MODEL PREDICTIVE CONTROL

HYBRID MODELS FOR MPC

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`imt.lu/ab`

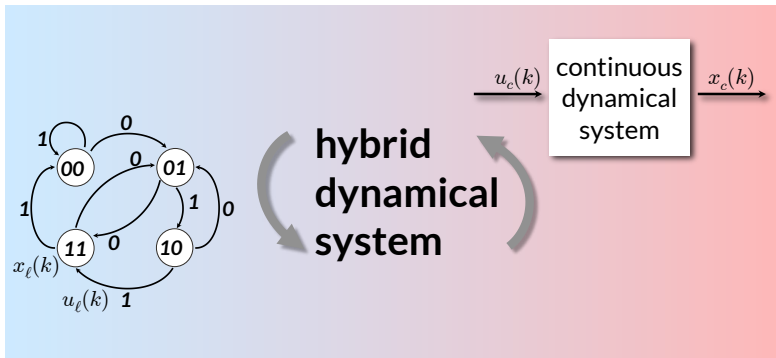# COURSE STRUCTURE

✔ **Basic concepts** of model predictive control (MPC) and **linear MPC**

✔ **Linear time-varying** and **nonlinear** MPC

✔ **Quadratic programming** (QP) and **explicit MPC**

- **Hybrid** MPC

- **Stochastic** MPC

- **Learning-based** MPC

Course page:
http://cse.lab.imtlucca.it/~bemporad/mpc_course.html

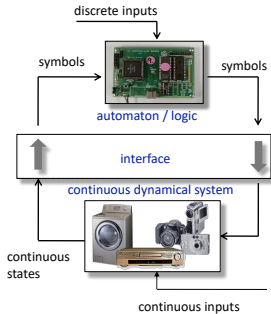# HYBRID MODELS

- Variables are **binary-valued**
  $x_\ell \in \{0,1\}^{n_\ell}, \ u_\ell \in \{0,1\}^{m_\ell}$

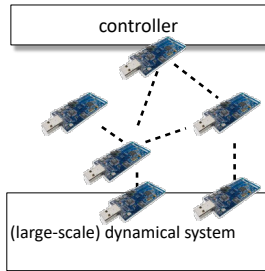- Dynamics = **finite state machine**

- **Logic constraints**

- Variables are **real-valued**
  $x_c \in \mathbb{R}^{n_c}, \ u_c \in \mathbb{R}^{m_c}$

- **Difference/differential equations**

- **Linear inequality** constraints

TECHNOLOGICAL PUSH FOR STUDYING HYBRID SYSTEMS
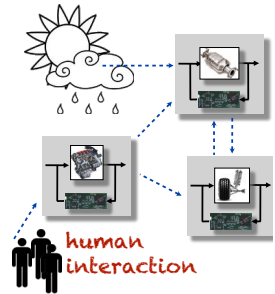

| embedded systems | networked control systems | cyber-physical systems |

>1995   >2005   >2010

- Vehicle

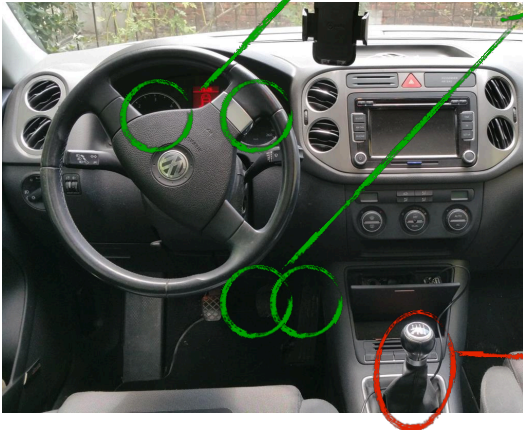**continuous dynamical variables (speed, torque, ...)**

continuous commands (brake & gas pedal)

+

discrete command (R,N,1,2,3,4,5)

KEY REQUIREMENTS FOR HYBRID MODELS

- **Descriptive** enough to capture the behavior of the system

  – **continuous** dynamics (physical systems)
  – **logic** components (switches, automata)
  – **interconnection** between logic and dynamics

- **Simple** enough for solving analysis and synthesis problems

$$\begin{cases} x' & = & Ax + Bu \\ y & = & Cx + Du \end{cases}$$

linear hybrid systems

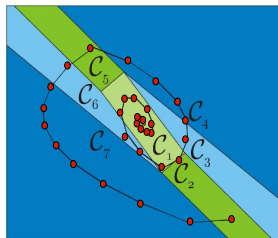$$\begin{cases} x' & = & f(x, u, t) \\ y & = & g(x, u, t) \end{cases}$$

"Perfection is achieved not when there is nothing more to add, but when there is nothing left to take away."

A. de Saint-Exupéry
(1900–1944)
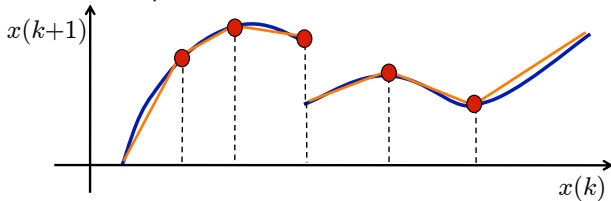
PIECEWISE AFFINE SYSTEMS

$$
\begin{aligned}
x(k+1) &= A_{i(k)}x(k) + B_{i(k)}u(k) + f_{i(k)} \\
y(k) &= C_{i(k)}x(k) + D_{i(k)}u(k) + g_{i(k)} \\
i(k) \quad &\text{s.t.} \quad H_{i(k)}x(k) + J_{i(k)}u(k) \le K_{i(k)}
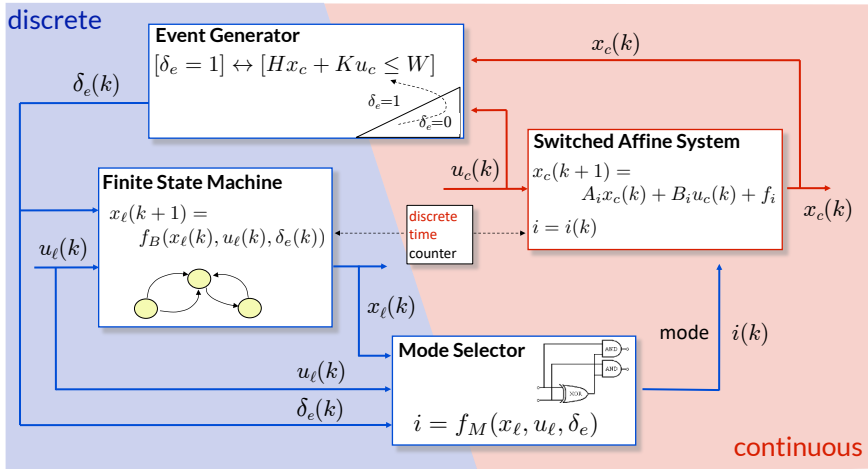\end{aligned}
$$

- PWA systems can approximate nonlinear dynamics arbitrarily well (even discontinuous ones)

- Model **descriptiveness** vs model **complexity tradeoff**

DISCRETE HYBRID AUTOMATON (DHA)

$x_\ell \in \{0,1\}^{n_\ell}$ = **binary state** $\qquad$ $x_c \in \mathbb{R}^{n_c}$ $\qquad$ = **real-valued state**

$u_\ell \in \{0,1\}^{m_\ell}$ = **binary input** $\qquad$ $u_c \in \mathbb{R}^{m_c}$ $\qquad$ = **real-valued input**

$\delta_e \in \{0,1\}^{n_e}$ = **event variable** $\qquad$ $i \in \{1,\ldots,s\}$ $\qquad$ = **current mode**
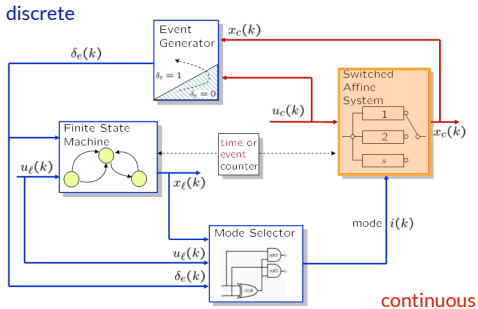
- The **affine dynamics** depend on the current mode $i(k)$:

$$x_c(k+1) = A_{i(k)}x_c(k) + B_{i(k)}u_c(k) + f_{i(k)}$$

$x_c \in \mathbb{R}^{n_c}, u_c \in \mathbb{R}^{m_c}$

EVENT GENERATOR

discrete
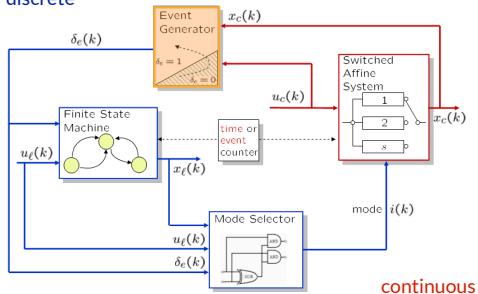
continuous

- **Event variables** are generated by **linear threshold conditions** over continuous

$$[\delta_e^i(k) = 1] \leftrightarrow [H^i x_c(k) + K^i u_c(k) \leq W^i]$$

$x_c \in \mathbb{R}^{n_c}, \quad u_c \in \mathbb{R}^{m_c}$
$\delta_e \in \{0,1\}^{n_e}$

- Example: $[\delta_e(k) = 1] \leftrightarrow [x_c(k) \geq 0]$

FINITE STATE MACHINE

discrete
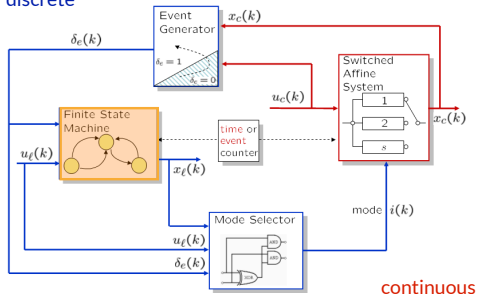
continuous

- The binary state of the **finite state machine** evolves according to a Boolean state update function $f_B : \{0,1\}^{n_\ell + m_\ell + n_e} \to \{0,1\}^{n_\ell}$:

$$x_\ell(k+1) = f_B(x_\ell(k), u_\ell(k), \delta_e(k)) \qquad \begin{array}{l} x_\ell \in \{0,1\}^{n_\ell}, \quad u_\ell \in \{0,1\}^{m_\ell} \\ \delta_e \in \{0,1\}^{n_e} \end{array}$$

- Example: $x_\ell(k+1) = \neg \delta_e(k) \vee (x_\ell(k) \wedge u_\ell(k))$

MODE SELECTOR

discrete

The mode selector can be seen as the output function of the discrete dynamics
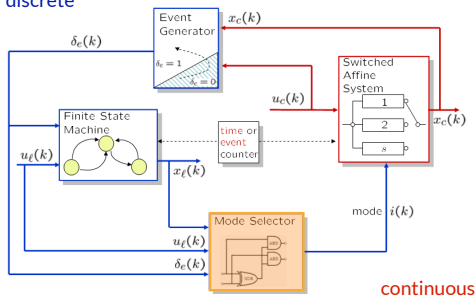
continuous

- The active **mode** $i(k)$ is selected by a Boolean function of the current binary

$$i(k) = f_M(x_\ell(k), u_\ell(k), \delta_e(k))$$

$x_\ell \in \{0,1\}^{n_\ell}, \quad u_\ell \in \{0,1\}^{m_\ell}$
$\delta_e \in \{0,1\}^{n_e}$

- Example:

$$i(k) = \begin{bmatrix} \neg u_\ell(k) \vee x_\ell(k) \\ u_\ell(k) \wedge x_\ell(k) \end{bmatrix}$$

u_ℓ/x_ℓ	0	1
0	$i = \begin{bmatrix} 1 \\ 0 \end{bmatrix}$	$i = \begin{bmatrix} 1 \\ 0 \end{bmatrix}$
1	$i = \begin{bmatrix} 0 \\ 0 \end{bmatrix}$	$i = \begin{bmatrix} 1 \\ 1 \end{bmatrix}$

the system has 3 modes

CONVERSION OF LOGIC FORMULAS TO LINEAR INEQUALITIES

- Key observation: $X_1 \lor X_2 = \texttt{true}$ $\implies$ $\delta_1 + \delta_2 \geq 1, \delta_1, \delta_2 \in \{0, 1\}$
- We want to impose the Boolean statement

$$F(X_1, \ldots, X_n) = \texttt{true}$$

- Convert the formula to **Conjunctive Normal Form (CNF)**

$$\bigwedge_{j=1}^{m} \left(\bigvee_{i \in P_j} X_i \bigvee_{i \in N_j} \bar{X}_i \right) = \texttt{true}, \quad P_j \cup N_j \subseteq \{1, \ldots, n\}$$

- Transform the CNF into the equivalent linear inequalities

$$\left\{ \begin{array}{ccc} \sum_{i \in P_1} \delta_i + \sum_{i \in N_1} (1 - \delta_i) & \geq & 1 \\ & \vdots & \vdots \\ \sum_{i \in P_m} \delta_i + \sum_{i \in N_m} (1 - \delta_i) & \geq & 1 \end{array} \right. \qquad \implies \qquad \begin{array}{c} A\delta \leq b, \ \delta \in \{0, 1\}^n \\ \textbf{polyhedron} \end{array}$$

Any logic proposition can be translated into integer linear inequalities

LOGIC → INEQUALITIES: SYMBOLIC APPROACH

- Example: we want to impose the following condition on X_1, X_2, X_3

$$F(X_1, X_2, X_3) = [X_3 \leftrightarrow X_1 \wedge X_2] = \texttt{true}$$

- Convert F to Conjunctive Normal Form (CNF):

  (see e.g. `https://www.wolframalpha.com` or just google "CNF + converter" ...)

$$(X_3 \vee \neg X_1 \vee \neg X_2) \wedge (X_1 \vee \neg X_3) \wedge (X_2 \vee \neg X_3)$$

- Transform into inequalities:

$$\left\{ \begin{array}{rcl} \delta_3 + (1 - \delta_1) + (1 - \delta_2) & \geq & 1 \\ \delta_1 + (1 - \delta_3) & \geq & 1 \\ \delta_2 + (1 - \delta_3) & \geq & 1 \end{array} \right.$$
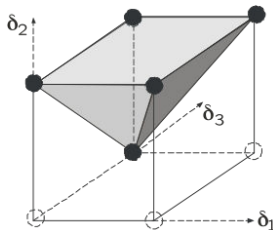
- Consider the Boolean statement $F(X_1, \ldots, X_n) = \texttt{true}$ and collect the rows of the **truth table** $T(F)$ of F

> The **convex hull** $P = \{\delta \in \mathbb{R}^n : A\delta \leq b\}$ of the **rows in** $T(F)$ is the smallest polytope equivalent to the Boolean statement F

(Mignone, Bemporad, Morari, 1999)

	X_1	X_2	...	X_n
	0	0	...	1
$T(F):$	0	1	...	1
	$\vdots$	$\vdots$	$\vdots$	$\vdots$
	1	1	...	0

- Convex hull packages: `cdd`, `lrs`, `qhull`, `chD`, `Hull`, `Porto`
  `CDDMEX` package by K. Fukuda included in the Hybrid Toolbox

LOGIC → INEQUALITIES: GEOMETRIC APPROACH

- Example: $F(X_1, X_2, X_3) = [X_3 \leftrightarrow X_1 \wedge X_2]$ (logic **and**)

$$T(F): \quad \begin{array}{c|c|c} X_1 & X_2 & X_3 \\ \hline 0 & 0 & 0 \\ 0 & 1 & 0 \\ 1 & 0 & 0 \\ 1 & 1 & 1 \end{array}$$
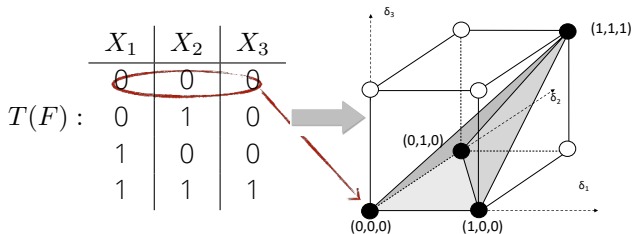
- **Key idea**: white points cannot be inside the convex hull of black points

$$\mathrm{conv}\left(\begin{bmatrix} 0 \\ 0 \\ 0 \end{bmatrix}, \begin{bmatrix} 0 \\ 1 \\ 0 \end{bmatrix}, \begin{bmatrix} 1 \\ 0 \\ 0 \end{bmatrix}, \begin{bmatrix} 1 \\ 1 \\ 1 \end{bmatrix}\right) = \left\{ \delta \in \mathbb{R}^3 : \begin{array}{rcl} -\delta_1 + \delta_3 & \leq & 0 \\ -\delta_2 + \delta_3 & \leq & 0 \\ \delta_1 + \delta_2 - \delta_3 & \leq & 1 \end{array} \right\}$$

```
>> V=struct('V',[0 0 0;0 1 0;1 0 0;1 1 1]);
>> H=cddmex('hull',V);A=H.A,b=H.B
```

GEOMETRIC VS SYMBOLIC APPROACH

- The polyhedron obtained via convex hull is the smallest one
- The one obtained via CNF may be larger. Example:

$$(X_1 \vee X_2) \wedge (X_1 \vee X_3) \wedge (X_2 \vee X_3) = \texttt{true}$$

convex hull   CNF

spurious vertex
in (0.5,0.5,0.5)
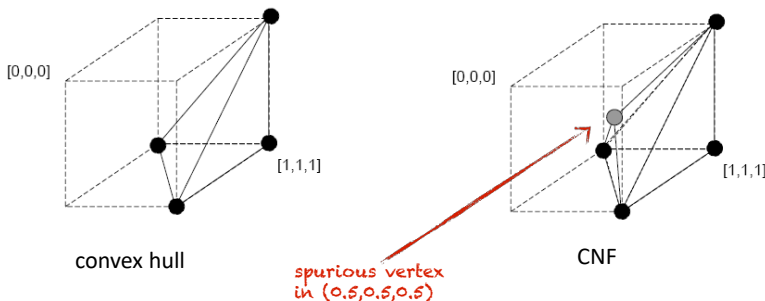
- **Note**: no other example with 3 vars but
  $(X_1 \vee X_2) \wedge (X_1 \vee X_3) \wedge (X_2 \vee X_3) \wedge (\neg X_1 \vee \neg X_2 \vee \neg X_3) = \texttt{true}$

BIG-M TECHNIQUE (IFF)

- Consider the **if-and-only-if** condition

$$[\delta = 1] \leftrightarrow [a'x_c - b \leq 0]$$

$$x_c \in \mathcal{X}$$
$$\delta \in \{0, 1\}$$

- Assume $\mathcal{X} \subset \mathbb{R}^{n_c}$ bounded. Let M and m such that $\forall x_c \in \mathcal{X}$

$$
\begin{aligned}
M &> a'x_c - b \\
m &< a'x_c - b
\end{aligned}
$$

- The if-and-only-if condition is equivalent to

$$
\left\{
\begin{aligned}
a'x_c - b &\leq M(1 - \delta) \\
a'x_c - b &> m\delta
\end{aligned}
\right.
$$

- We can replace the second constraint with $a'x_c - b \geq \epsilon + (m - \epsilon)\delta$ to avoid strict inequalities, where $\epsilon > 0$ is a small number (e.g., the machine precision)

COMPUTING THE BIG-M

- If $\mathcal{X}$ is a polyhedron, we can use linear programming

$$\begin{array}{rcl} m & < & \min_{x_c \in \mathcal{X}} \{a'x_c\} - b \\ M & > & -\min_{x_c \in \mathcal{X}} \{-a'x_c\} - b \end{array}$$

- If $\mathcal{X}$ is a box, $\mathcal{X} = [\ell, u]$, we can use the following simpler method

(Lee, Kouvaritakis, 2000) (Bemporad, 2022)

$$\begin{array}{rcl} a_+ & = & \max\{a, 0\} \\ a_- & = & a_+ - a = \max\{-a, 0\} \\ m & < & a'_+ \ell - a'_- u - b \\ M & > & a'_+ u - a'_- \ell - b \end{array}$$

BIG-M TECHNIQUE (IF-THEN-ELSE)

- Consider the **if-then-else** condition

$$
z = \begin{cases} a_1' x_c - b_1 & \text{if } \delta = 1 \\ a_2' x_c - b_2 & \text{otherwise} \end{cases}
\qquad
\begin{array}{l} x_c \in \mathcal{X} \\ \delta \in \{0,1\} \\ z \in \mathbb{R} \end{array}
$$

- Assume $\mathcal{X} \subset \mathbb{R}^{n_c}$ bounded. Let M_1, M_2 and m_1, m_2 such that $\forall x_c \in \mathcal{X}$

$$
\begin{array}{ccccc}
M_1 & > & a_1' x_c - b_1 & > & m_1 \\
M_2 & > & a_2' x_c - b_2 & > & m_2
\end{array}
$$

- The if-then-else condition is equivalent to

$$
\begin{cases}
z & \leq & a_1' x_c - b_1 - (m_1 - M_2)(1 - \delta) \\
z & \geq & a_1' x_c + b_1 + (m_2 - M_1)(1 - \delta) \\
z & \leq & a_2' x_c - b_2 - (m_2 - M_1)\delta \\
z & \geq & a_2' x_c - b_2 + (m_1 - M_2)\delta
\end{cases}
$$

SWITCHED AFFINE SYSTEM

- The state-update equation of a SAS can be rewritten as

Switched Affine System
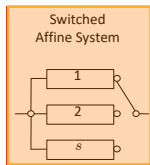
$$x_c(k+1) = \sum_{i=1}^{s} z_i(k) \quad z_i(k) \in \mathbb{R}^{n_c}$$

with

$$z_1(k) = \begin{cases} A_1 x_c(k) + B_1 u_c(k) + f_1 & \text{if } \delta_1(k) = 1 \\ 0 & \text{otherwise} \end{cases}$$

$$\vdots$$

$$z_s(k) = \begin{cases} A_s x_c(k) + B_s u_c(k) + f_s & \text{if } \delta_s(k) = 1 \\ 0 & \text{otherwise} \end{cases}$$

and with $\delta_i(k) \in \{0, 1\}$ subject to the **exclusive or** condition

$$\sum_{i=1}^{s} \delta_i(k) = 1 \text{ or equivalently } \begin{cases} \sum_{i=1}^{s} \delta_i(k) & \geq & 1 \\ \sum_{i=1}^{s} \delta_i(k) & \leq & 1 \end{cases}$$

- Output eq. $y_c(k) = C_i x_c(k) + D_i u_c(k) + g_i$ admits a similar transformation

TRANSFORMATION OF A DHA INTO LINEAR (IN)EQUALITIES

$X_1 \lor X_2 = \text{TRUE}$ $\implies$ $\delta_1 + \delta_2 \geq 1, \qquad \delta_1, \delta_2 \in \{0, 1\}$

Any logic statement
$f(X) = \text{TRUE}$

$\bigwedge_{j=1}^{m} \left(\lor_{i \in P_j} X_i \lor_{i \in N_j} \neg X_i \right)$ (CNF)
$N_j, P_j \subseteq \{1, \ldots, n\}$

$$\begin{cases} 1 \leq \sum_{i \in P_1} \delta_i + \sum_{i \in N_1} (1 - \delta_i) \\ \vdots \\ 1 \leq \sum_{i \in P_m} \delta_i + \sum_{i \in N_m} (1 - \delta_i) \end{cases}$$
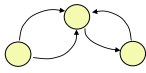
$[\delta_e^i(k) = 1] \leftrightarrow [H^i x_c(k) \leq W^i]$

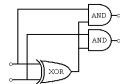$$\begin{cases} H^i x_c(k) - W^i \leq M^i(1 - \delta_e^i(k)) \\ H^i x_c(k) - W^i > m^i \delta_e^i(k) \end{cases}$$

IF $[\delta = 1]$ THEN $z = a_1^T x + b_1^T u + f_1$
ELSE $z = a_2^T x + b_2^T u + f_2$

$$\begin{cases} (m_1 - M_2)(1 - \delta) + z \leq a_1 x + b_1 u + f_1 \\ (m_2 - M_1)(1 - \delta) - z \leq -a_1 x - b_1 u - f_1 \\ (m_2 - M_1)\delta + z \leq a_2 x + b_2 u + f_2 \\ (m_1 - M_2)\delta - z \leq -a_2 x - b_2 u - f_2 \end{cases}$$
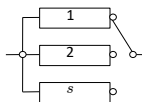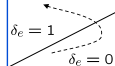
Finite State Machine
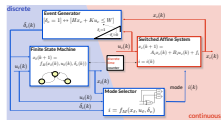
Mode Selector

Switched Affine System

1

2

s

Event Generator

$\delta_e = 1$

$\delta_e = 0$

MIXED LOGICAL DYNAMICAL (MLD) SYSTEMS

- By converting logic relations into mixed-integer linear inequalities
  a DHA can be rewritten as the **Mixed Logical Dynamical (MLD)** system

$$\begin{cases} x(k+1) &=& Ax(k) + B_1 u(k) + B_2 \delta(k) + B_3 z(k) + B_5 \\ y(k) &=& Cx(k) + D_1 u(k) + D_2 \delta(k) + D_3 z(k) + D_5 \\ E_2 \delta(k) &+& E_3 z(k) \leq E_4 x(k) + E_1 u(k) + E_5 \end{cases}$$

$$x \in \mathbb{R}^{n_c} \times \{0,1\}^{n_b}, \; u \in \mathbb{R}^{m_c} \times \{0,1\}^{m_b}$$
$$y \in \mathbb{R}^{p_c} \times \{0,1\}^{p_b}, \; \delta \in \{0,1\}^{r_b}, \; z \in \mathbb{R}^{r_c}$$

- The translation from DHA to MLD can be automatized, see e.g. the language
  HYSDEL (HYbrid Systems DEscription Language) (Torrisi, Bemporad, 2004)

- MLD models allow solving MPC, verification, state estimation, and fault
  detection problems via **mixed-integer programming**

A SIMPLE EXAMPLE OF MLD SYSTEM

- PWA system[1]:
$$x(k+1) = \begin{cases} 0.8x(k) + u(k) & \text{if} \quad x(k) \geq 0 \\ -0.8x(k) + u(k) & \text{if} \quad x(k) < 0 \end{cases}$$

- Introduce event variable $[\delta(k) = 1] \leftrightarrow [x(k) \geq 0]$ and use big-M technique:

$$\Rightarrow \quad \begin{aligned} x(k) &\geq m(1 - \delta(k)) \\ x(k) &\leq -\epsilon + (M + \epsilon)\delta(k) \end{aligned} \qquad \begin{aligned} M &= -m = 10 \\ \epsilon &> 0 \text{ "small"} \end{aligned}$$
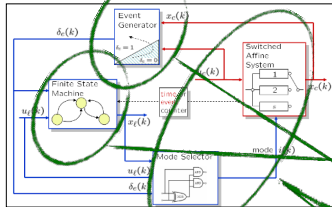
- Since $x(k+1) = 1.6\delta(k)x(k) - 0.8x(k) + u(k)$, introduce the aux variable

$$z(k) = \delta(k)x(k) \Rightarrow \quad \begin{aligned} z(k) &\leq M\delta(k) \\ z(k) &\geq m\delta(k) \\ z(k) &\leq x(k) - m(1 - \delta(k)) \\ z(k) &\geq x(k) - M(1 - \delta(k)) \end{aligned} \qquad \delta(k) \in \{0, 1\}$$

- Linear state update: $\quad x(k+1) = -0.8x(k) + 1.6z(k) + u(k)$

---
[1]This is the nonlinear system $x(k+1) = 0.8|x(k)| + u(k)$

DHA AND HYSDEL MODELS


```
SYSTEM name {
    INTERFACE {
        STATE {
            REAL xc [xmin,xmax];
            BOOL xl; }
        INPUT {
            REAL uc [umin,umax];
            BOOL ul; }
        PARAMETER {
            REAL param1 = 1;}
    } /* end of interface */

    IMPLEMENTATION {
        AUX { BOOL d;
              REAL z; }

        AUTOMATA { xl = xl & ~ul; }

        AD { d = xc - 1 <= 0; }

        DA { z = { IF d THEN 2*xc ELSE -xc }; }

        CONTINUOUS {
            xc = z; }

        MUST {
            xc + uc <= 2;
            ~(xl & ul); }
    } /* end implementation */
} /* end system */
```
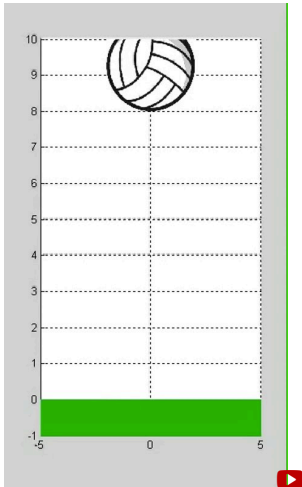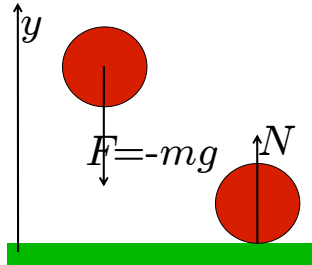
Additional relations
constraining system's
variables

$$\ddot{y} = -g$$
$$y \leq 0 \Rightarrow \dot{y}(t^+) = -(1-\alpha)\dot{y}(t^-)$$
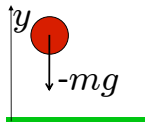
$$\alpha \in [0,1]$$

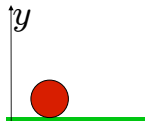How to model the bouncing ball as a discrete-time hybrid system ?

BOUNCING BALL – TIME DISCRETIZATION

- Case $y(k) > 0$ (ball falling):
$$\begin{array}{rcl} v(k) & \approx & \frac{y(k)-y(k-1)}{T_s} \\ -g & \approx & \frac{v(k)-v(k-1)}{T_s} \end{array}$$

$$\left\{\begin{array}{rcl} v(k+1) & = & v(k) - T_s g \\ y(k+1) & = & y(k) + T_s v(k+1) \\ & = & y(k) + T_s v(k) - T_s^2 g \end{array}\right.$$

- Case $y(k) \leq 0$ (ground level):
$$\begin{array}{rcl} v(k+1) & = & -(1-\alpha)v(k) \\ y(k+1) & = & y(k-1) = y(k) - T_s v(k) \end{array}$$

$$\left\{\begin{array}{rcl} v(k+1) & = & -(1-\alpha)v(k) \\ y(k+1) & = & y(k) - T_s v(k) \end{array}\right.$$

- We need a binary variable $[\delta(k) = 1] \leftrightarrow [y(k) \leq 0]$

```
SYSTEM bouncing_ball {
INTERFACE {
/* Description of variables and constants */
        STATE { REAL height [-10,10];
                REAL velocity [-100,100];    }

        PARAMETER {
                REAL g;
                REAL alpha;  /* 0=elastic, 1=completely anelastic */
                REAL Ts; }
}
IMPLEMENTATION {
        AUX {   BOOL negative;
                REAL hnext;
                REAL vnext;    }

        AD {    negative = height <= 0; }

        DA {    hnext = {  IF negative THEN height-Ts*velocity
                        ELSE height+Ts*velocity-Ts*Ts*g};
                vnext = {  IF negative THEN -(1-alpha)*velocity
                        ELSE velocity-Ts*g};   }
        CONTINUOUS {
                height  = hnext;
                velocity = vnext;}
}}
```

go to demo `demos/hybrid/bball.m`

BOUNCING BALL - SIMULATION

```
>> Ts=0.05;
>> g=9.8;
>> alpha=0.3;

>> S=mld('bouncing_ball',Ts);

>> N=150;
>> U=zeros(N,0);
>> x0=[5 0]';

>> [X,T,D]=sim(S,x0,U);
```
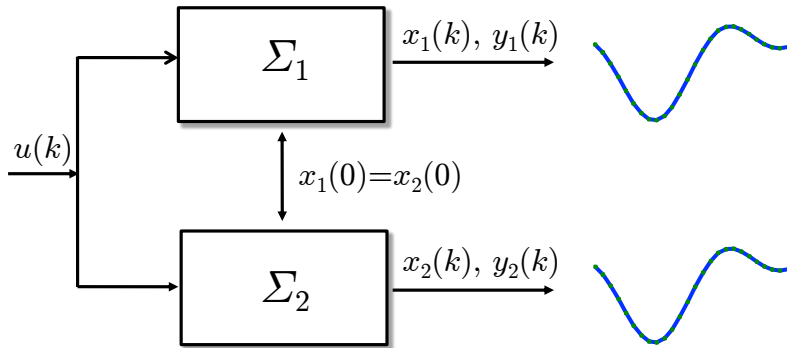
- **Note**: no **Zeno effect** in discrete time !

Height

Velocity

Event variable 'negative'
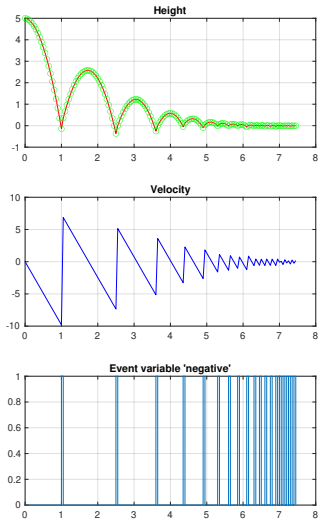
EQUIVALENCE OF HYBRID MODELS

EQUIVALENCE OF HYBRID MODELS

- Two hybrid models Σ_1, Σ_2 are **equivalent** if for all initial states $x_1(0) = x_2(0)$ and input excitations $u_1(k) \equiv u_2(k)$, the corresponding trajectories $x_1(k) \equiv x_2(k)$ and $y_1(k) \equiv y_2(k), \forall k = 0, 1, \ldots$

EQUIVALENCE OF HYBRID MODELS

- **MLD** and **PWA** systems are equivalent (Bemporad, Ferrari-Trecate, Morari, 2000)

  <u>Proof</u>: For a given combination (x_ℓ, u_ℓ, δ) of an MLD model, the state and output equation are linear and valid in a polyhedron.

  Conversely, a PWA system can be modeled as MLD system (see next slide)

- Efficient **conversion algorithms** from MLD to PWA form exist

  (Bemporad, 2004) (Geyer, Torrisi, Morari, 2003)

- Further equivalences exist with other classes of hybrid dynamical systems, such as **Linear Complementarity (LC)** systems (Heemels, De Schutter, Bemporad, 2001)

MODELING PWA SYSTEMS IN MLD FORM

- PWA system with bounded states $x \in \mathcal{X}$ and inputs $u \in \mathcal{U}$ with s regions

$$
\begin{array}{rcl}
x(k+1) & = & A_{i(k)}x(k) + B_{i(k)}u(k) + f_{i(k)} \\
y(k) & = & C_{i(k)}x(k) + D_{i(k)}u(k) + g_{i(k)} \\
i(k) & = & \text{such that } \begin{bmatrix} x(k) \\ u(k) \end{bmatrix} \in \mathcal{C}_{i(k)}
\end{array}
\qquad i(k) \in \{1, 2, \ldots, s\}
$$

- The sets $\mathcal{C}_i = \{ \begin{bmatrix} x \\ u \end{bmatrix} : H_i x + J_i u \leq K_i \}$ make a **polyhedral partition** of $\mathcal{X} \times \mathcal{U}$
  ($\cup_{i=1}^{s} \mathcal{C}_i = \mathcal{X} \times \mathcal{U}$ and $\overset{\circ}{\mathcal{C}}_i \cap \overset{\circ}{\mathcal{C}}_j = \emptyset, \forall i \neq j, i, j = 1, \ldots, s$)

- Introduce s binary variables $\delta_i, i = 1, \ldots, s$ and set the logic constraints

$$
\begin{aligned}
&[\delta_i = 1] \to [H_i x + J_i u \leq K_i] \\
&\bigoplus_{i=1}^{s} [\delta_i = 1] = \texttt{true}
\end{aligned}
\qquad\Longrightarrow\qquad
\begin{aligned}
&H_i x + J_i u \leq K_i + M_i(1 - \delta_i) \\
&\sum_{i=1}^{s} \delta_i = 1
\end{aligned}
$$

where the vector M_i of upper-bounds can be computed, e.g., via LP

MODELING PWA SYSTEMS IN MLD FORM

- Introduce auxiliary real vectors z_i, w_i defined by if-then-else rules

$$z_i = \begin{cases} A_i x + B_i u + f_i & \text{if } \delta_i = 1 \\ 0 & \text{otherwise} \end{cases} \qquad w_i = \begin{cases} C_i x + D_i u + g_i & \text{if } \delta_i = 1 \\ 0 & \text{otherwise} \end{cases}$$

and convert into mixed-integer inequalities (**big-M technique**):

$$\begin{cases} z_i & \leq & A_i x + B_i u + f_i - m_i(1 - \delta_i) \\ z_i & \geq & A_i x + B_i u + f_i - M_i(1 - \delta_i) \\ z_i & \leq & M_i \delta \\ z_i & \geq & m_i \delta \end{cases} \qquad m \leq A_i x + B_i u + f_i \leq M_i$$

- Finally, write the state update and output equations

$$x(k+1) = \sum_{i=1}^{s} z_i(k) \qquad y(k) = \sum_{i=1}^{s} w_i(k)$$

MODELING PWA SYSTEMS VIA DISJUNCTIVE PROGRAMMING

- A PWA system with bounded states and inputs is equivalent to the **disjunction**

$$\bigvee_{i=1}^{s} \left[\begin{array}{c} H_i x(k) + J_i u(k) \leq K_i \\ x(k+1) = A_i x(k) + B_i u(k) + f_i \end{array} \right] \quad \begin{array}{c} x_{\ell b} \leq x(k) \leq x_{ub} \\ u_{\ell b} \leq u(k) \leq u_{ub} \end{array}$$

- Introduce s binary variables $\delta_1(k), \ldots, \delta_s(k)$ subject to $\displaystyle\sum_{i=1}^{s} \delta_i(k) = 1$

- Introduce the **convex hull relaxation** of the disjunction

$$x(k) = \sum_{i=1}^{s} v_i(k), \quad x_{\ell b}\delta_i(k) \leq v_i(k) \leq x_{ub}\delta_i(k)$$

$$u(k) = \sum_{i=1}^{s} w_i(k), \quad u_{\ell b}\delta_i(k) \leq w_i(k) \leq u_{ub}\delta_i(k)$$

and set

$$x(k+1) = \sum_{i=1}^{s} A_i v_i(k) + B_i w_i(k) + f_i \delta_i(k), \quad H_i v_i(k) + J_i w_i(k) \leq K_i \delta_i(k)$$

MODELING PWA SYSTEMS WITHOUT AUX CONTINUOUS VARIABLES

(Bemporad, 2022)

- Only introduce s binary variables $\delta_i, i = 1, \ldots, s$ and set:

$$\underbrace{m_i^x(1 - \delta_i(k)) \leq x(k+1) - A_i x(k) - B_i u(k) - f_i \leq M_i^x(1 - \delta_i(k))}_{[\delta_i(k) = 1] \rightarrow [x(k+1) = A_i x(k) + B_i u(k) + f_i]}$$

$$\underbrace{m_i^y(1 - \delta_i(k)) \leq y(k) - C_i x(k) - D_i u(k) - g_i \leq M_i^y(1 - \delta_i(k))}_{[\delta_i(k) = 1] \rightarrow [y(k) = C_i x(k) + D_i u(k) + g_i]}$$

$$H_i x(k) + J_i u(k) \leq K_i + M_i(1 - \delta_i(k)) \qquad {\scriptstyle [\delta_i(k) = 1] \rightarrow [H_i x(k) + J_i u(k) \leq K_i]}$$
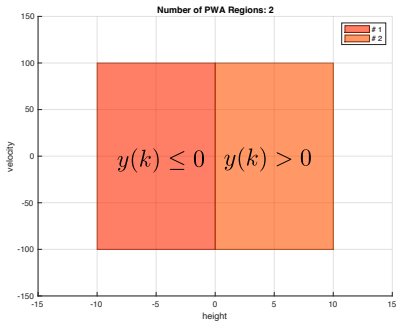
$$\sum_{i=1}^{s} \delta_i(k) = 1 \qquad\qquad {\scriptstyle \bigoplus_{i=1}^{s} [\delta_i(k) = 1] = \texttt{true}}$$

where $m_i^x, M_i^x, m_i^y, M_i^y$ are suitably defined upper and lower bounds

```
>> P=pwa(S);
>> plot(P)

>> [X,T,I]=sim(P,x0,U);
```
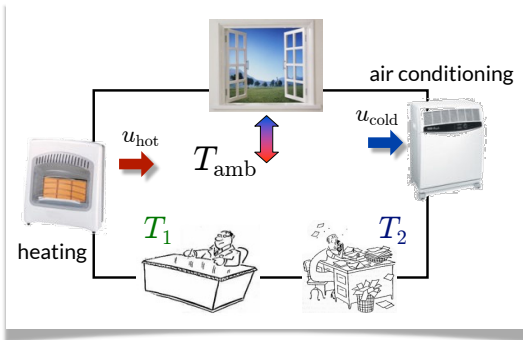(Bemporad, 2004)

Number of PWA Regions: 2

$y(k) \leq 0 \quad y(k) > 0$

Height

Velocity

PWA mode
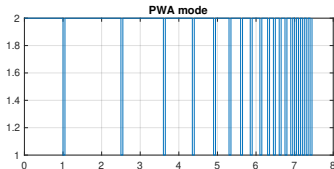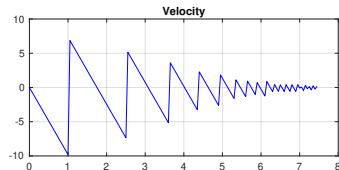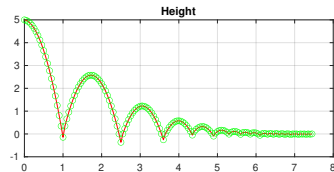
EXAMPLE: ROOM TEMPERATURE CONTROL

discrete dynamics

- #1 = cold → heater = on
- #2 = cold → heater = on **unless** #1 hot
- A/C activation has similar rules

continuous dynamics

$$\frac{dT_i}{dt} = -\alpha_i(T_i - T_{\mathrm{amb}}) + k_i(u_{\mathrm{hot}} - u_{\mathrm{cold}})$$

$$i = 1, 2$$

go to demo `demos/hybrid/heatcool.m`

EXAMPLE: ROOM TEMPERATURE CONTROL

```
SYSTEM heatcool {

INTERFACE {
    STATE { REAL T1 [-10,50];
            REAL T2 [-10,50]; }
    INPUT { REAL Tamb [-50,50]; }
    PARAMETER {
        REAL Ts, alpha1, alpha2, k1, k2;
        REAL Thot1, Tcold1, Thot2, Tcold2, Uc, Uh; }
    }

IMPLEMENTATION {
    AUX { REAL uhot, ucold;
          BOOL hot1, hot2, cold1, cold2; }

    AD { hot1 = T1>=Thot1;
         hot2 = T2>=Thot2;
         cold1 = T1<=Tcold1;
         cold2 = T2<=Tcold2; }

    DA { uhot = { IF cold1 | (cold2 & ~hot1) THEN Uh ELSE 0 };
         ucold = { IF hot1 | (hot2 & ~cold1) THEN Uc ELSE 0 }; }

    CONTINUOUS { T1 = T1+Ts*(-alpha1*(T1-Tamb)+k1*(uhot-ucold));
                 T2 = T2+Ts*(-alpha2*(T2-Tamb)+k2*(uhot-ucold)); }
    }
}
```

```
>> S=mld('heatcoolmodel',Ts);
```
get the MLD model in MATLAB

```
>> [XX,TT]=sim(S,x0,U);
```
simulate the MLD model

EXAMPLE: ROOM TEMPERATURE CONTROL

- MLD model of the room temperature system

$$\begin{cases} x(k+1) &=& Ax(k) + B_1u(k) + B_2\delta(k) + B_3z(k) + B_5 \\ y(k) &=& Cx(k) + D_1u(k) + D_2\delta(k) + D_3z(k) + D_5 \\ E_2\delta(k) &+& E_3z(k) \leq E_4x(k) + E_1u(k) + E_5 \end{cases}$$

- 2 continuous states   (temperature T_1, T_2)

- 1 continuous input   (room temperature T_{amb})

- 2 auxiliary continuous vars   (power flows $u_{\mathrm{hot}}, u_{\mathrm{cold}}$)

- 6 auxiliary binary vars   (4 threshold events + 2 for the OR condition)

- 20 mixed-integer inequalities

- In principle, we have $2^6 = 64$ possible combinations of binary variables
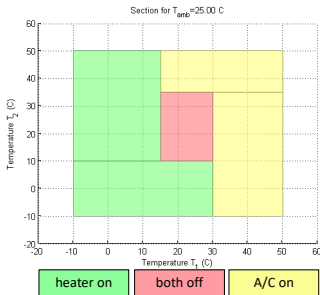
EXAMPLE: ROOM TEMPERATURE CONTROL

- PWA model of the room temperature system

$$
\begin{aligned}
x(k+1) &= A_{i(k)}x(k) + B_{i(k)}u(k) + f_{i(k)} \\
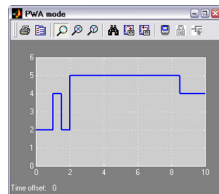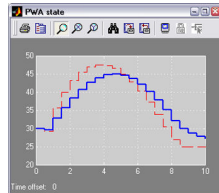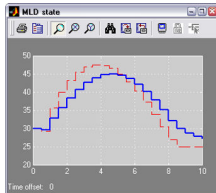y(k) &= C_{i(k)}x(k) + D_{i(k)}u(k) + g_{i(k)}
\end{aligned}
$$

```
>> P=pwa(S);
```

$$
i(k) \quad \text{s.t.} \quad H_{i(k)}x(k) + J_{i(k)}u(k) \leq K_{i(k)}
$$

Section for T_{amb}=25.00 C

| heater on | both off | A/C on |
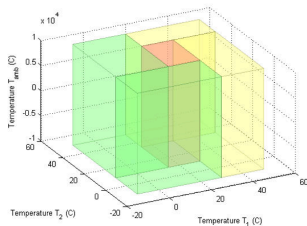
5 polyhedral regions

(partition does not depend on input)

2 continuous states (T_1, T_2)
1 continuous input (T_{amb})

EXAMPLE: ROOM TEMPERATURE CONTROL
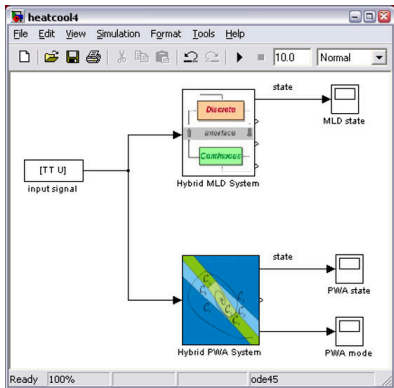
- MLD and PWA models are equivalent, hence simulated states are the same

USING PWA EQUIVALENCE FOR MODEL ANALYSIS

- Assume plant + controller can be modeled as DHA:

    – **plant** = approximated as PWA system (e.g.: nonlinear switched model)

    – **controller** = switched linear controller (e.g: combination of threshold conditions, logic, linear feedback laws, …)

- Convert DHA to MLD form, then to PWA form

- The resulting closed-loop PWA model reveals how the closed-loop system behaves in different regions of the state-space

- Can analyze **closed-loop stability** analysis using **piecewise quadratic Lyapunov functions** (Johansson, Rantzer, 1998) (Mignone, Ferrari-Trecate, Morari, 2000)

OTHER EXISTING HYBRID MODELS

- **Linear complementarity (LC)** systems (Heemels, 1999)

$$
\begin{aligned}
x(k+1) &= Ax(k) + B_1 u(k) + B_2 w(k) \\
y(k) &= Cx(k) + D_1 u(k) + D_2 w(k) \\
v(k) &= E_1 x(k) + E_2 u(k) + E_3 w(k) + E_4 \\
0 &\leq v(k) \perp w(k) \geq 0
\end{aligned}
$$

Examples:
mechanical systems,
electrical circuits

- **Min-max-plus-scaling (MMPS)** systems (De Schutter, Van den Boom, 2000)

$$
\begin{aligned}
x(k+1) &= M_x(x(k), u(k), w(k)) \\
y(k) &= M_y(x(k), u(k), w(k)) \\
0 &\geq M_c(x(k), u(k), w(k))
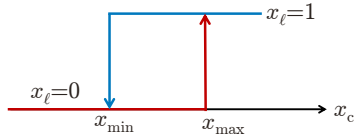\end{aligned}
$$

Example:
discrete-event system
k = event counter

where $M_{()}$ are MMPS functions defined by the grammar

$$M := x_i|\alpha| \max(M_1, M_2)| \min(M_1, M_2)|M_1 + M_2|\beta M_1$$

Example: $x(k+1) = 2\max(x(k), 0) + \min(\frac{1}{2}u(k), 1)$

MODELING HYSTERESIS

- Hysteresis between $x_{\min} \leq x_c(k) \leq x_{\max}$

- Introduce two binary variables

$$[\delta_{\min}(k) = 1] \quad \leftrightarrow \quad [x_c(k) \leq x_{\min}]$$
$$[\delta_{\max}(k) = 1] \quad \leftrightarrow \quad [x_c(k) \geq x_{\max}]$$

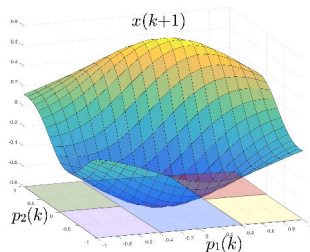- Introduce logic state $x_\ell \in \{0, 1\}$ with dynamics

$$x_\ell(k + 1) = (x_\ell(k) \land \neg\delta_{\min}(k)) \lor (\neg x_\ell(k) \land \delta_{\max}(k))$$

IDENTIFICATION OF HYBRID SYSTEMS

HYBRID SYSTEM IDENTIFICATION

- A hybrid model of the process may not be available from physical principles
- Therefore, a model must be either
  - **estimated from data** (model is unknown)
  - or **hybridized** (model is known but nonlinear)
- If one linear model is enough: easy problem (SYS-ID TBX) (Ljung, 1999)
- If switching sequence known: easy, just identify one linear model per mode
- If modes & dynamics must be identified simultaneously, we need **hybrid system identification** (or **piecewise affine regression**)

In industrial MPC most effort is spent in **identifying (multiple) linear prediction models** from data
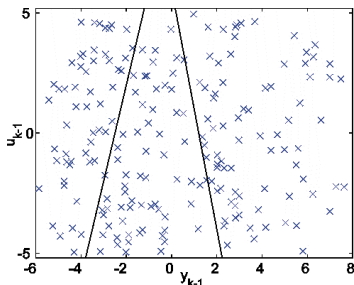
Problem: Estimate from data **both** the **parameters** of the affine submodels and the **partition** of the PWA map

Example: Let the data be generated by the PWARX system

$$
y_k = \begin{cases}
\begin{bmatrix} -0.4 & 1 & 1.5 \end{bmatrix} \phi_k + \epsilon_k \\
\quad \text{if } \begin{bmatrix} 4 & -1 & 10 \end{bmatrix} \phi_k < 0 \\
\begin{bmatrix} 0.5 & -1 & -0.5 \end{bmatrix} \phi_k + \epsilon_k \\
\quad \text{if } \begin{bmatrix} -4 & 1 & 10 \\ 5 & 1 & -6 \end{bmatrix} \phi_k \le 0 \\
\begin{bmatrix} -0.3 & 0.5 & -1.7 \end{bmatrix} \phi_k + \epsilon_k \\
\quad \text{if } \begin{bmatrix} -5 & -1 & 6 \end{bmatrix} \phi_k < 0
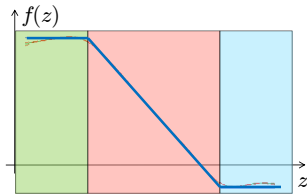\end{cases}
$$

with $\phi_k = [y_{k-1} \ u_{k-1} \ 1]'$, $|u_k| \le 5$, and $|\epsilon_k| \le 0.1$

PWA REGRESSION PROBLEM

- **Problem**: Given input/output pairs $\{x(k), y(k)\}, k = 1, \ldots, N$ and number s of models, compute a **piecewise affine** (PWA) approximation $y \approx f(x)$

$$v(k) = \begin{cases} F_1 z(k) + g_1 & \text{if } H_1 z(k) \leq K_1 \\ \vdots \\ F_s z(k) + g_s & \text{if } H_s z(k) \leq K_s \end{cases}$$

$$v(k) = \begin{bmatrix} x(k+1) \\ y(k) \end{bmatrix}, \quad z(k) = \begin{bmatrix} x(k) \\ u(k) \end{bmatrix}$$

- Need to learn **both** the parameters $\{F_i,\ g_i\}$ of the affine submodels **and** the partition $\{H_i,\ K_i\}$ of the PWA map from data (**offline learning**)

- Possibly update model+partition as new data are available (**online learning**)

APPROACHES TO PWA IDENTIFICATION

- Mixed-integer linear or quadratic programming (Roll, Bemporad, Ljung, 2004)

- Partition of infeasible set of inequalities (Bemporad, Garulli, Paoletti, Vicino, 2005)

- K-means clustering in a feature space (Ferrari-Trecate, Muselli, Liberati, Morari, 2003)

- Bayesian approach (Juloski, Wieland, Heemels, 2004)

- Kernel-based approaches (Pillonetto, 2016)

- Hyperplane clustering in data space (Münz, Krebs, 2002)

- **Recursive multiple least squares & PWL separation** (Breschi, Piga, Bemporad, 2016)

- Jump models (Bemporad, Breschi, Piga, Boyd, 2018)

- **Piecewise affine regression and classification** (PARC) (Bemporad, 2022)

- Any **machine learning** technique producing a PWA model, such as (leaky)ReLU-NNs, decision trees, softmax regression, K-nearest neighbors, ...

PWA REGRESSION ALGORITHM

1. Estimate models $\{F_i, \ g_i\}$ **recursively**. Let $e_i(k) = v(k) - F_i z(k) - g_i$ and only update model $i(k)$ such that

$$i(k) \leftarrow \arg\min_{i=1,\ldots,s} \underbrace{e_i(k)' \Lambda_e^{-1} e_i(k)}_{\substack{\text{one-step prediction error} \\ \text{of model } \#i}} + \underbrace{(z(k) - c_i)' R_i^{-1} (z(k) - c_i)}_{\substack{\text{proximity to centroid} \\ \text{of cluster } \#i}}$$
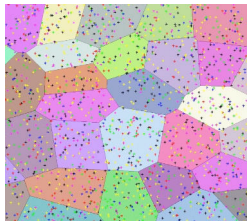
using **recursive LS** and **inverse QR decomposition** (Alexander, Ghirnikar, 1993)

This also splits the data points $z(k)$ in **clusters** $C_i = \{z(k) : \ i(k) = i\}$

2. Compute a polyhedral partition $\{H_i, \ K_i\}$ of the regressor space via **multi-category linear separation**

$$\phi(z) = \max_{i=1,\ldots,s} \{w_i' z - \gamma_i\} \qquad i(z) = \arg\max_i \{w_i' z - \gamma_i\}$$

3. Repeat the entire procedure M times

PWA REGRESSION EXAMPLES

- Identification of **piecewise-affine ARX** model

$$\begin{bmatrix} y_1(k) \\ y_2(k) \end{bmatrix} = \begin{bmatrix} -0.83 & 0.20 \\ 0.30 & -0.52 \end{bmatrix} \begin{bmatrix} y_1(k-1) \\ y_2(k-1) \end{bmatrix} + \begin{bmatrix} -0.34 & 0.45 \\ -0.30 & 0.24 \end{bmatrix} \begin{bmatrix} u_1(k-1) \\ u_2(k-1) \end{bmatrix}$$

$$+ \begin{bmatrix} 0.20 \\ 0.15 \end{bmatrix} + \max \left\{ \begin{bmatrix} 0.20 & -0.90 \\ 0.10 & -0.42 \end{bmatrix} \begin{bmatrix} y_1(k-1) \\ y_2(k-1) \end{bmatrix} \right.$$

$$\left. + \begin{bmatrix} 0.42 & 0.20 \\ 0.50 & 0.64 \end{bmatrix} \begin{bmatrix} u_1(k-1) \\ u_2(k-1) \end{bmatrix} + \begin{bmatrix} 0.40 \\ 0.30 \end{bmatrix}, \begin{bmatrix} 0 \\ 0 \end{bmatrix} \right\} + e(k),$$

- **Quality of fit**: best fit rate (BFR) $= \max \left\{ 1 - \frac{\|y_i - \hat{y}_i\|_2}{\|y_i - \bar{y}_i\|_2}, 0 \right\}, i = 1, 2$

  y = measured, $\hat{y}$ = open-loop simulated, $\bar{y}$ = sample mean of y

		$N = 4000$	$N = 20000$	$N = 100000$
y_1	(offline) RLP	96.0 %	96.5 %	99.0 %
	(Offline) RPSN	96.2 %	96.4 %	98.9 %
	(Online) ASGD	86.7 %	95.0 %	96.7 %
y_2	(offline) RLP	96.2 %	96.9 %	99.0 %
	(offline) RPSN	96.3 %	96.8 %	99.0 %
	(online) ASGD	87.4 %	95.2 %	96.4 %

BFR on validation data, open-loop validation

RLP = Robust linear programming
(Bennett, Mangasarian, 1994)

RPSN = Piecewise-smooth Newton method
(Bemporad, Bernardini, Patrinos, 2015)

ASGD = Averaged stochastic gradient descent
(Bottou, 2012)

- **CPU time for computing the partition**:
  (i7 2.40-GHz Intel core)

	$N = 4000$	$N = 20000$	$N = 100000$
(Offline) RLP	0.308 s	3.227 s	112.435 s
(Offline) RPSN	0.016 s	0.086 s	0.365 s
(Online) ASGD	0.013 s	0.023 s	0.067 s

PWA REGRESSION EXAMPLES

- Identification of **linear parameter varying ARX** model

$$\begin{bmatrix} y_1(k) \\ y_2(k) \end{bmatrix} = \begin{bmatrix} \bar{a}_{1,1}(p(k)) \; \bar{a}_{1,2}(p(k)) \\ \bar{a}_{2,1}(p(k)) \; \bar{a}_{2,2}(p(k)) \end{bmatrix} \begin{bmatrix} y_1(k-1) \\ y_2(k-1) \end{bmatrix}$$
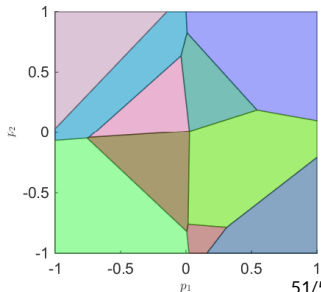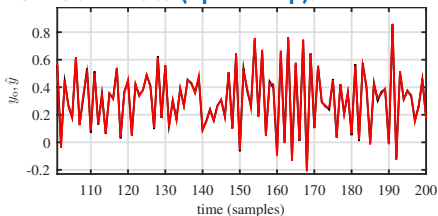$$+ \begin{bmatrix} \bar{b}_{1,1}(p(k)) \; \bar{b}_{1,2}(p(k)) \\ \bar{b}_{2,1}(p(k)) \; \bar{b}_{2,2}(p(k)) \end{bmatrix} \begin{bmatrix} u_1(k-1) \\ u_2(k-1) \end{bmatrix} + e_\mathrm{o}(k)$$

$\bar{a}(p)$ = PWA function of p
$\bar{b}(p)$ has **quadratic** and **sin** terms

- **Quality of fit (BFR)**:

	y_1	y_2
PWA regression	87 %	84 %
parametric LPV*	80 %	70 %

* (Bamieh, Giarré, 2002)

- **Validation data (open-loop)**:

IDENTIFICATION OF HYBRID SYSTEMS WITH LOGIC STATES

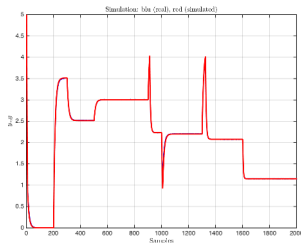- Identification of a **hybrid model** with logic states
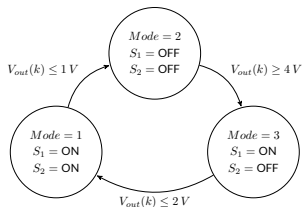
true system
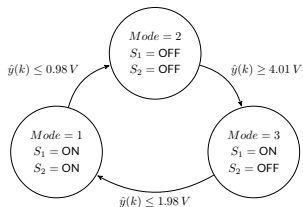
identified system
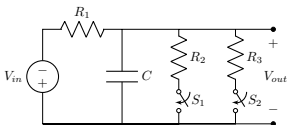
Quality of fit: BFR=96.64 % (validation)

CPU time for identification: 78 ms

(2000 samples, MacBook Pro 2.8 GHz)

# PARC - PIECEWISE AFFINE REGRESSION AND CLASSIFICATION

- **Piecewise Affine Regression and Classification** (**PARC**) algorithm

- Training dataset:
    - **feature vector** $z \in \mathbb{R}^n$ (categorical features **one-hot encoded** in $\{0, 1\}$)
    - **target vector** $v_c \in \mathbb{R}^{m_c}$ (numeric), $v_{di} \in \{w_{di}^1, \ldots, w_{di}^{m_i}\}$ (categorical)

- PARC iteratively **clusters** training data in K sets and **fits** linear predictors:
    1. fit $v_c = a_j z + b_j$ by **ridge regression** (=ℓ_2-regularized least squares)
    2. fit $v_{di} = w_{di}^{h_*}$, $h_* = \arg\max\{a_{dih}^h z + b_{di}^h\}$ by **softmax regression**
    3. fit a convex **PWL separation function** by **softmax regression**

$$\Phi(z) = \omega^{j(z)} z + \gamma^{j(z)}, \qquad j(z) = \min\left\{\arg\max_{j=1,\ldots,K}\{\omega^j z + \gamma^j\}\right\}$$

- Data reassigned to clusters based on weighted fit/PWL separation criterion

- PARC is a **block-coordinate descent** algorithm $\Rightarrow$ (local) convergence ensured

- Simple PWA regression example:

  - 1000 samples of $y = \sin(4x_1 - 5(x_2 - 0.5)^2) + 2x_2$ (use 80% for training)

 - PWA approximation over $K = 10$ polyhedral regions

Nonlinear function

PARC ($K = 10$)

PARC ($K = 10$)
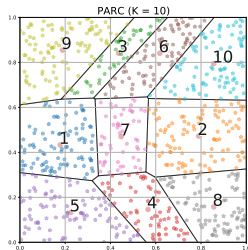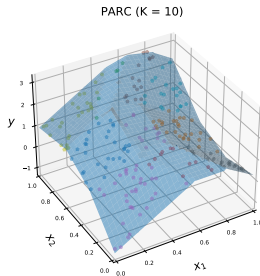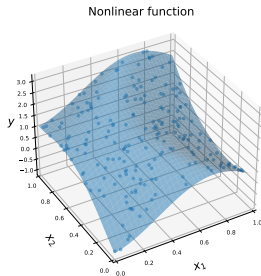
- Code: `pip install pyparc`   **github.com/bemporad/PyPARC**

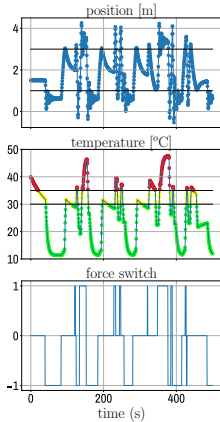- **Example**: moving cart and bumpers + heat transfer during bumps.

  Spring and viscous forces are **nonlinear**.

- Categorical input $F \in \{-\bar{F}, 0, \bar{F}\}$ and categorical output $c \in \{green, yellow, red\}$

- 4000 training samples

- Feature vector $z_k = [y_k, \dot{y}_k, T_k, F_k]$

- Target vector $v_k = [y_{k+1}, \dot{y}_{k+1}, T_{k+1}, c_k]$

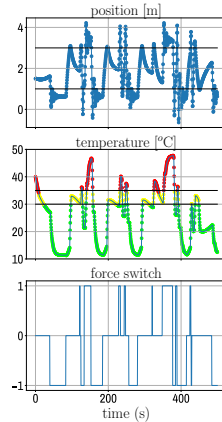- Hybrid model learned by **PARC** ($K = 5$ regions)
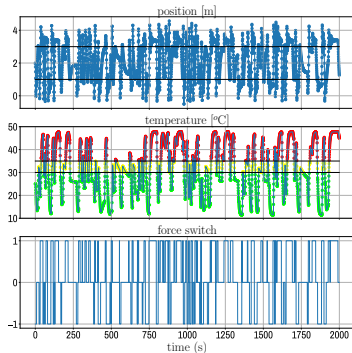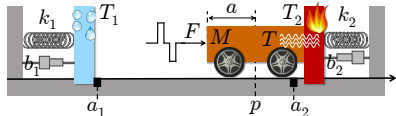
PARC - CART & BUMPERS EXAMPLE

- **Open-loop** simulation on 500 s **test** data:

continuous-time system

discrete-time PWA model

- Model fit is good enough for MPC design purposes (see later ...)