#### LTI State Variable Solutions

September 17, 2002 September 18, 2003 Sept 9, Sept 11, 2015/2016

#### The Problem

State variable model

$$\dot{\mathbf{x}} = \mathbf{A}\mathbf{x} + \mathbf{B}\mathbf{u} \tag{LTI}$$

Initial Condition

$$\mathbf{x}(t_0) = \mathbf{x}_0$$

• Desire  $\mathbf{x}(t)$  for  $t \ge t_0$ 

#### Lesson Goals

#### State variable model solutions by:

- 1. Laplace Transform
- 2. Series Approximation
- 3. Cayley-Hamilton Theorem
- 4. Similarity Transformation Reference: Brogan, Chap. 8 and 9

## Laplace Transform Approach

Apply Laplace Transform

$$sX(s)-x_0 = AX(s)+BU(s)$$

Some matrix algebra

$$[sI - A]X(s) = x_0 + BU(s)$$

$$\mathbf{X}(s) = \left[s\mathbf{I} - \mathbf{A}\right]^{-1} \mathbf{x}_0 + \left[s\mathbf{I} - \mathbf{A}\right]^{-1} \mathbf{B}\mathbf{U}(s)$$

$$\mathbf{X}(s) = \left[ s\mathbf{I} - \mathbf{A} \right]^{-1} \mathbf{x}_0 + \left[ s\mathbf{I} - \mathbf{A} \right]^{-1} \mathbf{B} \mathbf{U}(s)$$

"Natural"
"zero-input"
response

$$\mathbf{X}(s) = \left[ s\mathbf{I} - \mathbf{A} \right]^{-1} \mathbf{x}_0 + \left[ s\mathbf{I} - \mathbf{A} \right]^{-1} \mathbf{B}\mathbf{U}(s)$$

"Natural"
"zero-input"
response

"Forced"
"zero-state"
response

$$\mathbf{X}(s) = \left[ s\mathbf{I} - \mathbf{A} \right]^{-1} \mathbf{x}_0 + \left[ s\mathbf{I} - \mathbf{A} \right]^{-1} \mathbf{B} \mathbf{U}(s)$$

# Call this matrix term $\Phi(s)$

#### The State Transition Matrix

$$\Phi(s) = [sI - A]^{-1}$$
Inverse Laplace
Transform

#### The State Transition Matrix

$$\Phi(s) = \begin{bmatrix} s\mathbf{I} - \mathbf{A} \end{bmatrix}^{-1}$$
Inverse Laplace
Transform
$$\Phi(t) = L^{-1} \left\{ \begin{bmatrix} s\mathbf{I} - \mathbf{A} \end{bmatrix}^{-1} \right\}$$

#### Solution in Time-Domain

$$\mathbf{x}(t) = \Phi(t - t_0)\mathbf{x}_0 + \int_{t_0}^{t} \Phi(t - \tau)\mathbf{B}\mathbf{u}(\tau)d\tau$$
"Natural" response "Forced" response

$$\mathbf{A} = \begin{bmatrix} 0 & 1 \\ 8 & -2 \end{bmatrix}$$

$$\mathbf{A} = \begin{bmatrix} 0 & 1 \\ 8 & -2 \end{bmatrix} \begin{bmatrix} s\mathbf{I} - \mathbf{A} \end{bmatrix} = \begin{bmatrix} s & -1 \\ -8 & s+2 \end{bmatrix}$$

$$\mathbf{A} = \begin{bmatrix} 0 & 1 \\ 8 & -2 \end{bmatrix} \begin{bmatrix} s\mathbf{I} - \mathbf{A} \end{bmatrix} = \begin{bmatrix} s & -1 \\ -8 & s+2 \end{bmatrix}$$

$$\left[ s \mathbf{I} - \mathbf{A} \right]^{-1} = \begin{bmatrix} \frac{s+2}{\left(s+4\right)\left(s-2\right)} & \frac{1}{\left(s+4\right)\left(s-2\right)} \\ \frac{8}{\left(s+4\right)\left(s-2\right)} & \frac{s}{\left(s+4\right)\left(s-2\right)} \end{bmatrix}$$

$$[sI-A]^{-1} = \begin{bmatrix} \frac{s+2}{(s+4)(s-2)} & \frac{1}{(s+4)(s-2)} \\ \frac{8}{(s+4)(s-2)} & \frac{s}{(s+4)(s-2)} \end{bmatrix}$$

$$\begin{bmatrix} s\mathbf{I} - \mathbf{A} \end{bmatrix}^{-1} = \begin{bmatrix} \frac{s+2}{\left(s+4\right)\left(s-2\right)} & \frac{1}{\left(s+4\right)\left(s-2\right)} \\ \frac{8}{\left(s+4\right)\left(s-2\right)} & \frac{s}{\left(s+4\right)\left(s-2\right)} \end{bmatrix}$$

$$\Phi(t) = \begin{bmatrix} \frac{1}{3}e^{-4t} + \frac{2}{3}e^{2t} & -\frac{1}{6}e^{-4t} + \frac{1}{6}e^{2t} \\ -\frac{4}{3}e^{-4t} + \frac{4}{3}e^{2t} & \frac{2}{3}e^{-4t} + \frac{1}{3}e^{2t} \\ \frac{1}{3}e^{-4t} + \frac{4}{3}e^{2t} & \frac{2}{3}e^{-4t} + \frac{1}{3}e^{2t} \end{bmatrix}$$

#### Comment

#### Laplace Transform approach:

- conceptually easy to grasp
- not easy for high-order systems

#### The Transfer Function Matrix

Substituting X(s) in the output

$$\mathbf{Y}(s) = \mathbf{CX}(s) + \mathbf{DU}(s)$$

$$= \mathbf{C}[s\mathbf{I} - \mathbf{A}]^{-1}\mathbf{BU}(s) + \mathbf{DU}(s)$$

$$= \{\mathbf{C}[s\mathbf{I} - \mathbf{A}]^{-1}\mathbf{B} + \mathbf{D}\}\mathbf{U}(s)$$

#### The Transfer Function Matrix

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#### The Impulse Response Matrix

$$L^{-1}\left\{\mathbf{C}\left[s\mathbf{I}-\mathbf{A}\right]^{-1}\mathbf{B}+\mathbf{D}\right\}$$

### Series Approximation Method

#### Consider the *matrix exponential* of **A**:

$$e^{\mathbf{A}t} = \sum_{k=0}^{\infty} \frac{t^k}{k!} \mathbf{A}^k$$
$$= \mathbf{I} + \mathbf{A}t + \mathbf{A}^2 \frac{t^2}{2} + \mathbf{A}^3 \frac{t^3}{3!} + \cdots$$

#### Claim ...

The solution of

$$\dot{\mathbf{x}} = \mathbf{A}\mathbf{x}$$
 ,  $\mathbf{x}(0) = \mathbf{x}_0$ 

is given by

$$\mathbf{x}(t) = e^{\mathbf{A}t}\mathbf{x}_0$$

and e<sup>At</sup> is the state transition matrix!

#### **Check initial condition**

Verify solution at t = 0:

$$\mathbf{x}(t=0) = e^{\mathbf{A}0}\mathbf{x}_{0}$$
$$= \mathbf{I}\mathbf{x}_{0}$$
$$= \mathbf{x}_{0}$$

## Check state equation

Expand dx/dt term-by-term

$$\dot{\mathbf{x}} = \left(\mathbf{A} + \mathbf{A}^2 t + \mathbf{A}^3 \frac{t^2}{2!} + \cdots \right) \mathbf{x}_0$$

$$= \mathbf{A} \left( \mathbf{I} + \mathbf{A}t + \mathbf{A}^2 \frac{t^2}{2!} + \cdots \right) \mathbf{x}_0$$

$$=Ax$$

#### Comment

#### Infinite series approach:

- also conceptually easy to grasp
- practically difficult to implement

#### Cayley-Hamilton Theorem

Yields a method to replace the infinite series solution by a FINITE sum!

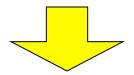
### Cayley-Hamilton Theorem

# Consider the *characteristic equation* of the matrix **A**:

$$\det(s\mathbf{I}-\mathbf{A})=0$$
 
$$\mathbf{S}^n+\alpha_{n-1}\mathbf{S}^{n-1}+\ldots+\alpha_1\mathbf{S}+\alpha_0=0$$

#### Cayley-Hamilton Theorem

# The matrix **A** satisfies its own characteristic equation:



$$\mathbf{A}^{n} + \alpha_{n-1}\mathbf{A}^{n-1} + \dots + \alpha_{1}\mathbf{A} + \alpha_{0}\mathbf{I} = \mathbf{0}$$

## Consequences of C-H

- A<sup>n</sup> can be expressed as a linear combination of lower order powers of A
- 2. e<sup>At</sup> can be written as a *finite* sum!

$$e^{\mathbf{A}t} = \sum_{k=0}^{n-1} \beta_k(t) \mathbf{A}^k$$

- Same A matrix as before (n = 2)
- By C-H Theorem

$$e^{\mathbf{A}t} = \beta_0(t)\mathbf{I} + \beta_1(t)\mathbf{A}$$

Q: How to find functions  $\beta_k$ ?

# Finding functions $\beta(t)$

#### Apply C-H to eigenvalues of A

$$e^{-4t} = \beta_0(t) + \beta_1(t)(-4)$$
$$e^{2t} = \beta_0(t) + \beta_1(t)(2)$$

# Finding functions $\beta(t)$

#### Apply C-H to eigenvalues of A

$$e^{-4t} = \beta_0(t) + \beta_1(t)(-4)$$

$$e^{2t} = \beta_0(t) + \beta_1(t)(2)$$



$$\beta_0(t) = \frac{2}{3}e^{2t} + \frac{1}{3}e^{-4t}$$

$$\beta_1(t) = \frac{1}{6}e^{2t} - \frac{1}{6}e^{-4t}$$

#### <u>Homework</u>

#### Confirm

$$e^{\mathbf{A}t} = \left(\frac{2}{3}e^{2t} + \frac{1}{3}e^{-4t}\right)\mathbf{I} + \left(\frac{1}{6}e^{2t} - \frac{1}{6}e^{-4t}\right)\mathbf{A}$$

= same as first example??

### The Similarity Transformation

#### Consider

$$\mathbf{A} = \begin{bmatrix} 0 & 1 \\ 8 & -2 \end{bmatrix}$$

## The Similarity Transformation

#### Consider

$$\mathbf{A} = \begin{bmatrix} 0 & 1 \\ 8 & -2 \end{bmatrix}$$

similar to

$$\mathbf{S} = \begin{bmatrix} -4 & 0 \\ 0 & 2 \end{bmatrix}$$

#### Review & Homework

# Use Matlab to find the modal matrix M that satisfies

MS=AM

or

A=MSM-1

#### State Transition Matrix for S

- Rows are decoupled
- Each row is easy to solve

$$e^{\mathsf{S}t} = \left[ \begin{array}{cc} e^{-4t} & 0 \\ 0 & e^{2t} \end{array} \right]$$

#### Solution from Modal Solution

#### State model

$$\dot{\mathbf{x}} = \mathbf{A}\mathbf{x}$$
 ,  $\mathbf{x}(0) = \mathbf{x}_0$ 

similar to

$$\dot{\mathbf{z}} = \mathbf{S}\mathbf{z}$$
 ,  $\mathbf{z}(0) = \mathbf{z}_0$ 

by similarity transformation **x=Mz** 



#### Solution from Modal Solution

#### State solution

$$\mathbf{x}(t) = e^{\mathbf{A}t}\mathbf{x}_0$$

is similar to

$$\mathbf{z}(t) = e^{\mathbf{S}t}\mathbf{z}_0$$

by similarity transformation **x=Mz** 



#### Solution from Modal Solution

#### Therefore, the state solution is

$$\mathbf{x}(t) = \mathbf{Mz}(t)$$

$$= \mathbf{M}e^{\mathbf{S}t}\mathbf{z}_{0}$$

$$= \mathbf{M}e^{\mathbf{S}t}\mathbf{M}^{-1}\mathbf{x}_{0}$$

$$e^{\mathbf{A}t}$$

## Summary of Modal Approach

- Find eigenvalues and eigenvectors
- Write state transition matrix e<sup>St</sup>
- Then

$$\mathbf{x}(t) = \mathbf{M}e^{\mathbf{S}t}\mathbf{M}^{-1}\mathbf{x}_0$$

where **M** is the modal matrix

#### **Homework**

# Confirm e<sup>At</sup> by transformation from the modal form ...

# Thought for the day

- "No act of kindness, no matter how small, is ever wasted."
  - Aesop, "The Lion and the Mouse"Greek fabulist, 550 BC

# Thought for the day

- "No act of kindness, no matter how small, is ever wasted."
  - Aesop, "The Lion and the Mouse" Greek fabulist, 550 BC
- "And to godliness, (add) brotherly kindness..."

-2 Peter 1:7