac{E}{I_2^*} = \frac{I_2^* - I_2}{I_2^*} ; \quad sI_2 = \frac{\hat{U}_2}{L_2} - \frac{R_5}{L_2} I_2$$

$$\frac{E}{I_2^*} = 1 - \frac{I_2}{I_2^*}$$

$$\frac{E}{I_2^*} = 1 - \frac{U_c^2}{sL_2 + U_c^2 + R_5}$$

$$\frac{E}{I_2^*} = \frac{sL_2 + R_5}{sL_2 + U_c^2 + R_5}$$

$$sI_2 = \frac{U_c^2}{L_2} (I_2^* - I_2) - \frac{R_5}{L_2} I_2$$

$$\frac{I_2}{I_2^*} = \frac{U_c^2/L_2}{s + \frac{U_c^2}{L_2} + \frac{R_5}{L_2}} = \frac{U_c^2}{sL_2 + U_c^2 + R_5}$$

$$M_{em} = \frac{3}{2} \pi \left[\phi_{mg} i_2 + \underbrace{(L_d - L_g) i_d i_g}_{=0 \text{ SMUPH}} \right]$$

~~=0 SMUPH~~

AI NON SMUPH

DER SE i_2^* POSTAVI
U 0

$$I_2^* = \frac{2 M_{em}^*}{3 \pi \phi_{mg}}$$

$$\text{WE } M_{em}^* = \frac{1}{s}$$

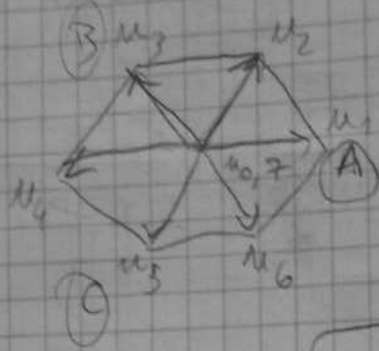
(SMOHOVITA
FJA $\varnothing \rightarrow 1$)

$$\bar{E} = \frac{sL_2 + R_5}{sL_2 + U_c^2 + R_5} \cdot \frac{2}{3 \pi s \phi_{mg}}$$

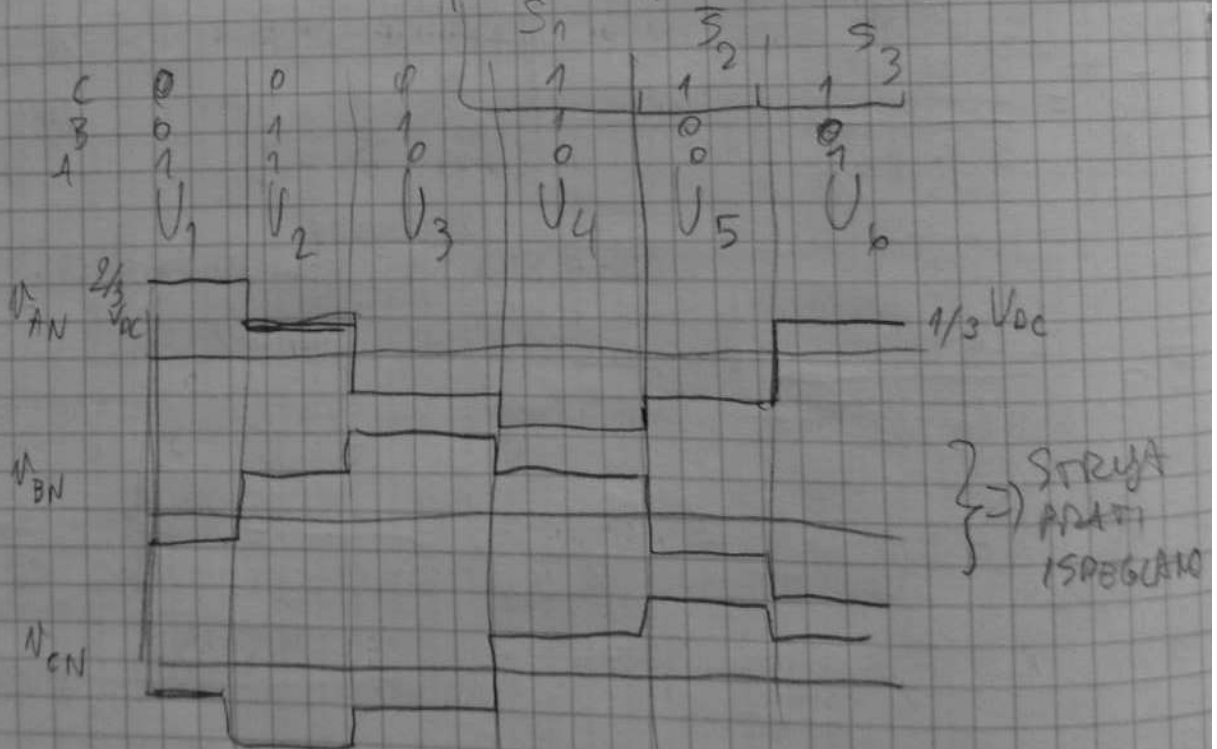
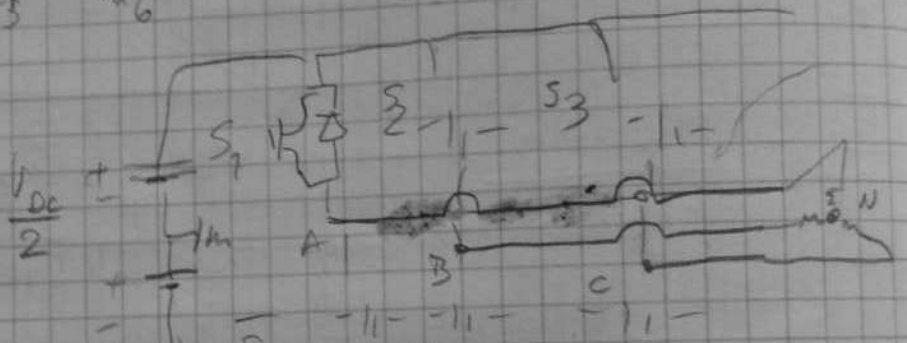
$$\lim_{t \rightarrow \infty} e(t) = \lim_{s \rightarrow 0} sE(s) = \frac{R_5}{U_c^2 + R_5} \cdot \frac{2}{3 \pi \phi_{mg}} = \frac{R_5}{11 R_5} \cdot \frac{2}{3 \pi \phi_{mg}} = 0.2424$$

[3]
 5. NAVESTI OSNOVNE RAZLIKE MODULACIJE SIX STEP,
 SINUSNE I VEKTORSKE PWM MOD PRETVARAOA NAPONA
 I FREKVENCJE ZA DOBIVANJE ŽELJENE AMPLITUDE I FREKV.
 NAPONA NA STEZALJAMA STROJA

SIX STEP:



	u_0	u_1	u_2	u_3	u_4	u_5	u_6
S_1	0	0	0	1	1	1	1
S_2	0	0	1	1	1	0	0
S_3	0	1	1	0	0	0	1



→ NIJE PWM, DEFINIRA SAMO FREKV. OSNOVNOG HARMONIKA (OVISNO O ω_e), NE MOŽE MJEŃJATI AMPLITUDU (TREBALO BI MJEŃJATI

V_{dc}), MAX. AMPL. $\frac{2}{\pi} V_{dc}$

→ JE DOKTORNA, MALI GUBICI, ALI LOŠ HARMONISKI SASTAV (MENA 3. HARM. I VIŠE)

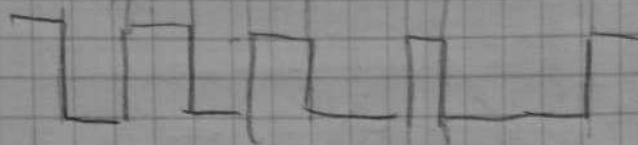
SINUSNA PWM \rightarrow KIJENJA AMPL. & frekv.

REFERENTNI \sim
NOSIOČ \sim } USPOREDBA

$\sim > \sim \Rightarrow$ IZLAZ TE FAZE ($S_{a,1,2} = 1$)



S_a



\rightarrow veća dubina zbog većeg slepovanja, bolji spektral,
u linearnom području rada ($m_a = \frac{U_o}{U_{dc}/2} < 1$) IZLAZ

PROPORCIONALAN REFERENCI

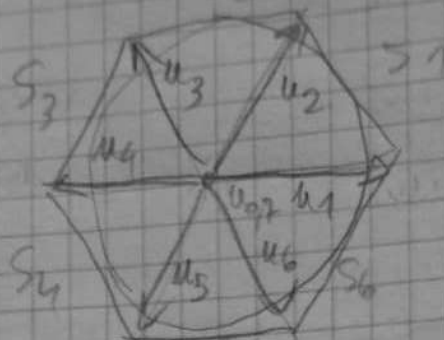
\rightarrow INJEKCIJA SE ZERO-SEQUENCE $\left(\frac{1}{6} u_m \sin(3\omega_e t) \right)$, DOKLE
 $\frac{1}{6}$
 $\frac{1}{4} \sim \rightarrow$ SMANJUJE THD

REFERENTNOM DA BI SE POVEĆAO LINEARNI RASPON MODULACIJE
(& NE BILAO OGRANIČENO NA $U_{dc}/2$)

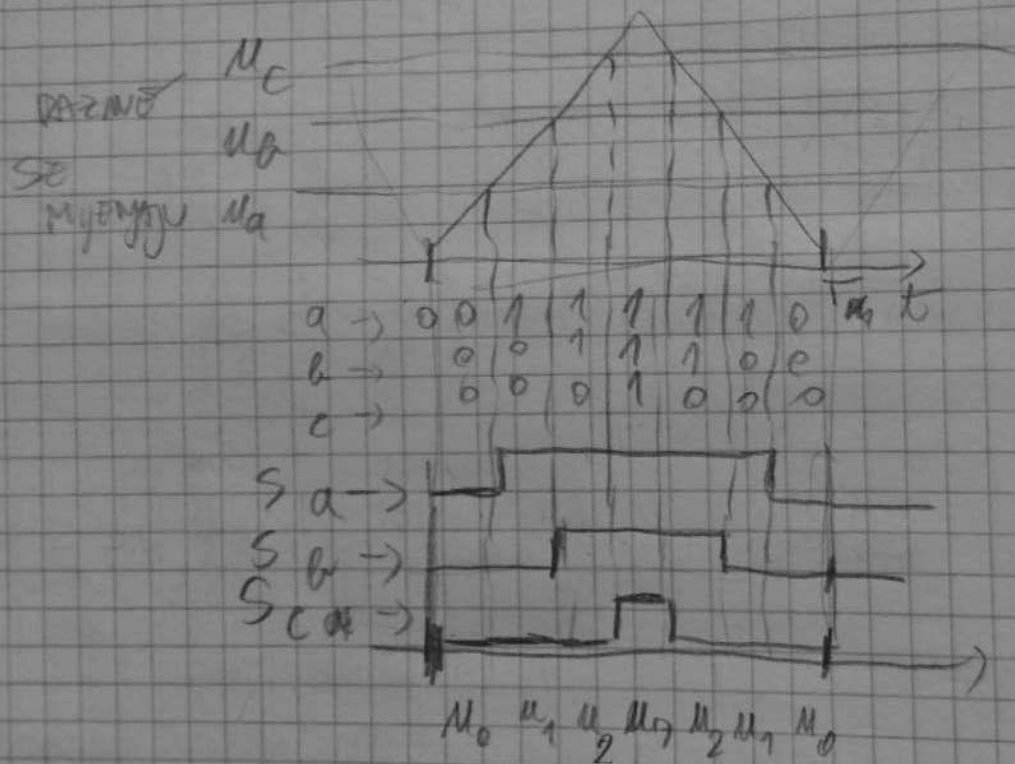
$I_{lim} \rightarrow$ neposredni uslovovalni referentne frekv. DA SE SMANJI
UTJECAJ VIŠIH HARMONIKA

VEKTORSKA PWN (SVPWM)

→ ima 2 karakterist. vektora dugo 0 sektoru

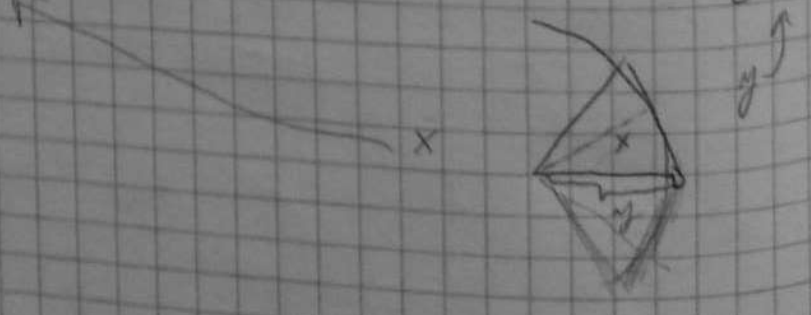


→ vremenski odjeganje se koristi s 2 vektora 2
2 NULTA NPK $S_1 \rightarrow u_1, u_2, u_0, u_0$



SEKTOR 1

→ MAX. AMPLITUDA $\frac{V_{dc}}{\sqrt{3}}$ jer su maximumi lin. vektora $\frac{2V_{dc}}{3}$



6. [3] OPREDI UKUPNO STATORSKO PROJEKCIJE $\vec{F}(t)^\theta$ ZA
 POLOŽAJ $\Theta = 90^\circ$ U TREĆEJ FAZI $t = 15, 4 \mu s$.

$$i_a(t) = 5 \cos(314t)$$

$$i_b(t) = 5 \cos(314t - \frac{2\pi}{3})$$

$$i_c(t) = 5 \cos(314t - \frac{4\pi}{3})$$

BRJ ZAVOJA PO

POJEDINOJ FAZI JE

$$N_1 = N_2 = N_3 = N_S = 100$$

$$F_a(t) = N_S \cdot I_S \cdot \cos(314t) = 500 \cos(314t)$$

$$F_b(t) = 500 \cos(314t - \frac{2\pi}{3})$$

$$F_c(t) = 500 \cos(314t - \frac{4\pi}{3})$$

$$F_a(t)^\theta = F_a(t) \cos(0^\circ - \theta)$$

$$F_b(t)^\theta = F_b(t) \cos(\frac{2\pi}{3} - \theta)$$

$$F_c(t)^\theta = F_c(t) \cos(\frac{4\pi}{3} - \theta)$$

$$F(t)^\theta = \frac{3}{2} \operatorname{Re} \left\{ \frac{2}{3} [F_a(t) + F_b(t) e^{j\frac{2\pi}{3}} + F_c(t) e^{j\frac{4\pi}{3}}] e^{-j\theta} \right\}$$

$\vec{F}(t)$ PROSTORNI VEKTOR
 PROJ.

$$F(t)^\theta = \frac{3}{2} F_m \cos(\omega t + \phi_0 - \theta)$$

$$\phi_0 = 0^\circ \quad \theta = 90^\circ$$

$$F(t)^\theta = \frac{3}{2} \cdot 500 \cos(314 - 90^\circ) = \frac{3}{2} \cdot 500 \cdot \sin(314) = 118,95$$

AMR
 ZAVOJA

7. [4] ZA SUSTAV UPRAVLJANJA BRZINOM VRTNJE SINHRONIZIRANOG MOTORA S UNUTARNJIM PERM. MAGN. NAPIŠATI IZRAZE ZA UPRAVLJAČKE SIGNALS u_d I u_q (U VREM. DOMENI) IZ REGULATORA STRUJA i_d I i_q (UZETI U OBRZ NASPREZANJE). ZA DOBIVENE IZRAZE ZA u_d I u_q NACRTATI BLOKOVSKU SHEMU. ODABRATI TIP REGULATORA STRUJE i_d , ODNOSNO i_q TAKAV DA JE STATIČNA GREŠKA (ODSTUPANJE STVARNE OD ŽELJENE VRTJ. U STAL. STANJU) BRZINE VRTNJE STROJA JEDNAKA NULI.

SMUPM

$$\frac{di_d}{dt} = \frac{1}{L_d} (u_d - R_s i_d + L_q \omega_e i_q)$$

$$\frac{di_q}{dt} = \frac{1}{L_q} (u_q - R_s i_q - L_d \omega_e i_d - \omega_e \Phi_m)$$

IZLAZI
IZ REG. ~~UNUTARNJIM~~
~~UNUTARNJIM~~

$$\begin{cases} \hat{u}_d = u_d + L_q \omega_e i_q \\ \hat{u}_q = u_q - L_d \omega_e i_d - \omega_e \Phi_m \end{cases}$$

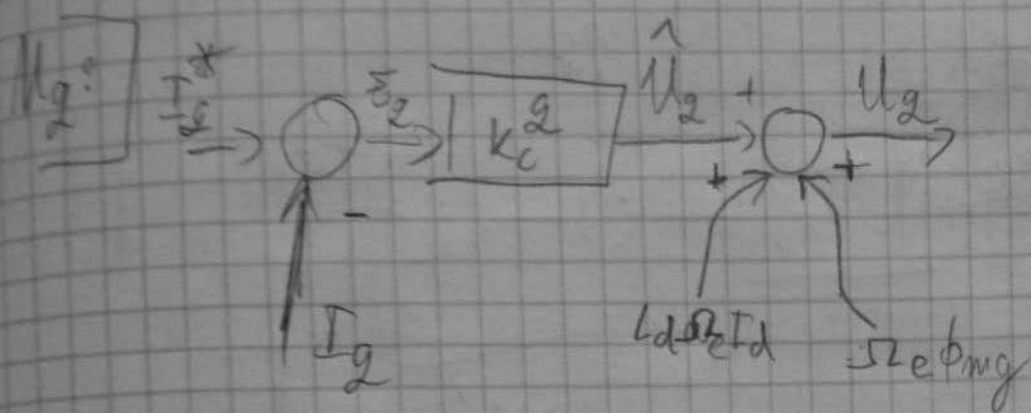
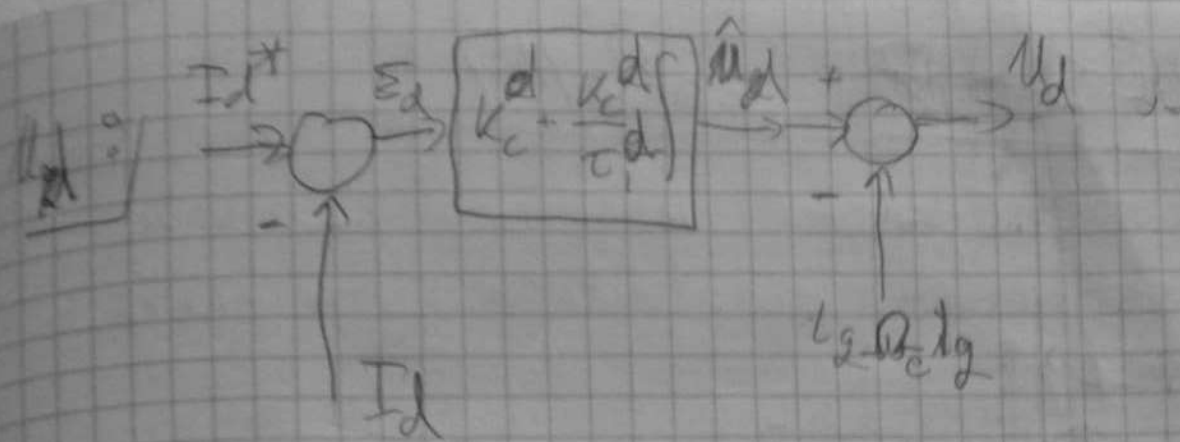
PI

$$\hat{u}_d = k_c (i_d^* - i_d(t)) + \frac{k_c}{\omega_c} \int_0^t (i_d^*(\tau) - i_d(\tau)) d\tau$$

P (MOŽE I PI)

$$\hat{u}_q = k_c (i_q^* - i_q(t))$$

→ NADREBENI REGULATORI MORA BITI PI TIPA DA STATIČNA GREŠKA = 0
(BRZINA ω_e)



ERZETL:

$$u_d = \hat{u}_d - L_g \omega_e I_g$$

$$u_d(t) = \hat{u}_d(t) - L_g \omega_e i_g(t)$$

$$= L_d \frac{di_d(t)}{dt} + R_s i_d(t) - L_g \omega_e(t) i_g(t)$$

$$(2) \quad u_q(t) = \hat{u}_q(t) + L_d \omega_e(t) i_d(t) + \omega_e(t) \phi_{mg}$$

$$u_q(t) = \frac{di_q(t)}{dt} L_g + R_s i_q(t) + L_d \omega_e(t) i_d(t) + \omega_e(t) \phi_{mg}$$