TEA152x

SMPS ICs for low-power systems Rev. 04 — 14 September 2010

Product data sheet

General description 1.

The TEA152x family STARplug is a Switched Mode Power Supply (SMPS) controller IC that operates directly from the rectified universal mains. It is implemented in the high-voltage EZ-HV SOI process, combined with a low-voltage Bipolar Complementary Metal-Oxide Semiconductor (BiCMOS) process. The device includes a high-voltage power switch and a circuit for start-up directly from the rectified mains voltage.

A dedicated circuit for valley switching is built in, which makes a very efficient slim-line electronic power-plug concept possible.

In its most basic version of application, the TEA152x family acts as a voltage source. Here, no additional secondary electronics are required. A combined voltage and current source can be realized with minimum costs for external components. Implementation of the TEA152x family renders an efficient and low cost power supply system.

Features and benefits 2.

- Designed for general purpose supplies up to 30 W
- Integrated power switch:
 - ◆ TEA1520x: 48 Ω; 650 V
 - TEA1521x: 24 Ω; 650 V
 - ♦ TEA1522x: 12 Ω; 650 V
 - ♦ TEA1523P: 6.5 Ω; 650 V
- Operates from universal AC mains supplies (80 V to 276 V)
- Adjustable frequency for flexible design
- RC oscillator for load insensitive regulation loop constant
- Valley switching for minimum switch-on loss
- Frequency reduction at low power output makes low standby power possible (< 100 mW)
- Adjustable overcurrent protection
- Undervoltage protection
- Temperature protection
- Short-circuit winding protection
- Simple application with both primary and secondary (opto) feedback
- Available in DIP8 and SO14 packages



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3. Applications

- Chargers
- Adapters
- Set-Top Box (STB)
- DVD
- CD(R)
- TV/monitor standby supplies
- PC peripherals
- Microcontroller supplies in home applications and small portable equipment, etc.

4. Quick reference data

Table 1. Quick reference data

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
V_{drain}	voltage on pin DRAIN	T _j > 0 °C	-0.4	-	+650	V
R_{DSon}	drain-source on-state resistance					
	TEA1520x	$I_{\text{source}} = -0.06 \text{ A}$				
		T _j = 25 °C	-	48	55.2	Ω
		T _j = 100 °C	-	68	78.2	Ω
	TEA1521x	$I_{\text{source}} = -0.125 \text{ A}$				
		T _j = 25 °C	-	24	27.6	Ω
		T _j = 100 °C	-	34	39.1	Ω
	TEA1522x	$I_{\text{source}} = -0.25 \text{ A}$				
		T _j = 25 °C	-	12	13.8	Ω
		T _j = 100 °C	-	17	19.6	Ω
	TEA1523P	$I_{\text{source}} = -0.50 \text{ A}$				
		T _j = 25 °C	-	6.5	7.5	Ω
		T _j = 100 °C	-	9.0	10.0	Ω
V_{CC}	supply voltage	continuous	-0.4	-	+40	V
f _{osc}	oscillator frequency		10	100	200	kHz
I _{drain}	current on pin DRAIN	V _{drain} > 60 V; no auxiliary supply	-	1.5	2	mA

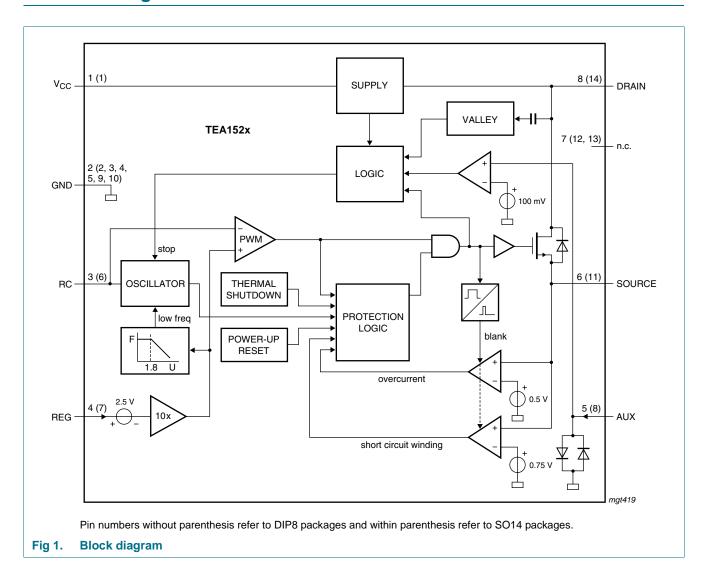
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5. Ordering information

Table 2. Ordering information

Type number	Package							
	Name	Description	Version					
TEA1520P	DIP8	plastic dual in-line package; 8 leads (300 mil)	SOT97-1					
TEA1521P	DIP8	plastic dual in-line package; 8 leads (300 mil)	SOT97-1					
TEA1522P	DIP8	plastic dual in-line package; 8 leads (300 mil)	SOT97-1					
TEA1523P	DIP8	plastic dual in-line package; 8 leads (300 mil)	SOT97-1					
TEA1520T	SO14	plastic small outline package; 14 leads; body width 3.9 mm	SOT108-1					
TEA1521T	SO14	plastic small outline package; 14 leads; body width 3.9 mm	SOT108-1					
TEA1522T	SO14	plastic small outline package; 14 leads; body width 3.9 mm	SOT108-1					

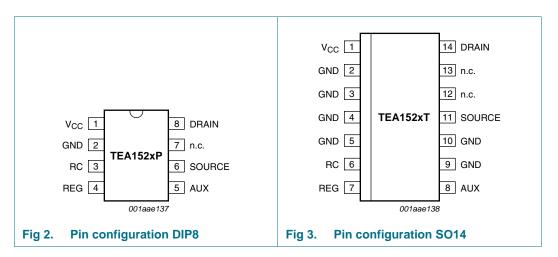
6. Block diagram



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7. Pinning information

7.1 Pinning



7.2 Pin description

Table 3. Pin description

Symbol	Pin		Description
	DIP8	SO14	
V_{CC}	1	1	supply voltage
GND	2	2, 3, 4, 5, 9, 10	ground
RC	3	6	frequency setting
REG	4	7	regulation input
AUX	5	8	input for voltage from the auxiliary winding for timing (demagnetization)
SOURCE	6	11	source of the internal MOS switch
n.c.	7	12, 13	not connected
DRAIN	8	14	drain of the internal MOS switch; input for the start-up current and valley sensing

8. Functional description

The TEA152x family is the heart of a compact flyback converter, with the IC placed at the primary side. The auxiliary winding of the transformer can be used for indirect feedback to control the isolated output. This additional winding also powers the IC. A more accurate control of the output voltage and/or current can be implemented with an additional secondary sensing circuit and optocoupler feedback.

The TEA152x family uses voltage mode control. The frequency is determined by the maximum transformer demagnetizing time and the time of the oscillator. In the first case, the converter operates in the Self-Oscillating Power Supply (SOPS) mode. In the latter case, it operates at a constant frequency, which can be adjusted with external

SMPS ICs for low-power systems

components R_{RC} and C_{RC} . This mode is called Pulse Width Modulation (PWM). Furthermore, a primary stroke is started only in a valley of the secondary ringing. This valley switching principle minimizes capacitive switch-on losses.

8.1 Start-up and Underoltage lockout

Initially, the IC is self supplying from the rectified mains voltage. The IC starts switching as soon as the voltage on pin V_{CC} passes the $V_{CC(startup)}$ level. The supply is taken over by the auxiliary winding of the transformer as soon as V_{CC} is high enough and the supply from the line is stopped for high efficiency operation.

When for some reason the auxiliary supply is not sufficient, the high-voltage supply also supplies the IC. As soon as the voltage on pin V_{CC} drops below the $V_{CC(stop)}$ level, the IC stops switching and restarts from the rectified mains voltage.

8.2 Oscillator

The frequency of the oscillator is set by the external resistor and capacitor on pin RC. The external capacitor is charged rapidly to the $V_{RC(max)}$ level and, starting from a new primary stroke, it discharges to the $V_{RC(min)}$ level. Because the discharge is exponential, the relative sensitivity of the duty factor to the regulation voltage at low duty factor is almost equal to the sensitivity at high duty factors. This results in a more constant gain over the duty factor range compared to PWM systems with a linear sawtooth oscillator. Stable operation at low duty factors is easily realized. For high efficiency, the frequency is reduced as soon as the duty factor drops below a certain value. This is accomplished by increasing the oscillator charge time.

To ensure that the capacitor can be charged within the charge time, the value of the oscillator capacitor should be limited to approximately 1 nF.

8.3 Duty factor control

The duty factor is controlled by the internal regulation voltage and the oscillator signal on pin RC. The internal regulation voltage is equal to the external regulation voltage (–2.5 V) multiplied by the gain of the error amplifier (typically 20 dB which is 10×).

8.4 Valley switching

A new cycle is started when the primary switch is switched on (see <u>Figure 4</u>). After a certain time (determined by the oscillator voltage RC and the internal regulation level), the switch is turned off and the secondary stroke starts. The internal regulation level is determined by the voltage on pin REG.

After the secondary stroke, the drain voltage shows an oscillation with a frequency of approximately:

$$\frac{1}{2 \times \pi \times \sqrt{(L_p \times C_p)}}\tag{1}$$

where:

 L_p = primary self-inductance

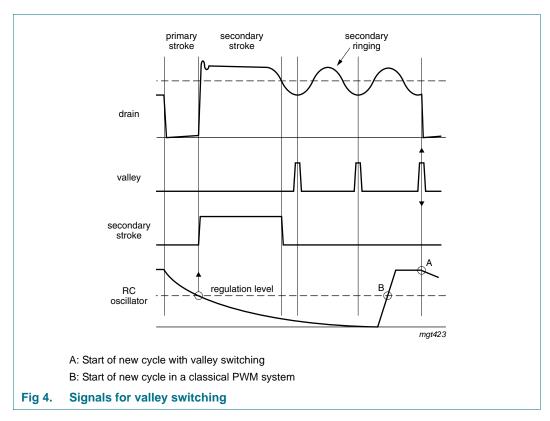
 C_p = parasitic capacitance on drain node

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As soon as the oscillator voltage is high again and the secondary stroke has ended, the circuit waits for a low drain voltage before starting a new primary stroke.

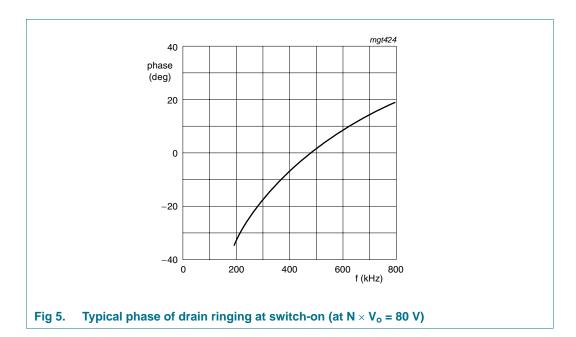
<u>Figure 4</u> shows the drain voltage together with the valley signal, the signal indicating the secondary stroke and the RC voltage.

The primary stroke starts some time before the actual valley at low ringing frequencies, and some time after the actual valley at high ringing frequencies.



<u>Figure 5</u> shows a typical curve for a reflected output voltage N \times V_o of 80 V. This voltage is the output voltage V_o (see <u>Figure 6</u>) transferred to the primary side of the transformer with the factor N (determined by the turns ratio of the transformer). <u>Figure 5</u> shows that the system switches exactly at minimum drain voltage for ringing frequencies of 480 kHz, thus reducing the switch-on losses to a minimum. At 200 kHz, the next primary stroke is started at 33° before the valley. The switch-on losses are still reduced significantly.

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8.5 Demagnetization

The system operates in discontinuous conduction mode all the time. As long as the secondary stroke has not ended, the oscillator will not start a new primary stroke. During the first t_{suppr} seconds, demagnetization recognition is suppressed. This suppression may be necessary in applications where the transformer has a large leakage inductance and at low output voltages.

8.6 Minimum and maximum duty factor

The minimum duty factor of the switched mode power supply is 0 %. The maximum duty factor is set to 75 % (typical value at 100 kHz oscillation frequency).

8.7 OverCurrent Protection (OCP)

The cycle-by-cycle peak drain current limit circuit uses the external source resistor R_l to measure the current. The circuit is activated after the leading edge blanking time t_{leb} . The protection circuit limits the source voltage to $V_{source(max)}$, and thus limits the primary peak current.

8.8 Short-circuit winding protection

The short-circuit winding protection circuit is also activated after the leading edge blanking time. If the source voltage exceeds the short-circuit winding protection voltage V_{swp} , the IC stops switching. Only a power-on reset will restart normal operation. The short-circuit winding protection also protects in case of a secondary diode short circuit.

8.9 OverTemperature Protection (OTP)

An accurate temperature protection is provided in the device. When the junction temperature exceeds the thermal shutdown temperature, the IC stops switching. During thermal protection, the IC current is lowered to the start-up current. The IC continues normal operation as soon as the overtemperature situation has disappeared.

SMPS ICs for low-power systems

8.10 OverVoltage Protection (OVP)

Overvoltage protection can be achieved in the application by pulling pin REG above its normal operation level. The current primary stroke is terminated immediately, and no new primary stroke is started until the voltage on pin REG drops to its normal operation level. Pin REG has an internal clamp. The current feed into this pin must be limited.

8.11 Output characteristics of complete power-plug

Typical characteristics:

- Output power: A wide range of output power levels can be handled by choosing the R_{DS(on)} and package of the TEA152x family. Power levels up to 30 W can be realized.
- Accuracy: The accuracy of the complete converter, functioning as a voltage source
 with primary sensing, is approximately 8 % (mainly dependent on the transformer
 coupling). The accuracy with secondary sensing is defined by the accuracy of the
 external components. For safety requirements in case of optocoupler feedback loss,
 the primary sensing remains active when an overvoltage circuit is connected.
- Efficiency: An efficiency of 75 % at maximum output power can be achieved for a complete converter designed for universal mains.
- Ripple: A minimum ripple is obtained in a system designed for a maximum duty factor
 of 50 % under normal operating conditions, and a minimized dead time. The
 magnitude of the ripple in the output voltage is determined by the frequency and duty
 factor of the converter, the output current level and the value and ESR of the output
 capacitor.

8.12 Input characteristics of complete power-plug

Typical characteristics:

The input voltage range comprises the universal AC mains (80 V to 276 V)

SMPS ICs for low-power systems

9. Limiting values

Table 4. Limiting values

In accordance with the Absolute Maximum Rating System (IEC 60134). All voltages are measured with respect to ground; positive currents flow into the device; pins $V_{\rm CC}$ and RC are not allowed to be current driven and pins REG and AUX are not allowed to be voltage driven.

Symbol	Parameter	Conditions	Min	Max	Unit
Voltages					
V _{CC}	supply voltage	continuous	-0.4	+40	V
V_{RC}	voltage on pin RC		-0.4	+3	V
V _{source}	voltage on pin SOURCE	DMOS power transistor	-0.4	+5	V
V_{drain}	voltage on pin DRAIN	T _j > 0 °C	-0.4	+650	V
Currents					
I _{REG}	current on pin REG		-	6	mΑ
I _{AUX}	current on pin AUX		-10	+5	mA
I _{source}	source current				
	TEA1520x		-0.25	+0.25	Α
	TEA1521x		-0.5	+0.5	Α
	TEA1522x		-1	+1	Α
	TEA1523P		-2	+2	Α
I _{drain}	drain current				
	TEA1520x		-0.25	+0.25	Α
	TEA1521x		-0.5	+0.5	Α
	TEA1522x		-1	+1	Α
	TEA1523P		-2	+2	Α
General					
P _{tot}	total power dissipation				
	DIP8 package	T _{amb} < 45 °C	-	1.0	W
	SO14 package	T _{amb} < 50 °C	-	1.0	W
T _{stg}	storage temperature		-55	+150	°C
Tj	junction temperature		-40	+145	°C
V _{esd}	electrostatic discharge voltage	human body model	<u>[1]</u> -	±2500	V
		machine model	[2] _	±200	V

^[1] Human body model: equivalent to discharging a 100 pF capacitor through a 1.5 kΩ series resistor. All pins are 2500 V maximum, except pin DRAIN, which is 1000 V maximum.

^[2] Machine model: equivalent to discharging a 200 pF capacitor through a 0.75 μ H coil and a 10 Ω series resistor.

SMPS ICs for low-power systems

10. Thermal characteristics

Table 5. Thermal characteristics

Symbol	Parameter	Conditions	Тур	Unit
$R_{th(j-a)}$	thermal resistance from junction to ambient	in free air	<u>[1]</u>	
	DIP8 package		100	K/W
	SO14 package		91	K/W

^[1] Thermal resistance $R_{th(j-a)}$ can be lower when the GND pins are connected to sufficient copper area on the printed-circuit board. See the *TEA152x* application note for details.

11. Characteristics

Table 6. Characteristics

 T_{amb} = 25 °C; no overtemperature; all voltages are measured with respect to ground; currents are positive when flowing into the IC; unless otherwise specified.

Symbol	Parameter	Conditions	Min	T_	May	Unit
Symbol	Parameter	Conditions	IVIII	Тур	Max	Unit
Supply						
I _{CC(oper)}	operating supply current	normal operation	-	1.3	1.9	mA
I _{CC(startup)}	start-up supply current	start-up	-	180	400	μΑ
I _{CC(ch)}	charge supply current	V _{drain} > 60 V	-6	-4	-3	mΑ
V _{CC(startup)}	start-up supply voltage		9	9.5	10	V
V _{CC(stop)}	stop supply voltage	undervoltage lockout	7.0	7.5	8.0	V
I _{drain}	current on pin DRAIN	V _{drain} > 60 V				
		no auxiliary supply	-	1.5	2	mΑ
		with auxiliary supply	-	30	125	μΑ
Pulse-width	modulator					
δ_{min}	minimum duty factor		-	0	-	%
δ_{max}	maximum duty cycle	f = 100 kHz	-	75	-	%
SOPS						
V _{det(demag)}	demagnetization detection voltage		50	100	150	mV
$t_{sup(xfmr_ring)}$	transformer ringing suppression time	start of 2 nd stroke	1.0	1.5	2.0	μS
RC oscillator	•					
V _{RC(min)}	minimum voltage on pin RC		60	75	90	mV
V _{RC(max)}	maximum voltage on pin RC		2.4	2.5	2.6	V
t _{ch(RC)}	charge time on pin RC		-	1	-	μS
f _{osc}	oscillator frequency		10	100	200	kHz
Duty factor r	egulator: pin REG					
V_{REG}	voltage on pin REG		2.4	2.5	2.6	V
G _v	voltage gain	error amplifier	-	20	-	dB
$V_{clamp(REG)}$	clamp voltage on pin REG	I _{REG} = 6 mA	-	-	7.5	V

SMPS ICs for low-power systems

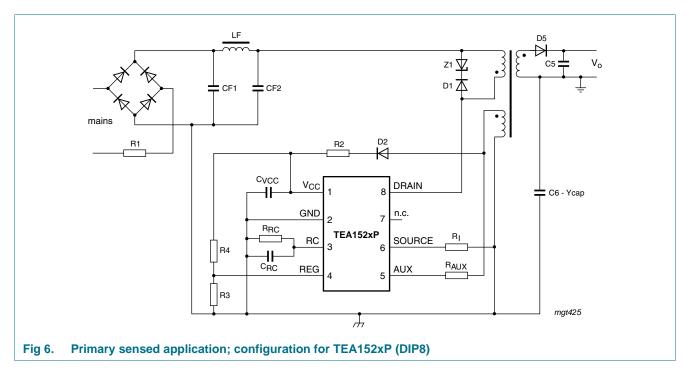
 Table 6.
 Characteristics ...continued

 T_{amb} = 25 °C; no overtemperature; all voltages are measured with respect to ground; currents are positive when flowing into the IC; unless otherwise specified.

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
Valley switc	hing					
$(\Delta V/\Delta t)_{Vrec}$	valley recognition voltage change with time		-102	-	+102	V/µs
f _{valley}	ringing frequency for valley switching	$N \times V_0 = 100 \text{ V}$	200	550	800	kHz
t _{d(vrec-swon)}	valley recognition to switch-on delay time		-	150	-	ns
Current and	short-circuit winding protec	tion				
$V_{\text{source(max)}}$	maximum voltage on pin SOURCE	$\Delta V/\Delta t = 0.1 \text{ V/}\mu\text{s}$	0.47	0.50	0.53	V
t _{d(prop)}	delay from detecting V _{source(max)} to switch-off	$\Delta V/\Delta t = 0.5 V/\mu s$	-	160	185	ns
V_{swp}	short-winding protection voltage	$\Delta V/\Delta t = 0.5 V/\mu s$	0.7	0.75	8.0	V
t _{leb}	leading edge blanking time		250	350	450	ns
FET output	stage					
I _{L(drain)}	leakage current on pin DRAIN	$V_{drain} = 650 \text{ V}$	-	-	125	μΑ
V _{BR(DRAIN)}	breakdown voltage on pin DRAIN	T _j > 0 °C	650	-	-	V
R_{DSon}	drain-source on-state resistance					
	TEA1520x	$I_{\text{source}} = -0.06 \text{ A}$				
		T _j = 25 °C	-	48	55.2	Ω
		T _j = 100 °C	-	68	78.2	Ω
	TEA1521x	$I_{\text{source}} = -0.125 \text{ A}$				
		T _j = 25 °C	-	24	27.6	Ω
		T _j = 100 °C	-	34	39.1	Ω
	TEA1522x	$I_{\text{source}} = -0.25 \text{ A}$				
		T _j = 25 °C	-	12	13.8	Ω
		T _j = 100 °C	-	17	19.6	Ω
	TEA1523P	$I_{\text{source}} = -0.50 \text{ A}$				
		T _j = 25 °C	-	6.5	7.5	Ω
		T _j = 100 °C	-	9.0	10.0	Ω
t _{f(DRAIN)}	fall time on pin DRAIN	V _i = 300 V; no external capacitor at drain	-	75	-	ns
Temperature	protection					
$T_{prot(max)}$	maximum threshold temperature		150	160	170	°C
T _{prot(hys)}	hysteresis of protection temperature		-	2	-	°C

SMPS ICs for low-power systems

12. Application information



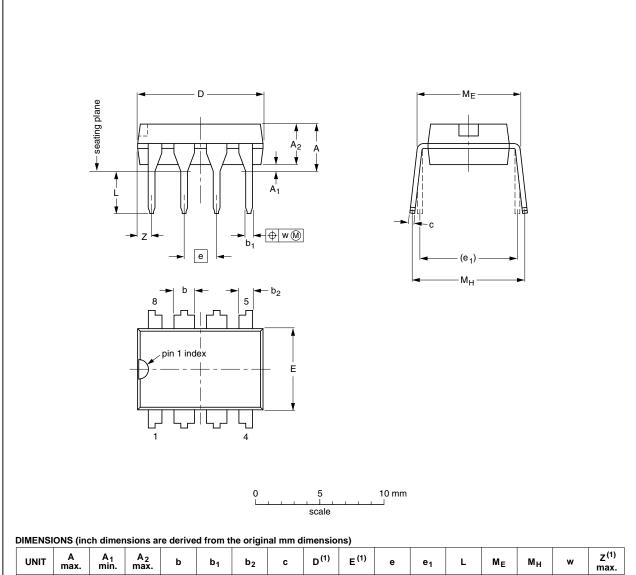
Further application information can be found in the *TEA152x* application note.

SMPS ICs for low-power systems

13. Package outline

DIP8: plastic dual in-line package; 8 leads (300 mil)

SOT97-1



UNIT	A max.	A ₁ min.	A ₂ max.	b	b ₁	b ₂	С	D ⁽¹⁾	E ⁽¹⁾	е	e ₁	L	ME	M _H	w	Z ⁽¹⁾ max.
mm	4.2	0.51	3.2	1.73 1.14	0.53 0.38	1.07 0.89	0.36 0.23	9.8 9.2	6.48 6.20	2.54	7.62	3.60 3.05	8.25 7.80	10.0 8.3	0.254	1.15
inches	0.17	0.02	0.13	0.068 0.045	0.021 0.015	0.042 0.035	0.014 0.009	0.39 0.36	0.26 0.24	0.1	0.3	0.14 0.12	0.32 0.31	0.39 0.33	0.01	0.045

Note

1. Plastic or metal protrusions of 0.25 mm (0.01 inch) maximum per side are not included.

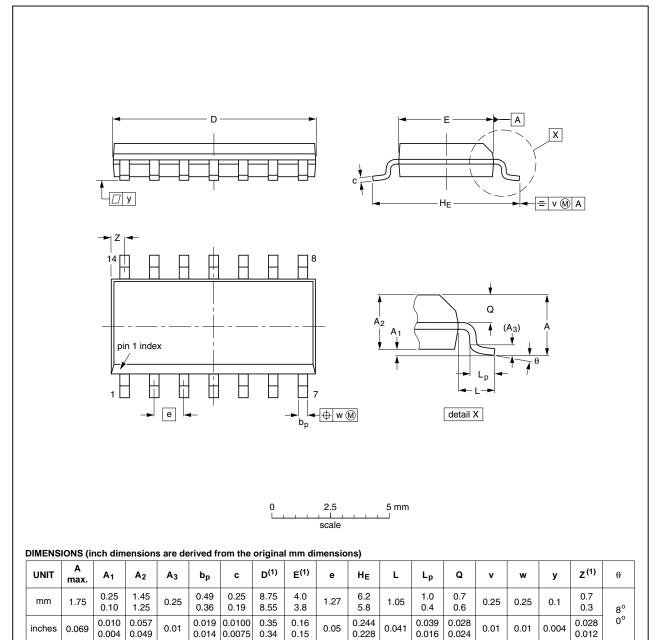
OUTLINE		REFER	REFERENCES		EUROPEAN	ISSUE DATE
VERSION	IEC	JEDEC	JEITA		PROJECTION	ISSUE DATE
SOT97-1	050G01	MO-001	SC-504-8			99-12-27 03-02-13

Fig 7. Package outline SOT97-1 (DIP8)

SMPS ICs for low-power systems

SO14: plastic small outline package; 14 leads; body width 3.9 mm

SOT108-1



Note

1. Plastic or metal protrusions of 0.15 mm (0.006 inch) maximum per side are not included.

OUTLINE		REFER	ENCES	EUROPEAN	ISSUE DATE	
VERSION	IEC	JEDEC	JEITA	PROJECTION		
SOT108-1	076E06	MS-012			99-12-27 03-02-19	

Fig 8. Package outline SOT108-1 (SO14)

SMPS ICs for low-power systems

14. Abbreviations

Table 7. Abbreviations

Acronym	Description
BiCMOS	Bipolar Complementary Metal-Oxide Semiconductor
DMOS	Diffusion Metal-Oxide Semiconductor
ESR	Equivalent Series Resistance
EZ-HV SOI	Easy High Voltage Silicon-On-Insulator
FET	Field-Effect Transistor
PWM	Pulse Width Modulation
SMPS	Switched Mode Power Supply
SOPS	Self-Oscillating Power Supply

15. Revision history

Table 8. Revision history

	•			
Document ID	Release date	Data sheet status	Change notice	Supersedes
TEA152X v4.0	20100914	Product data sheet	-	TEA152X_3
Modifications:	• <u>Table 1</u> "Qu	iick reference data" updated		
	 <u>Table 4</u> "Lin 	niting values" updated		
TEA152X_3	20090323	Product data sheet	-	TEA152X_FAM_2
Modifications:		of this data sheet has been of NXP Semiconductors.	redesigned to comply	with the new identity
	 Legal texts 	have been adapted to the no	ew company name wh	ere appropriate.
	 The minimum 	ım value of T _{amb} has change	ed in Table 1 and Table	. 4
	 The minimum 	ım value of T _j has changed i	n Table 4	
TEA152X_FAM_2	20060125	Product data sheet	-	TEA152X_FAMILY_1
TEA152X_FAMILY_1	20000908	Product specification	-	-

SMPS ICs for low-power systems

16. Legal information

16.1 Data sheet status

Document status[1][2]	Product status[3]	Definition
Objective [short] data sheet	Development	This document contains data from the objective specification for product development.
Preliminary [short] data sheet	Qualification	This document contains data from the preliminary specification.
Product [short] data sheet	Production	This document contains the product specification.

- [1] Please consult the most recently issued document before initiating or completing a design.
- [2] The term 'short data sheet' is explained in section "Definitions"
- [3] The product status of device(s) described in this document may have changed since this document was published and may differ in case of multiple devices. The latest product status information is available on the Internet at URL http://www.nxp.com.

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SMPS ICs for low-power systems

18. Tables

Table 1.	Quick reference data	2
Table 2.	Ordering information	3
Table 3.	Pin description	4
Table 4.	Limiting values	9
Table 5.	Thermal characteristics	.10
Table 6.	Characteristics	.10
Table 7.	Abbreviations	.15
Table 8	Revision history	15

continued >>

SMPS ICs for low-power systems

19. Figures

Fig 1.	Block diagram
Fig 2.	Pin configuration DIP84
Fig 3.	Pin configuration SO14
Fig 4.	Signals for valley switching 6
Fig 5.	Typical phase of drain ringing at switch-on
	(at N \times V ₀ = 80 V)
Fig 6.	Primary sensed application; configuration
	for TEA152xP (DIP8)12
Fig 7.	Package outline SOT97-1 (DIP8)
Fig 8.	Package outline SOT108-1 (SO14)14

continued >>

SMPS ICs for low-power systems

20. Contents

1	General description 1
2	Features and benefits 1
3	Applications 2
4	Quick reference data 2
5	Ordering information
6	Block diagram 3
7	Pinning information 4
7.1	Pinning
7.2	Pin description 4
8	Functional description 4
8.1	Start-up and Underoltage lockout 5
8.2	Oscillator 5
8.3	Duty factor control 5
8.4	Valley switching 5
8.5	Demagnetization
8.6	Minimum and maximum duty factor
8.7	OverCurrent Protection (OCP)
8.8 8.9	Short-circuit winding protection
8.10	OverVoltage Protection (OVP)
8.11	Output characteristics of complete power-plug. 8
8.12	Input characteristics of complete power-plug. 8
9	Limiting values
10	Thermal characteristics
11	Characteristics
12	Application information
13	Package outline
14	Abbreviations
15	Revision history
16	Legal information
16.1	Data sheet status
16.2 16.3	Definitions
16.4	Disclaimers 16 Trademarks 17
10. 4	Contact information
18	Tables
19	Figures
20	Contents

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