

7.0 (a) Let  $v(t) = e^{j\omega_0 t}$  be the input to an LTI system with frequency response  $H(\omega)$ . Show that the output is  $y(t) = e^{j\omega_0 t} H(\omega_0)$ .

Proof: Since the system is LTI,

$$y(t) = (v * h)(t)$$

$$= \int_{-\infty}^{\infty} v(t-\lambda) h(\lambda) d\lambda$$

$$= \int_{-\infty}^{\infty} e^{j\omega_0(t-\lambda)} h(\lambda) d\lambda$$

$$= \int_{-\infty}^{\infty} e^{j\omega_0 t} e^{-j\omega_0 \lambda} h(\lambda) d\lambda$$

$$= e^{j\omega_0 t} \int_{-\infty}^{\infty} h(\lambda) e^{-j\omega_0 \lambda} d\lambda$$

$$= e^{j\omega_0 t} H(\omega_0)$$

as required.

(72) we have  $H(\omega) = \frac{1}{j\omega + 1}$

(a) For the input  $v_1(t) = \cos(t)$ ,  $\omega_0 = 1$   
 we have  $A_k^v = \begin{cases} 1, & k=1 \\ 0, & \text{otherwise} \end{cases}$

and  $\theta_k^v = 0$  for all  $k$ .

Let  $y_1$  be the output of the system. Then

$$\begin{aligned} A_k^y &= A_k^v |H(k\omega_0)| \\ \text{So } A_1^y &= |H(\omega_0)| \\ &= |H(1)| \\ &= \left| \frac{1}{j+1} \right| \end{aligned}$$

$$\theta_k^y = \theta_k^v + \angle H(k\omega_0)$$

$$\begin{aligned} \text{So } \theta_1^y &= 0 + \angle H(1) \\ &= 0 + \angle \left( \frac{1}{j+1} \right) \\ &= -\angle(j+1) \\ &= -\pi/4 \end{aligned}$$

$$\begin{aligned} \text{So } y_1(t) &= A_1^y \cos(t + \theta_1^y) \\ &= \frac{1}{\sqrt{2}} \cos\left(t - \frac{\pi}{4}\right). \end{aligned}$$

(b) For  $v_2(t) = \cos\left(t + \frac{\pi}{4}\right)$   
 $= v_1\left(t + \frac{\pi}{4}\right)$

by time-invariance

$$\begin{aligned} y_2(t) &= \text{output from } v_2(t) \\ &= y_1\left(t + \frac{\pi}{4}\right) \\ &= \frac{1}{\sqrt{2}} \cos(t). \end{aligned}$$



(73) An LTI continuous time system has frequency response

$$H(j\omega) = \frac{j\omega}{j\omega + 2}$$

$$(a) \text{ The } |H(j\omega)| = \frac{|j\omega|}{|2 + j\omega|}$$
$$= \frac{|\omega|}{\sqrt{4 + \omega^2}}$$

and

$$\angle H(j\omega) = \angle(j\omega) - \angle(j\omega + 2)$$

Case (1): Assume  $\omega > 0$ . Then

$$\angle j\omega = \frac{\pi}{2}$$

$$\angle(2 + j\omega) = \tan^{-1}\left(\frac{\omega}{2}\right)$$

$$\text{So } \angle H(j\omega) = \frac{\pi}{2} - \tan^{-1}\left(\frac{\omega}{2}\right)$$

Case (2):

$$\text{Then } \angle(j\omega) = -\frac{\pi}{2}$$

Assume  $\omega < 0$ :

$$\angle(2 + j\omega) = \tan^{-1}\left(\frac{\omega}{2}\right)$$

$$\text{So } \angle H(j\omega) = -\frac{\pi}{2} - \tan^{-1}\left(\frac{\omega}{2}\right)$$

$$\text{So } |H(j\omega)| = \begin{cases} \frac{\omega}{\sqrt{4 + \omega^2}}, & \omega > 0 \\ \frac{-\omega}{\sqrt{4 + \omega^2}}, & \omega < 0 \end{cases}$$

$$\angle H(j\omega) = \begin{cases} \frac{\pi}{2} - \tan^{-1}\left(\frac{\omega}{2}\right), & \omega > 0 \\ -\frac{\pi}{2} - \tan^{-1}\left(\frac{\omega}{2}\right), & \omega < 0 \end{cases}$$

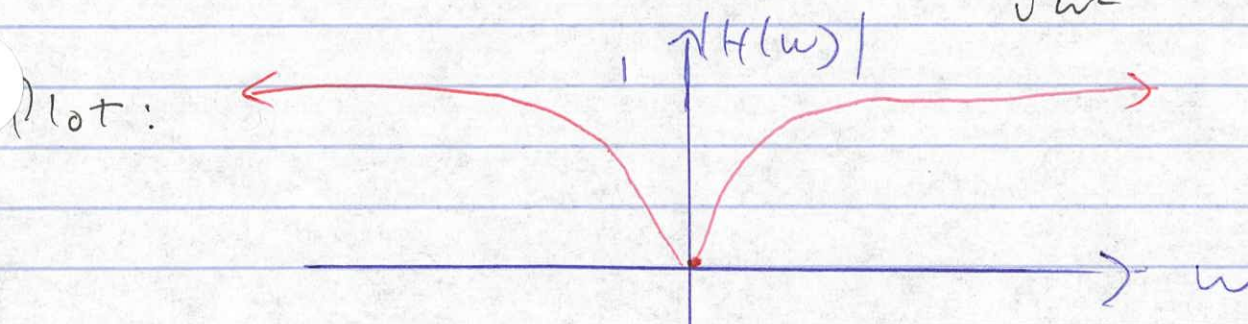


### 73 Plot graphs (without MATLAB)

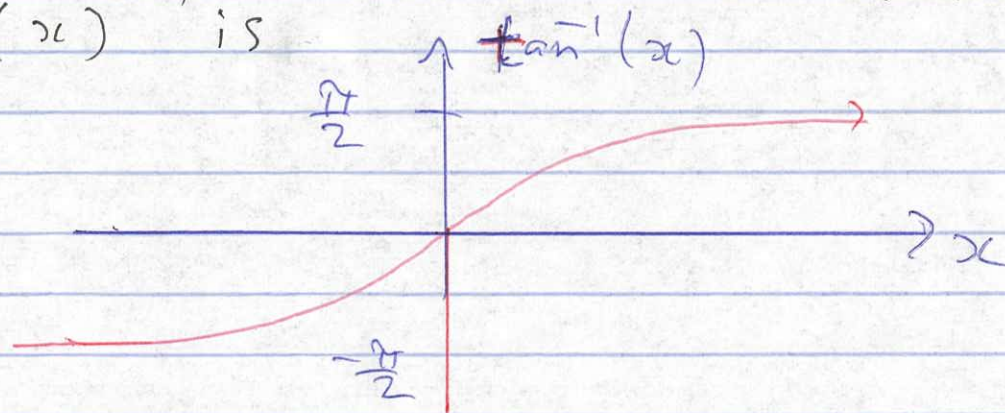
For  $|H(\omega)|$  we have

- (i)  $|H(0)| = 0$
- (ii) even function as  $|\omega|$  is even
- (iii) As  $\omega \rightarrow \infty$ ,  $|H(\omega)| \rightarrow \frac{\omega}{\sqrt{\omega^2}} = 1$

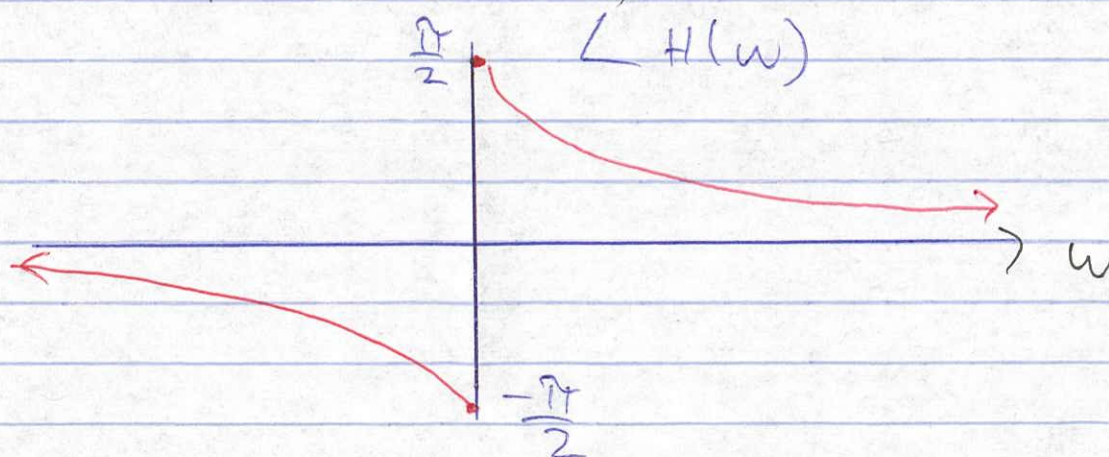
As  $\omega \rightarrow -\infty$ ,  $|H(\omega)| \rightarrow \frac{-\omega}{\sqrt{\omega^2}} = 1$

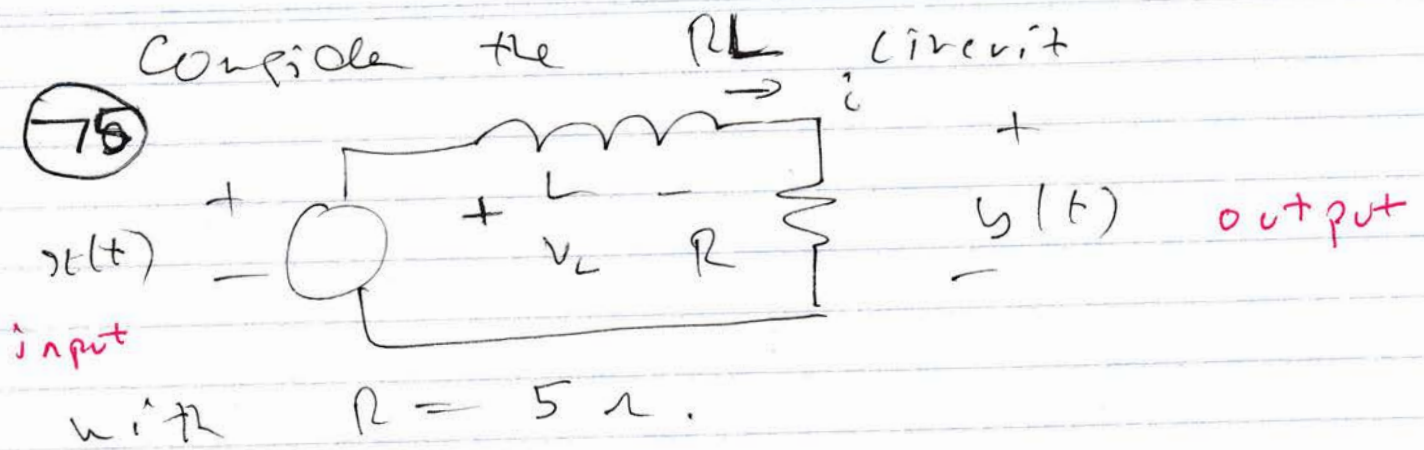


For  $\angle H(\omega)$ , we know that graph of  $\tan^{-1}(x)$  is



So plot of  $\angle H(\omega)$  is





(a) Find the frequency response of the circuit.

from KVL:  $x(t) = L \frac{di}{dt} + Ri(t)$

OHMS LAW  $\Rightarrow y(t) = Ri$   
 $\Rightarrow \frac{dy}{dt} = R \frac{di}{dt}$

So  $x(t) = \frac{L}{R} \frac{dy}{dt} + y(t)$

Assume  $x(t) = e^{j\omega t}$   
 then  $y(t) = H(\omega) e^{j\omega t}$   
 $\frac{dy}{dt} = j\omega H(\omega) e^{j\omega t}$

Substitute:

$e^{j\omega t} = \frac{L}{R} j\omega H(\omega) e^{j\omega t} + H(\omega) e^{j\omega t}$

rearranging gives

$e^{j\omega t} = e^{j\omega t} \left( \frac{L}{R} j\omega + 1 \right) H(\omega)$

$\Rightarrow H(\omega) = \frac{1}{1 + \frac{jL\omega}{R}} = \frac{5}{5 + jL\omega}$

using  $R = 5 \Omega$ .



(75)

Suppose the input is

$$x(t) = 10 |\sin(377t)| \quad \text{for } 0 \leq t < \frac{\pi}{377}$$

and  $x(t) = x(t + \pi/377)$ .

Find the dc term in the complex Fourier series for  $y(t)$ .

Solution : Here  $T = \pi/377$   
and  $\omega_0 = \frac{2\pi}{T} = 754 \text{ rad/s}$

Also

$$C_0^x = \frac{1}{T} \int_0^T |\sin(377t)| dt$$
$$= \frac{20}{\pi}.$$

So the dc (constant) term in the output  $y(t)$  is

$$C_0^y = H(0) C_0^x$$
$$= \left( \frac{5}{5 + 0} \right) \left( \frac{20}{\pi} \right)$$
$$= \frac{20}{\pi}.$$

Note this does not depend on  $L$ .

Consider the 2-point MAU filter

$$(77) \quad y[n] = \frac{1}{2} [v[n] + v[n-1]]$$

Obtain the frequency response and show that it has linear phase for  $0 < \Omega < \pi$ .

Solution: We saw in Lectures that

$$\begin{aligned} (a) \quad H(\Omega) &= \frac{1}{2} \left[ \frac{1 - e^{-j2\Omega}}{1 - e^{-j\Omega}} \right] \\ &= \left( \frac{1}{2} \right) \left[ \frac{(e^{-j\Omega}) \left( \frac{1}{2j} \right) [e^{j\Omega} - e^{-j\Omega}]}{(e^{-j\Omega/2}) \left( \frac{1}{2j} \right) [e^{j\Omega/2} - e^{-j\Omega/2}]} \right] \\ &= \left( \frac{1}{2} \right) \frac{\sin(\Omega)}{\sin(\Omega/2)} e^{-j\Omega/2} \end{aligned}$$

(b) Assume  $0 \leq \Omega \leq \pi$ :

$$\begin{aligned} \text{Then } \angle H(\Omega) &= \angle \frac{\sin(\Omega)}{\sin(\Omega/2)} + \angle e^{-j\Omega/2} \\ &= \angle (a) + \angle e^{-j\Omega/2} \\ &= -\frac{\Omega}{2} \quad \text{for some } a > 0 \end{aligned}$$

Assume  $\pi \leq \Omega < 2\pi$ .

$$\begin{aligned} \text{Then } \angle H(\Omega) &= \angle \frac{\sin(\Omega)}{\sin(\Omega/2)} + \angle e^{-j\Omega/2} \\ &= \angle (-a) + \angle e^{-j\Omega/2} \\ \text{affine, } \rightarrow &= \pi - \frac{\Omega}{2} \quad \text{for some } a > 0 \end{aligned}$$

Thus for  $0 < \Omega < \pi$

$$\angle H(\Omega) = -\Omega t_d, \quad \text{where } t_d = \frac{1}{2}$$

So it has linear phase for these  $\Omega$ .

(80) We know that

$$H(\omega) = \frac{3 \sin(1.5\omega)}{(3 - e^{-j\omega}) \sin(\omega/2)}$$

$$= H_1(\omega) H_2(\omega)$$

where  $H_1(\omega) = \text{FT}(h_1[n])$

$$= \text{FT}\left(\left(\frac{1}{3}\right)^n u[n]\right)$$

$$= \frac{1}{1 - \left(\frac{1}{3}\right)e^{-j\omega}}$$

$$= \frac{3}{3 - e^{-j\omega}}$$

Hence  $H_2(\omega) = \frac{\sin(1.5\omega)}{\sin(\omega/2)}$ .

Recall that

$$p_L[n] \leftrightarrow \frac{\sin\left[\left(q + \frac{1}{2}\right)\omega\right]}{\sin(\omega/2)}$$

We see that here  $q = 1$

so 
$$L = 2q + 1$$
$$= 3$$

Hence  $h_2[n] = p_3[n]$ .