

Characteristics of 3 band orthogonal wavelet system

Mayur Nawal

Communication and Signal Processing

Guide : Prof. V. M. Gadre

Department of Electrical Engineering

IIT Bombay

June 4, 2013

- Introduction
- Basic concepts
- 3 band filter bank
- Parameterization of 3 band orthogonal wavelet system
- Generating high tfp localized wavelets
- Sobolev smoothness of scaling function
- New method to calculate the smoothness of scaling functions
- Conclusion and future work
- References

- In the past 2 decades, there has been an indepth study of the 2 band wavelet system and its associated properties. We continue the research by extending the ideas to 3-band wavelet system and its properties.

Basic concepts

Polyphase decomposition

- Polyphase representation permits a great level of simplification of theoretical results for both FIR and IIR filters
- There are 2 ways in which we can represent a filter in its polyphase form, they are as follows:

$$\text{Type : 1, } H(z) = \sum_{l=0}^{M-1} z^{-l} E_l(z^M) \quad (1)$$

$$\text{where, } E_l(z^M) = \sum_{n=-\infty}^{\infty} h(Mn + l) z^{-n} \quad (2)$$

$$\text{Type : 2, } H(z) = \sum_{l=0}^{M-1} z^{(-M-1-l)} R_l(z^M) \quad (3)$$

$$\text{where, } R_l(z^M) = \sum_{n=-\infty}^{\infty} h(M(n + l) - (l + 1)) z^{-l} \quad (4)$$

Basic concepts

Paraunitary Perfect Reconstruction Filter Banks

- The filter bank with FIR filters satisfies P.R property then the following(Eqn. 5) condition on the polyphase component of analysis filter must hold true:

$$|E(z)| = z^{-n} \quad (5)$$

- A transfer matrix is said to be paraunitary if :

$$\tilde{E}(z)E(z) = cI, \quad \text{for all } z \quad (6)$$

$$\tilde{E}(z) = E_*^T(z^{-1}) \quad (7)$$

- Paraunitary property automatically satisfies Eqn. 5, hence we can generate P.R filter bank using transfer functions that satisfies paraunitarity

3 band filter bank

Paraunitary Perfect Reconstruction Filter Banks

- A general structure of how 3 channel filter bank works:

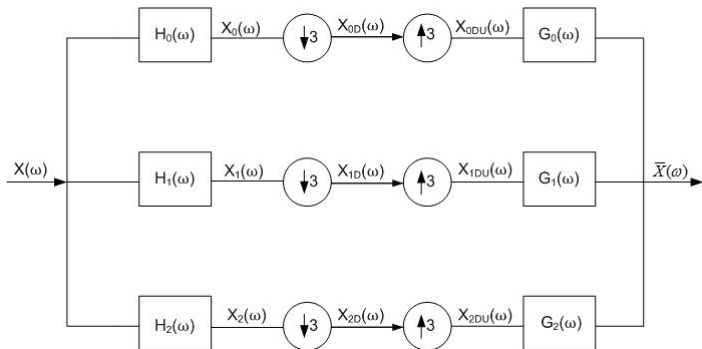


Figure : 3 channel filter bank

3 band filter bank

Generating scaling and wavelet functions via cascade algorithm

- $a^0(n)$, $a^1(n)$ and $a^2(n)$ are the filter coefficients of analysis filters
- Dilation equation in terms of the scaling function and filter coefficients is as follows:

$$\phi(t) = \sum_{n=0}^{N-1} a^0(n) * \phi(3t - 1) \quad (8)$$

$$\psi_1(t) = \sum_{n=0}^{N-1} a^1(n) * \phi(3t - 1) \quad (9)$$

$$\psi_2(t) = \sum_{n=0}^{N-1} a^2(n) * \phi(3t - 1) \quad (10)$$

3 band filter bank

Generating scaling and wavelet functions via cascade algorithm

- The above equations can be equivalently explained in the following 3 ways:
- In time domain:

$$\phi(t) = \lim_{k \rightarrow \infty} a^0(3t) \otimes a^0(3^2t) \dots \otimes a^0(3^k t) \quad (11)$$

- In Frequency domain:

$$\hat{\phi}(\Omega) = \lim_{k \rightarrow \infty} \hat{\phi}(0) \prod_{m=1}^k \left(\frac{1}{3}\right) * A^0\left(\frac{\Omega}{3^m}\right) \quad (12)$$

3 band filter bank

Generating scaling and wavelet functions via cascade algorithm

- It can be also understood in the following way:

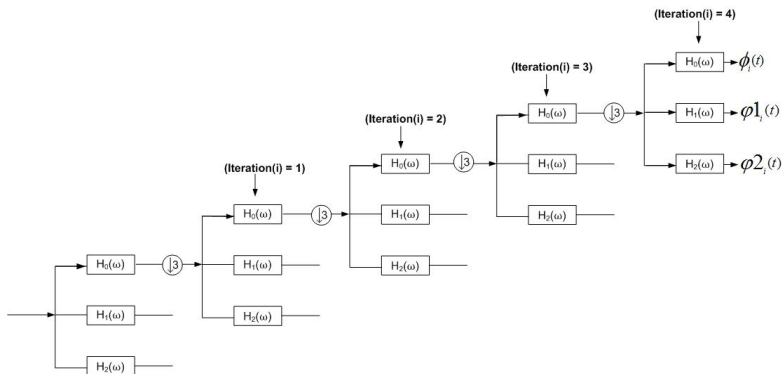


Figure : Diagrammatic view of how the cascade algorithm works

3 band filter bank

variation of scaling and wavelet function for different iterations

- Following figure shows how scaling function varies with different iterations

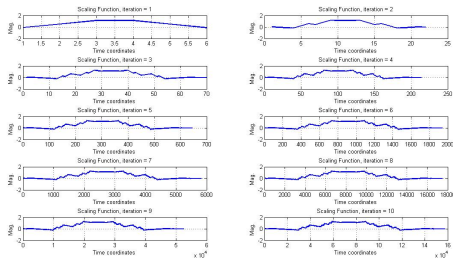


Figure : Evolution of scaling function under different levels of iteration

- As evident from the iterations, the scaling function don't change after 8 iterations

3 band filter bank

Parameterizing 3 band orthogonal wavelet system

- $P[a](z)$ is called 3 – band orthogonal wavelet matrix if it satisfies:

$$P[a](z) * P[a](z^{-1}) = 3I \quad (13)$$

$$\sum_{k=0}^2 p[a]_k^s(1) = \delta_{s,0} \quad (14)$$

- 1st (Eqn. 13) equation is the similarity between P.R and Orthogonal filter banks, 2nd equation (14) is the essential difference between them
- 2nd equation(Eqn. (14)) ensures that the generated wavelet and scaling functions will converge in L_2 norm

3 band filter bank

Parameterizing 3 band orthogonal wavelet system

- Peng L. and Wang Y. in [9] gave a parameterization of subset of 3 – *band* orthogonal wavelet systems in which scaling filter has length $3 * g$ and the McMillan degree is $g - 1$. Following is the algebraic structure of the system:

$$P[a](z) = V_{g-1} V_{g-2} V_{g-3} \dots V_1 H_0 \quad (15)$$

$$V_k = [I - (1 - z^{-1}) u_k u_k^T] \quad (16)$$

where u_k is any 3×1 unit norm vector and H_0 is a real Haar Wavelet Matrix

Results: Characteristics of Filter Banks

Method to generate highly localized wavelets

- In theory, Time Bandwidth Product (T.B.P) is calculated both from frequency and time domain. Equations governing them are as follows:
- In time domain:

$$\sigma_t^2 = \frac{\int_{-\infty}^{\infty} t^2 |x(t)|^2 dt}{\int_{-\infty}^{\infty} |x(t)|^2 dt} \quad (17)$$

$$\sigma_{\Omega}^2 = \frac{\int_{-\infty}^{\infty} |dx(t)/dt|^2 dt}{\int_{-\infty}^{\infty} |x(t)|^2 dt} \quad (18)$$

$$TBP = \frac{\int_{-\infty}^{\infty} t^2 |x(t)|^2 dt}{\int_{-\infty}^{\infty} |x(t)|^2 dt} * \frac{\int_{-\infty}^{\infty} |dx(t)/dt|^2 dt}{\int_{-\infty}^{\infty} |x(t)|^2 dt} \quad (19)$$

Results: Characteristics of Filter Banks

Method to generate highly localized wavelets

- In Frequency Domain:

$$\sigma_t^2 = \frac{\int_{-\infty}^{\infty} \left| \frac{d\hat{x}(\Omega)}{d\Omega} \right| d\Omega}{\int_{-\infty}^{\infty} 2 * \pi * |\hat{x}(\Omega)|^2 d\Omega} \quad (20)$$

$$\sigma_{\Omega}^2 = \frac{\int_{-\infty}^{\infty} \Omega^2 |\hat{x}(\Omega)|^2 d\Omega}{\int_{-\infty}^{\infty} |\hat{x}(\Omega)|^2 d\Omega} \quad (21)$$

$$TBP = \frac{\int_{-\infty}^{\infty} \left| \frac{d\hat{x}(\Omega)}{d\Omega} \right| d\Omega}{\int_{-\infty}^{\infty} 2 * \pi * |\hat{x}(\Omega)|^2 d\Omega} * \frac{\int_{-\infty}^{\infty} \Omega^2 |\hat{x}(\Omega)|^2 d\Omega}{\int_{-\infty}^{\infty} |\hat{x}(\Omega)|^2 d\Omega} \quad (22)$$

Results: Characteristics of Filter Banks

Method to generate highly localized wavelets

- Method to calculate the Time Bandwidth Product for discrete signals:
 - 1 Calculate the FFT of the normalized $f(n)$
 - 2 The number of iterations used in cascade algorithm, provides the sampling distance in discrete time.
 - 3 The integrals calculated using summations
 - 4 In frequency domain we replace the variable Ω by an equivalent variable proportional to the FFT sequence index

Results: Characteristics of Filter Banks

Understanding the TFP behavior for various input parameters

- The TFP calculations for length 6 filters involve 3 variables $var1$, $var2$, Θ .
- Variation in Θ does not change the value of TFP.
- Hence, we create a grid of the first 2 variables $var1$, $var2$,
- We calculate the TBP of the scaling and wavelet functions on all these points of the grid

Results: Characteristics of Filter Banks

Observations from the above algorithm

- The variation of the TBP values of the scaling and wavelet functions are similar

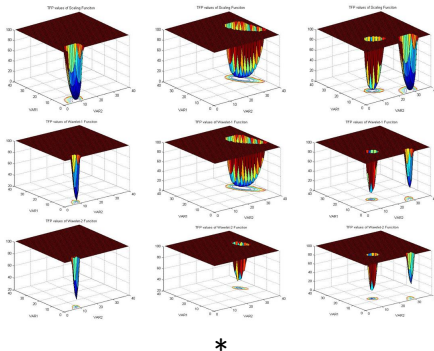


Figure : Scaling and Wavelet Function TBP (Row-1 Corresponds to Scaling functions, Row 2-3 corresponds to wavelet function, column 1,2 and 3 corresponds to grid-1,2 and 5 respectively

Results: Characteristics of Filter Banks

Observations from the above algorithm

- The variation of the TBP values of the scaling and wavelet functions are similar

Correlation & Count → Function ↓	Count of Low TFP points	Correlation with Scaling function	Correlation with Wavelet-1 function	Correlation with Wavelet-2 function
Scaling function	4489	4489	1293	964
Wavelet-1 function	1293	1293	1293	964
Wavelet-2 function	964	964	964	964

Figure : TBP variation for Scaling and both Wavelet Functions

Results: Characteristics of Filter Banks

Observations from the above algorithm

- The variation of the TBP values of the scaling and wavelet functions repeat itself at regular intervals

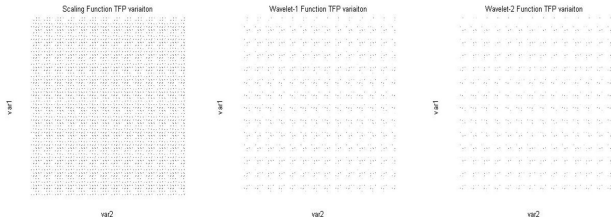


Figure : Similarity in TFP variation for both Scaling and Wavelet Functions

Results: Characteristics of Filter Banks

calculating the input parameters that give highly localized wavelets

- Create a cost function that calculates the sum of TFP values of the 2 wavelets
- Further divided the grid into 9 sub sections
- For each subsection we make an even finer grid of $var1$, $var2$
- At all points of this grid we calculated the TFP values and selected a set (S_1) of points that give us low TFP values
- Optimize the TBP value using the above points as input variables
- Compare them and calculate the global minima

Results: Characteristics of Filter Banks

Results for highly optimized wavelets

- Following are the results of TBP values for 5 grids out of the 9 grids that we had constructed

	var1	var2	Theta	Scaling TFP	Wavelet-1 TFP	Wavelet-2 TFP	Sum TFP Wav-1 & Wav-2
GRID-1	-3.4736	-2.4712	0.5771	2.3981	12.3171	13.1788	25.4959
	-2.6908	-3.0832	1.1692	2.4050	13.2829	17.8599	31.1428
	-2.6908	-3.0832	1.1692	2.4050	13.2829	17.8599	31.1428
	-2.6908	-3.0832	1.1692	2.4050	13.2829	17.8599	31.1428
	-3.4736	-2.4712	0.5771	2.3981	12.3171	13.1788	25.4959
GRID-2	-0.4508	0.0584	1.1692	2.4050	13.2829	17.8599	31.1428
	-0.4508	0.0584	1.1692	2.4050	13.2829	17.8599	31.1428
	-0.4508	0.0584	1.1692	2.4050	13.2829	17.8599	31.1428
	-0.4508	0.0584	1.1692	2.4050	13.2829	17.8599	31.1428
	-3.5924	0.0584	1.1692	2.4050	13.2829	17.8599	31.1428
GRID-5	0.3320	0.6704	0.5771	2.3981	12.3171	13.1788	25.4959
	-0.4508	0.0584	-0.4016	2.4050	17.8599	13.1788	31.1428
	-0.4508	0.0584	-0.4016	2.4050	17.8599	13.1788	31.1428
	-0.4508	0.0584	-0.4016	2.4050	17.8599	13.1788	31.1428
	-0.4508	0.0584	-1.9724	2.4050	13.2829	17.8599	31.1428
GRID-7	-0.4508	-3.0832	1.1692	2.4050	13.2829	17.8599	31.1428
	0.4508	0.0584	1.1692	2.4050	13.2829	17.8599	31.1428
	-0.4508	0.0584	1.1692	2.4050	13.2829	17.8599	31.1428
	0.4508	-3.0832	1.1692	2.4050	13.2829	17.8599	31.1428
	0.4508	-3.0832	1.1692	2.4050	13.2829	17.8599	31.1428
GRID-9	0.3320	0.6704	0.5771	2.3981	12.3171	13.1788	25.4959
	0.3320	0.6704	0.5771	2.3981	12.3171	13.1788	25.4959
	0.3320	0.6704	0.5771	2.3981	12.3171	13.1788	25.4959
	0.3320	0.6704	0.5771	2.3981	12.3171	13.1788	25.4959
	0.3320	0.6704	0.5771	2.3981	12.3171	13.1788	25.4959

Results: Characteristics of Filter Banks

Generating smooth wavelets with regularity = 1

- Wavelet functions will be a sum of the linear shifted scaling functions
- We will study only the smoothness of scaling functions and generalize the results
- Smoothness of a function is generally related with its regularity
- The parameterization imposes a regularity of order - 1 on all the scaling functions
- We will hence calculate the value of sobolev smoothness for all the filters

Results: Characteristics of Filter Banks

Calculating sobolev smoothness of a filter

- Sobolev smoothness of the scaling filter tells us about the smoothness of the scaling function which is generated after cascade iterations. Calculated in the following way:
- Normalize $Q(z)$, such that $Q(1) = 1$
- Calculate the transition operator (T^Q) of $Q(z)$
- The transition operator(T^Q) of $h(n)$ captures how the cascade algorithm converges and hence the stability of scaling function under iterations
- Sobolev smoothness (S_{max}) is dependent on the spectral radius of T^Q
- Smaller the spectral radius, better is the Sobolev smoothness

Results: Characteristics of Filter Banks

New method to calculate the smoothness of scaling function

- General unwanted characteristics of the scaling functions:
 - 1 A lot of small peaks
 - 2 Surface very rough, slope of the function is small but changes quickly
 - 3 Single/multiple points where function changes its values very quickly

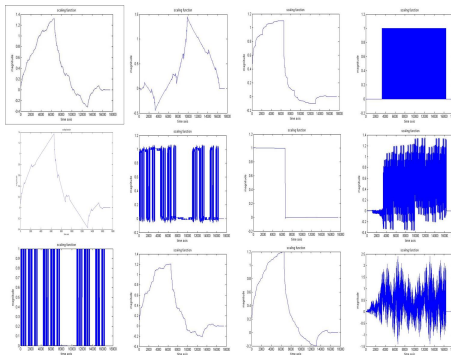


Figure : General Figures of the scaling filters

Results: Characteristics of Filter Banks

Parameterizing smoothness of scaling functions

- *cost1* is first parameter

$$cost1 = \sum_{n=1}^N |f''(n)|, \text{ where } f''(n) \text{ is the double derivative} \quad (23)$$

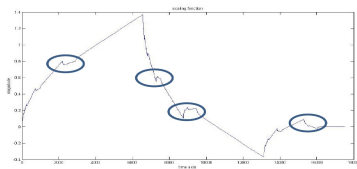


Figure : Most smooth plot when compared in the metric of *cost1*

- The circled regions are the point of concern, which we will refer to as small peaks. They don't get compensated by this cost function

Results: Characteristics of Filter Banks

Parameterizing smoothness of scaling functions

- $cost2$ is the second parameter

$$cost2 = \sum_{n=1}^{Threshold} |sf''(n)|, \text{ where } Threshold \text{ is carefully selected} \quad (24)$$

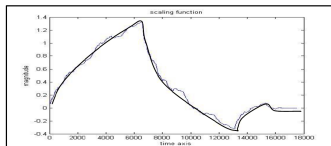


Figure : Most smooth plot when compared in the metric of $cost2$

- Threshold selected from (.025 to .50 of the total length), Threshold values subjectively analyzed
- Curve is very rough, slope not changing drastically but changing at almost all points hence leading to this roughness.

Results: Characteristics of Filter Banks

Parameterizing smoothness of scaling functions

- *cost3* is the second parameter

$$\text{cost3} = \text{count of small peaks} \quad (25)$$

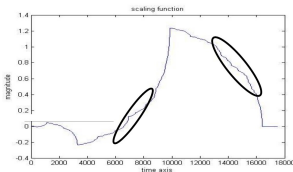


Figure : Most smooth plot when compared in the metric of *cost2*

- The function is not as rough as the previous one, highlighted some portion in the curve which are still rough.

Results: Characteristics of Filter Banks

Parameterizing smoothness of scaling functions

- Cost functions have their own advantages and disadvantages
- Need to have a function with appropriately weighted sum of all these 3 cost functions

$$\text{costfunction} = w1*\text{norm}(\text{cost1})+w2*\text{norm}(\text{cost2})+w3*\text{norm}(\text{cost3}) \quad (26)$$

where, $\text{norm}(\text{cost1}/2/3)$ are the normalized cost functions

- Aim is to decide what will be these weights $w1, w2, w3$.

Results: Characteristics of Filter Banks

Parameterizing smoothness of scaling functions

- To calculate the weights, I have adopted the following method:
- calculate the cost functions for the grid, and normalize them respectively
- With the constraint $w_1 + w_2 + w_3 = 1$, calculate the *costfunction* for the grid
- For a particular set of w_1, w_2, w_3 , calculate the input which gives the least value of cost function
- Check the rank in the sorted $cost_1, cost_2, cost_3$
- Weight corresponding to least sum of ranks will give us the smoothest functions.

Results: Characteristics of Filter Banks

Results for weights for different values of Theta

- I have applied the above method for $\Theta = \pi/2, \pi/3, \pi/4, \pi/6$ and $\pi/9$

Table : Comparison of Weights for various Theta

$\Theta \downarrow$ Weights \rightarrow	W1	W2	W2
$\frac{\pi}{2}$	0.09	0.90	0.09
$\frac{\pi}{3}$	0.09	0.90	0.09
$\frac{\pi}{4}$	0.09	0.90	0.09
$\frac{\pi}{6}$	0.09	0.90	0.09
$\frac{\pi}{9}$	0.09	0.90	0.09

- Value of Smoothness does not change with Theta

Results: Characteristics of Filter Banks

Parameterizing smoothness of scaling functions

- Using $(0.09, 0.90, 0.01)$ as the weights in my function
- Comparison of smoothness variation for my function and sobolev's

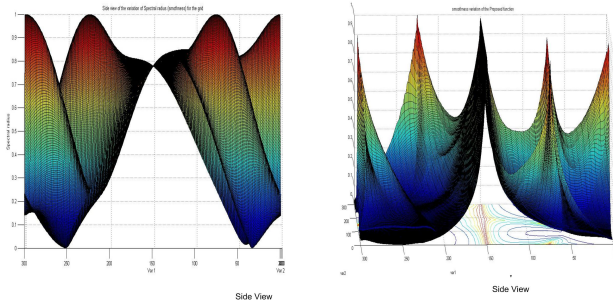


Figure : Comparison of variation of smoothness over grid for sobolev (left) and my function (right)

- The similarity of minima and maxima matches in both

Results: Characteristics of Filter Banks

Parameterizing smoothness of scaling functions

- comparison of the most smooth plots obtained from my method and sobolev's regularity

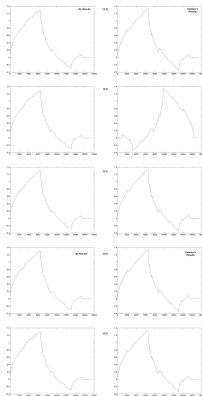


Figure : Comparision of smoothness of scaling functions for sobolev (right) and my function (left)

Results: Characteristics of Filter Banks

Parameterizing smoothness of scaling functions

- comparison of the most smooth plots obtained from my method and sobolev's regularity

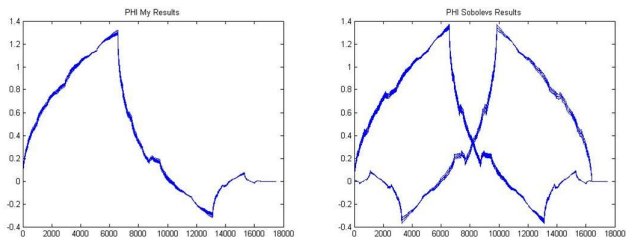


Figure : Comparision of smoothness of scaling functions for sobolev (right) and my function (left)

- In the first section of our investigation we have developed an algorithm to get the wavelets that are highly localized in time and frequency domain
- In the second section of our investigation we have developed a novel method to generate smooth scaling functions
- At a preliminary stage we have compared our results with the Sobolev's measure of regularity and we have found our results to be better than those obtained from his measure.

- For the algorithm giving TFP localized wavelets we need to implement a few more wavelet system designs and verify our results for all of them.
- For the Smoothness criteria we can expand our research on generating the smoothness for filters of varied lengths and compare our results.

References I

- [1] P.P. Vaidyanathan, *Multirate Systems and Filter Banks*. New Delhi, India: Dorling Kindersley(India) Pvt. Ltd., 2006
- [2] Ying-Jui Chen, Soontorn Oraintara, K.S Amaratunga, "Dyadic-based factorizations for regular paraunitary filterbanks and M-band orthogonal wavelets with structural vanishing moments," *Signal Processing, IEEE Transactions on* , vol.53, no.1, pp.193-207, Jan. 2005
- [3] P. Steffen, P. N. Heller, R. A. Gopinath, and C. S. Burrus, "Theory of regular M-band wavelet bases," *Signal Process., IEEE Trans.*, vol. 41, no. 12, pp. 3497 - 3510, Dec. 1993
- [4] R.A.Gopinath, C.S Burrus, "Factorization approach to unitary time-varying filter bank trees and wavelets," *Signal Processing, IEEE Transactions on* , vol.43, no.3, pp.666-680, Mar 1995

References II

- [5] Andre Tkacenko, P.P. Vaidyanathan, T.Q Nguyen, "On the eigenfilter design method and its applications: a tutorial," *Circuits and Systems II: Analog and Digital Signal Processing, IEEE Transactions on* , vol.50, no.9, pp.497 - 517, Sept. 2003
- [6] Zou Hehong, A.H Tewfik, "Discrete orthogonal M-band wavelet decompositions," *Acoustics, Speech, and Signal Processing, IEEE International Conference on* , vol.4, no., pp.605-608, 23-26 Mar 1992
- [7] Hui Xie, Joel M. Morris, "Design of orthonormal wavelets with better time-frequency resolution", *Proc. SPIE*, vol. 2242, no., pp. 878-887, 15 March, 1994
- [8] Qingtang Jiang, "Orthogonal multiwavelets with optimum time-frequency resolution," *Signal Processing, IEEE Transactions on*, vol.46, no.4, pp.830-844, Apr 1998

References III

- [9] Peng Lizhong, Wang Yongge, "Parameterization and algebraic structure of 3-band orthogonal wavelet systems," *Science in China (Series A)*, vol. 44, no. 12, pp. 1531-1543, Dec. 2001
- [10] Ritesh Kolte, Pushkar Patwardhan, Vikram M Gadre, "A class of time-frequency product optimized biorthogonal wavelet filter banks," *National Conference on Communications (NCC)*, vol., no., pp.1-5, 29-31 Jan. 2010
- [11] Damiana Lazzaro, "Biorthogonal M-band filter construction using the lifting scheme," *Numeric Algorithms*, vol. 22, no.1, pp. 53-72, June 2006
- [12] Ning Bi, Xinrong Dai, Qiyu Sun, "Construction of compactly supported M-Band Wavelets," *Applied and Computational Harmonic Analysis*, vol. , no., pp. 113-131, June 1995

References IV

- [13] Tony Lin, Shufang Xu, Qingyun Shi, Pengwei Hao, An algebraic construction of orthonormal M-band wavelets with perfect reconstruction, *Applied Mathematics and Computation*, Volume 172, Issue 2, pp. 717-730, 15 January 2006
- [14] M. Vetterli, C. Herley, "Wavelets and filter banks: theory and design," *Signal Processing, IEEE Transactions on*, vol.40, no.9, pp.2207-2232, Sep 1992
- [15] Martin Vetterli, Jelena Kovačević, *Wavelets and Subband Coding*. Engelwood Cliffs, New Jersey: Prentice Hall of India Pvt. Limited., 1995
- [16] M Smith, T.P Barnwell, "A procedure for designing exact reconstruction filter banks for tree-structured subband coders," *Acoustics, Speech, and Signal Processing, IEEE International Conference*, Vol.9, pp. 421-424, Mar. 1984

References V

- [17] P.P Vaidyanathan, P.Q. Hoang, "Lattice structures for optimal design and robust implementation of two-channel perfect-reconstruction QMF banks," *Acoustics, Speech and Signal Processing, IEEE Transactions on* , vol.36, no.1, pp.81,94, Jan 1988
- [18] Oktay Alkin, Hakan Caglar, "Design of efficient M-band coders with linear-phase and perfect-reconstruction properties," *Signal Processing, IEEE Transactions on*, vol.43, no.7, pp.1579-1590, Jul 1995
- [19] Anand K. Soman, P.P. Vaidyanathan, Truong Q. Nguyen, "Linear Phase Paraunitary Filter Banks: Theory, Factorizations and Designs," *Signal Processing, IEEE Transactions on* , vol. 41, no. 12, pp. 3480-3496, Dec. 1993
- [20] G. Pau, B. Pesquet-Popescu, G. Piella, "Modified M-band synthesis filter bank for fractional scalability of images," *Signal Processing Letters, IEEE*, vol.13, no.6, pp.345-348, June 2006

- [21] R.A. Gopinath, C.S. Burrus, "On upsampling, downsampling and rational sampling rate filter banks," *Signal Processing, IEEE Transactions on*, vol.42, no.4, pp.812 -824, Apr 1994
- [22] J. Kovacevic, M. Vetterli, "Perfect reconstruction filter banks with rational sampling factors," *Signal Processing, IEEE Transactions on*, vol.41, no.6, pp.2047-2066, June 1993