

Casio VZ Virtual Instrument:

A replica of the Casio VZ-1/VZ-10M music synthesizer

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September 2, 2019

See <https://github.com/matthias-wolff/Casio-VZ-virtual-instrument/blob/master/Casio-VZ-virtual-instrument.pdf> for the latest version of this document.

Abstract

In this project I try to rebuild the vintage Casio VZ-1/VZ-10M music synthesizer in Reaktor 6 [6]. The primary goal is a fully functional player which is compatible with MIDI editor/librarian software like Midi Quest [8] or the like. My workplan is

1. make some debugging and development tools (waveform validator, envelope validator, etc.),
2. reproduce the 8 core waveforms of VZ-1/VZ-10M (1x sine, 5x sawtooth-like waveforms created by Casio's Phase Distortion Modulation, 1x white noise, 1x pitch-sensitive narrow-band noise),
3. implement the core sound engine (wavetable oscillators, phase and ring modulators, VCAs, oscillator circuits),
4. implement control signal generators (amplitude envelope, key following, layering, parametric sensitivity characteristics, etc.),
5. implement MIDI SysEx control capability, and
6. reproduce the factory voice and operation libraries.

I always strongly disliked the unpleasant—though most characteristic—aliasing and analog noise of the VZ-1. Hence, I will not attempt to reproduce this. Insofar, the remake is not intended to be perfect.

As a secondary goal I may want to reproduce the GUI of the original instrument. This would be a nice-to-have, however not necessarily of much practical use.

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1 Goals and Prerequisites

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1.1 The VZ-1/VZ-10M Music Synthesizer

[TODO: ...]

1.2 The Reaktor 6 Modular DSP Lab

[TODO: ...] [6]

1.3 The Midi Quest 12 Editor/Librarian Software

[TODO: ...] [8]

2 Development and Debugging Tools

2.1 Waveform Validator

[TODO: ...]

2.2 Envelope Validator

[TODO: ...]

3 Development and Debugging Tools

4 Core Sound Engine

VZ's sound engine is called “iPD (interactive phase distortion) modular sound engine” [2, p. 12]. According to Casio, this is an enhanced version of the phase distortion modulation introduced by the predecessor of the VZ-1 called CZ-1 [3, p. 39]. Actually, the VZ synths series seem to feature wavetable oscillators—with waveforms partly generated by the original PD synthesis—together with ordinary phase modulation and ring modulation.

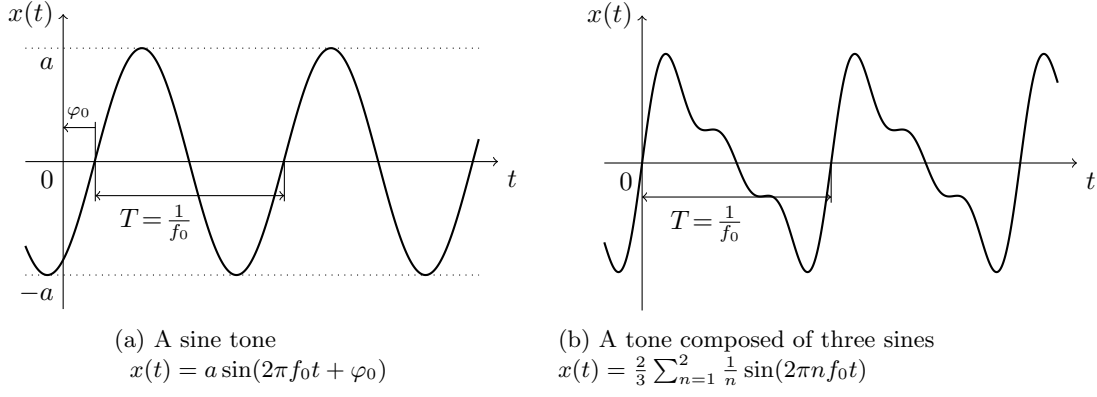


Figure 1: Examples for basic musical tones. The tone frequency—measured in Hertz—is denoted by f_0 . We denote time by t and the duration of one cycle by T . Both are measured in seconds. The argument of the sine functions are called phase angles φ . One cycle corresponds to a phase difference of $\Delta\varphi = 2\pi$.

4.1 Theory

I assume that the reader is familiar with the basics of signal processing. Introductions can be found in [5] and [7].

4.1.1 Tones, Timbre, Frequency, and Pitch

A musical tone is a periodic sound. Fig. 1a shows an example of the most basic musical tone, a sine. Its mathematical representation is

$$x(t) = a \sin(2\pi f_0 t + \varphi_0), \quad (1)$$

where a denotes the amplitude, f_0 denotes the tone (or fundamental) frequency, and φ_0 denotes a phase offset.

Remark: Throughout this report I will assume continuous time signals $x(t)$ which, unless stated otherwise, take real values in $-1 \leq x(t) \leq 1$, even though we actually talk about digital signals of course. However, this simplification saves some notational trouble.

Fig. 1b shows an example of another basic tone composed by the mixture of three sine tones. The timbre of a musical tone can be described mathematically by the coefficients of its FOURIER series, i.e., the amplitudes and phase offsets of sine tones with frequencies $f = n \cdot f_0$ constituting the composed tone (see, e.g., [5, p.102–113]). More complex tones comprise aperiodic components (noise) and varying timbre over time.

As human perception of sound frequency is nearly logarithmic, Western music theory commonly uses the pitch

$$p = 1200 \cdot \log_2 \left(\frac{f_0}{f_{a^1}} \right) \quad (2)$$

measured in Cents (ct). The quantity in the denominator is a reference frequency. I will use the concert pitch $f_{a^1} = 440$ Hz. Musical intervals can be measured in terms of pitch. E.g., a semitone spans 100 ct, a fifth spans 500 ct, and an octave spans 1200 ct. Solving Eq. (2) for f_0 , we obtain a formula for converting pitch to frequency:

$$f_0 = f_{a^1} \cdot 2^{\frac{p}{1200}}. \quad (3)$$

4.1.2 Phase Distortion Modulation

Phase distortion modulation was invented in the early 1980's by I. TOMITA, Y. TAKAHASHI and colleagues of Casio Computer Ltd [4]. It is based on the characteristic curve shown in Fig. 2a [1, p.20–21]. We see that the curve consists of two straight sections

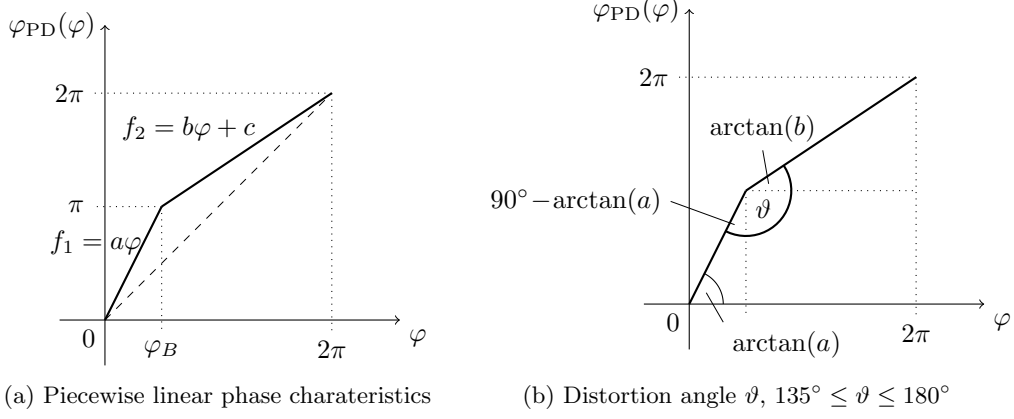


Figure 2: Characteristic curve of phase distortion

$$f_1(\varphi) = a\varphi, \quad 0 \leq \varphi < \varphi_B, \quad \text{connecting points } (0,0) \text{ and } (\varphi_B, \pi), \quad (4)$$

$$f_2(\varphi) = b\varphi + c, \quad \varphi_B \leq \varphi \leq 2\pi, \quad \text{connecting points } (\varphi_B, \pi) \text{ and } (2\pi, 2\pi). \quad (5)$$

The slope a of the first section is variable and can take real values ≥ 1 . The start and end points of the sections define the following constraints

$$f_1(\varphi_B) = a\varphi_B = \pi \quad \rightsquigarrow \varphi_B = \frac{\pi}{a} \quad (6)$$

$$f_2(\varphi_B) = \frac{\pi}{a}b + c = \pi \quad (7)$$

$$f_2(2\pi) = 2\pi b + c = 2\pi \quad (8)$$

From Eqs. (6) to (8) follows

$$b = \frac{a}{2a-1} \quad \text{and} \quad c = \frac{2a-2}{2a-1}\pi, \quad (9)$$

and finally

$$\varphi_{PD}(\varphi) = \begin{cases} a\varphi & \text{for } 0 \leq \varphi < \frac{\pi}{a} \\ \frac{a\varphi + 2\pi(a-1)}{2a-1} & \text{for } \frac{\pi}{a} \leq \varphi \leq 2\pi \end{cases} \quad \text{with } a \geq 1. \quad (10)$$

Modulation depth can also be expressed by the angle between the straight sections

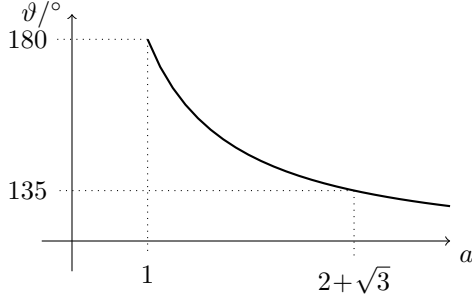
$$\vartheta/^\circ = 180 - \arctan(a) + \arctan(b) = 180 - \arctan(a) + \arctan\left(\frac{a}{2a-1}\right) \quad (11)$$

measured in degrees (see Fig. 2b). Literature states that the permissible range is $135^\circ \leq \vartheta \leq 180^\circ$ [4]. However, for the wavetable of the VZ series this restriction does not seem to hold. Measurement yielded angles down to 123° corresponding to slope parameters $a > 10$ (see `matlab/PDMwaves.mlx`). As I did not manage to solve Eq. (11) for a , I just plot the function and print a lookup table in Fig. 3.

Phase distortion modulation was applied in the Casio CZ synthesizer series. Fig. 4 shows an example from the CZ manual. The VZ series uses wavetables basing on PDM.

Remark: An alternate phase distortion characteristics $f(x)$ with domain $0 \leq x \leq 1$ taking values $0 \leq f(x) \leq 1$ may be needed for Reaktor implementation:

$$f(x) = \begin{cases} ax & 0 \leq x < \frac{1}{2a} \\ \frac{ax + a - 1}{2a - 1} & \frac{1}{2a} \leq x < 1 \end{cases} \quad \text{with } a \geq 1. \quad (12)$$

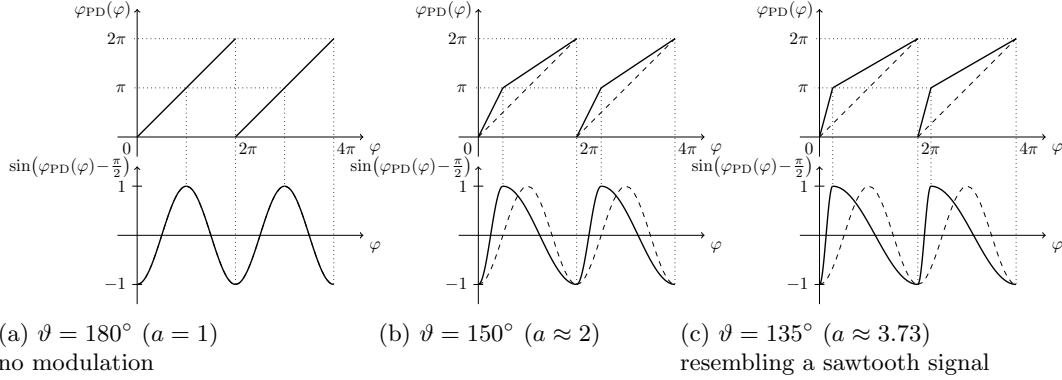


(a) Distortion angle parameter ϑ as a function of slope parameter a

$\vartheta/^\circ$	a	$\vartheta/^\circ$	a
135	$2 + \sqrt{3} \approx 3.73$	160	1.52
140	2.92	165	1.35
145	2.39	170	1.21
150	2.02	175	1.10
155	1.74	180	1.00

(b) Lookup table

Figure 3: Relation between distortion angle parameter ϑ and slope parameter a , see Eq. (11)



(a) $\vartheta = 180^\circ$ ($a = 1$)
no modulation

(b) $\vartheta = 150^\circ$ ($a \approx 2$)

(c) $\vartheta = 135^\circ$ ($a \approx 3.73$)
resembling a sawtooth signal

Figure 4: Example of phase distortion modulation of the function $\sin(\varphi - \frac{\pi}{2})$ adapted from [1, p. 20–21, Figs. 1–3]

4.1.3 Wavetable Oscillators

The original VZ-1/VZ-10M is apparently based on wavetable oscillators [2, p. 34]. Such oscillators contain a set of digital waveforms, each comprising exactly one signal period. Depending on the pitch, these waveforms are played back at different rates. **[TODO: ??]** shows an example of a VZ waveform (SAW4).

Formally, a single waveform can be expressed by

$$w(\varphi) \quad \text{with } 0 \leq \varphi < 2\pi, \quad (13)$$

where the argument φ is a phase angle. A oscillator based on this waveform uses a time-varying phase angle $\varphi(t)$. Its output signal can be written as

$$x(t) = w(\varphi(t)) = w(2\pi(f_0 t \bmod 1)) \quad \text{with } -1 \leq x(t) \leq 1 \quad (14)$$

where w denotes a waveform, $f_0 \in \mathbb{R}^{>0}$ denotes the note frequency in Hertz, and t denotes the time in seconds. Further, we denote by

$$x \bmod 1 := x - \lfloor x \rfloor \quad \text{with } 0 \leq (x \bmod 1) < 1 \quad (15)$$

the decimal fraction of the real valued quantity x . Hence, the argument $\varphi(t) = 2\pi(f_0 t \bmod 1)$ on the right side of Eq. (14) involves resetting the phase angle of the waveform to zero whenever the cycle length 2π is surpassed. **[TODO: ??]** shows an example.

We will use the circuit symbol **[TODO: ...]** for a wavetable oscillator.

4.1.4 Phase Modulation

4.1.5 Ring Modulation

4.2 VZ Hardware

The VZ series features eight waveforms: sine (SINE), five sawtooth-shaped waveforms created by phase distortion modulation (SAW1–SAW5), white noise (NOISE1), and narrow-band noise (NOISE2).

4.3 Replica

4.3.1 Voltage-Controlled Oscillators (VCO)

4.3.2 Voltage-Controlled Amplifiers (VCA)

5 Control Signal Generators

6 MIDI SysEx Control

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