Estimation and Removal of Clock Skew from Network Delay Measurements

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Abstract-Packet delay and loss traces are frequently used by network engineers, as well as network applications, to analyze network performance. The clocks on the end-systems used to measure the delays, however, are not always synchronized, and this lack of synchronization reduces the accuracy of these measurements. Therefore, estimating and removing relative skews and offsets from delay measurements between sender and receiver clocks are critical to the accurate assessment and analysis of network performance. In this paper we introduce a linear programming-based algorithm to estimate the clock skew in network delay measurements and compare it with three other algorithms. We show that our algorithm has time complexity of O(N), leaves the delay after the skew removal positive, and is robust in the sense that the error margin of the skew estimate is independent of the magnitude of the skew. We use traces of real Internet delay measurements to assess the algorithm, and compare its performance to that of three other algorithms. Furthermore, we show through simulation that our algorithm is unbiased, and that the sample variance of the skew estimate is better (smaller) than existing algorithms.

Keywords--- clock skew, clock ratio, end-to-end delay, delay measurement.

I. Introduction

End-to-end delay and loss traces are frequently used in analyzing network performance. The accuracy of such measurements is important for several reasons. First, end-to-end measurements may be the only way of measuring network performance, especially when there is no provision inside the network to provide end-systems with information about the current status of the network. The current Internet has no mechanism for providing feedback on network congestion to end-systems at the IP layer, and neither does IPv6 [1]. Second, protocols and applications that behave adaptively at the end-system base their control on observed network performance, and it is critical that they obtain correct measurements.

Packet loss can be detected if a sender puts a sequence number on every packet it sends out, and the receiver sees a gap in the sequence numbers of packets arriving within a reasonable amount of time. For delay measurements, a sender needs to add timestamps to packets for a receiver to gather delay information [2]. Since the clocks at both end-systems are involved in measuring delay, the synchronization of the two clocks becomes an issue in the accuracy of delay measurement. The Network Time Protocol(NTP) [3] is widely used in the Internet for clock synchronization. It provides an accuracy of the order of milliseconds under reasonable circumstances. The accuracy, however, is not guaranteed, and not all hosts on the Internet support it.

Packet loss and delay are crucial in understanding the performance and reliability of the Internet. To provide unbiased and quantitative measures of performance, there has been consider-

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able effort to define one-way loss and delay metrics [4]. To obtain an accurate measurement of one-way delay, errors and uncertainties related to clocks need to be accounted for. When two clocks involved in the measurement run at different frequencies (that is, have a clock skew), inaccuracies are introduced in the measurement. In this paper we focus on filtering out the effects of clock skew specifically in one-way delay measurements.

The rest of the paper is organized as follows. In Section II we present a typical delay trace that motivated us to design a skew estimation algorithm. In Section III we define the terms needed to describe clock behavior, and introduce the notation used in the remainder of the paper. In Section IV we formalize the clock synchronization problem between two hosts, and show how the skew and offset affect the delay measurements. Then we list several desirable properties expected of a skew estimation algorithm. We introduce our skew estimation algorithm based on a linear programming technique in Section V, and three existing algorithms in Section VI. In Section VII we compare the four algorithms presented in this paper with respect to three desirable properties, as well as their performance in measurements and simulations. We conclude the paper in Section VIII.

II. MOTIVATION

Before we introduce the skew estimation algorithms, let us first examine a sample of one-way delay measurements. This sample, illustrated in Fig. 1, is taken from a trace described in Section VII-D (Trace 1 in Table II). The x-axis is the sender timestamp, and the y-axis is the delay calculated by subtracting the sender timestamp from the receiver timestamp of each packet. The measured delay lies in the range of 31.15 to 31.5 seconds. The measured delay is not the actual end-to-end delay, but includes the clock offset between the two clocks plus the end-to-end delay. Clock offset is the difference in time, and skew is the difference in clock speed. We defer the formal definitions of offset and skew to Section III.

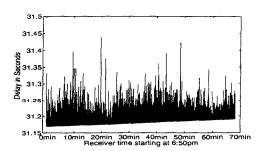


Fig. 1. Trace 1

In Fig. 1 the delay shows an increasing trend of about 100

milliseconds over the duration of 70 minutes at the receiver. It is significant enough to distort performance metrics such as the average and autocorrelation of end-to-end delay. While one might imagine that this is due to increasing congestion and queueing delay, it is unlikely as the minimum observed delay increases over time. Instead, the linear increase in delay attests to a constant speed difference between the sender and receiver clocks.

The end-to-end delay consists of transmission and propagation delays plus variable queueing delay. When all of the packets go through the same route to the receiver, they have the same propagation delay, and, if they have the same size, the transmission delay also is the same. Even if the packets go through the same route, and are of the same size, the packets experience different levels of queueing inside the network. This is what causes the variability in the end-to-end delay.

Previous work by Paxson [5], [6] addresses problems in delay measurements due to clock adjustments and rate mismatches. In his work, Paxson uses forward and reverse path measurements of delay between a pair of hosts to deal with clock synchronization problems, such as relative offset and skew. Many applications, however, see only one-way delay, and still have to deal with the clock synchronization problems in packet delay. Unfortunately one-way measurements alone are not enough to infer the clock offset, and we cannot distinguish the clock offset from the fixed portion of end-to-end delay. For example, in the figure shown above, it is difficult to tell how much of the 31.15 seconds is due to the time difference between clocks and the fixed transmission and propagation delay, without the availability of more information. Due to this lack of information in one-way delay, we focus on the variable portion in one-way delay measurements.

The variable queueing delay serves a very important role in network and application design. Continuous-media applications such as audio and video need to absorb the delay jitter perceived at the receiver for smooth playout of the original stream [7], [8], [9]. Determining the correct amount of buffering, and reconstructing the original timing is crucial to the performance of continuous-media applications. The variable queueing delay is also useful in monitoring the network performance at the edges of the network; the transmission and propagation delay is fixed per route, and does not convey any information about the dynamic changes inside the network when packets follow a fixed route.

III. BACKGROUND

A. Clock terminology

In this section we introduce the terminology we use to describe clock behavior. A *clock* is a piecewise continuous function that is twice differentiable except on a finite set of points:

$$C: \mathcal{R} \longrightarrow \mathcal{R}$$

where $C'(t) \equiv dC(t)/dt$ and $C''(t) \equiv d^2C(t)/dt^2$ exist everywhere except for $t \in P \subset \mathcal{R}$ where |P| is finite.

A "true" clock reports "true" time at any moment, and runs at a constant rate. Let C_t denote the "true" clock; it is the identity function given below,

$$C_t(t) = t$$
 and $P_t = \emptyset$

We use the following nomenclature from [3] and [10] to describe clock characteristics. Let C_a and C_b be two clocks:

• offset: the difference between the time reported by a clock and the "true" time; the offset of C_a is $(C_a(t) - t)$. The offset of the clock C_a relative to C_b at time $t \ge 0$ is $C_a(t) - C_b(t)$.

- frequency: the rate at which the clock progresses. The frequency at time t of C_a is $C'_a(t)$.
- skew: the difference in the frequencies of a clock and the "true" clock. The skew of C_a relative to C_b at time t is $(C'_a(t) C'_b(t))$.
- drift: The drift of clock C_a is $C_a''(t)$. The drift of C_a relative to C_b at time $t \ge 0$ is $(C_a''(t) C_b''(t))$.

Two clocks are said to be *synchronized* at a particular moment if both the relative offset and skew are zero. When it is clear that we refer to two clocks, neither of which is the true clock in our discussion, we simply refer to relative offset and relative skew as offset and skew, respectively.

It is sometimes convenient to compare the frequency ratio between two clocks instead of the skew. This is captured by the following definition.

• clock ratio: the frequency ratio between a clock and the "true" clock; the ratio of C_a is C'(a). The ratio of C_a relative to C_b at time t is $C'_a(t)/C'_b(t)$.

Let C_a and C_b have constant frequencies, and α and δ be the clock ratio and skew of C_b relative to C_a , respectively. $\alpha = C_b'/C_a'$ and $\delta = C_b' - C_a'$. Then the relation between the clock ratio and the skew is:

$$\delta = C_b' - C_a' = \alpha C_a' - C_a' = (\alpha - 1)C_a' \tag{1}$$

From now on, we assume that the sender and receiver clocks have constant frequencies, and their skew and clock ratio are constant over time; we use them interchangeably, and use (1) whenever necessary to convert from one to the other.

B. Time duration consistent with a clock

In the previous section, we have defined a *clock* and terms relevant to its behavior. In this section we look at how a *time duration* is measured according to a clock. Let $\Delta(t_1, t_2, C_a)$ denote the time that has passed according to C_a between t_1 and t_2 of the "true" clock. Since a clock is a piecewise continuous function, we define the time duration as:

$$\begin{split} \Delta(t_1,t_2,C_a) &\equiv \int_{t_1}^{t_2} C_a' dt \\ &= \int_{t_1}^{p_1} C_a' dt + \int_{p_1}^{p_2} C_a' dt + \dots + + \int_{p_n}^{t_2} C_a' dt \\ \text{where } P_a \bigcap (t_1,t_2) &= \{p_1,p_2,\dots,p_n\} \text{ and } \\ t_1 &< p_1 < p_2 < \dots < p_n < t_2, \ 1 \leq i \leq n. \end{split}$$

If $P_a \cap (t_1, t_2) = \phi$, then

$$\Delta(t_1, t_2, C_a) = \int_{t_1}^{t_2} C'_a dt = C_a(t_2) - C_a(t_1)$$
 (2)

When two clocks are not synchronized and, more specifically, have different frequencies, time duration measured with one clock will be different from the other. We say that a time duration measured with a clock is **consistent** with any other clock of the same frequency and any offset. If two clocks have a nonzero skew, time measured on one clock will not be consistent with the other clock.

We have modeled a clock as a piecewise continuous function in order to take into account the restrictions of real clocks. The resolution of a clock on a computer system is the smallest unit by which the clock's time is updated, and is greater than zero. At best, a clock in a computer is a step function with increments at every unit of its time resolution. We consider the time reports by a real clock with a fixed minimum resolution as samples of a continuous function at specific moments, and thus circumvent the discretization effect of the real clock. Another problem a real clock poses is the abrupt time adjustment possible through a time resetting system call. Some systems that do not run NTP [3] have a very coarse-grain (in the order of hours) synchronization mechanism in the cron table. The time adjustment in such a case can be several orders of magnitude larger than the usual increment of the clock resolution, and the time can even be set backward. The piecewise nature of a clock function accommodates the abrupt time adjustment.

When a delay measurement involves more than one clock, the synchronization between those clocks has a tremendous impact on the accuracy of the measurement. Let us consider a case of measuring a packet delay between two hosts. The sender adds a timestamp to a packet when it leaves the sender, and the receiver records the time the packet arrives at the receiver. When the two host clocks are perfectly synchronized, the difference between the two timestamps is the end-to-end network delay experienced by that packet. If the clocks on the two hosts have a non-zero offset, but no skew, the difference between two timestamps includes not only the end-to-end delay, but also the offset. Given only a one-way measurement, we cannot distinguish the offset from the measurement, unless we are given the network delay, which is what we intended to measure in the first place. If the clocks have a non-zero skew, not only is the end-to-end delay measurement off by an amount equal to the offset, but it also gradually increases or decreases over time depending on whether the sender clock runs slower or faster than the receiver

In the following sections we formalize the clock synchronization problem outlined above, and show how to remove the clock skew in measurements.

IV. BASICS OF A SKEW ESTIMATION ALGORITHM

In the previous section we have defined a clock, and what is meant for a delay to be consistent with a clock. In this section we discuss the estimation and removal of the effects of clock skew in delay measurements. We first derive how much the clock skew contributes to the measured end-to-end delay when the skew is non-zero and constant. This derivation provides a basis for the discussion of several desirable properties for skew estimation algorithms.

A. Delay measured between two clocks

From Section IV-A, if the clock ratio between the sender and receiver clocks is greater than or less than 1, network delays will appear to become longer or shorter over the course of a measurement period. The purpose of removing this effect of skew on the delay measurements is to transform the delay measurements so that they are consistent with a single clock. In our work, we have chosen to make the delay measurements consistent with the receiver clock. When there is no access at the receiver to the "true" clock, the only clock the receiver has access to is its own clock. It is thus natural to measure one-way delay according to the receiver clock. For the simplicity of derivation, we assume the receiver clock to be the true clock, i.e. $C_r'(t) = 1$ and $\alpha = C_s'(t)$. However, this assumption does not lead to loss of generality; the same derivation leads to the delay consistent with the receiver clock whether it is the "true" clock or not.

For different size packets, the clock skew may not be distinguishable from the delay trend, if any. For example, if the packet

size grows over time and the route from the sender to the receiver is fixed, then the transmission delay gradually increases, and it is hard to distinguish a skew from this delay trend. Thus we assume all the packets have the same size.

Let us now introduce the terminology for clocks, timestamps, and delays used in measurements.

- C_s : sender clock.
- C_r : receiver clock.
- N: number of packets that arrive at the receiver.
- l_i : timestamp of the *i*-th packet leaving the sender according to C_r , i = 1, 2, ..., N.
- t_i^s : timestamp of the *i*-th packet leaving the sender according to C_s , i = 1, 2, ..., N; $t_i^s = C_s(l_i)$.
- t_i^r : timestamp of the *i*-th packet arriving at the receiver according to C_r , i = 1, 2, ..., N.
- d_i : end-to-end delay measurement of the *i*-th packet, using timestamps t_i^s and t_i^r , i = 1, 2, ..., N;

$$d_i = t_i^r - t_i^s \tag{3}$$

Fig. 2 shows the timing between C_s and C_r when C_s runs at half the speed of C_r and all the packets experience the same network delay. The end-to-end delay of the i-th packet consistent with C_r is $t_i^r - l_i$. However, l_i is not known at the receiver, and d_i is estimated using t_i^s and t_i^r . As a result, in this case, the end-to-end delay is consistent with neither C_s nor C_r . To make it consistent with C_r , we need to determine the skew of C_r relative to C_s , and remove it from the measurement d_i .

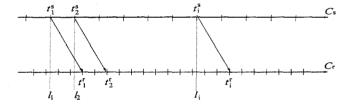


Fig. 2. Timing chart showing constant delay

B. Clock Skew in Delay Measurements

When there is a constant clock skew between two clocks, the clock offset between them gradually increases or decreases over time, depending on the sign of the skew. The amount of increase or decrease in the clock offset is proportional to the time duration of observation. We use the change in offset to estimate the clock skew. Thus it is more convenient to use timestamps relative to a specific point in time, such as the departure and arrival times of the first packet, than absolute timestamps. Below we introduce relative timestamps at the sender and the receiver.

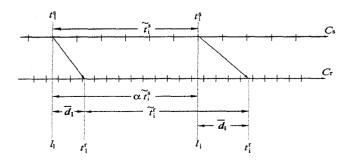


Fig. 3. Timing chart showing variable delay

 ⁱ/_s: time duration between the first and the i-th packets' departures at the sender consistent with C_s.

$$\tilde{t}_1^s = 0$$
 and $\tilde{t}_i^s = \Delta(l_1, l_i, C_s) \equiv t_i^s - t_1^s$.

• \tilde{t}_i^r : time duration between the first and the *i*-th packets' arrivals at the receiver consistent with C_r .

$$\tilde{t}_1^r = 0$$
 and $\tilde{t}_i^r = t_i^r - t_1^r$.

By (1) and (2),

$$\Delta(l_1, l_i, C_r) = l_i - l_1 = \alpha \Delta(l_1, l_i, C_s) = \alpha \tilde{t}_i^s. \tag{4}$$

Fig. 3 illustrates the relationship between $\Delta(l_1, l_i, C_r)$ and \tilde{t}_i^s on a timing chart. The quantities \bar{d}_1 and \bar{d}_i shown in the figure are defined below.

• \bar{d}_i : end-to-end delay consistent with C_r .

$$\bar{d}_{1} = \Delta(l_{1}, t_{1}^{r}, C_{r}) = t_{1}^{r} - l_{1},
\bar{d}_{i} = \Delta(l_{i}, t_{i}^{r}, C_{r}) = t_{i}^{r} - l_{i},
= t_{i}^{r} - l_{1} - \alpha \tilde{t}_{i}^{s} = (t_{i}^{r} - t_{1}^{r}) + (t_{1}^{r} - l_{1}) - \alpha \tilde{t}_{i}^{s},
= \tilde{t}_{i}^{r} + \bar{d}_{1} - \alpha \tilde{t}_{i}^{s}.$$
(5)

The quantity \bar{d}_i , however, is not obtainable directly from measured timestamps, due to the skew between the sender and receiver clocks. The quantity that is obtainable from actual timestamps is the following.

• \tilde{d}_i : delay calculated from \tilde{t}_i^s and \tilde{t}_i^r .

$$\tilde{d}_{i} = \tilde{t}_{i}^{r} - \tilde{t}_{i}^{s} = \tilde{t}_{i}^{r} + \bar{d}_{1} - \alpha \tilde{t}_{i}^{s} + (\alpha - 1)\tilde{t}_{i}^{s} - \bar{d}_{1}
= \bar{d}_{i} + (\alpha - 1)\tilde{t}_{i}^{s} - \bar{d}_{1}$$
(6)

The goal of estimating and removing the clock skew is to obtain \bar{d}_i from the actual delay measurement, \tilde{d}_i . From (3) and (6), we note that the difference between d_i and \tilde{d}_i is:

$$d_i - \tilde{d}_i = t_1^r - t_1^s.$$

Also note in (6) that \tilde{d}_i differs from \bar{d}_i by $(\alpha-1)\tilde{t}_i^s$ minus a constant \bar{d}_1 . If $\alpha>1$, $(\alpha-1)\tilde{t}_i^s$ grows linearly with \tilde{t}_i^s , and thus \tilde{d}_i gets larger. This is what contributed to an increasing trend we observed in Fig. 1. Finally, from (6), it is obvious that the measured network delays can be made consistent with C_r given α and \bar{d}_i according to:

$$\bar{d}_i = \tilde{d}_i - (\alpha - 1)\tilde{t}_i^s + \bar{d}_1. \tag{7}$$

Let $\hat{\alpha}$ and $\hat{\beta}$ be the estimates for α and \bar{d}_1 . Then the delay after the skew removal, \hat{d}_i , is:

$$\hat{d}_i = \tilde{d}_i - (\hat{\alpha} - 1)\tilde{t}_i^s + \hat{\beta} \tag{8}$$

C. Desirable Properties for Skew Estimation Algorithms

Before we delve into the details of the skew estimation algorithm, we first state the desirable properties that any such algorithm should exhibit. We use these properties later as a basis for comparing different estimation algorithms.

We begin by introducing the notations for an estimation algorithm and estimates parametrized by the estimation algorithm. Let \mathcal{A} be a skew estimation algorithm. We make the same assumption as in Section IV-A that the skew between the sender

and the receiver clocks is constant, and the receiver clock is the "true" clock. Given a set of end-to-end delays, $\mathcal{D} = \{\bar{d}_i\}_{i=1}^N$, which are predetermined and fixed, and also consistent with the "true" clock, we know that the delay measurements, \tilde{d}_i 's, are equal to \bar{d}_i if the clocks on the two hosts are synchronized $(\alpha=0)$; \tilde{d}_i differs from \bar{d}_i if the clock skew is not zero $(\alpha\geq 0)$. In that sense \tilde{d}_i depends on the clock skew, α , and \bar{d}_i , and is noted $\tilde{d}_i(\alpha,\mathcal{D})$.

We define $\hat{\alpha}_{\mathcal{A}}(\alpha, \mathcal{D})$ and $\hat{\beta}_{\mathcal{A}}(\alpha, \mathcal{D})$ to be the estimates of α and \bar{d}_1 , respectively, delivered by algorithm \mathcal{A} , when given $\bar{d}_i, 1 \leq i \leq N$ and α . Below is a list of desirable properties that should be exhibited by algorithm \mathcal{A} .

Property 1 The time and space complexity of algorithm \mathcal{A} should be linear in N. The computational complexity of an algorithm in terms of time and space is an important metric in assessing the performance and applicability of the algorithm. We will compare the *time complexity* of skew estimation algorithms as a function of the number of delay measurements.

Property 2 Since the purpose of the skew estimation is to remove the skew from delay measurements, it is desirable that the delays be non-negative after the skew is removed,

$$\hat{d}_i = \tilde{d}_i(\alpha, \mathcal{D}) - (\hat{\alpha}_{\mathcal{A}}(\alpha, \mathcal{D}) - 1)\tilde{t}_i^s + \hat{\beta}_{\mathcal{A}}(\alpha, \mathcal{D}) \ge 0$$

Property 3 The skew estimation algorithm should be robust in the sense that it is not affected by the magnitude of the actual skew. That is, the difference between the estimate and the actual skew should be independent of the actual skew. Under the same network condition, the skew estimate for different skews should exhibit the same margin of error from the actual skew, no matter how small or large the skew is. We state this property as follows:

$$\hat{\alpha}_{\mathcal{A}}(\alpha, \mathcal{D}) - \alpha = \hat{\alpha}_{\mathcal{A}}(1, \mathcal{D}) - 1, \forall \alpha > 0. \tag{9}$$

In the following section, we introduce a new algorithm based on linear programming to estimate α in delay measurements, and use the result to remove the skew from one-way delay measurements to make them consistent with the receiver clock. In this paper, we focus on a simple case where the clock skew is constant, and defer the discussion of time-varying skew to [11].

V. LINEAR PROGRAMMING ALGORITHM

Fig. 1 illustrates a trace where the skew between two clocks was nearly constant over the measurement duration. Looking at the figure, one is tempted to pick up a ruler, draw a line that skims through the bottom of the mass of the scatter-plot, measure the angle between the line and the x-axis, and calculate the skew using simple trigonometry. This approach is hard to automate, and invites human errors that are untraceable. A second thought would be to pick the first and last data points, and draw a line between them. The accuracy of this approach, however, can be easily thrown off, since delay has formidable variability that is in the order of magnitude bigger than the skew all through the measurement duration. Our approach is to fit a line that lies under all the data points, but as closely to them as possible.

We have formulated the above idea as a linear programming problem. The condition that the line should lie under all the data points forms the first part of our linear programming problem, and defines the feasible region for a solution; the objective function of the linear programming problem is to minimize the sum of the distances between the line and all the data points on the

A. Algorithm

Having presented our intuition behind the algorithm, we now formally introduce the algorithm. The goal of the skew estimation algorithm is to estimate the clock ratio α given \tilde{t}_i^s and \tilde{d}_i . The output of the skew estimation algorithm is: $\hat{\alpha}$ and $\hat{\beta}$, where $\hat{\alpha}$ is the estimate of α , and $\hat{\beta}$ is the estimate of \bar{d}_1 . We return to the interpretation of $\hat{\beta}$ at the end of this section. If we estimate both α and \bar{d}_1 correctly, then we can subtract $(\alpha-1)\tilde{t}_i^s-\bar{d}_1$ from \tilde{d}_i , and obtain \bar{d}_i , which is the end-to-end delay consistent with C_r and free of clock skew. Even when the estimates $\hat{\alpha}$ and $\hat{\beta}$ are not exactly the same as α and \bar{d}_1 , we still want the resulting end-to-end delay to be non-negative, after the skew is removed. When we formulate our skew estimation as a linear programming problem, this condition defines the feasible region where a solution should lie,

$$\tilde{d}_i - (\hat{\alpha} - 1)\tilde{t}_i^s + \hat{\beta} \ge 0, \quad 1 \le i \le N \tag{10}$$

There are infinitely many pairs of $\hat{\alpha}$ and $\hat{\beta}$ that satisfy the condition above, if the feasible region defined by (10) is not trivial. Our objective function to minimize the distance between the line and all the delay measurements is stated as

$$\min\{\sum_{i=1}^{N} \left(\tilde{d}_i - (\hat{\alpha} - 1)\tilde{t}_i^s + \hat{\beta} \right) \}$$
 (11)

and is used to determine the solution to $\hat{\alpha}$ and $\hat{\beta}$ from (10).

One important point to note in (10) is that the estimated end-to-end delay of \bar{d}_i , calculated as $(\tilde{d}_i - (\hat{\alpha} - 1)\tilde{t}_i^s + \hat{\beta})$, once $\hat{\alpha}$ and $\hat{\beta}$ are obtained, will be greater than zero instead of being greater than $\min_i \bar{d}_i$. Thus $\hat{\beta}$ is actually an estimate of $(\bar{d}_1 + \min_i \bar{d}_i)$. The resulting delay of $\tilde{d}_i - (\hat{\alpha} - 1)\tilde{t}_i^s + \hat{\beta}$ is not the end-to-end delay, but rather the variable portion of the end-to-end delay.

In the following sections, we look into other algorithms that can be used in skew estimation, and compare them with our linear programming algorithm in terms of the properties listed in Section IV, and their performance in actual and synthetic measurements.

VI. OTHER ALGORITHMS

A. Paxson's algorithm

In [5], [6], Paxson designed an algorithm for removing a clock skew from a set of forward and reverse path delay measurements. In this section we briefly describe how we use his algorithm when given delays in only one direction. Assume that the input to the algorithm is the same as in the previous linear programming algorithm. Readers are referred to [5], [6] for more detail. Paxson's algorithm is as follows:

- Step 1. Partition \tilde{d}_i 's into \sqrt{N} segments, and pick the minimum delay measurement from each segment. The selected measurements are called the "de-noised" one-way transit times (OTTs).
- Step 2. Pick the median of the slopes of all possible pairs of the "de-noised" OTTs. If the median slope is negative, assume that the OTTs have a decreasing trend (here we assume a decreasing trend is detected).
- Step 3. Select the cumulative minima test from the "denoised" OTTs (see [5]), and test if the number of cumulative minima is large enough to show that the decreasing trend found in Step 2 is probabilistically not likely, if there is no trend.

• Step 4. If it passes the cumulative minima test, pick the median from the slopes of all possible pairs of the cumulative minima: output it as the estimate of $\alpha - 1$. Otherwise, the algorithm concludes that there is no skew, and outputs $\hat{\alpha} = 0$.

The core of Paxson's algorithm is the robust line fitting technique based on robust statistics[12]. It uses the median as a robust estimate for the slope. As mentioned in [5], [6], robust line fitting alone fails in estimating the slope of the trend due to the high variability in OTTs, and that is why the "de-noised" OTTs and cumulative minima are used in his algorithm.

B. Linear regression algorithm

Linear regression is a standard technique for fitting a line to a set of data points. It is optimal in the mean square sense if the network delays are normally distributed, but is not robust in the presence of outliers. As pointed out in [5], [6], it is not a good choice for a skew estimation, even when applied to the "de-noised" OTTs above. Here we use it only as a reference algorithm that requires no knowledge of the underlying behavior of delay measurements.

C. Piecewise minimum algorithm

There is another simple algorithm to illustrate the difficulty in estimating the skew. It partitions the delay measurements into segments, picks a minimum from each segment, and connects them to obtain a concatenation of line segments. The minima are the same as the "de-noised" OTTs in Section VI-A. The resulting concatenation of line segments is the estimate of the skew, and is very unlikely to be a straight line.

When the skew is as obvious as in Fig. 1, the resulting concatenation of line segments is close to a straight line, and can be used as a rough estimate.

VII. COMPARISON OF THE FOUR ALGORITHMS

In this section we compare the four algorithms based on the desirable properties from Section IV. We also apply the algorithms to actual delay measurements, and compare their performance by looking at the adjusted delays. Last we use delay measurements from simulation to compare our approach to Paxson's algorithm.

A. Computational complexity

The time complexity of a two-variable linear programming problem is proven to be O(N) [13], [14]. We have implemented a simple and efficient O(N) algorithm that exploits the fact that \tilde{t}_i^s 's are sorted in our specific problem [11]. The other three algorithms also have complexity of O(N).

B. Non-negative delay after the skew removal

In order to guarantee that the delay remains positive after the skew is removed, a skew estimation algorithm must estimate \bar{d}_1 correctly. The linear programming algorithm, however, is the only one that estimates \bar{d}_1 (or $\bar{d}_1 + \min_i \bar{d}_i$), as explained in Section V. Paxson's original algorithm for skew estimation is for two-way measurements after the clock offset has been removed. The linear regression algorithm provides an estimate of $\hat{\beta}$. However, this is just the y-intercept of the regression line which bears no relevance to the correct estimation of \bar{d}_1 . The piecewise minimum algorithm outputs a concatenation of line segments, and the slopes of those line segments are skew estimates. The algorithm does not have any provision to guarantee that all the data points lie above the concatenation of line segments.

For the three algorithms that do not provide an estimate for d_1 that ensures that delays are non-negative after the skew removal, we choose a $\hat{\beta}$ that satisfies the following condition for all $\hat{\alpha}$'s in each algorithm:

$$\max_{1 \le i \le N} \{ \hat{\beta}_i : \tilde{d}_i - (\hat{\alpha}_i - 1)\tilde{t}_i^s + \hat{\beta}_i > 0 \}$$
 (12)

where $\hat{\alpha_i} = \hat{\alpha}$ and $\hat{\beta_i} = \hat{\beta}$ for $1 \leq i \leq N$ in Paxson's and linear regression algorithms; in the piecewise minimum algorithm $\hat{\alpha_i}$ and $\hat{\beta_i}$ are determined by the line segment to which $\tilde{d_i}$ and \tilde{t}_i^s belong to.

C. Robustness

We focus on the performance of an algorithm, as measured by the difference between the estimate and the actual skew, and whether the difference depends on the variability of the network delays alone, and not on the magnitude of the clock skew. This property guarantees that the estimation algorithm performs reliably, in the sense that the margin of error remains the same, no matter how large the skew is.

We first show that the linear programming algorithm satisfies this property, and follow it with a discussion about other algorithms.

C.1 Linear Programming Algorithm

Using the same assumptions and notations for the skew estimation algorithm and estimates as in Section IV-C, we consider two different clock skews that varies from one to some constant for a set of delays, $\mathcal{D} = \{\bar{d}_i\}_{i=1}^N$, where \bar{d}_i 's are fixed. From one set of measurements to the other, nothing changes except for the frequency of the sender clock relative to the receiver clock. The receiver observes that the delay measurements, \tilde{d}_i 's, are different from one skew to the other, but the end-to-end delays \bar{d}_i consistent with the receiver clock remain the same in both sets. We also note that \tilde{t}_i^s 's remain the same in both sets.

Consider the sender and receiver clocks are "true" clocks, and a set of packet delays, $\mathcal{D}=\{\bar{d}_i\}_{i=1}^N$, is consistent with the "true clock." Suppose that we measure those delays when the frequency of the sender clock changes so that the skew is $\alpha \neq 1$. From (6) we have

$$\tilde{d}_i(1,\mathcal{D}) = \bar{d}_i - \bar{d}_1 \tag{13}$$

$$\tilde{d}_i(\alpha, \mathcal{D}) = \bar{d}_i + (\alpha - 1)\tilde{t}_i^s - \bar{d}_1 \tag{14}$$

Let \mathcal{A} be the linear programming algorithm, and consider the problem of determining $\hat{\alpha}$ and $\hat{\beta}$ when both clocks are "true" clocks. By (10), (11), and (13), the problem becomes that of minimizing $\sum_{i=1}^{N} \{\bar{d}_i - \bar{d}_1 - (\hat{\alpha} - 1)\tilde{t}_i^s + \hat{\beta}\}$ such that $\hat{\beta} \geq (\hat{\alpha} - 1)\tilde{t}_i^s - \bar{d}_i + \bar{d}_1$, for $1 \leq i \leq N$.

Let $\hat{\alpha}_{\mathcal{A}}(1,\mathcal{D})$ and $\hat{\beta}_{\mathcal{A}}(1,\mathcal{D})$ be the values that solve this problem.

Now define $\alpha^* = (\alpha + \hat{\alpha}_{\mathcal{A}}(1,\mathcal{D}) - 1)$, and substitute $\alpha^* - \alpha$ with $\hat{\alpha}_{\mathcal{A}}(1,\mathcal{D}) - 1$ above, and the above problem is now equivalent to choosing α^* and $\hat{\beta}$ that minimize $\sum_{i=1}^N \{\bar{d}_i - \bar{d}_1 - (\alpha^* - \alpha)\tilde{t}_i^s - \hat{\beta}\}$ such that $\hat{\beta} \geq (\alpha^* - \alpha)\tilde{t}_i^s - \bar{d}_i + \bar{d}_1$, for $1 \leq i \leq N$. By (14), it is equivalent to choosing α^* and $\hat{\beta}$ to minimize $\sum_{i=1}^N \{\bar{d}_i(\alpha,\mathcal{D}) - (\alpha^* - 1)\tilde{t}_i^s + \hat{\beta}\}$ such that $\hat{\beta} \geq (\alpha^* - 1)\tilde{t}_i^s - \bar{d}_i(\alpha,\mathcal{D})$, for $1 \leq i \leq N$, which solves the case when the skew is $\alpha \neq 1$. Let $\hat{\alpha}_{\mathcal{A}}(\alpha,\mathcal{D})$ and $\hat{\beta}_{\mathcal{A}}(\alpha,\mathcal{D})$ be the values that solve the above problem. Then we can conclude: $\hat{\alpha}_{\mathcal{A}}(\alpha,\mathcal{D}) - \alpha = \hat{\alpha}_{\mathcal{A}}(1,\mathcal{D}) - 1$ and $\hat{\beta}_{\mathcal{A}}(\alpha,\mathcal{D}) = \hat{\beta}_{\mathcal{A}}(1,\mathcal{D})$.

TABLE I
SAMPLE VARIANCE OF SIMULATION RESULTS TO TEST PROPERTY 3

		$\mu = 10$	msec	$\mu = 100msec$		
	lpha -1	Linear prog.	Paxson	Linear prog.	Paxson	
	0.01	0.9666e-5	0.0373e-3	0.8405e-4	0.0002	
	0.1	0.9666e-5	0.1073e-3	0.8405e-4	0.0004	
- [4	0.9666e-5	0.3945e-3	0.8405e-4	0.002	

C.2 Other algorithms

It is clear that the linear regression algorithm satisfies Property 3. In the piecewise minimum algorithm, the increase in measured delay due to a clock skew is a function of the sender timestamp, but not of the end-to-end delay as stated in (6). As the skew gets larger, the increase due to the clock skew becomes the dominant part of a measured delay, and the minimum of a segment is more likely to be found near the beginning of the segment. Depending on the magnitude of the skew, a minimum-based algorithm uses different minima, and clearly the differences between the estimate and the actual skew are not the same.

Since Paxson's algorithm employs a robust line fitting technique after local minima are obtained, we choose to simulate Paxson's algorithm to examine its robustness. In the simulation, the number of packets is 600, and α changes from 0.01 to 0.1 and 4. The end-to-end delay consistent with C_r , \bar{d}_i , is assumed to have an exponential distribution with the means, $\mu = 10 msec$ and $\mu = 100 msec$. The purpose of the simulation is to show the variability of the difference between the estimate and the actual skew over a range of clock skews. Thus we use the same set of \bar{d}_i for all three values of α , and calculate the sample mean and the variance of the estimates. As shown in Section VII-C.1, the difference between the actual skew and the estimate does not change in the linear programming algorithm case, and thus the sample variance of the estimates from the algorithm remains the same for the three values of α in the simulation. The sample variance of Paxson's algorithm, however, increases as the skew increases. We list only the sample variances in Table I. This illustrates that the difference between the actual skew and the estimate of Paxson's algorithm grows as the skew grows, and thus the algorithm does not satisfy Property 3.

D. Measurement

In this section we determine how each algorithm performs when applied to actual measurements. We collected several traces of delay measurements on the Internet and MBone [15] between November 14, 1997, and December 21, 1997. Table II provides a brief description of these traces. The clocks of the end-hosts were not synchronized in all traces. We used constantlength UDP packets whose payloads consisted of a sequence number and a timestamp, and they were sent out at periodic intervals.

Trace 1 in Table II exhibits an increasing trend in delay measurements, which translates to a constant skew. All four algorithms estimate the skew well, and we show only the results of linear programming and Paxson's algorithms in Fig. 4 to 5. On the x-axis the sender timestamp, \tilde{t}_i^s is plotted, and on the y-axis, the delay before and after the skew estimation and removal, \tilde{d}_i and $\tilde{d}_i - (\hat{\alpha} - 1)\hat{t}_i^s$, are plotted.* The gray foreground is the delay before the skew is estimated and removed, and the black

^{*}Here we actually plot $\tilde{d}_i - \min_i \tilde{d}_i$ and $\tilde{d}_i - (\hat{\alpha} - 1)\hat{t}_i^s - \min_i \tilde{d}_i$ to plot the delays before and after the skew estimation and removal in the same range of values on the y-axis.

TABLE II TRACE DESCRIPTIONS

Trace	Date	Type	Source	Destination	Interval	Time	Duration
1	14Nov97	<u> </u>	im.cs.umass.edu	anhur.sics.se	80ms	12:50	4hr 10min
2	20Nov97	uni	im.cs.umass.edu	anhur.sics.se	160ms	16:03	5hr 54min
3	20Nov97		eraser.cs.umass.edu	spiff.sics.se	160ms	15:54	10hr

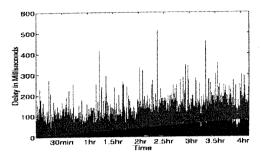


Fig. 4. Linear Programming Algorithm

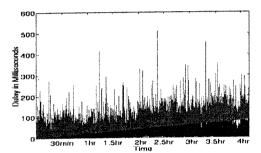


Fig. 5. Paxson's Algorithm

background is the delay after the skew is removed.

Fig. 6 to 9 are from Trace 3 in Table II. The same is on the x and y axes, as in Fig. 4 to 5. The linear skew trend is apparent as in Trace 1, but the delay behavior changes significantly in the later half of the trace, and more losses are detected. This is a multicast packet delay measurement, and the overall loss rate of the trace is very high: 42%. The losses are more pronounced as the jagged bottom of the gray plot in the second half of the trace. Considering the extraordinary high loss rate of the trace, we think that the clock skew was constant, but due to heavy congestion inside the network, queueing delays increased significantly over an extended period of time, and is shown in measurements.

In Fig. 8 the linear regression algorithm fails miserably in estimating the skew. The large delays between 6 and 8 hours on the x-axis produce outliers which have a significant impact on the linear regression. After the skew is filtered out, the delay has a decreasing trend, which is the opposite of the original increasing trend. Since the linear programming and Paxson's algorithms come up with one estimate for the constant skew, the resulting delays from both algorithms are close to the x-axis, while keeping the increased delay trend intact. The piecewise minimum algorithm calculates too high a minimum over the increased delay period, and ends up interpreting the increased delay trend as a skew. The result is that the effect of network congestion on delay is removed along with the skew.

Fig. 4 to 9 visually demonstrate the relative performance of each algorithm. The linear programming and Paxson's algorithms estimate the skew accurately in the presence of different levels of network congestion. In contrast, the linear regression and piecewise minimum algorithms perform poorly when the network is heavily congested, and the delay fluctuates signif-

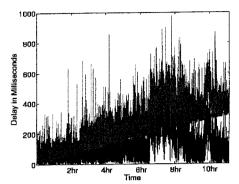


Fig. 6. Linear Programming Algorithm

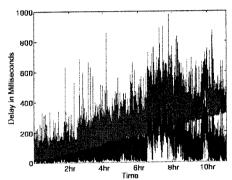


Fig. 7. Paxson's Algorithm

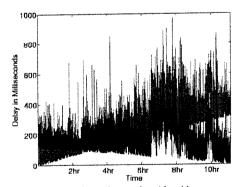


Fig. 8. Linear Regression Algorithm

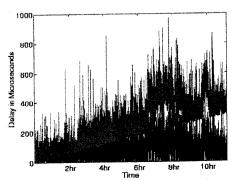


Fig. 9. Piecewise Minimum Algorithm

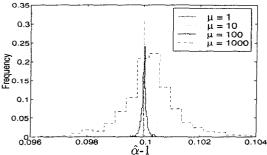
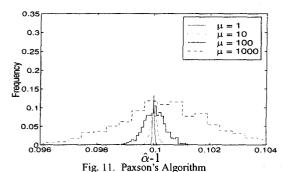


Fig. 10. Linear Programming Algorithm



icantly. In order to investigate the relative performance more precisely, we simulate the linear programming and Paxson's algorithms with synthetic delay, and compare the sample mean and variance of the estimate in the next section.

E. Simulation

The purpose of the simulation is to examine the average performance of a skew estimation algorithm in terms of the sample mean and variance of the estimate. We have simulated the linear programming and Paxson's algorithms on the same set of delay data varying the following parameters: the mean delay (μ) , the packet inter-departure time, the number of packets, and the skew. We assume an exponential distribution for end-to-end delays in our simulation and vary the mean (μ) of the exponential distribution from 1 to 10, 100, and 1000msec. The packet interdeparture time is either 20msec or lsec; the number of packets is either 600 or 3000; the clock ratio(α) varies from 1.0001 to 1.1 and 5. Only one set of results, however, are presented in this paper, and readers are referred to [11] for further results and

Fig. 10 and 11 are from 1000 runs when the packet interdeparture time is 20msec, the number of packets is 600, and the skew is 0.1. The mean delay (μ) varies from 1msec to 1000msec. We plot $\hat{\alpha} - 1$ on the x-axis, and the frequency on the y-axis in Fig. 10 and 11. The histograms have a fixed bin size of 30; the greater distance between the minimum and maximum of the estimates is, the wider the bin is. Most histograms are symmetric centered at the mean delay, and their estimates are very close to the true α values with a sample variance less than $\pm 4\%$ of α . The histograms of the linear programming algorithm are consistently less spread out than those of Paxson's algorithm for the given range of parameter values. We have shown through simulation that the estimate of our linear programming algorithm is unbiased against a small skew, and is likely to have less variance than that of Paxson's.

VIII. CONCLUSION

In this paper, we have presented a framework for understanding the systematic errors introduced in one-way network delay

measurements by unsynchronized clocks, and discuss several properties desirable of a skew estimation algorithm. We have developed an linear-programming-based algorithm, and compared it with three other existing algorithms. The linear regression and piecewise minimum algorithms demonstrated a poor performance over traces of Internet delay measurements. We generated synthetic delay measurements, and analyzed the sample mean and variance of the estimates of our and Paxson's algorithms. The results show that the estimate of the linear programming algorithm is likely to be unbiased and have less variance. In conclusion, the linear programming algorithm addresses all the desirable properties, and is simple, fast, and robust.

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