

Capacity of Multi-antenna Gaussian Channels (I. E. Telatar, 1999)

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Overview

- 1 Introduction
 - Abstract
 - Notation and Assumptions
 - Properties and Lemmas
- 2 Gaussian Channel with Deterministic H
 - Capacity Derivation
 - Formal Derivation of Capacity
- 3 Gaussian Channel with Ergodic \mathbf{H}
 - Introduction to the problem
 - Capacity-achieving Distribution
 - Evaluation of the Capacity
 - Examples
- 4 *Bonus*: Multiaccess Channel

Table of Contents

- 1 Introduction
 - Abstract
 - Notation and Assumptions
 - Properties and Lemmas
- 2 Gaussian Channel with Deterministic H
 - Capacity Derivation
 - Formal Derivation of Capacity
- 3 Gaussian Channel with Ergodic \mathbf{H}
 - Introduction to the problem
 - Capacity-achieving Distribution
 - Evaluation of the Capacity
 - Examples
- 4 *Bonus*: Multiaccess Channel

In this presentation I will talk about the capacity of a single-user Gaussian channel with multiple receiving and/or transmitting antennas [1] (also known as **MIMO** channel).

I will talk about 3 cases:

- Deterministic Channel
- Random i.i.d. Ergodic Channel
- *Bonus*: Multi-User Case

Notation I

The notation adopted is halfway between the one used during the course and the one from the original paper.

Let's denote with t the number of transmitting and with r the number of receiving antennas.

We will consider the classical linear model, where $\mathbf{x} \in \mathbb{C}^t$ is the transmitted vector and $\mathbf{y} \in \mathbb{C}^r$ is the received vector, $H \in \mathbb{C}^{r \times t}$ is the complex channel matrix and $\mathbf{n} \in \mathbb{C}^r$ is the noise

$$\mathbf{y} = H\mathbf{x} + \mathbf{n}$$

Notation II

We assume noise at different receivers to be independent and normalized, i.e. $E[\mathbf{n}\mathbf{n}^\dagger] = I_r$.

The dag (\dagger) notation is used for the conjugate-transpose operation.

We have the power constraint

$$E[\mathbf{x}^\dagger \mathbf{x}] = E[\text{tr}(\mathbf{x}\mathbf{x}^\dagger)] = \text{tr}(E[\mathbf{x}\mathbf{x}^\dagger]) \leq P$$

A complex random vector (r.v.e.) $\mathbf{x} \in \mathbb{C}^n$ is said to be Complex Gaussian if its real extension $\hat{\mathbf{x}} \triangleq \begin{bmatrix} \text{Re}\{\mathbf{x}\} \\ \text{Im}\{\mathbf{x}\} \end{bmatrix} \in \mathbb{R}^{2n}$ is Gaussian.

The r.v.e. \mathbf{x} will have *mean* and *covariance* respectively $\mu = E[\mathbf{x}]$ and $Q = \text{Cov}(Q) = E[(\mathbf{x} - \mu)(\mathbf{x} - \mu)^\dagger]$

Notation III

Defining for a complex matrix A

$$\hat{A} = \begin{bmatrix} \text{Re}(A) & -\text{Im}(A) \\ \text{Im}(A) & \text{Re}(A) \end{bmatrix}$$

we say that a Complex Gaussian r.v. is *circularly symmetric* if

$$\text{Cov}(\hat{\mathbf{x}}) = E[(\hat{\mathbf{x}} - \hat{\mu})(\hat{\mathbf{x}} - \hat{\mu})^T] = \frac{1}{2}\hat{Q}$$

pdf of circularly symmetric Complex Gaussian

The pdf of a circularly symmetric Complex Gaussian is

$$\begin{aligned} \gamma_{\mu, Q} &= \det(\pi \hat{Q})^{-\frac{1}{2}} e^{-(\hat{x} - \hat{\mu})^T \hat{Q} (\hat{x} - \hat{\mu})} \\ &= \det(\pi Q)^{-1} e^{-(x - \mu)^\dagger Q (x - \mu)} \end{aligned}$$

Lemma 1

The following properties hold:

$$C = AB \iff \hat{C} = \hat{A}\hat{B} \quad (1a)$$

$$C = A + B \iff \hat{C} = \hat{A} + \hat{B} \quad (1b)$$

$$C = A^\dagger \iff \hat{C} = \hat{A}^T \quad (1c)$$

$$C = A^{-1} \iff \hat{C} = \hat{A}^{-1} \quad (1d)$$

$$\det(\hat{A}) = |\det(A)|^2 = \det(AA^\dagger) \quad (1e)$$

$$z = x + y \iff \hat{z} = \hat{x} + \hat{y} \quad (1f)$$

$$y = Ax \iff \hat{y} = \hat{A}\hat{x} \quad (1g)$$

$$\operatorname{Re}\{x^\dagger y\} = \hat{x}^T \hat{y} \quad (1h)$$

Properties and Lemmas II

Corollary 1

A matrix $U \in \mathbb{C}^{n \times n}$ is unitary if and only if $\hat{U} \in \mathbb{R}^{2n \times 2n}$ is orthonormal.

Corollary 2

If $Q \in \mathbb{C}^{n \times n}$ is positive semi-definite, then so is $\hat{Q} \in \mathbb{R}^{2n \times 2n}$.

Lemma 2

Suppose the complex r.v.e. $\mathbf{x} \in \mathbb{C}^n$ is zero-mean and $\text{Cov}(\mathbf{x}) = Q$. Then the differential entropy of \mathbf{x} satisfies $h(\mathbf{x}) \leq \log \det(\pi e Q)$ with equality if and only if \mathbf{x} is circularly symmetric Complex Gaussian.

In other words, circularly symmetric Complex Gaussian r.v.e. are entropy maximizers for the class of complex random vectors.

Lemma 3

If $\mathbf{x} \in \mathbb{C}^n$ is a circularly symmetric Complex Gaussian then so is $\mathbf{y} = A\mathbf{x}$ for any $A \in \mathbb{C}^{m \times n}$.

Lemma 4

If \mathbf{x} and \mathbf{y} are independent circularly symmetric Complex Gaussians, then $\mathbf{z} = \mathbf{x} + \mathbf{y}$ is also circularly symmetric Complex Gaussian.

Table of Contents

- 1 Introduction
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 - Properties and Lemmas
- 2 Gaussian Channel with Deterministic H
 - Capacity Derivation
 - Formal Derivation of Capacity
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 - Introduction to the problem
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 - Examples
- 4 *Bonus*: Multiaccess Channel

Capacity Derivation I

We first consider the case of a deterministic (fixed) transfer function H . We consider the case where **both receiver and transmitter know H** .

Singular Value Decomposition (SVD)

Any matrix $H \in \mathbb{C}^{r \times t}$ can be decomposed as

$$H = UDV^\dagger$$

where $U \in \mathbb{C}^{r \times r}$ and $V \in \mathbb{C}^{t \times t}$ are unitary, $D \in \mathbb{C}^{r \times t}$ is diagonal with non-negative real entries.

In fact, the columns of U are the eigenvectors of HH^\dagger , the columns of V are the eigenvectors of $H^\dagger H$ and the diagonal entries of D , called *singular values* of H , are the eigenvalues (which coincide for the two cases) and given the hermitianity of such matrices, they are real and non-negative.

Note: typically the eigenvalues are ordered from the largest to the smallest.

Capacity Derivation II

Thus, the problem can be seen as follows

$$\mathbf{y} = UDV^\dagger \mathbf{x} + \mathbf{n}$$

Therefore, preprocessing the transmitted symbol as $\mathbf{x} = V\tilde{\mathbf{x}}$ and $\tilde{\mathbf{y}} = U^\dagger \mathbf{y}$ and defining $\tilde{\mathbf{n}} = U^\dagger \mathbf{n}$, we get

$$\tilde{\mathbf{y}} = D\tilde{\mathbf{x}} + \tilde{\mathbf{n}}$$

as an equivalent channel.

Calling $\sigma_i = \lambda_i^{\frac{1}{2}}$ the singular values for $i = 1, \dots, \min(r, t)$, we have

$$\begin{cases} \tilde{\mathbf{y}}_i = \sigma_i \tilde{\mathbf{x}}_i + \tilde{\mathbf{n}}_i & 1 \leq i \leq \min(r, t) \\ \tilde{\mathbf{y}}_i = \tilde{\mathbf{n}}_i & i > \min(r, t) \end{cases}$$

Capacity Derivation III

It's clear that in order to maximize the capacity we need to choose $\tilde{\mathbf{x}}_i$ for $1 \leq i \leq \min(r, t)$ (actually only up to the number of non-zero singular values) to be independent, zero-mean, circularly symmetric Complex Gaussian.

It can be shown that the optimal variances must follow once again the “water-filling” equations

$$\begin{cases} E[\text{Re}(\tilde{\mathbf{x}}_i)] = E[\text{Im}(\tilde{\mathbf{x}}_i)] = \frac{1}{2}(\mu - \lambda_i^{-1})^+ \\ P = \sum_i (\mu - \lambda_i^{-1})^+ \\ C = \sum_i (\log(\mu \lambda_i))^+ \end{cases}$$

Formal Derivation of Capacity I

I will now proceed in the derivation of the capacity in a more formal way. This also will be used with some slight differences in the derivation of the capacity when \mathbf{H} is not deterministic.

Recall, the mutual information that we are interested in is

$$I(\mathbf{x}; \mathbf{y}) = h(\mathbf{y}) - h(\mathbf{y}|\mathbf{x}) = h(\mathbf{y}) - h(\mathbf{n})$$

Note that if \mathbf{x} satisfies $E[\mathbf{x}^\dagger \mathbf{x}] \leq P$, so does $\mathbf{x} - E[\mathbf{x}]$, thus we will only focus on zero-mean \mathbf{x} .

Furthermore, if \mathbf{x} is zero-mean with covariance $E[\mathbf{x}\mathbf{x}^\dagger] = Q$, then \mathbf{y} is zero-mean with covariance $E[\mathbf{y}\mathbf{y}^\dagger] = HQH^\dagger + I_r$, and by Lemma 2 among such \mathbf{y} the entropy is largest when \mathbf{y} is circularly symmetric Complex Gaussian, which is the case when \mathbf{x} is circularly symmetric Complex Gaussian (Lemmas 3 and 4).

Formal Derivation of Capacity II

Considering, then, \mathbf{x} circularly symmetric Complex Gaussian, we have

$$\begin{aligned} I(\mathbf{x}; \mathbf{y}) &= h(\mathbf{y}) - h(\mathbf{n}) = \log \det \left(\pi e (I_r + H Q H^\dagger) \right) - \log \det (\pi e I_r) \\ &= \log \det \left(I_r + H Q H^\dagger \right) = \log \det \left(I_t + Q H^\dagger H \right) \end{aligned}$$

where the last equality is given by *Sylvester's determinant identity*, that states $\det(I + AB) = \det(I + BA)$.

Since it occurs many times during the paper, we define

$$\Psi(Q, H) = \log \det \left(I_r + H Q H^\dagger \right)$$

Formal Derivation of Capacity III

Since we're dealing with zero-mean circularly symmetric Complex Gaussian, we only need to find a positive semi-definite matrix Q to fully determine the distribution, such that $\text{tr}(Q) \leq P$.

Notice that $H^\dagger H$ is hermitian, thus it can be decomposed as $H^\dagger H = V^\dagger \Lambda V$ with unitary V and diagonal positive semi-definite Λ . Thus, using Sylvester's identity again we obtain

$$\begin{aligned}\log \det(I_t + QH^\dagger H) &= \log \det(I_t + \Lambda^{\frac{1}{2}} V Q V^\dagger \Lambda^{\frac{1}{2}}) \\ &= \log \det(I_t + \Lambda^{\frac{1}{2}} \tilde{Q} \Lambda^{\frac{1}{2}})\end{aligned}$$

where we noticed that Q and \tilde{Q} are equivalent for both definiteness and power constraint, so we can as well optimize over \tilde{Q} .

Formal Derivation of Capacity IV

A special case of *Hadamard's inequality* for positive semi-definite matrices, states that

$$I(\mathbf{x}; \mathbf{y}) = \log \det \left(I_t + \Lambda^{\frac{1}{2}} \tilde{Q} \Lambda^{\frac{1}{2}} \right) \leq \prod_i (1 + \tilde{Q}_{ii} \lambda_i)$$

with equality for diagonal matrices.

Thus, we see that the maximizing $\tilde{Q} = VQV^\dagger$ is diagonal, and the optimal entries follow the water-filling equations, i.e.

$$\begin{cases} \tilde{Q}_{ii} = (\mu - \lambda_i^{-1})^+ \\ P = \sum_i \tilde{Q}_{ii} \\ C = \sum_i (\log(\mu \lambda_i))^+ \end{cases}$$

Table of Contents

- 1 Introduction
 - Abstract
 - Notation and Assumptions
 - Properties and Lemmas
- 2 Gaussian Channel with Deterministic H
 - Capacity Derivation
 - Formal Derivation of Capacity
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 - Capacity-achieving Distribution
 - Evaluation of the Capacity
 - Examples
- 4 *Bonus*: Multiaccess Channel

We now consider the case when the matrix \mathbf{H} is i.i.d. random and ergodic. Ergodicity comes into play to ensure that the capacity is the actual bound as matrix is randomly changing over time.

We assume that \mathbf{H} is independent of both \mathbf{x} and \mathbf{n} , and that the specific realization H is known at the receiver (while the transmitter only knows its distribution). Thus, we can equivalently say that the channel yields the couple (\mathbf{y}, \mathbf{H}) .

We will further assume that the entries of \mathbf{H} are i.i.d. zero-mean Complex Gaussian, with independent real and imaginary parts, each with variance $\frac{1}{2}$.

In other words we are considering a Rayleigh fading channel for each component and enough physical separation within transmitting and receiving antennas to achieve independence between the entries of \mathbf{H} .

Capacity-achieving Distribution I

This time, we can compute the mutual information as follows

$$\begin{aligned} I(\mathbf{x}; (\mathbf{y}, \mathbf{H})) &= I(\mathbf{x}; \mathbf{H}) + I(\mathbf{x}; \mathbf{y} | \mathbf{H}) \\ &= I(\mathbf{x}; \mathbf{y} | \mathbf{H}) \\ &= E_{\mathbf{H}}[I(\mathbf{x}; \mathbf{y} | \mathbf{H} = H)] \end{aligned}$$

We know that for a deterministic H , the maximizer \mathbf{x} takes the form of a zero-mean circularly symmetric Complex Gaussian with covariance Q , and $\Psi(Q, H) = \log \det(I_r + H Q H^\dagger)$ is the corresponding maximal mutual information.

We thus need to maximize

$$\Psi(Q) = E[\Psi(Q, \mathbf{H})] = E\left[\log \det\left(I_r + \mathbf{H} Q \mathbf{H}^\dagger\right)\right]$$

over the choice of positive semi-definite Q , such that $\text{tr}(Q) \leq P$.

Capacity-achieving Distribution II

Since Q is positive semi-definite, we can write it as $Q = W\Delta W^\dagger$, where W is unitary and Δ is non-negative diagonal. Thus

$$\Psi(Q) = E\left[\log \det\left(I_r + (\mathbf{H}W)\Delta(\mathbf{H}W)^\dagger\right)\right]$$

It can be shown that $\mathbf{H}W$ has the same distribution as \mathbf{H} (since W is unitary and \mathbf{H} has i.i.d. entries), and thus $\Psi(Q) = \Psi(\Delta)$. It follows that we can focus on positive semi-definite diagonal Q .

Consider the permutation matrix Π (which is orthogonal, thus unitary) and define $Q^\Pi \triangleq \Pi Q \Pi^T$. Again, we note that $\Psi(Q^\Pi) = \Psi(Q)$.

Capacity-achieving Distribution III

Note that the transformation $Q \rightarrow I_r + HQH^\dagger$ is linear and preserves positive definiteness, and know that the $\log \det$ function is concave over the closed cone of positive semi-definite matrices.

Thus $Q \rightarrow \Psi(Q)$ is concave.

Let's define

$$\tilde{Q} = \frac{1}{t!} \sum_{\Pi} Q^{\Pi}$$

Note that $\tilde{Q} = \alpha I_t$ for some α for any Q , that $\text{tr}(\tilde{Q}) = \text{tr}(Q)$ and $\Psi(\tilde{Q}) \geq \Psi(Q)$.

Capacity-achieving Distribution IV

Theorem

The capacity of the channel is achieved when \mathbf{x} is a circularly symmetric Complex Gaussian with zero-mean and covariance $Q = \frac{P}{t} I_t$.

The capacity is given by

$$C(r, t, P) = E \left[\log \det \left(I_r + \frac{P}{t} \mathbf{H} \mathbf{H}^\dagger \right) \right]$$

Note that $Q = \frac{P}{t} I_t$ means that the transmitted symbol \mathbf{x} is i.i.d.

Note also that for fixed r , by the Law of Large Numbers $\frac{1}{t} \mathbf{H} \mathbf{H}^\dagger = I_r$, almost surely as t gets large.

Thus, the capacity in the limit of large t equals

$$r \log(1 + P)$$

which scales linearly with the number of receivers!

Evaluation of the Capacity I

First of all, from the Sylvester's determinant identity we know that

$$\det\left(I_r + \frac{P}{t} \mathbf{H} \mathbf{H}^\dagger\right) = \det\left(I_t + \frac{P}{t} \mathbf{H}^\dagger \mathbf{H}\right)$$

We can then define

$$\mathbf{W} = \begin{cases} \mathbf{H} \mathbf{H}^\dagger & r < t \\ \mathbf{H}^\dagger \mathbf{H} & r \geq t \end{cases}$$

and also $n = \min(r, t)$, $N = \max(r, t)$.

Clearly, $\mathbf{W} \in \mathbb{C}^{n \times n}$ is a random positive semi-definite matrix, thus it has random, real, non-negative eigenvalues $\lambda_1, \dots, \lambda_n$.

Evaluation of the Capacity II

From the literature, we know that the entries of \mathbf{W} follow a *Wishart distribution with parameters* n, N and the joint density of the *ordered* eigenvalues is known to be [2]

$$p_{\lambda, \text{ord}} = \frac{1}{K_{n,N}} \prod_i e^{-\lambda_i} \lambda_i^{N-n} \prod_{j>i} (\lambda_i - \lambda_j)^2, \quad \lambda_1 \geq \dots \geq \lambda_n \geq 0$$

where $K_{n,N}$ is a normalizing factor.

After many complex calculations and transformations we reach...

Evaluation of the Capacity III

Theorem

The capacity of the Gaussian Channel with ergodic matrix \mathbf{H} , with t transmitters and r receivers under power constraint P equals

$$C(r, t, P) = \int_0^\infty \log\left(1 + \frac{P}{t} \lambda\right) \sum_{k=0}^{n-1} \frac{k!}{(k + N - n)!} \left[L_k^{N-n}(\lambda)\right]^2 \lambda^{N-n} e^{-\lambda} d\lambda \quad (2)$$

where $n = \min(r, t)$, $N = \max(r, t)$, and

$$L_k^{N-n}(x) = \frac{1}{k!} e^x x^{n-N} \frac{d^k}{dx^k} \left(e^{-x} x^{N-n+k} \right)$$

are the *associated Laguerre polynomials*.

Example ($t = 1$)

In this case $n = 1$ and $N = r$. Note that $L_0^{r-1}(\lambda) = 1$, thus from Eq. (2) we get

$$C(r, 1, P) = \frac{1}{(r-1)!} \int_0^\infty \log(1 + P\lambda) \lambda^{r-1} e^{-\lambda} d\lambda$$

As r gets large, so does the capacity.

It can be shown that

$$C \xrightarrow{r \rightarrow \infty} \log(1 + rP)$$

Examples II

Example ($t = 1$)

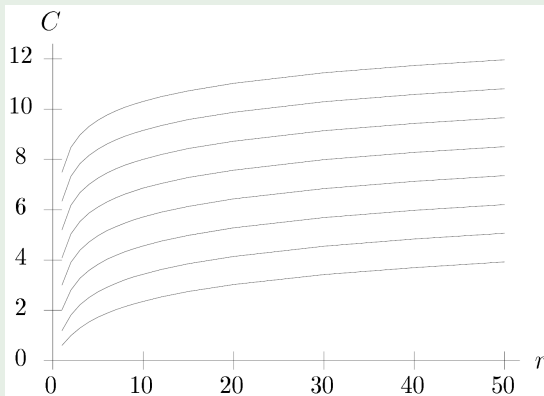


Figure: The value of the capacity (in *nats*) for $0\text{dB} \leq P \leq 35\text{dB}$ in 5dB increments

Examples III

Example ($r = 1$)

In this case $n = 1$ and $N = t$. Differently from before, though, the capacity is

$$C(1, t, P) = \frac{1}{(t-1)!} \int_0^\infty \log\left(1 + \frac{P}{t}\lambda\right) \lambda^{t-1} e^{-\lambda} d\lambda$$

Quite differently from before, the asymptotic capacity is

$$C \xrightarrow{t \rightarrow \infty} \log(1 + P)$$

Thus, just increasing the number of transmitting antennas will not help getting more capacity.

Examples IV

Example ($r = 1$)

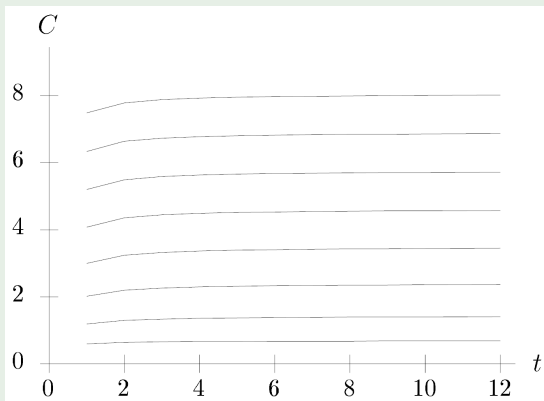


Figure: The value of the capacity (in *nats*) for $0\text{dB} \leq P \leq 35\text{dB}$ in 5dB increments

Example ($r = t$)

In this case $n = N = r = t$. From Eq. (2), then, we get

$$C(n, n, P) = \int_0^\infty \log\left(1 + \frac{P}{n}\lambda\right) \sum_{k=0}^{n-1} L_k(\lambda)^2 e^{-\lambda} d\lambda$$

There actually exists a close form formula for this integral, which is

$$\lim_{n \rightarrow \infty} \frac{C(n, n, P)}{n} = \log P - 1 + \frac{\sqrt{1 + 4P} - 1}{2P} + 2 \tanh^{-1}\left(\frac{1}{\sqrt{1 + 4P}}\right) \quad (3)$$

which clearly means that the capacity is a linear function of the number of transmitting/receiving antennas.

Examples VI

Example ($r = t$)

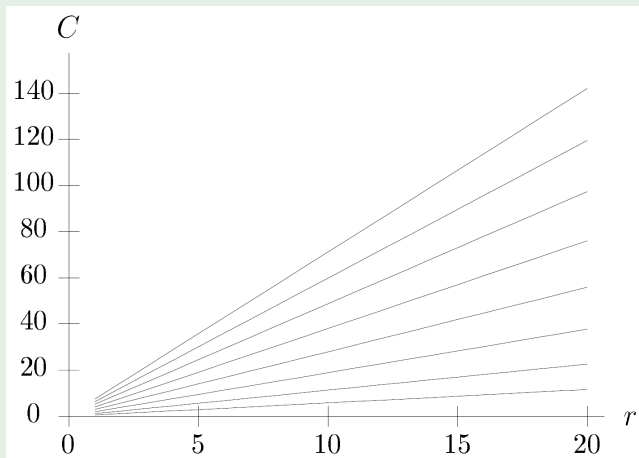


Figure: The value of the capacity (in *nats*) for $0\text{dB} \leq P \leq 35\text{dB}$ in 5dB increments

Table of Contents

- 1 Introduction
 - Abstract
 - Notation and Assumptions
 - Properties and Lemmas
- 2 Gaussian Channel with Deterministic H
 - Capacity Derivation
 - Formal Derivation of Capacity
- 3 Gaussian Channel with Ergodic \mathbf{H}
 - Introduction to the problem
 - Capacity-achieving Distribution
 - Evaluation of the Capacity
 - Examples
- 4 *Bonus*: Multiaccess Channel

Bonus: Multiaccess Channel I

Consider now a number of transmitters, say K , each with t transmitting antennas, and each subject to a power constraint P . There is a single receiver with r antennas.

The received signal \mathbf{y} is given by

$$\mathbf{y} = [\mathbf{H}_1 \dots \mathbf{H}_K] \begin{bmatrix} \mathbf{x}_1 \\ \vdots \\ \mathbf{x}_K \end{bmatrix} + \mathbf{n}$$

where \mathbf{x}_k is the signal sent by the k -th transmitter, \mathbf{n} is independent Gaussian noise, and $\mathbf{H}_k \in \mathbb{C}^{r \times t}$ is the random channel from transmitter k to the receiver.

Bonus: Multiaccess Channel II

We assume, as for the ergodic case, that the receiver knows all the matrices \mathbf{H}_k , and their entries are i.i.d. circularly symmetric Complex Gaussian with zero-mean and unit variance.

Since we know that the optimal single user transmission yields an i.i.d. solution for each antenna, it doesn't matter if the transmitters in the multiuser scenario cannot cooperate.

Thus, a rate vector (R_1, \dots, R_K) will be achievable if

$$\sum_{i=1}^k R_{[i]} \leq C(r, kt, kP) \quad \text{for all } k = 1, \dots, K$$

where $(R_{[1]}, \dots, R_{[K]})$ is the ordering of the rate vector from the largest to the smallest, and $C(a, b, P)$ is the single user a receiver b transmitter capacity under power constraint P .



E. Telatar, “Capacity of multi-antenna gaussian channels,” *European Transactions on Telecommunications*, vol. 10, no. 6, pp. 585–595, 1999.



A. T. James, “Distributions of matrix variates and latent roots derived from normal samples,” *Ann. Math. Statist.*, vol. 35, pp. 475–501, 06 1964.