

Casio VZ Virtual Instrument:

A replica of the Casio VZ-1/VZ-10M music synthesizer

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See <https://github.com/matthias-wolff/Casio-VZ-virtual-instrument/blob/master/Casio-VZ-virtual-instrument.pdf> for the latest version of this document.

Abstract

In this project I try to rebuild the vintage Casio VZ-1/VZ-10M music synthesizer in Reaktor 6 [6]. The primary goal is a fully functional player which is compatible with MIDI editor/librarian software like Midi Quest [8] or the like. My workplan is

1. make some debugging and development tools (waveform validator, envelope validator, etc.),
2. reproduce the 8 core waveforms of VZ-1/VZ-10M (1x sine, 5x sawtooth-like waveforms created by Casio's Phase Distortion Modulation, 1x white noise, 1x pitch-sensitive narrow-band noise),
3. implement the core sound engine (wavetable oscillators, phase and ring modulators, VCAs, oscillator circuits),
4. implement control signal generators (amplitude envelope, key following, layering, parametric sensitivity characteristics, etc.),
5. implement MIDI SysEx control capability, and
6. reproduce the factory voice and operation libraries.

I always strongly disliked the unpleasant—though most characteristic—aliasing and analog noise of the VZ-1. Hence, I will not attempt to reproduce this. Insofar, the remake is not intended to be perfect.

As a secondary goal I may want to reproduce the GUI of the original instrument. This would be a nice-to-have, however not necessarily of much practical use.

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1 Goals and Prerequisites

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1.1 The VZ-1/VZ-10M Music Synthesizer

[TODO: ...]

1.2 The Reaktor 6 Modular DSP Lab

[TODO: ...] [6]

1.3 The Midi Quest 12 Editor/Librarian Software

[TODO: ...] [8]

2 Development and Debugging Tools

2.1 Waveform Validator

[TODO: ...]

2.2 Envelope Validator

[TODO: ...]

3 Development and Debugging Tools

4 Core Sound Engine

VZ's sound engine is called “iPD (interactive phase distortion) modular sound engine” [2, p. 12]. According to Casio, this is an enhanced version of the phase distortion modulation introduced by the predecessor of the VZ-1 called CZ-1 [3, p. 39]. Actually, the VZ synths series seem to feature wavetable oscillators—with waveforms partly generated by the original PD synthesis—together with ordinary phase modulation and ring modulation.

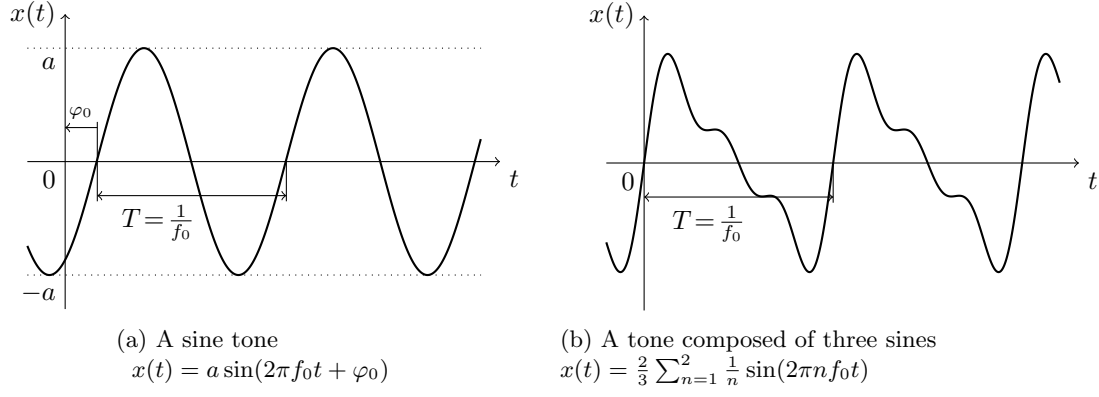


Figure 1: Examples for basic musical tones. The tone frequency—measured in Hertz—is denoted by f_0 . We denote time by t and the duration of one cycle by T . Both are measured in seconds. The argument of the sine functions are called phase angles φ . One cycle corresponds to a phase difference of $\Delta\varphi = 2\pi$.

4.1 Theory

I assume that the reader is familiar with the basics of signal processing. Introductions can be found in [5] and [7].

4.1.1 Tones, Timbre, Frequency, and Pitch

A musical tone is a periodic sound. Fig. 1a shows an example of the most basic musical tone, a sine. Its mathematical representation is

$$x(t) = a \sin(2\pi f_0 t + \varphi_0), \quad (1)$$

where a denotes the amplitude, f_0 denotes the tone (or fundamental) frequency, and φ_0 denotes a phase offset.

Remark: Throughout this report I will assume continuous time signals $x(t)$ which, unless stated otherwise, take real values in $-1 \leq x(t) \leq 1$, even though we actually talk about digital signals of course. However, this simplification saves some notational trouble.

Fig. 1b shows an example of another basic tone composed by the mixture of three sine tones. The timbre of a musical tone can be described mathematically by the coefficients of its FOURIER series, i.e., the amplitudes and phase offsets of sine tones with frequencies $f = n \cdot f_0$ constituting the composed tone (see, e.g., [5, p.102–113]). More complex tones comprise aperiodic components (noise) and varying timbre over time.

As human perception of sound frequency is nearly logarithmic, Western music theory commonly uses the pitch

$$p = 69 + 12 \log_2 \left(\frac{f_0}{440 \text{ Hz}} \right) \quad p = f2p(f) \quad (2)$$

measured in semitones. The value of 440 Hz in the denominator is the concert pitch. The offset of 69 makes the pitch computed by Eq. (2) compatible with a MIDI frequency data value. According to the MIDI tuning standard [TODO:], p takes values from the interval $0.00 \leq p \leq 127.99$ with the integer part corresponding to a note number and the decimal fraction, with two digits precision, to a detune. MIDI note number $p = 0$ stands for C_{-1} , $p = 69$ for the concert pitch A_4 , and $p = 127$ stands for B_9 .

Solving Eq. (2) for f_0 , we obtain a formula for converting pitch to frequency:

$$f_0 = 2^{\frac{p-69}{12}} \cdot 440 \text{ Hz} \quad f = p2f(p). \quad (3)$$

4.1.2 Phase Distortion Modulation

Phase distortion modulation (PDM) was invented in the early 1980's by I. TOMITA, Y. TAKAHASHI and colleagues of Casio Computer Ltd. for the CZ music synthesizer series [4]. Fig. 4 shows an

example from the CZ manual. The VZ series uses wavetables basing on PDM.

Core element of PDM is a phase distortion characteristic curve as shown in Fig. 2a [1, p. 20–21]. We see that the curve consists of two straight sections

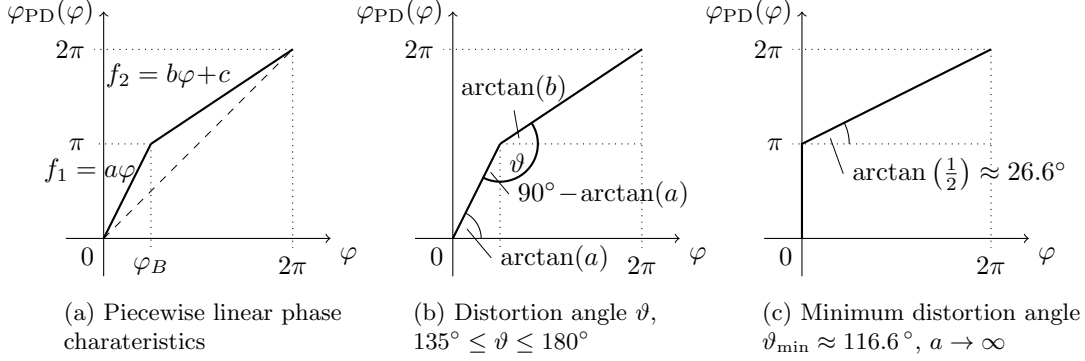


Figure 2: Characteristic curve of phase distortion

$$f_1(\varphi) = a\varphi, \quad 0 \leq \varphi < \varphi_B, \quad \text{connecting points } (0,0) \text{ and } (\varphi_B, \pi), \text{ and} \quad (4)$$

$$f_2(\varphi) = b\varphi + c, \quad \varphi_B \leq \varphi < 2\pi, \quad \text{connecting points } (\varphi_B, \pi) \text{ and } (2\pi, 2\pi). \quad (5)$$

The slope a of the first section is variable and can take real values ≥ 1 . The start and end points of the sections define the following constraints

$$f_1(\varphi_B) = a\varphi_B = \pi \quad \rightsquigarrow \varphi_B = \frac{\pi}{a} \quad (6)$$

$$f_2(\varphi_B) = \frac{\pi}{a}b + c = \pi \quad (7)$$

$$f_2(2\pi) = 2\pi b + c = 2\pi \quad (8)$$

From Eqs. (6) to (8) follows

$$b = \frac{a}{2a-1} \quad \text{and} \quad c = \frac{2a-2}{2a-1}\pi, \quad (9)$$

and finally

$$\varphi_{PD}(\varphi) = \begin{cases} a\varphi & \text{for } 0 \leq \varphi < \frac{\pi}{a} \\ \frac{a\varphi + 2\pi(a-1)}{2a-1} & \text{for } \frac{\pi}{a} \leq \varphi < 2\pi \end{cases} \quad \text{with } a \geq 1. \quad (10)$$

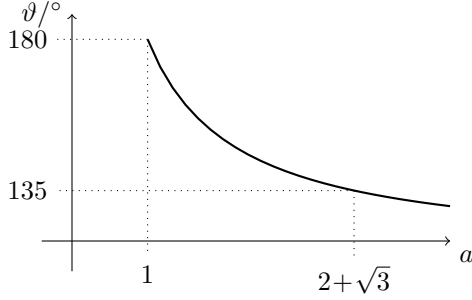
Modulation depth can also be expressed by the angle between the straight sections

$$\vartheta/^\circ = 180 - \arctan(a) + \arctan(b) = 180 - \arctan(a) + \arctan\left(\frac{a}{2a-1}\right) \quad (11)$$

measured in degrees (see Fig. 2b). Literature states that the permissible range is $135^\circ \leq \vartheta \leq 180^\circ$ [4]. However, for the wavetable of the VZ series this restriction does not seem to hold. Measurement yielded angles down to 123° corresponding to slope parameters $a > 10$ (see `matlab/PDMwaves.mlx`). As I did not manage to solve Eq. (11) for a , I just plot the function and print a lookup table in Fig. 3.

Six out of eight VZ-1 wavetable samples are phase-shifted and phase-distorted sine waves:

$$x(\varphi) = \sin\left(\varphi_{PD}(\varphi) - \frac{\pi}{2}\right). \quad (12)$$



(a) Distortion angle parameter ϑ as a function of slope parameter a

$\vartheta/^\circ$	a	$\vartheta/^\circ$	a
116.6	∞	150	2.02
120	20.07	155	1.74
125	8.19	160	1.52
130	5.14	165	1.35
135	$2+\sqrt{3} \approx 3.73$	170	1.21
140	2.92	175	1.10
145	2.39	180	1.00

(b) Lookup table

Figure 3: Relation between distortion angle parameter ϑ and slope parameter a , see Eq. (11)

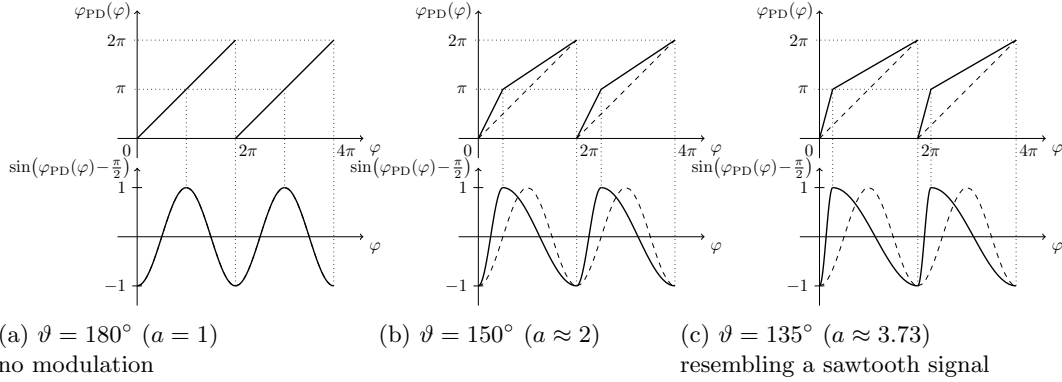


Figure 4: Example of phase distortion modulation of the function $\sin(\varphi - \frac{\pi}{2})$ adapted from [1, p. 20–21, Figs. 1–3]. Phase angles $\varphi < 0$ and $\varphi \geq 2\pi$ are wrapped into the base interval $0 \leq \varphi < 2\pi$.

Fig. 4 shows some examples.

Issue: Actually, VZ-1 seems to produce the inverted waveforms:

$$x_{\text{VZ-1}}(\varphi) = -\sin\left(\varphi_{\text{PD}}(\varphi) - \frac{\pi}{2}\right) = \sin\left(\varphi_{\text{PD}}(\varphi) + \frac{\pi}{2}\right). \quad (13)$$

Remark: An alternate phase distortion characteristic $f(x)$ with domain $0 \leq x < 1$ taking values $0 \leq f(x) < 1$ may be needed for Reaktor implementation:

$$f(x) = \begin{cases} ax & 0 \leq x < \frac{1}{2a} \\ \frac{ax + a - 1}{2a - 1} & \frac{1}{2a} \leq x < 1 \end{cases} \quad \text{with } a \geq 1. \quad (14)$$

4.1.3 Wavetable Oscillators

The original VZ-1/VZ-10M is apparently based on wavetable oscillators [2, p. 34]. Such oscillators contain a set of digital waveforms, each comprising exactly one signal period. Depending on the pitch, these waveforms are played back at different rates. [TODO: ??] shows an example of a VZ waveform (SAW4).

Formally, a single waveform can be expressed by

$$w(\varphi) \quad \text{with } 0 \leq \varphi < 2\pi, \quad (15)$$

where the argument φ is a phase angle. A oscillator based on this waveform uses a time-varying phase angle $\varphi(t)$. Its output signal can be written as

$$x(t) = w(\varphi(t)) = w(2\pi(f_0 t \bmod 1)) \quad \text{with } -1 \leq x(t) \leq 1 \quad (16)$$

where w denotes a waveform, $f_0 \in \mathbb{R}^{>0}$ denotes the note frequency in Hertz, and t denotes the time in seconds. Further, we denote by

$$x \bmod 1 := x - \lfloor x \rfloor \quad \text{with } 0 \leq (x \bmod 1) < 1 \quad (17)$$

the decimal fraction of the real valued quantity x . Hence, the argument $\varphi(t) = 2\pi(f_0 t \bmod 1)$ on the right side of Eq. (16) involves resetting the phase angle of the waveform to zero whenever the cycle length 2π is surpassed. **[TODO: ??]** shows an example.

We will use the circuit symbol **[TODO: ...]** for a wavetable oscillator.

4.1.4 Phase Modulation

4.1.5 Ring Modulation

4.2 VZ Hardware

The VZ series features eight waveforms: sine (SINE), five sawtooth-shaped waveforms created by phase distortion modulation (SAW1–SAW5), white noise (NOISE1), and narrow-band noise (NOISE2).

4.3 Replica

4.3.1 Voltage-Controlled Oscillators (VCO)

In order to read the wavetable out of my VZ-1, I proceeded as follows

1. Set the master tune of VZ-1 to 440 Hz in MENU 3-00,
2. Initialize a new voice,
3. Mute all oscillator modules except M1 and configure line A as MIX in MENU 1/00,
4. In MENU 1/02, set M1 to a fix pitch of OCT=00, NOTE=00, FINE=08 to generate tones with a frequency of $46^7/8$ Hz (see below),
5. Configure all amplitude controls of M1 for maximum signal output and set a trivial amplitude envelope mimicking a mere gate signal,
6. Disable all LFOs and MIDI controllers on M1.

I recorded with the Komplete Audio 6 sound interface at a sampling frequency of $f_S = 48$ kHz. Each waveform in the wavetable shall contain $K = 1024$ samples for one full cycle. Hence, the “natural” frequency of the waveforms—i.e., when reading out at 48 kSamples per second—is

$$f_0 = \frac{f_A}{K} = \frac{48000 \text{ Hz}}{1024} = 46^7/8 \text{ Hz} \quad (18)$$

which corresponds to the musical note F#1 + 23 ct. Using the fix pitch settings above, one cycle of the waveform being recorded is exactly 1024 samples long and no resampling is necessary.

4.3.2 Voltage-Controlled Amplifiers (VCA)

5 Control Signal Generators

6 MIDI SysEx Control

Issue: Reaktor 6 does not seem to support MIDI SysEx. Circumvent by control automation?

7 GUI Replica

- LCD dot matrix display 96×64 pixels

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