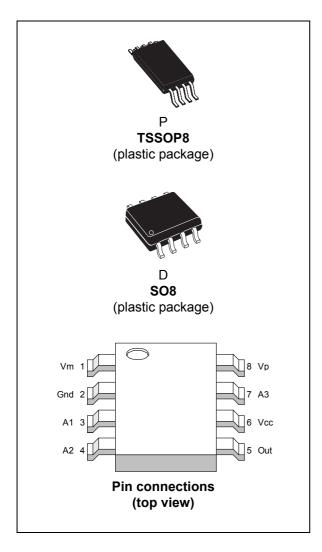


High-side current sense amplifier plus signal conditioning amplifier

Datasheet - production data



Features

- Independent supply and input common-mode voltages
- Wide common-mode operating range: 2.8 to 30 V
- Wide common-mode surviving range:

 16 to 60 V (reversed battery and load-dump conditions)
- Low current consumption: I_{CC} max = 420 μA

- Output amplifier for tailor-made signal conditioning
- -40 °C to 125 °C operating temperature range
- 4 kV ESD protection

Applications

- · Battery chargers
- · Automotive current monitoring
- · Notebook computers
- DC motor control
- Photo-voltaic systems
- Precision current sources
- Uninterruptible power supplies
- · High-end power supplies

Description

The TSC102 measures a small differential voltage on a high-side shunt resistor and translates it into a ground-referenced output voltage.

The device's wide input common-mode voltage range, low quiescent current and tiny TSSOP8 packaging enable use in a wide variety of applications (also available in SO8 package).

The input common-mode and power supply voltages are independent. The common-mode voltage can range from 2.8 to 30 V in operating conditions.

The TSC102 is rugged against abnormal conditions on the input pins: Vp and Vm can withstand up to 60 V in case of voltage spikes, as little as -16 V in case of reversed battery, and up to 4 kV in case of electrostatic discharge.

In addition to the current sensing amplifier, the TSC102 offers a fully accessible amplifier for output signal conditioning. The device's overall current consumption is lower than 420 μ A.

Contents TSC102

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1 Application schematic and pin description

The TSC102 high-side current sense amplifier features a 2.8 V to 30 V input common-mode range that is independent of the supply voltage. The main advantage of this feature is that it allows high-side current sensing at voltages much greater than the supply voltage (V_{CC}).

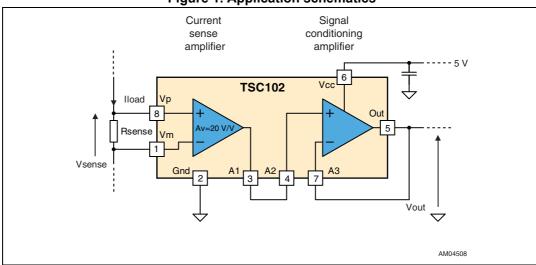


Figure 1. Application schematics

Table 1 describes the function of each pin. Their position is shown in the illustration on the cover page and in *Figure 1* above.

Symbol	Type	Function		
Out	Analog output	Out voltage is proportional to the magnitude of the sense voltage $V_p\hbox{-} V_m$		
Gnd	Power supply	Ground line		
V _{CC}	Fower suppry	Positive power supply line		
V _p		Connection for the external sense resistor. The measured current enters the shunt on the $\rm V_{\rm p}$ side		
V _m	Analog input	Connection for the external sense resistor. The measured current exits the shunt on the $\rm V_{m}$ side		
A1	, , , , , , , , , , , , , , , , , , ,	Connection to current sensing amplifier output		
A2		Connection to signal conditioning amplifier non-inverting input		
А3		Connection to signal conditioning amplifier inverting input		

Table 1. Pin description

2 Absolute maximum ratings and operating conditions

Table 2. Absolute maximum ratings

Symbol	Parameter	Value	Unit
V _{id}	Input pins differential voltage (V _p -V _m)	±20	
Vi	Current sensing input pin voltages $(V_p \text{ and } V_m)^{(1)}$	-16 to 60	V
V ₁	Voltage for pins A1, A2, A3, Out, Vcc ⁽¹⁾	-0.3 to 7	
T _{stg}	Storage temperature	-55 to 150	°C
T _j	Maximum junction temperature	150	C
D	TSSOP8 thermal resistance junction to ambient	120	°C/W
R _{thja}	SO8 thermal resistance junction to ambient	125	C/VV
	HBM: human body model for V_m and V_p pins ⁽²⁾	4	kV
ESD	HBM: human body model ⁽³⁾	2.5	NV
ESD	MM: machine model ⁽⁴⁾	200	V
	CDM: charged device model ⁽⁵⁾	1.5	kV

- 1. These voltage values are measured with respect to the GND pin.
- 2. Human body model for Vm and Vp: a 100 pF capacitor is charged to the specified voltage, then discharged through a 1.5 k Ω resistor between the Vp or Vm pin and Gnd while the other pins are floating.
- 3. Human body model: a 100 pF capacitor is charged to the specified voltage, then discharged through a 1.5 k Ω resistor between two pins of the device. This is done for all couples of connected pin combinations while the other pins are floating.
- 4. Machine model: a 200 pF capacitor is charged to the specified voltage, then discharged directly between two pins of the device with no external series resistor (internal resistor < 5 Ω). This is done for all couples of connected pin combinations while the other pins are floating.
- Charged device model: all pins plus package are charged together to the specified voltage and then discharged directly to ground.

Table 3. Operating conditions

Symbol	Parameter	Value	Unit
V _{CC}	DC supply voltage from T _{min} to T _{max}	3.5 to 5.5	V
T _{oper}	Operational temperature range (T _{min} to T _{max})	-40 to 125	°C
V _{icm}	Common-mode voltage range (V _m pin voltage)	2.8 to 30	V

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3 Electrical characteristics

Unless otherwise specified, the electrical characteristics given in the following tables have been measured under the following test conditions.

- $T_{amb} = 25 \text{ °C}, V_{CC} = 5 \text{ V}, V_{sense} = V_{p} V_{m} = 50 \text{ mV}, V_{m} = 12 \text{ V}.$
- No load on Out pin.
- Signal conditioning amplifier used as a buffer (pin A3 connected to pin Out and pin A1 connected to pin A2).

Table 4. Supply

Symbol	Parameter	Test conditions	Min.	Тур.	Max.	Unit
I _{CC}	Total supply current	V_{sense} = 0 V, pin A1 open, pin A2 shorted to Gnd T_{min} < T_{amb} < T_{max}	_	240	420	μA
I _{CC1}	Total supply current	V_{sense} = 50 mV, pin A1 connected to pin A2 T_{min} < T_{amb} < T_{max}		420	700	μΛ

Table 5. Current sensing amplifier input stage

Symbol	Parameter	Test conditions	Min.	Тур.	Max.	Unit
DC CMR1	DC common-mode rejection Variation of V _{a1} versus V _{icm} referred to input ⁽¹⁾	2.8 V < V _m < 30 V -40 °C < T _{amb} < 150 °C	90	100		
AC	AC common-mode rejection Variation of V _{a1} versus V _{icm}	2.8 V< V _m < 30 V 1 kHz sine wave		75		dB
CMR1	referred to input (peak-to-peak voltage variation)	2.8 V < V _m < 30 V 10 kHz sine wave		60		
SVR1	Supply voltage rejection Variation of V _{a1} versus V _{CC} ⁽²⁾	3.5 V< V _{CC} < 5.5 V -40 °C < T _{amb} < 125 °C	85	90		
V _{os}	Input offset voltage ⁽³⁾	T _{amb} = 25 ° C -40 °C < T _{amb} < 125 °C			±1.5 ±2.3	mV
dV _{os} /dT	Input offset drift versus T	-40 °C < T _{amb} < 125 °C		±3	±8	μV/°C
I _{lk}	Input leakage current	$V_{CC} = 0 V$ $T_{min} < T_{amb} < T_{max}$			1	μA
I _{ib}	Input bias current	$V_{\text{sense}} = 0 \text{ V}$ $T_{\text{min}} < T_{\text{amb}} < T_{\text{max}}$		5	7	μΛ

- 1. See Section 6: Parameter definitions for the definition of CMR
- 2. See Section 6 for the definition of SVR
- 3. See Section 6 for the definition of V_{os}



Electrical characteristics TSC102

Table 6. Current sensing amplifier output stage

Symbol	Parameter	Test conditions	Test conditions Min.			Unit	
Av	Gain (variation of V _{a1} versus V _{sense})			20		V/V	
V _{oh1}	A1 node high-level saturation voltage $V_{oh1} = V_{cc} - V_{a1}$	V _{sense} = 1 V I _{a1} = 1 mA -40 °C< T _{amb} < 125 °C		85	185	m\/	
V _{ol1}	A1 node low-level saturation voltage	V _{sense} =-1 V I _{a1} = 1 mA -40 °C< T _{amb} < 125 °C		75 165		- mV	
I _{sc1}	Short-circuit current	A1 connected to V _{CC} or Gnd	10	30		mA	
$\Delta V_{a1}/\Delta T$	Output voltage drift versus T ⁽¹⁾	T _{min} < T _{amb} < T _{max}			±400	ppm/°C	
$\Delta V_{a1}/\Delta I_{a1}$	Output stage load regulation	-5 mA < I _{a1} < +5 mA I _{a1} sink or source current		0.4	±2	mV/mA	
ΔV _{a1}	Total output voltage accuracy ⁽²⁾	V_{sense} = 50 mV T_{amb} = 25 ° C T_{min} < T_{amb} < T_{max}			±2.5 ±4		
ΔV_{a1}	Total output voltage accuracy ⁽²⁾	V_{sense} = 100 mV T_{amb} = 25 ° C T_{min} < T_{amb} < T_{max}			±2.5 ±4	%	
ΔV_{a1}	Total output voltage accuracy ⁽²⁾	V_{sense} = 20 mV T_{amb} = 25 ° C T_{min} < T_{amb} < T_{max}			±8 ±10	76	
ΔV_{a1}	Total output voltage accuracy ⁽²⁾	V_{sense} = 10 mV T_{amb} = 25 ° C T_{min} < T_{amb} < T_{max}			±13 ±16		

^{1.} See Section 6: Parameter definitions for the definition of output voltage drift versus temperature.

Table 7. Current sensing amplifier frequency response

Symbol	Parameter Test conditions I		Min.	Тур.	Max.	Unit
ts	V _{a1} settling to 1% final value	V _{sense} = 10 mV to 100 mV, C _{load} = 47 pF	-	7	-	μs
SR	Slew rate	V _{sense} = 10 mV to 100 mV	0.2	0.4	-	V/µs
BW	3 dB bandwidth	C _{load} = 47 pF	-	800	-	kHz

Table 8. Current sensing amplifier noise

Symbol	Parameter	Test conditions	Min.	Тур.	Max.	Unit
e _N	Equivalent input noise voltage	f = 1 kHz	-	50	ı	nV/√ Hz



^{2.} Output voltage accuracy is the difference with the expected theoretical output voltage V_{a1-th} = Av * V_{sense}. See Section 6 for a more detailed definition.

Table 9. Signal conditioning amplifier

Symbol	Parameter	Test conditions	Min.	Тур.	Max.	Unit
V _{icm}	Common-mode voltage range	$T_{min} < T_{amb} < T_{max}$	0		Vcc	
V _{IO}	Input offset voltage	V _{a2} = 1 V T _{amb} = 25 ° C -40° C < T _{amb} < 150 ° C			±3.5 ±4.5	mV
ΔV_{IO}	Input offset voltage drift	$T_{min} < T_{amb} < T_{max}$		5		μV/°C
lib	Input bias current	$V_{a2} = V_{a3} = V_{CC}/2$		10		рА
V _{oh2}	Output high-level saturation voltage (V _{oh2} = V _{CC} -V _{out})	V _{a2} = 1 V V _{a3} = 0 V I _{out} = 1 mA -40° C< T _{amb} < 125° C		85	185	
V _{ol2}	Output low-level saturation voltage	V _{a2} = 0 V V _{a3} = 1 V I _{out} = 1 mA -40 °C< T _{amb} < 125 °C		75	165	mV
I _{sc2}	Short-circuit current	Out connected to V _{CC} or Gnd	12	30		mA
ΔV _{out} /ΔI _{out}	Output stage load regulation	-10 mA < I_{out} < +10 mA V_{a2} = 1 V I_{out} sink or source current			300	μV/mA
CMR2	DC common-mode rejection Variation of V _{IO} versus V _{icm}	$T_{min} < T_{amb} < T_{max}$ 0 V <v<sub>a2<3 V 0 V<v<sub>a2<5 V</v<sub></v<sub>	70 60	95 80		dB
SVR2	Supply voltage rejection Variation of V _{IO} versus V _{CC}	3.5 V <v<sub>CC<5.5 V V_{a2} = 1 V -40 °C < T_{amb} < 125 °C</v<sub>	85	105		dБ
GBP	Gain bandwidth product	$R_L = 10 \text{ k}\Omega \text{ C}_{load} = 100 \text{ pF},$ f = 100 kHz		1		MHz
PM	Phase margin	R_L = 10 kΩ, C_{load} = 100 pF		65		deg
SR	Slew rate	R_L = 10 k Ω C _{load} = 100 pF V_{a2} = 0.5 V to 4.5 V A3 connected to OUT (follower configuration) Slew rate measured from 10% to 90% of V _{out} step	0.2	0.4		V/µs

4 Electrical characteristics curves: current sense amplifier

Unless otherwise specified, the test conditions for the following curves are:

- $T_{amb} = 25 \text{ °C}, V_{CC} = 5 \text{ V}, V_{sense} = V_p V_m = 50 \text{ mV}, V_m = 12 \text{ V}.$
- no load on Out pin.
- signal conditioning amplifier used as a buffer (pin A3 connected to pin Out and pin A1 connected to pin A2).

Figure 2. Output voltage vs. Vsense

Figure 3. A1 pin voltage accuracy vs. Vsense

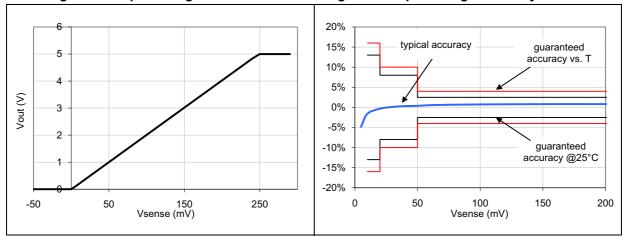
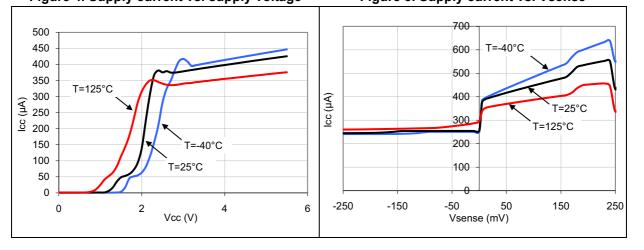


Figure 4. Supply current vs. supply voltage

Figure 5. Supply current vs. Vsense



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Figure 6. Vp pin input bias current vs. Vsense Figure 7. Vm pin input bias current vs. Vsense

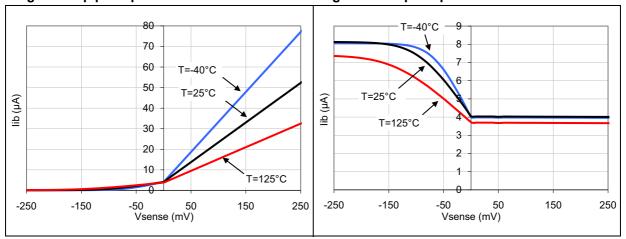


Figure 8. Output stage low-state saturation voltage versus output current (Vsense = -1 V)

Figure 9. Output stage high-state saturation voltage versus output current (Vsense = +1 V)

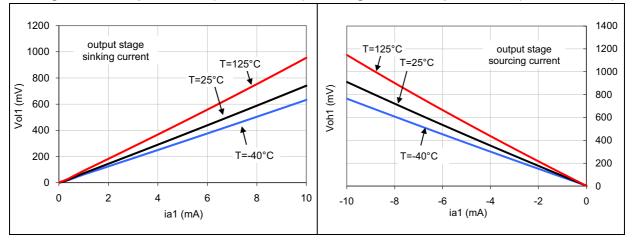
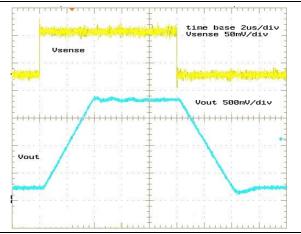


Figure 10. Output stage load regulation

output stage output stage 6 sourcing current sinking current Vsense 5 Va1-Va1@ia1=0 (mV) T=125°C T=25°C 3 2 T=-40°C Vout 10 ia1(mA)

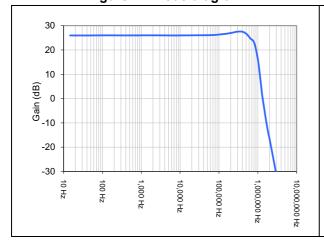
Figure 11. Step response

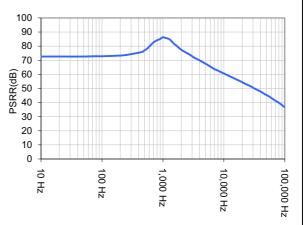


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Figure 12. Bode diagram

Figure 13. Power supply rejection ratio





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5 Electrical characteristics curves: signal conditioning amplifier

Unless otherwise specified, the test conditions for the following curves are:

- T_{amb} = 25 °C, V_{CC} = 5 V
- no load on Out.
- signal conditioning amplifier tested as standalone amplifier.

Figure 14. Input offset voltage versus input common-mode voltage

Figure 15. Input offset voltage versus supply voltage (Vicm = Vcc/2)

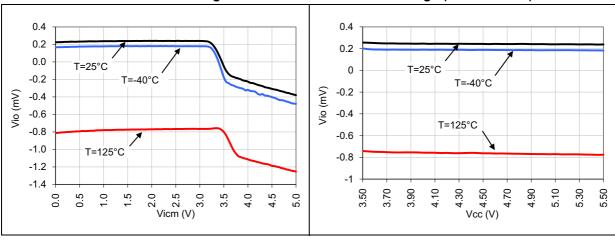
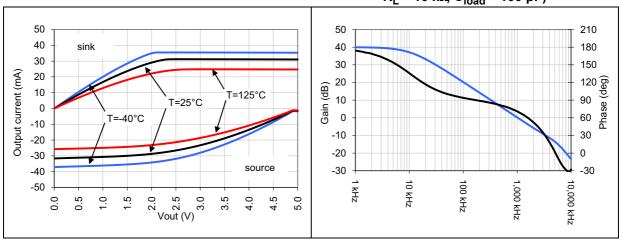


Figure 16. Output current versus output voltage

Figure 17. Bode diagram (Vout = Vcc/2, R_L = 10 k Ω , C_{load} = 100 pF)



Parameter definitions TSC102

6 Parameter definitions

6.1 Common-mode rejection ratio (CMR)

The common-mode rejection ratio (CMR) measures the ability of the current sensing amplifier to reject any DC voltage applied on both inputs V_p and V_m . The CMR is referred back to the input so that its effect can be compared with the applied differential signal. The CMR is defined by the formula:

$$CMR = -20 \cdot log \frac{\Delta V_{a1}}{\Delta V_{icm} \cdot Av}$$

6.2 Supply voltage rejection ratio (SVR)

The supply voltage rejection ratio (SVR) measures the ability of the current sensing amplifier to reject any variation of the supply voltage V_{CC} . The SVR is referred back to the input so that its effect can be compared with the applied differential signal. The SVR is defined by the formula:

$$SVR = -20 \cdot \log \frac{\Delta V_{a1}}{\Delta V_{cc} \cdot Av}$$

6.3 Gain (Av) and input offset voltage (V_{os})

The input offset voltage is defined as the intersection between the linear regression of the V_{a1} versus V_{sense} curve with the X-axis (see *Figure 18*). If V_{a11} is the output voltage with $V_{sense} = V_{sense1} = 50$ mV and V_{a12} is the output voltage with $V_{sense} = V_{sense2} = 5$ mV, then V_{os} can be calculated with the formula:

$$V_{os} = V_{sense1} - \left(\frac{V_{sense1} - V_{sense2}}{V_{a11} - V_{a12}} \cdot V_{out1} \right)$$

The amplification gain Av is defined as the ratio between the output voltage and the input differential voltage.

$$Av = \frac{V_{out}}{V_{sense}}$$

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TSC102 Parameter definitions

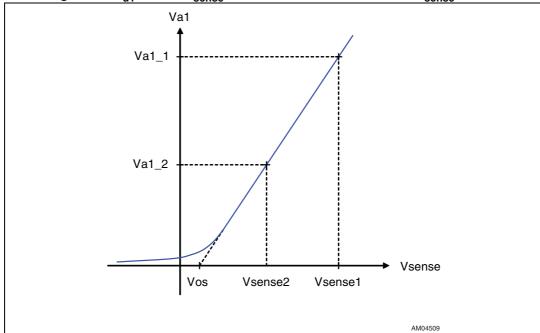


Figure 18. V_{a1} versus V_{sense} characteristics: detail for low V_{sense} values

6.4 Output voltage drift versus temperature

The output voltage drift versus temperature is defined as the maximum variation of V_{a1} with respect to its value at 25 ° C, over the temperature range. It is calculated as follows:

$$\frac{\Delta V_{a1}}{\Delta T} = max \frac{V_{a1}(T_{amb}) - V_{a1}(25^{\circ}C)}{T_{amb} - 25^{\circ}C}$$

with $T_{min} < T_{amb} < T_{max}$.

Figure 19 provides a graphical definition of the output voltage drift versus temperature. On this chart, V_{a1} is always within the area defined by the maximum and minimum variation of V_{a1} versus T, and T = 25 °C is considered to be the reference.

Parameter definitions TSC102

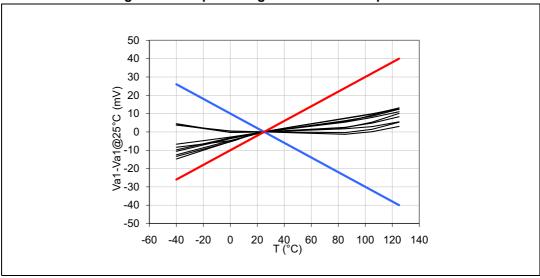


Figure 19. Output voltage drift versus temperature

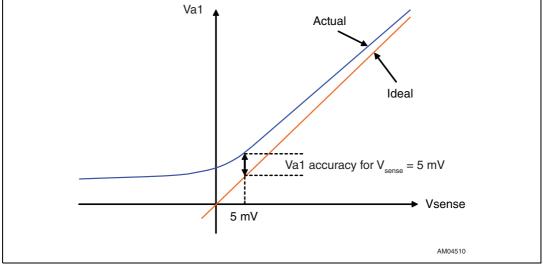
TSC102 Parameter definitions

Output voltage accuracy 6.5

The output voltage accuracy is the difference between the actual output voltage and the theoretical output voltage. Ideally, the current sensing output voltage should be equal to the input differential voltage multiplied by the theoretical gain, as in the following formula.

The actual value is very slightly different, mainly due to the effects of the input offset voltage Vos and the non-linearity.

Figure 20. V_{a1} vs. V_{sense} theoretical and actual characteristics Va1



The output voltage accuracy, expressed as a percentage, can be calculated with the following formula:

$$\Delta V_{a1} = \frac{abs(V_{a1} - (Av \cdot V_{sense}))}{Av \cdot V_{sense}}$$

with Av = 20 V/V.

7 Application information

The TSC102 can be used to measure current and feed back the information to a microcontroller, as shown in *Figure 21*.

TSC102 Vcc Microcontroller

ADC Microcontroller

Vsense load

AM04511

Figure 21. Typical application schematic

This fully-accessible output amplifier offers wide schematic possibilities, as shown in the following examples.

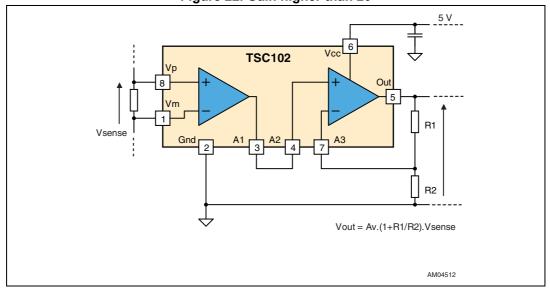


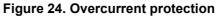
Figure 22. Gain higher than 20

TSC102 Vcc 6

Vsense R1

Vout = Av.R2.Vsense/(R1+R2)

Figure 23. Gain lower than 20



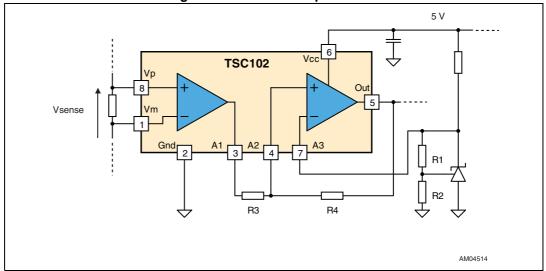
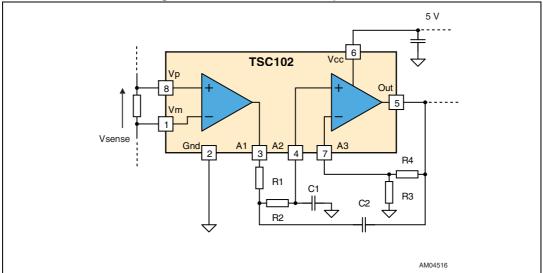


Figure 25. First-order low-pass filter





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TSC102 Package information

8 Package information

In order to meet environmental requirements, ST offers these devices in different grades of ECOPACK[®] packages, depending on their level of environmental compliance. ECOPACK[®] specifications, grade definitions and product status are available at: www.st.com. ECOPACK[®] is an ST trademark.



Package information TSC102

8.1 TSSOP8 package information

O.25 mm
GAGE PLANE

O.25 m

Figure 27. TSSOP8 package mechanical drawing

Table 10. TSSOP8 package mechanical data

	Dimensions									
Ref.		Millimeters		Inches						
	Min.	Тур.	Max.	Min.	Тур.	Max.				
Α			1.20			0.047				
A1	0.05		0.15	0.002		0.006				
A2	0.80	1.00	1.05	0.031	0.039	0.041				
b	0.19		0.30	0.007		0.012				
С	0.09		0.20	0.004		0.008				
D	2.90	3.00	3.10	0.114	0.118	0.122				
Е	6.20	6.40	6.60	0.244	0.252	0.260				
E1	4.30	4.40	4.50	0.169	0.173	0.177				
е		0.65			0.0256					
k	0°		8°	0°		8°				
L	0.45	0.60	0.75	0.018	0.024	0.030				
L1		1			0.039					
aaa			0.10			0.004				

TSC102 Package information

8.2 SO8 package information

Figure 28. SO8 package mechanical drawing

Table 11. SO8 package mechanical data

	Dimensions								
Ref.		Millimeters		Inches					
	Min.	Тур.	Max.	Min.	Тур.	Max.			
Α			1.75			0.069			
A1	0.10		0.25	0.004		0.010			
A2	1.25			0.049					
b	0.28		0.48	0.011		0.019			
С	0.17		0.23	0.007		0.010			
D	4.80	4.90	5.00	0.189	0.193	0.197			
Е	5.80	6.00	6.20	0.228	0.236	0.244			
E1	3.80	3.90	4.00	0.150	0.154	0.157			
е		1.27			0.050				
h	0.25		0.50	0.010		0.020			
L	0.40		1.27	0.016		0.050			
L1		1.04			0.040				
k	0		8°	1°		8°			
ccc			0.10			0.004			

Ordering information TSC102

9 Ordering information

Table 12. Order codes

Part number	Temperature range	Package	Packing	Marking
TSC102IPT	-40 °C, +125 °C	TSSOP8	Tape and reel	1021
TSC102IDT	-40 C, +125 C	SO8		TSC102I
TSC102IYPT	-40 °C, +125 °C	TSSOP8 ⁽¹⁾		102Y
TSC102IYDT	automotive grade	SO8 ⁽¹⁾		TSC102IY

^{1.} Qualification and characterization according to AEC Q100 and Q003 or equivalent, advanced screening according to AEC Q001 and Q 002 or equivalent.

10 Revision history

Table 13. Document revision history

Date	Revision	Changes	
09-Nov-2009	1	Initial release.	
03-Mar-2011	2	Added automotive grade qualification for SO8 package (note 2. under <i>Table 12</i>).	
31-Jan-2014	3	Table 12: Updated automotive-grade footnotes.	

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