



Nonlinear Systems & Control
Part II
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1. Given

$$\dot{x} = \begin{pmatrix} -1 & 1 & 1 \\ 0 & 0 & -1 \\ 1 & -1 & 0 \end{pmatrix} x + \begin{pmatrix} 1 & 1 \\ 0 & 1 \\ 0 & 0 \end{pmatrix} u$$

$$y = \begin{pmatrix} 1 & 0 & -1 \\ 1 & 0 & 0 \end{pmatrix} x$$

state whether it is possible to achieve input/output noninteraction under feedback.

2. Given the system

$$\begin{aligned} \dot{x}_1 &= x_3 \\ \dot{x}_2 &= x_3 + x_1^2 + \cos(x_2)d \\ \dot{x}_3 &= -x_3 + x_1^2 + (1+x_1^2)u \\ y &= x_1 \end{aligned}$$

where d denotes a measurable and bounded disturbance $d \in \mathbb{R}$.

(a) Assuming at first $d \equiv 0$

- i. Compute the feedback that solves the input-output linearization problem;
- ii. Compute the zero dynamics;
- iii. Discuss on the stabilizability of the closed-loop system.

(b) When $d \neq 0$

- i. Compute, if any, the feedback solving the output-disturbance decoupling problem;
- ii. Qualitatively discuss on the boundedness of the internal dynamics of the closed-loop system.

3. Given a system $\dot{x} = f(x) + g(x)u$ discuss and provide condition for deducing the maximal input-to-state feedback linearizable component.

4. Compute, if any, a stabilizing high gain feedback for the origin of the dynamics

$$\begin{aligned} \dot{x}_1 &= -x_1 - \cos(x_1)x_2 \\ \dot{x}_2 &= x_1 - e^{-x_2}u \\ y &= x_2 \end{aligned}$$

5. Given a input-output feedback linearizable system in input-affine form, discuss about the design of an asymptotic tracking feedback over a bounded reference output $y_r(t)$.

① State whether it is possible to achieve I/O noninteraction under feedback

$$\dot{\tilde{x}} = \begin{pmatrix} -1 & 1 & 1 \\ 0 & 0 & -1 \\ 1 & -1 & 0 \end{pmatrix} \tilde{x} + \begin{pmatrix} 1 & 1 \\ 0 & 1 \\ 0 & 0 \end{pmatrix} u \quad p=q=2$$

$$y = \begin{pmatrix} 1 & 0 & -1 \\ 1 & 0 & 0 \end{pmatrix} \tilde{x}$$

Rel. deg. :

$$r_1 = C_1 B_1 = \begin{pmatrix} 1 & 0 & -1 \end{pmatrix} \begin{pmatrix} 1 \\ 0 \\ 0 \end{pmatrix} = 1 \neq 0 \quad r_1 = 1$$

$$r_2 = C_2 B_2 = \begin{pmatrix} 1 & 0 & 0 \end{pmatrix} \begin{pmatrix} 1 \\ 0 \\ 0 \end{pmatrix} = 1 \neq 0 \quad r_2 = 1$$

$$\dots \quad \begin{matrix} 1 & 0 & 0 & \backslash & 1 & 1 & 1 \end{matrix}$$

- - - - - (b) - - -

$$U = \begin{pmatrix} C_1 B_1 & C_1 B_2 \\ C_2 B_1 & C_2 B_2 \end{pmatrix} = \begin{pmatrix} 1 & 1 \\ 1 & 1 \end{pmatrix}$$

$\det U = 1 - 1 = 0 \rightarrow U$ not invertible

$$U = F_x + Gv$$

$$G = U^{-1}$$

$$F = -U^{-1} \begin{pmatrix} C_1 A \\ C_2 A \end{pmatrix}$$

$$U = U^{-1} \left(v - \begin{pmatrix} C_1 A^{r_1} \\ C_2 A^{r_2} \end{pmatrix} x \right) = U^{-1} \left(v - \begin{pmatrix} C_1 A \\ C_2 A \end{pmatrix} x \right)$$

Since U is not invertible I try to put it in

$$\begin{pmatrix} * & 0 \\ * & 0 \end{pmatrix} \text{ form}$$

$$U' = U \cdot M = \begin{pmatrix} 1 & 1 \\ 1 & 1 \end{pmatrix} \begin{pmatrix} a & b \\ c & d \end{pmatrix} = \begin{pmatrix} * & 0 \\ * & 0 \end{pmatrix} = \begin{pmatrix} 1 & 1 \\ 1 & 1 \end{pmatrix} \begin{pmatrix} 1 & -1 \\ 0 & 1 \end{pmatrix} = \begin{pmatrix} 1 & 0 \\ 1 & 0 \end{pmatrix} \text{ at}$$

$$a+c = *$$

$$b+d = 0$$

$$a+c = * \rightarrow a = * - c$$

$$b+d = 0 \rightarrow b = -d$$

$$* = 1$$

$$U = Mv$$

$$B_U = BMv$$

$$BM = \begin{pmatrix} 1 & 1 \\ 0 & 1 \\ 0 & 0 \end{pmatrix} \begin{pmatrix} 1 & -1 \\ 0 & 1 \end{pmatrix} = \begin{pmatrix} 1 & 0 \\ 0 & 1 \\ 0 & 0 \end{pmatrix}$$

$$\dot{x} = Ax + BMv = \underbrace{Ax}_{\begin{pmatrix} 1 & 0 \\ 0 & 1 \\ 0 & 0 \end{pmatrix}} + \underbrace{\begin{pmatrix} 1 & 0 \\ 0 & 1 \\ 0 & 0 \end{pmatrix} \begin{pmatrix} v_1 \\ v_2 \end{pmatrix}}_{\begin{pmatrix} v_1 \\ v_2 \end{pmatrix}} \quad v_1 = \omega_1, v_2 = \omega_2$$

$$\dot{z} = \begin{pmatrix} \dot{x} \\ \dot{v}_1 \\ \dot{v}_2 \end{pmatrix} = \underbrace{\begin{pmatrix} A & 0 & 0 \\ 0 & 0 & 0 \\ 0 & 0 & 0 \end{pmatrix}}_{A} z + \underbrace{\begin{pmatrix} 0 & 0 & 0 \\ 0 & 0 & 1 \\ 0 & 1 & 0 \end{pmatrix}}_{B_1} w_1 + \underbrace{\begin{pmatrix} 0 & 0 & 0 \\ 0 & 1 & 0 \\ 0 & 0 & 0 \end{pmatrix}}_{B_2} w_2$$

$$y = \begin{pmatrix} 1 & 0 & -1 & 0 \\ 1 & 0 & 0 & 0 \end{pmatrix} z$$

$$\bar{A} = \begin{pmatrix} -1 & 1 & 1 & 1 \\ 0 & 0 & -1 & 0 \\ 1 & -1 & 0 & 0 \\ 0 & 0 & 0 & 0 \end{pmatrix}$$

rel. dep.

$$\textcircled{1} \Rightarrow C_1 B_1 = (1 \ 0 \ -1 \ 0) \begin{pmatrix} 0 \\ 0 \end{pmatrix} = 0$$

$$r_1 \Rightarrow c_1 \bar{A} b_1 = (1 \ 0 \ -1 \ 0) \begin{pmatrix} 0 \\ 0 \\ 0 \\ 1 \end{pmatrix} = 0$$

$$c_2 \bar{A} b_2 = (1 \ 0 \ 0 \ 0) \begin{pmatrix} 0 \\ 1 \\ 0 \\ 0 \end{pmatrix} = 0$$

$$c_1 \bar{A} b_1 = (-2 \ 2 \ 1 \ 1) \begin{pmatrix} 0 \\ 0 \\ 1 \\ 1 \end{pmatrix} = 1 \neq 0 \quad r_1 = 2$$

$$r_2 \Rightarrow c_2 \bar{A} b_1 = (1 \ 0 \ 0 \ 0) \begin{pmatrix} 1 \\ 0 \\ 0 \\ 0 \end{pmatrix} = 1 \neq 0 \quad r_2 = 2$$

I compute also $c_1 \bar{A} b_2 = (-2 \ 2 \ 1 \ 1) \begin{pmatrix} 0 \\ 1 \\ 0 \\ 0 \end{pmatrix} = 2$

and $c_2 \bar{A} b_2 = (1 \ 0 \ 0 \ 0) \begin{pmatrix} 1 \\ 0 \\ -1 \\ 0 \end{pmatrix} = 1$

therefore

$$U = \begin{pmatrix} c_1 \bar{A} b_1 & c_1 \bar{A} b_2 \\ c_2 \bar{A} b_1 & c_2 \bar{A} b_2 \end{pmatrix} = \begin{pmatrix} 1 & 2 \\ 1 & 1 \end{pmatrix}$$

$\det U = -1 \neq 0 \rightarrow U$ invertible

$$U^{-1} = -1 \begin{pmatrix} 1 & -2 \\ -1 & 1 \end{pmatrix} = \begin{pmatrix} -1 & 2 \\ 1 & -1 \end{pmatrix}$$

$w = Fz + G(t) \rightsquigarrow$ new input

$$w = U^{-1} \left(t - \begin{pmatrix} c_1 A^2 \\ c_2 A^2 \end{pmatrix} z \right)$$

$$A^2 = \begin{pmatrix} -1 & 1 & 1 & 1 \\ 0 & 0 & -1 & 0 \\ 1 & -1 & 0 & 0 \\ 0 & 0 & 0 & 0 \end{pmatrix} \begin{pmatrix} -1 & 1 & 1 & 1 \\ 0 & 0 & -1 & 0 \\ 1 & -1 & 0 & 0 \\ 0 & 0 & 0 & 0 \end{pmatrix} = \begin{pmatrix} 2 & -2 & -2 & -1 \\ -1 & 1 & 0 & 0 \\ -1 & 1 & 2 & 1 \\ 0 & 0 & 0 & 0 \end{pmatrix}$$

$$c_1 A^2 = (1 \ 0 \ -1 \ 0) \cdot A^2 = (3 \ -3 \ -1 \ -2)$$

$$c_2 \cdot A^2 = (1 \ 0 \ 0 \ 0) \cdot A^2 = (2 \ -2 \ -2 \ -1)$$

The resulting sys is:

$$\begin{cases} \dot{y} = \begin{pmatrix} 0 & 1 \\ 0 & 0 \end{pmatrix} y + \begin{pmatrix} 0 \\ 1 \end{pmatrix} t, \end{cases}$$

$$\begin{cases} \dot{\xi} = \begin{pmatrix} 0 & 1 \\ 0 & 0 \end{pmatrix} \xi + \begin{pmatrix} 0 \\ 1 \end{pmatrix} t_1, \\ \dot{\eta} = \begin{pmatrix} 0 & 1 \\ 0 & 0 \end{pmatrix} \eta + \begin{pmatrix} 0 \\ 1 \end{pmatrix} t_2 \\ \xi_1 = \xi, \\ \xi_2 = \eta, \end{cases}$$

(2) $\begin{cases} \dot{x}_1 = x_3 \\ \dot{x}_2 = x_3 + x_1^2 + \cos(x_2) d \\ \dot{x}_3 = -x_3 + x_1^2 + (1+x_1^2)v \\ y = x_1 \end{cases}$

d = measurable bounded $d \in \mathbb{R}$

(2.1) Assuming $d \equiv 0$

- Compute the fb that solves $v \rightarrow \text{FL}$

rel. deg.

$$r=1 \quad L_g L_h = \left(\begin{smallmatrix} \frac{\partial L_h}{\partial x} & \cdot & e \\ 1 & 0 & 0 \end{smallmatrix} \right) \begin{pmatrix} 0 \\ 0 \\ 1+x_1^2 \end{pmatrix} = 0$$

(*) $L_g L_h = \frac{\partial L_h}{\partial x} \cdot f = \left(\begin{smallmatrix} 1 & 0 & 0 \end{smallmatrix} \right) \begin{pmatrix} x_3 \\ x_3 + x_1^2 \\ -x_3 + x_1^2 \end{pmatrix} = x_3$

$$\begin{aligned} L_g L_g h &= \frac{\partial L_g h}{\partial x} \cdot g \\ &= \left(\begin{smallmatrix} 0 & 0 & 1 \end{smallmatrix} \right) \begin{pmatrix} 0 \\ 0 \\ 1+x_1^2 \end{pmatrix} = 1+x_1^2 \neq 0 \end{aligned}$$

Coord. test:

$$\begin{pmatrix} z \\ \eta \end{pmatrix} \begin{cases} n=2 \\ n-r=1 \end{cases} \rightarrow \begin{pmatrix} z_1 \\ z_2 \\ \eta \end{pmatrix}$$

$$z_1 = h = x_1$$

$$z_2 = L_g h = x_3$$

$$\eta \text{ s.t. } \nabla \varphi \cdot g = 0$$

$$M \text{ s.t. } \nabla \varphi \cdot \dot{x} = 0$$

$$\left(\frac{\partial \varphi}{\partial x_1} \quad \frac{\partial \varphi}{\partial x_2} \quad \frac{\partial \varphi}{\partial x_3} \right) \begin{pmatrix} 0 \\ 0 \\ 1+x_1^2 \end{pmatrix} = 0$$

$$\frac{\partial \varphi}{\partial x_3} + \frac{\partial \varphi}{\partial x_3} x_1^2 = 0$$

$$\eta = x_2$$

$$\begin{cases} \dot{z}_1 = \dot{x}_1 = x_3 = z_2 \\ \dot{z}_2 = \dot{x}_3 = -x_3 + x_1^2 + (1+x_1^2)v = -z_2 + z_1^2 + (1+z_1^2)v \\ \dot{\eta} = \dot{x}_2 = x_3 + x_1^2 = z_2 + z_1^2 \end{cases}$$

$$\text{Having } v = \frac{(z_2 - z_1^2 + v)}{(1+z_1^2)}$$

$$\begin{cases} \dot{z}_1 = z_2 \\ \dot{z}_2 = v \\ \dot{\eta} = z_2 + z_1^2 \end{cases}$$

• Zero dynamics

$$\begin{aligned} Z &= \{x \in \mathbb{R}^n \text{ s.t. } y(t_0) = 0 \text{ and } \dot{y}(t) = 0\} \\ &= \{x \in \mathbb{R}^n \text{ such that } x_1 = 0\} \end{aligned}$$

$$z = 0 \quad v = 0$$

$$\dot{\eta} = z_2 + z_1^2 = 0$$

$$Q = \left. \frac{\partial q(0, \eta)}{\partial \eta} \right|_0 = 0 \quad \text{No zero dynamics}$$

• Stabilizability of the closed loop sys

$\dot{\eta} = 0$ is stable since η is a constant

(2.2) Assuming $d \neq 0$

Converse if one the above solving the

(c) assuming $u \neq 0$

compute if only the output solving the output DDP problem

It's assured that d is a bounded disturbance.

Remember that v found in case $d=0$ is the fb that decouples the output from the disturbance if $r_0 < r_w$ and $L_{ph} = 0$

$$L_{ph} = 0$$

$$L_{ph} = 0$$

$$y = x_1, \dot{y} = x_3, \ddot{y} = x_3 + x_1^2 + (1+x_1^2)v, \text{ so } \boxed{\bar{r}_0 = 2}$$

since there's no w in the last derivative it's obvious that $r_0 < r_w$

The DDP problem admits a solution

$$\text{From } \ddot{y} \text{ I get } v = \frac{\sqrt{-x_1^2 - x_3}}{1 + x_1^2}$$

with this fb you just decouple the effect of the disturbance on the output.

$$\text{So I set } v = -k_1 x_1 - k_2 x_3$$

Discuss on the boundedness of the internal dynamics of the closed loop system

A necessary condition for boundedness is asymptotic stability.

Since $\ddot{y} = \cos(\eta)d$ ($= \dot{x}_2 = z_1^2 + z_2 + \cos(\eta)d|_{z=0} = \cos(\eta)d$) is the zero dynamics, which is unstable, nothing can be concluded.

(d) Stabilize if possible via high gain fb

$$\begin{cases} \dot{x}_1 = -x_1 - x_2 \cos(x_1) \\ \dot{x}_2 = x_1 - e^{-x_2} u \\ y = x_2 \end{cases}$$

$$f = \begin{pmatrix} -x_1 - x_2 \cos(x_1) \\ x_1 \end{pmatrix}$$
$$n \begin{pmatrix} 1 & 0 & 1 \end{pmatrix}$$

$$\left\{ \begin{array}{l} y = x_2 \\ \end{array} \right.$$

$$U \setminus \hat{x}_1$$

$$E = \begin{pmatrix} 0 \\ -e^{-x_2} \end{pmatrix}$$

$$h = x_2 \quad \frac{\partial h}{\partial x} = (0 \ 1)$$

$$\text{equilibrium at } \begin{pmatrix} x_1 \\ x_2 \end{pmatrix} = \begin{pmatrix} 0 \\ 0 \end{pmatrix}$$

Relative deg.

$$\textcircled{r=1} \quad L_g h = \frac{\partial h}{\partial x} \cdot E = (0 \ 1) \begin{pmatrix} 0 \\ -e^{-x_2} \end{pmatrix} = -e^{-x_2} \neq 0$$

then

$$v = -k(c_0 h + \dots + L_g^{r-1} h) \quad r > 1 \quad \text{where } (c_0, \dots, c_{m-2}) \text{ Hurwitz}$$

$v = -k h \text{ sign}(L_g h) \Rightarrow$ a fb of H-L's kind if $r = 1$
stabilizes the sys:

$$\text{sign}(x) = \begin{cases} -1 & \text{if } x < 0 \\ 0 & \text{if } x = 0 \\ 1 & \text{if } x > 0 \end{cases}$$

$$\text{In this case } \underbrace{\text{sign}(-e^{-x_2})}_{\text{pos}} = \text{sign}\left(-\frac{1}{e^{-x_2}}\right) = -1$$

so $v = +k h$ with k big enough

$$\bar{g} = \begin{pmatrix} -x_1 - x_2 \cos(x_1) \\ x_1 - k e^{-x_2} x_2 \end{pmatrix}$$

LTH:

$$\bar{A} = \frac{\partial \bar{g}}{\partial x} \Big|_{x=0} = \begin{pmatrix} -1 + x_2 \sin(x_1) & -\cos(x_1) \\ 1 & +k e^{-x_2} x_2 - k e^{-x_2} \end{pmatrix} \Big|_{x=0} = \begin{pmatrix} -1 & -1 \\ 1 & -k \end{pmatrix}$$

$$|\bar{A} - \lambda I| = \begin{pmatrix} -1-\lambda & -1 \\ 1 & -k-\lambda \end{pmatrix} = (-1-\lambda)(-k-\lambda) + 1 = 0$$

$$= +k + \lambda + \lambda k + \lambda^2 + 1 = 0$$

$$\lambda^2 + \lambda(k+1) + k+1 = 0$$

$$\lambda_{1,2} = \frac{-(\lambda+1) \pm \sqrt{(\lambda+1)^2 - 4(k+1)}}{2}$$

$$= \frac{-k-1}{2} \pm \frac{\sqrt{k^2 + 1 + 2k - 4k - 6}}{2}$$

with $k = 10$ (for instance)

$$\lambda_1 = -1.11$$

$$\lambda_2 = -9.88$$

So nonlinear sys is AS under high gain fb