

DSP HW2 Part 2: Digital Parametric Resonators

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1 Introduction

A parametric resonator is a second-order digital filter designed to produce sustained oscillations at a specific frequency when excited by a brief input signal. These filters are used as blocks in digital audio synthesis and can simulate the resonant behavior of physical instruments.

2 Mathematical Derivation

2.1 Time Domain Difference Equation

A second-order digital resonator can be described by the following difference equation:

$$y[n] = b_0x[n] - a_1y[n - 1] - a_2y[n - 2] \quad (1)$$

where:

- $x[n]$ is the input signal
- $y[n]$ is the output signal
- b_0 is the feedforward coefficient
- a_1, a_2 are the feedback coefficients

The difference equation $y[n] = Gx[n] - a_1y[n - 1] - a_2y[n - 2]$ creates oscillations because it implements a **digital memory system with delayed feedback**. When you input a brief pulse, the system “remembers” its recent outputs through the $y[n - 1]$ and $y[n - 2]$ terms, and the carefully chosen coefficients a_1 and a_2 cause these delayed versions to reinforce each other in a periodic pattern that naturally ‘rings’ at the designed frequency.

2.2 Z-Transform and Transfer Function

Taking the z-transform of equation (1):

$$Y(z) = b_0X(z) - a_1z^{-1}Y(z) - a_2z^{-2}Y(z) \quad (2)$$

Rearranging to solve for the transfer function:

$$Y(z)[1 + a_1z^{-1} + a_2z^{-2}] = b_0X(z) \quad (3)$$

$$H(z) = \frac{Y(z)}{X(z)} = \frac{b_0}{1 + a_1z^{-1} + a_2z^{-2}} \quad (4)$$

2.3 Pole Placement for Resonance

To create a resonator at frequency ω_0 with decay factor R , we place complex conjugate poles at:

$$p_{1,2} = Re^{\pm j\omega_0} \quad (5)$$

The denominator polynomial with these poles is:

$$D(z) = (1 - Re^{j\omega_0}z^{-1})(1 - Re^{-j\omega_0}z^{-1}) \quad (6)$$

$$= 1 - R(e^{j\omega_0} + e^{-j\omega_0})z^{-1} + R^2z^{-2} \quad (7)$$

$$= 1 - 2R\cos(\omega_0)z^{-1} + R^2z^{-2} \quad (8)$$

Therefore, the coefficients are:

$$a_1 = -2R\cos(\omega_0) \quad (9)$$

$$a_2 = R^2 \quad (10)$$

2.4 Frequency Domain Analysis

The frequency response is obtained by substituting $z = e^{j\omega}$:

$$H(e^{j\omega}) = \frac{b_0}{1 + a_1e^{-j\omega} + a_2e^{-j2\omega}} \quad (11)$$

Substituting our coefficients:

$$H(e^{j\omega}) = \frac{b_0}{1 - 2R\cos(\omega_0)e^{-j\omega} + R^2e^{-j2\omega}} \quad (12)$$

2.5 Gain Normalization

To ensure unity gain at the resonant frequency ω_0 , we set:

$$|H(e^{j\omega_0})| = 1 \quad (13)$$

The magnitude at resonance is:

$$|H(e^{j\omega_0})| = \frac{b_0}{|1 - 2R\cos(\omega_0)e^{-j\omega_0} + R^2e^{-j2\omega_0}|} \quad (14)$$

After algebraic manipulation, the normalization factor is:

$$b_0 = G = (1 - R)\sqrt{1 - 2R\cos(2\omega_0) + R^2} \quad (15)$$

2.6 Digital Frequency Conversion

For a desired analog frequency f Hz and sampling rate F_s :

$$\omega_0 = \frac{2\pi f}{F_s} \quad (16)$$

3 MATLAB Implementation

3.1 Resonator Coefficient Function

Listing 1: res_coeffs.m - Resonator coefficient generator

```

1 function [a, b] = res_coeffs(freq, R, Fs)
2 % RES_COEFFS Generate coefficients for parametric resonator
3 %
4 % Inputs:
5 %   freq - Desired resonant frequency in Hz
6 %   R     - Pole radius (0 < R < 1), controls decay rate
7 %   Fs    - Sampling frequency in Hz
8 %
9 % Outputs:
10 %   a - Denominator coefficients [1, a1, a2]
11 %   b - Numerator coefficient [G]
12 %
13 % Convert analog frequency to digital frequency
14 omega_0 = (2 * pi * freq) / Fs;
15 %
16 % Calculate denominator coefficients (feedback)
17 a1 = -2 * R * cos(omega_0);
18 a2 = R^2;
19 %
20 % Calculate gain normalization factor
21 G = (1 - R) * sqrt(1 - 2*R*cos(2*omega_0) + R^2);
22 %
23 % Return filter coefficients
24 b = [G];           % Numerator (feedforward)
25 a = [1, a1, a2];   % Denominator (feedback)
26 end

```

4 Example Figures

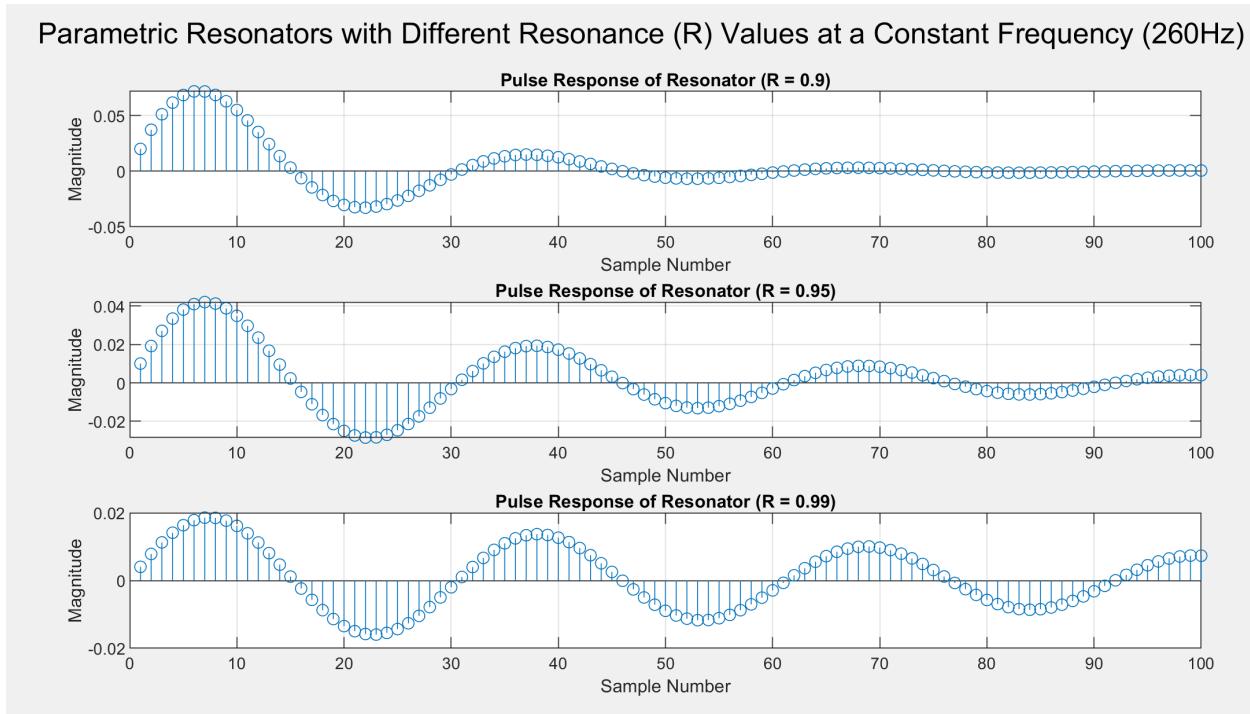


Figure 1: Resonance Sweep in time domain.

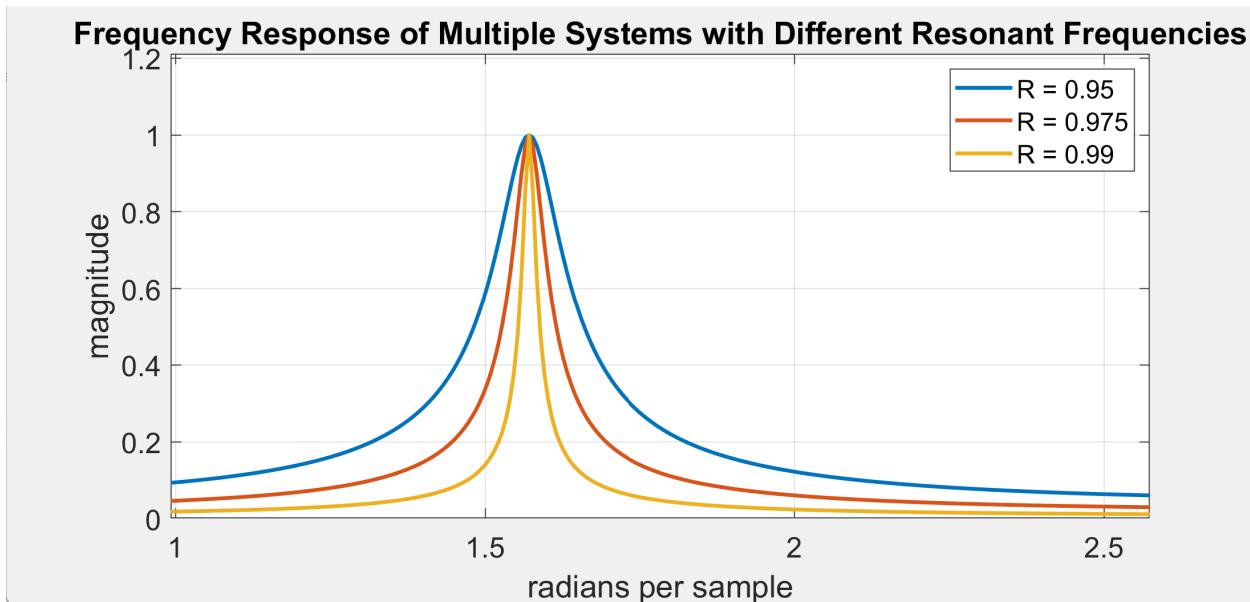


Figure 2: Resonance Sweep in frequency domain.

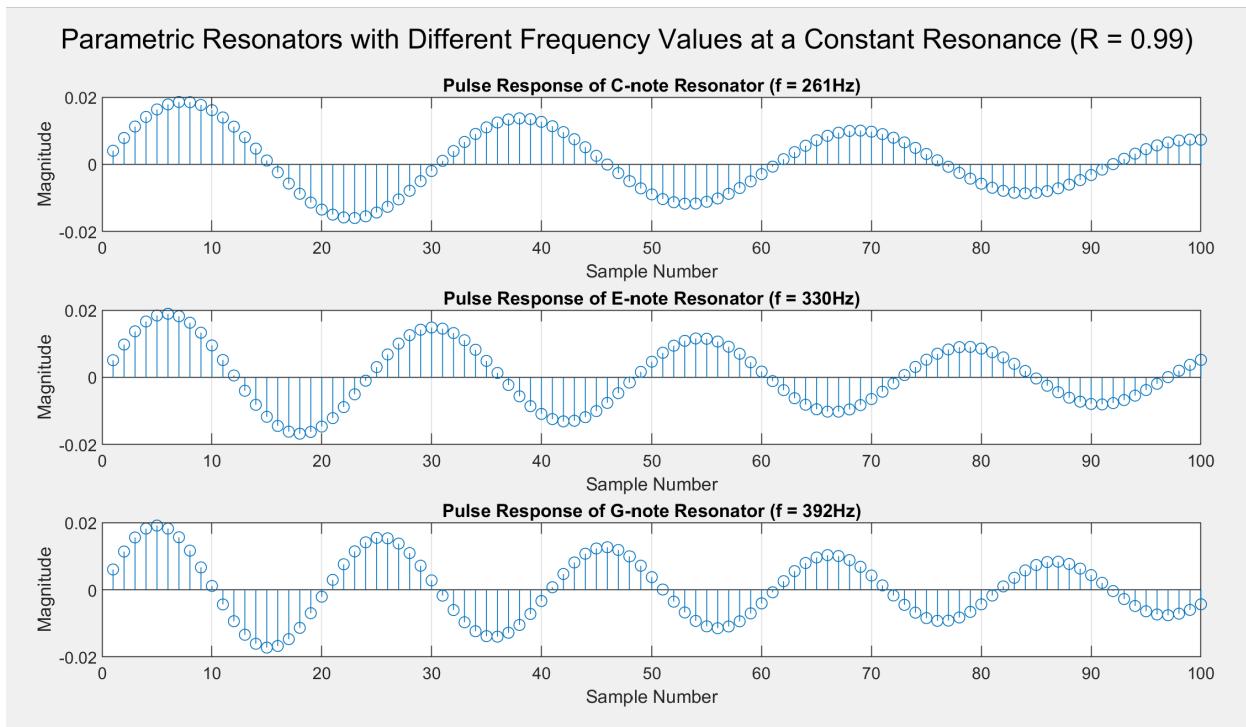


Figure 3: Resonance Frequency Sweep in time domain.

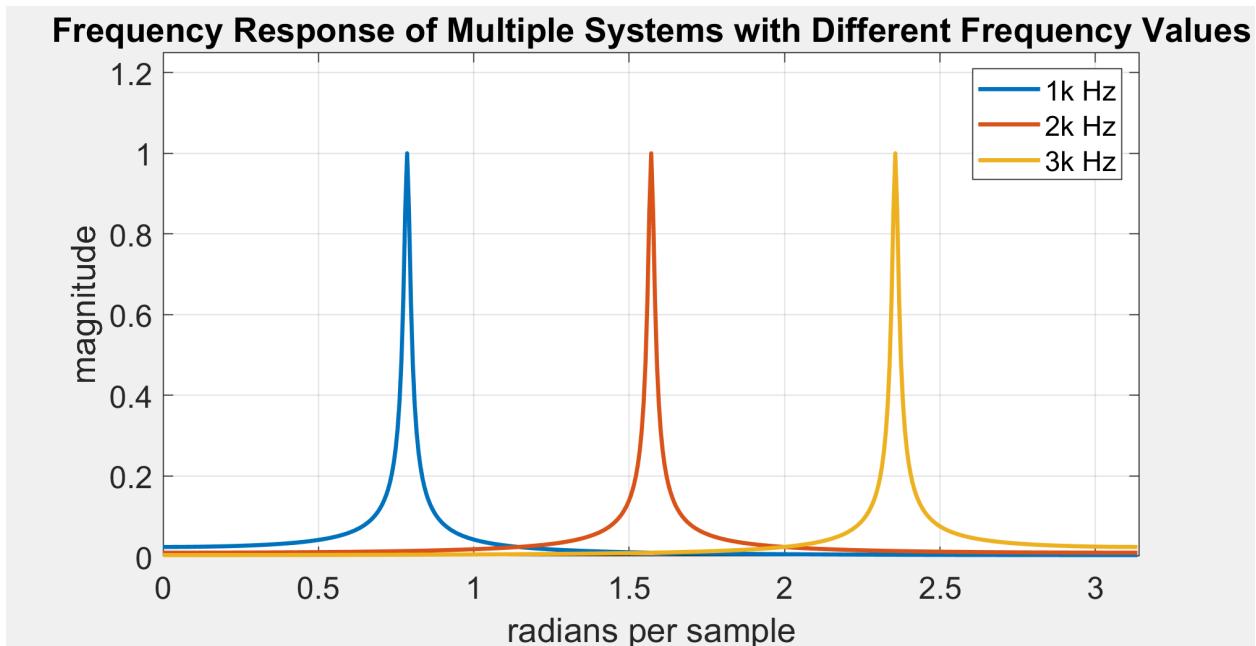


Figure 4: Resonance Frequency Sweep in frequency domain.