

## Features

### • Radio

- Frequency range: 150 MHz to 960 MHz
- Modulation: LoRa<sup>®</sup>, (G)FSK, (G)MSK and BPSK
- RX sensitivity: –125 dBm for 2-FSK (at 1.2 Kbit/s), –148 dBm for LoRa<sup>®</sup> (at 10.4 kHz, spreading factor 12)
- Transmitter high output power, programmable up to +22 dBm
- Transmitter low output power, programmable up to +15 dBm
- Compliant with the following radio frequency regulations: ETSI EN 300 220, EN 300 113, EN 301 166, FCC CFR 47 Part 15, 24, 90, 101 and the Japanese ARIB STD-T30, T-67, T-108

### • Ultra-low-power platform

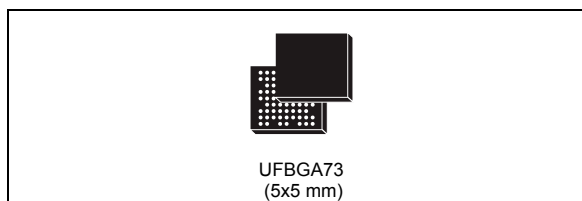
- 1.8 V to 3.6 V power supply
- –40 °C to +85 °C temperature range
- Shutdown mode: 31 nA ( $V_{DD} = 3$  V)
- Standby (+ RTC) mode: 360 nA ( $V_{DD} = 3$  V)
- Stop2 (+ RTC) mode: 1.07  $\mu$ A ( $V_{DD} = 3$  V)
- Active-mode MCU: < 72  $\mu$ A/MHz (CoreMark<sup>®</sup>)
- Active-mode RX: 4.82 mA
- Active-mode TX: 15 mA at 10 dBm and 87 mA at 20 dBm (LoRa<sup>®</sup> 125 kHz)

### • Core: 32-bit Arm<sup>®</sup>Cortex<sup>®</sup>-M4 CPU

- Adaptive real-time accelerator (ART Accelerator) allowing 0-wait-state execution from Flash memory, frequency up to 48 MHz, MPU and DSP instructions
- 1.25 DMIPS/MHz (Dhrystone 2.1)

### • Security and identification

- Hardware encryption AES 256-bit
- True random number generator (RNG)



- Sector protection against read/write operations (PCROP, RDP, WRP)
- CRC calculation unit
- Unique device identifier (64-bit UID compliant with IEEE 802-2001 standard)
- 96-bit unique die identifier
- Hardware public key accelerator (PKA)

### • Supply and reset management

- High-efficiency embedded SMPS step down converter
- SMPS to LDO smart switch
- Ultra-safe, low-power BOR (brownout reset) with 5 selectable thresholds
- Ultra-low-power POR/PDR
- Programmable voltage detector (PVD)
- $V_{BAT}$  mode with RTC and 20x32-byte backup registers

### • Clock sources

- 32 MHz crystal oscillator
- TCXO support: programmable supply voltage
- 32 kHz oscillator for RTC with calibration
- High-speed internal 16 MHz factory trimmed RC ( $\pm 1\%$ )
- Internal low-power 32 kHz RC
- Internal multi-speed low-power 100 kHz to 48 MHz RC
- PLL for CPU, ADC and audio clocks

### • Memories

- Up to 256-Kbyte Flash memory
- Up to 64-Kbyte RAM
- 20x32-bit backup register

- Bootloader supporting USART and SPI interfaces
- OTA (over-the-air) firmware update capable
- Sector protection against read/write operations
- **Rich analog peripherals** (down to 1.62 V)
  - 12-bit ADC 2.5 Msps, up to 16-bit with hardware oversampling, conversion range up to 3.6 V
  - 12-bit DAC, low-power sample-and-hold
  - 2x ultra-low-power comparators
- **System peripherals**
  - Semaphores for processor firmware process synchronization
- **Controllers**
  - 2x DMA controller (7 channels each) supporting ADC, DAC, SPI, I2C, LPUART, USART, AES and timers
  - 2x USART (ISO 7816, IrDA, SPI)
- 1x LPUART (low-power)
- 2x SPI 16 Mbit/s (1 over 2 supporting I2S)
- 3x I2C (SMBus/PMBus™)
- 2x 16-bit 1 channel timer
- 1x 16-bit 4 channels timer (supporting motor control)
- 1x 32-bit 4 channels timer
- 3x 16-bit ultra-low-power timer
- 1x RTC with 32-bit sub-second wakeup counter
- 1x independent SysTick
- 1x independent watchdog
- 1x window watchdog
- **Up to 43 I/Os**, most 5 V-tolerant
- **Development support**
  - Serial-wire debug (SWD), JTAG for the application processor
- All packages are ECOPACK2 compliant

Table 1. Device summary

Reference	Part number
STM32WLE5J8/JB/JC	STM32WLE5J8I6, STM32WLE5JB16, STM32WLE5UJC16

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# 1 Introduction

This document provides information on the STM32WLE5J8/JB/JC microcontrollers.

For information on the Arm<sup>®(a)</sup>Cortex<sup>®</sup>-M4 core, refer to the Cortex<sup>®</sup>-M4 Technical Reference Manual available from the [www.arm.com](http://www.arm.com) website.

For information on LoRa<sup>®</sup>, refer to the Semtech website (<https://www.semtech.com/technology/lora>).



# 2 Description

The STM32WLE5J8/JB/JC long-range wireless and ultra-low-power devices embed a powerful and ultra-low-power radio compliant LPWAN radio solution: LoRa<sup>®</sup>, (G)FSK, (G)MSK, and BPSK.

These devices are designed to be extremely low-power and are based on the high-performance Arm<sup>®</sup> Cortex<sup>®</sup>-M4 32-bit RISC core operating at a frequency of up to 48 MHz. This core implements a full set of DSP instructions and an independent memory protection unit (MPU) that enhances the application security.

The devices embed high-speed memories (Flash memory up to 256 Kbytes, SRAM up to 64 Kbytes), and an extensive range of enhanced I/Os and peripherals.

The devices also embed several protection mechanisms for embedded Flash memory and SRAM: readout protection, write protection and proprietary code readout protection.

These devices offer a 12-bit ADC, a 12-bit DAC low-power sample-and-hold, two ultra-low-power comparators associated with a high accuracy reference voltage generator.

The devices embed a low-power RTC with a 32-bit sub-second wakeup counter, one 16-bit single-channel timer, two 16-bit four-channel timers (supporting motor control), one 32-bit four-channel timer and three 16-bit ultra-low-power timers.

These devices also embed two DMA controllers (7 channels each) allowing any transfer combination between memory (Flash memory, SRAM1 and SRAM2) and peripheral, using the DMAMUX1 for flexible DMA channel mapping.

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a. Arm is a registered trademark of Arm Limited (or its subsidiaries) in the US and/or elsewhere.

The devices also feature the standard and advanced communication interfaces listed below:

- two USART (supporting LIN, smartcard, IrDA, modem control and ISO7816)
- one low-power UART (LPUART)
- three I2C (SMBus/PMBus)
- two SPIs (up to 16 MHz, one supporting I<sup>2</sup>S)
- semaphores for processor firmware process synchronization

The operating temperature/voltage ranges are –40 °C to +85 °C from a 1.8 V to 3.6 V power supply. A comprehensive set of power-saving modes allows the design of low-power applications.

The devices integrate a high-efficiency SMPS step-down converter and independent power supplies for ADC, DAC and comparator analog inputs.

A V<sub>BAT</sub> dedicated supply allows the LSE 32.768 kHz oscillator, the RTC and the backup registers to be backed up. The devices can maintain these functions even if the main V<sub>DD</sub> is not present, through a CR2032-like battery, a supercap or a small rechargeable battery.

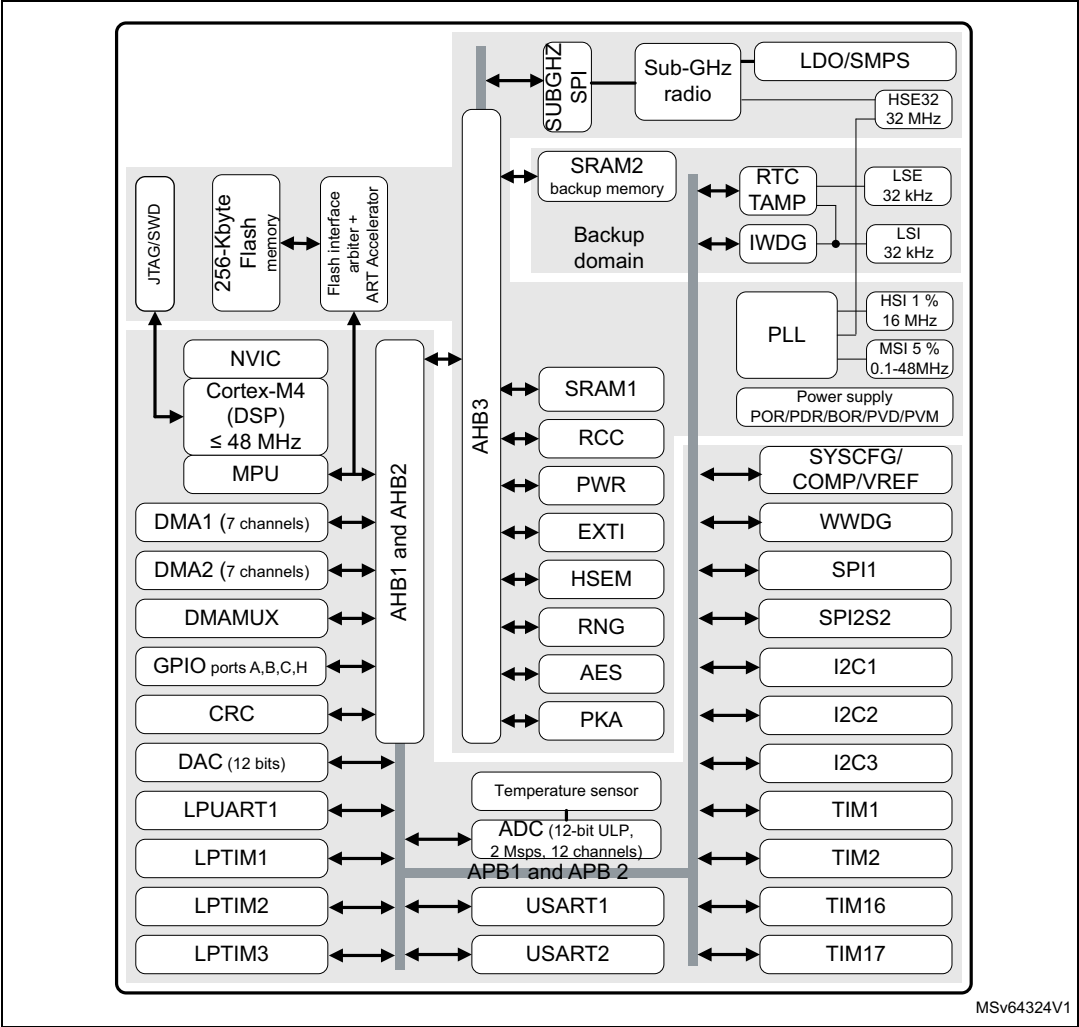
**Table 2. Main features and peripheral counts**

Feature		STM32WLE5JCI6	STM32WLE5JBI6	STM32WLE5J8I6
CPU		Arm Cortex-M4		
Maximum CPU frequency (MHz)		48		
Flash memory density (Kbytes)		256	128	64
SRAM density (Kbytes)	SRAM1	32	16	0
	SRAM2	32	32	20
Radio	LoRa	Yes		
	(G)FSK			
	(G)MSK			
	BPSK			
Radio power amplifier	Low-output power (up to 15 dBm)	Yes		
	High-output power (up to 22 dBm)			
Timer	General purpose	4		
	Low-power	3		
	SysTick	1		
Communication interface	SPI/I <sup>2</sup> S	2 (1 supporting I <sup>2</sup> S)		
	I <sup>2</sup> C	3		
	USART	2		
	LPUART	1		
Watchdog	Independent	1		
	Window	1		
RTC (with wakeup counter)		1		

Table 2. Main features and peripheral counts (continued)

Feature		STM32WLE5JCI6	STM32WLE5JBI6	STM32WLE5J8I6
DMA (7 channels)		2		
Semaphores		1		
Security	AES 256 bits	1		
	RNG	1		
	PKA	1		
	PCROP, RDP, WRP	1		
	CRC	1		
	64-bit UID compliant with IEEE 802-2001 standard	1		
	96-bit die ID	1		
Tamper pins		3		
Wakeup pins		3		
GPIOs		43		
ADC (number of channels, ext + int)		1 (12 + 4)		
DAC (number of channels)		1 (1)		
Internal $V_{REF}$		Yes		
Analog comparator		2		
Operating voltage		1.8 to 3.6 V		
Ambient operating temperature		−40 °C to +85 °C		
Junction temperature		−40 °C to +105 °C		
Package		UFBGA73 (5x5 mm)		

Figure 1. STM32WLE5J8/JB/JC block diagram



## 3 Functional overview

### 3.1 Architecture

The devices embed a sub-GHz RF subsystem that interfaces with a generic microcontroller subsystem using an Arm Cortex-M4 (called CPU).

An RF low-layer stack is needed and is to be run on CPU with the host application code.

The RF subsystem communication is done through an internal SPI interface.

### 3.2 Arm Cortex-M4 core

The Arm Cortex-M4 is a processor for embedded systems. It has been developed to provide a low-cost platform that meets the needs of MCU implementation, with a reduced pin count and low-power consumption, while delivering outstanding computational performance and an advanced response to interrupts.

The Arm Cortex-M4 32-bit RISC processor features exceptional code-efficiency, delivering the high-performance expected from an Arm core in the memory size usually associated with 8- and 16-bit devices.

This processor supports a set of DSP instructions that allow efficient signal processing and complex algorithm execution.

With its embedded Arm core, the STM32WLE5J8/JB/JC devices are compatible with all Arm tools and software.

*Figure 1* shows the general block diagram of the STM32WLE5J8/JB/JC devices.

### 3.3 Adaptive real-time memory accelerator (ART Accelerator)

The ART Accelerator is a memory accelerator that is optimized for STM32 industry-standard Arm Cortex-M4 processor. The ART Accelerator balances the inherent performance advantage of the Arm Cortex-M4 over Flash memory technologies, that normally require the processor to wait for the Flash memory at higher frequencies.

To release the processor near 60 DMIPS performance at 48 MHz, the ART Accelerator implements an instruction prefetch queue and branch cache, that increases the program execution speed from the 64-bit Flash memory. Based on CoreMark benchmark, the performance achieved thanks to the ART Accelerator is equivalent to 0 wait state program execution from Flash memory at a CPU frequency up to 48 MHz.

### 3.4 Memory protection unit (MPU)

The memory protection unit (MPU) is used to manage the CPU accesses to memory, to prevent one task to accidentally corrupt the memory or resources used by any other active task. This memory area is organized into up to eight protected areas that can in turn be divided up into eight subareas. The protection area sizes are between 32 bytes and the whole 4 Gbytes of addressable memory.

The MPU is especially helpful for applications where some critical or certified code must be protected against the misbehavior of other tasks. It is usually managed by an RTOS (real-time operating system). If a program accesses a memory location that is prohibited by the MPU, the RTOS can detect it and take action. In an RTOS environment, the kernel can dynamically update the MPU area setting, based on the process to be executed.

The MPU is optional and can be bypassed for applications that do not need it.

## 3.5 Memories

### 3.5.1 Embedded Flash memory

The Flash memory interface manages the CPU AHB ICode and DCode accesses to the Flash memory. It implements the access, the erase and program Flash memory operations, and the read and write protection.

The main features of the Flash memory are listed below:

- Memory organization: 1 bank
  - main memory: up to 256 Kbytes
  - page size: 2 Kbytes
- 72-bit wide data read (64 bits plus 8 ECC bits)
- 72-bit wide data write (64 bits plus 8 ECC bits)
- Page erase and mass erase

Flexible protections can be configured thanks to option bytes:

- Readout protection (RDP) to protect the whole memory. Three levels are available:
  - Level 0: no readout protection
  - Level 1: memory readout protection. The Flash memory cannot be read from or written to if either debug features are connected, boot in SRAM or bootloader is selected.
  - Level 2: chip readout protection. Debug features (JTAG and serial wire), boot in SRAM and bootloader selection are disabled (JTAG fuse). This selection is irreversible.

**Table 3. Access status versus RDP level and execution mode**

Area	RDP level	User execution			Debug, boot from SRAM or boot from system memory (loader)		
		Read	Write	Erase	Read	Write	Erase
Main memory	1	Yes	Yes	Yes	No	No	No
	2	Yes	Yes	Yes	NA	NA	NA
System memory	1	Yes	No	No	Yes	No	No
	2	Yes	No	No	NA	NA	NA
Option bytes	1	Yes	Yes	Yes	Yes	Yes	Yes
	2	Yes	No <sup>(1)</sup>	No <sup>(1)</sup>	NA	NA	NA

Table 3. Access status versus RDP level and execution mode (continued)

Area	RDP level	User execution			Debug, boot from SRAM or boot from system memory (loader)		
		Read	Write	Erase	Read	Write	Erase
Backup registers	1	Yes	Yes	NA <sup>(2)</sup>	No	No	NA <sup>(2)</sup>
	2	Yes	Yes	NA	NA	NA	NA
SRAM2	1	Yes	Yes	Yes <sup>(2)</sup>	No	No	No <sup>(2)</sup>
	2	Yes	Yes	Yes	NA	NA	NA

1. The option byte can be modified by the sub-GHz radio.

2. Erased when RDP changes from Level 1 to Level 0.

- Write protection (WRP): the protected area is protected against erasing and programming. Two areas can be selected, with 4-Kbyte granularity.
- Proprietary code readout protection (PCROP): two parts of the Flash memory can be protected against read and write from third parties. The protected area is execute-only: it can only be reached by the STM32 CPU, as an instruction code, while all other accesses (DMA, debug and CPU data read, write and erase) are strictly prohibited. Two areas can be selected, with 2-Kbyte granularity. An additional option bit (PCROP\_RDP) is used to select if the PCROP area is erased or not when the RDP protection is changed from Level 1 to Level 0.

The whole non-volatile memory embeds the error correction code (ECC) feature supporting:

- single error detection and correction
- double error detection
- address of the ECC fail can be read in the FLASH\_ECCR register

### 3.5.2 Embedded SRAM

The devices feature up to 64 Kbytes of embedded SRAM, split in two blocks:

- SRAM1: up to 32 Kbytes mapped at address 0x2000 0000
- SRAM2: up to 32 Kbytes located at address 0x2000 8000 (contiguous to SRAM1), also mirrored at 0x1000 0000, with hardware parity check (this SRAM can be retained in Standby mode)

The SRAMs can be accessed in read/write with 0 wait states for all CPU clock speeds.



## 3.6 Security memory management

The devices contain many security blocks both for the sub-GHz MAC layer and the Host application, such as:

- true random number generator (RNG)
- advance encryption standard hardware accelerators (128- and 256-bit AES, supporting ECB, CBC, CTR, GCM, GMAC and CCM chaining modes)
- private key acceleration (PKA):
  - modular arithmetic including exponentiation with maximum modulo size of 3136 bits
  - elliptic curves over prime field scalar multiplication, ECDSA signature, ECDSA verification with maximum modulo size of 521 bits
- cyclic redundancy check calculation unit (CRC)

## 3.7 Boot modes

At startup, BOOT0 pin and BOOT1 option bit are used to select one of the following boot options:

- Boot from user Flash memory
- Boot from boot system memory (where embedded bootloader is located)
- Boot from embedded SRAM

The devices always boot on CPU core. The embedded bootloader code makes it possible to boot from the USART or SPI peripheral.

## 3.8 Sub-GHz radio

### 3.8.1 Introduction

The sub-GHz radio is an ultra-low-power sub-GHz radio operating in the 150 - 960 MHz ISM band. LoRa, (G)FSK/(G)MSK modulation in transmit and receive, and (D)BPSK in transmit only, allow an optimal trade-off between range, data rate and power consumption. This sub-GHz radio is compliant with the LoRaWAN specification v1.0 and radio regulations including ETSI EN 300 220, EN 300 113, EN 301 166, FCC CFR 47 part 15, 24, 90, 101 and the ARIB STD-T30, T-67, T-108.

The sub-GHz radio consists of:

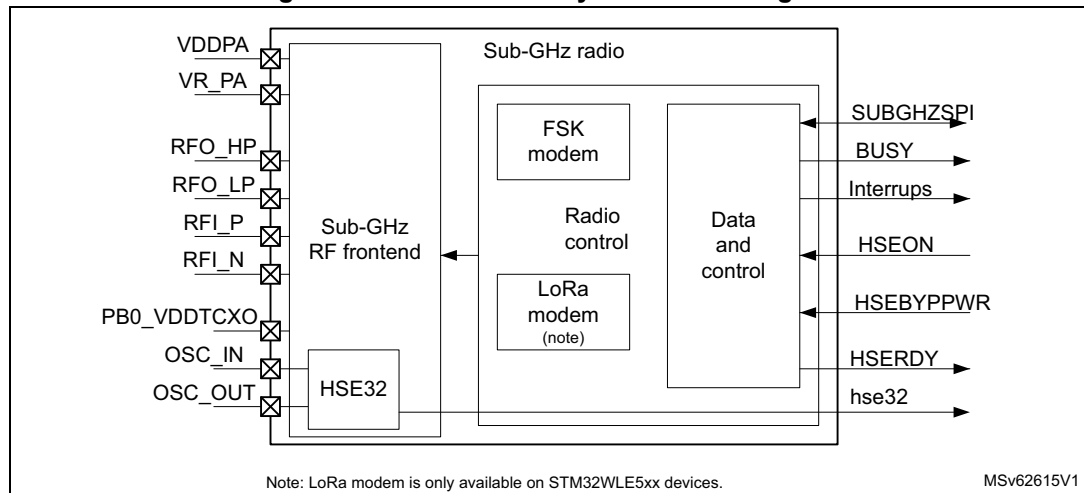
- an analog front-end transceiver, capable of outputting up to + 15 dBm maximum power on its RFO\_LP pin and up to + 22 dBm maximum power on RFO\_HP pin
- a digital modem bank providing the following modulation schemes:
  - LoRa Rx/Tx with bandwidth (BW) from 7.8 - 500 kHz, spreading factor (SF) 5 - 12, bit rate (BR) from 0.013 to 17.4 Kbit/s (real bitrate)
  - FSK and GFSK Rx/Tx with BR from 0.6 to 300 Kbit/s
  - (G)MSK Tx with BR from 0 to 10 Kbit/s
  - BPSK and DBPSK TX only with bitrate for 100 and 600 bit/s
- a digital control including all data processing and sub-GHz radio configuration control
- a high-speed clock generation

### 3.8.2 General description

The sub-GHz radio provides an internal processing unit to handle communication with the system CPU. Communication is handled by commands sent over the SPI interface, and a set of interrupts is used to signal events. BUSY information signals operation activity and is used to indicate when the sub-GHz radio commands cannot be received.

The block diagram of the sub-GHz radio system is shown in the figure below.

**Figure 2. sub-GHz radio system block diagram**



### 3.8.3 Transmitter

The transmit chain comprises the modulated output from the modem, that directly modulates the RF-PLL. An optional pre-filtering of the bit stream can be enabled to reduce the power in the adjacent channel also dependent on the selected modulation scheme. The modulated signal from the RF-PLL directly drives the high output power PA (HP PA) or low output power PA (LP PA).

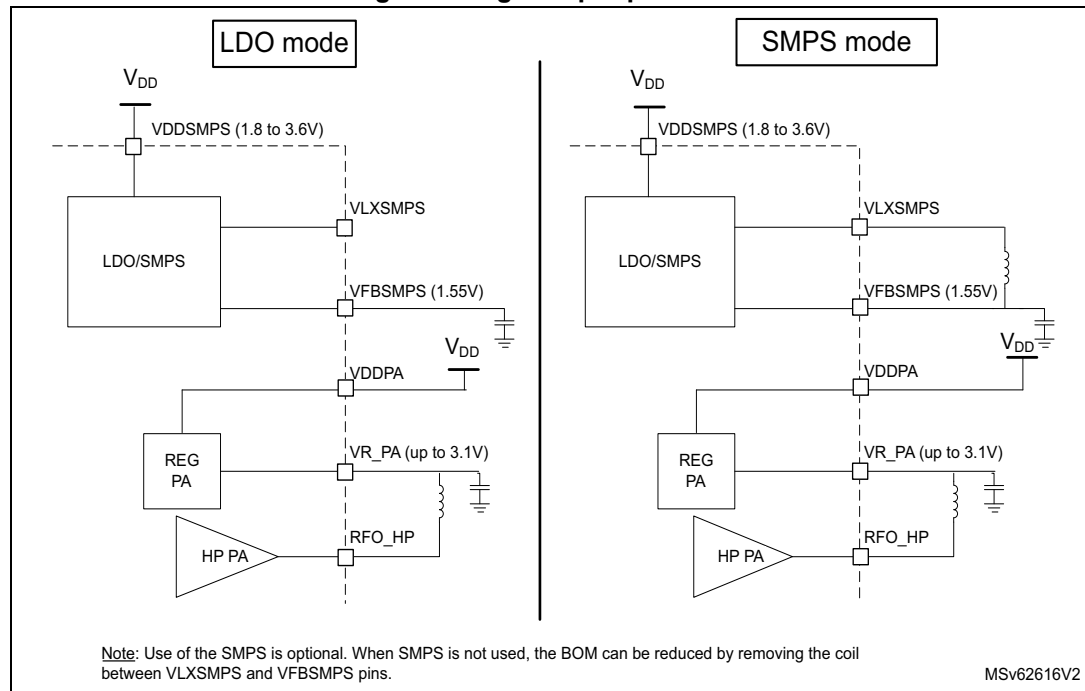
#### Transmitter high output power

Transmit high output power up to + 22 dBm, is supported through the RFO\_HP RF pin. The HP PA can be supplied from the PA regulator (REG PA) up to 3.1 V.

For this, the REG PA must be supplied directly from  $V_{DD}$  on VDDSMPS pin, as shown in the figure below.

The output power range is programmable in 32 steps of ~ 1 dB. The power amplifier ramping timing is also programmable. This allows adaptation to meet radio regulation requirements.

**Figure 3. High output power PA**



The table below gives the maximum transmit output power versus the  $V_{DDPA}$  supply level.

**Table 4. Sub-GHz radio transmit high output power**

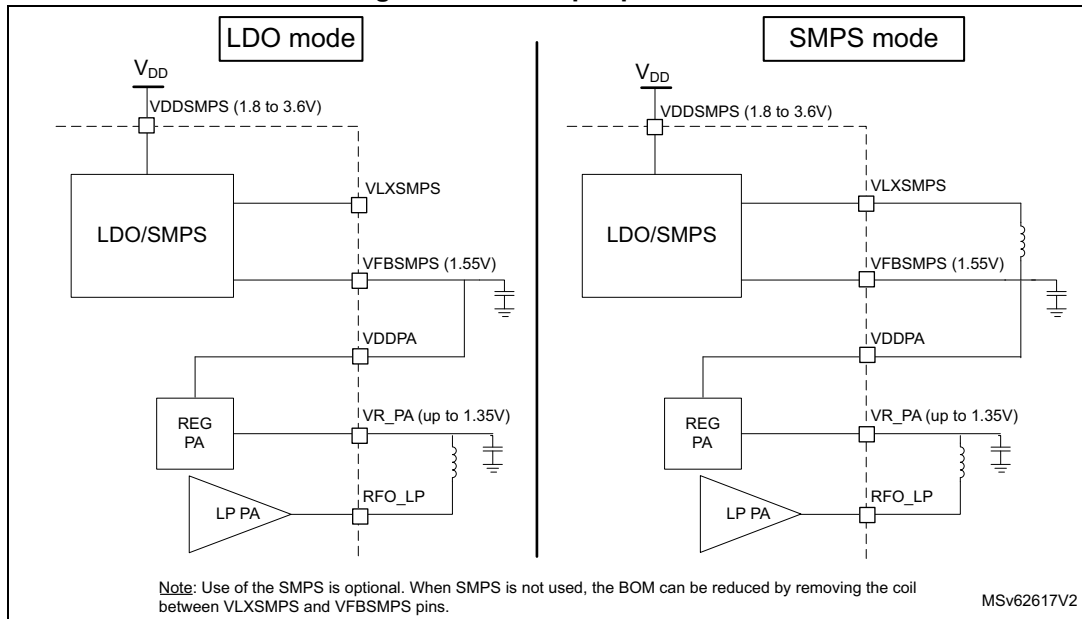
$V_{DDPA}$ supply (V)	Transmit output power (dBm)
3.3	+ 22
2.7	+ 20
2.0	+ 16

### Transmitter low output power

The transmit low output power up to + 15 dBm, is supported through the RFO\_LP RF pin. The LP PA can be supplied from the PA regulator (REG PA) up to 1.35 V. For this, the REG PA must be supplied from the regulated  $V_{FBSMPS}$  supply at 1.55 V, as shown in the figure below.

The output power range is programmable in 32 steps of ~1 dB. The power amplifier ramping timing is also programmable. This allows adaptation to meet radio regulation requirements.

Figure 4. Low output power PA



### 3.8.4 Receiver

The receive chain comprises a differential low-noise amplifier (LNA), a down-converter to low-IF by mixer operation in quadrature configuration. The I and Q signals are low pass filtered and a  $\Sigma\Delta$  ADC converts them into the digital domain. In the digital modems, the signals are decimated, further down converted and channel filtered. The demodulation is done according to the selected modulation scheme.

The down mixing to low-IF is done by mixing the receive signal with the local RF-PLL located in the negative frequency, where  $-f_{lo} = -f_{rf} + -f_{if}$ . (where  $f_{lo}$  is the local RF-PLL frequency,  $f_{rf}$  is the received signal and  $f_{if}$  is the intermediate frequency). The wanted signal is located at  $f_{rf} = f_{lo} + f_{if}$ .

The receiver features automatic I and Q calibration, that improves image rejection. The calibration is done automatically at startup before using the receiver, and can be requested by command.

The receiver supports LoRa, (G)MSK and (G)FSK modulations.

### 3.8.5 RF-PLL

The RF-PLL is used as the frequency synthesizer for the generation of the local oscillator frequency ( $f_{lo}$ ) for both transmit and receive chains. The RF-PLL uses auto calibration and uses the 32 MHz HSE32 reference. The sub-GHz radio covers all continuous frequencies in the range between 150 to 960 MHz.

### 3.8.6 Intermediate frequencies

The sub-GHz radio receiver operates mostly in low-IF configuration, except for specific high-bandwidth settings.

**Table 5. FSK mode intermediate frequencies**

Setting name	Bandwidth (kHz)	$f_{if}$ (kHz)
RX_BW_467	467.0	250
RX_BW_234	234.3	
RX_BW_117	117.3	
RX_BW_58	58.6	
RX_BW_29	29.3	
RX_BW_14	14.6	
RX_BW_7	7.3	
RX_BW_373	373.6	200
RX_BW_187	187.2	
RX_BW_93	93.8	
RX_BW_46	46.9	
RX_BW_23	23.4	
RX_BW_11	11.7	
RX_BW_5	5.8	
RX_BW_312	312.0	167
RX_BW_156	156.2	
RX_BW_78	78.2	
RX_BW_39	39.0	
RX_BW_19	19.5	
RX_BW_9	9.7	
RX_BW_4	4.8	

**Table 6. LoRa mode intermediate frequencies**

Setting name	Bandwidth (kHz)	$f_{if}$ (kHz)
LORA_BW_500	500	0
LORA_BW_250	250	250
LORA_BW_125	125	
LORA_BW_62	62.5	
LORA_BW_41	41.67	167
LORA_BW_31	31.25	250
LORA_BW_20	20.83	167

Table 6. LoRa mode intermediate frequencies (continued)

Setting name	Bandwidth (kHz)	$f_{if}$ (kHz)
LORA_BW_15	15.63	250
LORA_BW_10	10.42	167
LORA_BW_7	7.81	250

## 3.9 Power supply management

The devices embed two different regulators: one LDO and one DC/DC (SMPS). The SMPS can be optionally switched-on by software to improve the power efficiency. As LDO and SMPS operate in parallel, the SMPS switch-on is transparent to the user and only the power efficiency is affected.

### 3.9.1 Power supply schemes

The devices require a  $V_{DD}$  operating voltage supply between 1.8 V and 3.6 V. Several independent supplies ( $V_{DDSMPS}$ ,  $V_{FBSMPS}$ ,  $V_{DDA}$ ,  $V_{DDRF}$ ) can be provided for specific peripherals:

- **$V_{DD}$  = 1.8 V to 3.6 V**  
 $V_{DD}$  is the external power supply for the I/Os, the system analog blocks such as reset, power management, internal clocks and low-power regulator. It is provided externally through VDD pins.
- **$V_{DDSMPS}$  = 1.8 V to 3.6 V**  
 $V_{DDSMPS}$  is the external power supply for the SMPS step-down converter. It is provided externally through VDDSMPS supply pin and must be connected to the same supply as  $V_{DD}$ .
- **$V_{FBSMPS}$  = 1.45 V to 1.62 V (1.55 V typical)**  
 $V_{FBSMPS}$  is the external power supply for the main system regulator. It is provided externally through VFBSMPS pin and is supplied through the SMPS step-down converter.
- **$V_{DDA}$  = 0 V to 3.6 V (DAC/COMPs minimum voltage is 1.71 V, ADC minimum voltage is 1.8 V and VREFBUF minimum voltage is 2.4 V).**  
 $V_{DDA}$  is the external analog power supply for A/D converters, D/A converters, voltage reference buffer, and comparators. The  $V_{DDA}$  voltage level is independent from the  $V_{DD}$  voltage (see power-up and power-down limitations below) and must preferably be connected to  $V_{DD}$  when these peripherals are not used.
- **$V_{DDRF}$  = 1.8 V to 3.6 V**  
 $V_{DDRF}$  is an external power supply for the radio. It is provided externally through the VDDRF pin and must be connected to the same supply as  $V_{DD}$ .
- **$V_{DDRF1V5}$  = 1.45 V to 1.62 V**  
 $V_{DDRF1V5}$  is an external power supply for the radio. It is provided externally through the VDDRF1V5 pin and must be connected externally to VFBSMPS.
- **$V_{BAT}$  = 1.55 V to 3.6 V**  
 $V_{BAT}$  is the power supply for RTC, TAMP, external clock 32 kHz oscillator and backup registers (through power switch) when  $V_{DD}$  is not present.

- **VREF-, VREF+**

$V_{REF+}$  is the input reference voltage for ADC. It is also the output of the internal voltage reference buffer when enabled.

- When  $V_{DDA} < 2\text{ V}$ ,  $V_{REF+}$  must be equal to  $V_{DDA}$ .
- When  $V_{DDA} \geq 2\text{ V}$ ,  $V_{REF+}$  must be between  $2\text{ V}$  and  $V_{DDA}$ .

$V_{REF+}$  can be grounded when ADC is not active. The internal voltage reference buffer supports the following output voltages, configured with VRS bit in the VREFBUF\_CSR register:

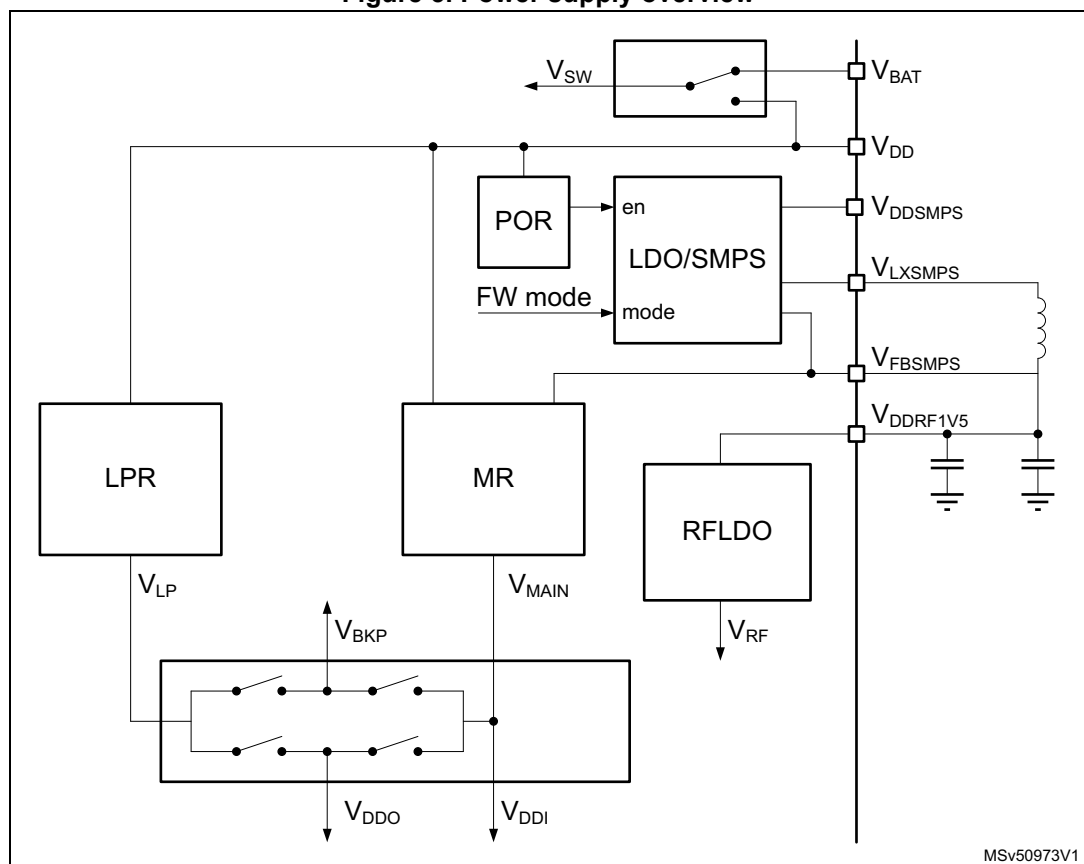
- $V_{REF+}$  around  $2.048\text{ V}$ : this requires  $V_{DDA} \geq 2.4\text{ V}$ .
- $V_{REF+}$  around  $2.5\text{ V}$ : this requires  $V_{DDA} \geq 2.8\text{ V}$ .

During power up and power down, the following power sequence is required:

1. When  $V_{DD} < 1\text{ V}$  other power supplies ( $V_{DDA}$ ) must remain below  $V_{DD} + 300\text{ mV}$ .  
During power down,  $V_{DD}$  can temporarily become lower than other supplies only if the energy provided to the device remains below  $1\text{ mJ}$ . This allows external decoupling capacitors to be discharged with different time constants during this transient phase.
2. When  $V_{DD} > 1\text{ V}$ , all other power supplies ( $V_{DDA}$ ) become independent.

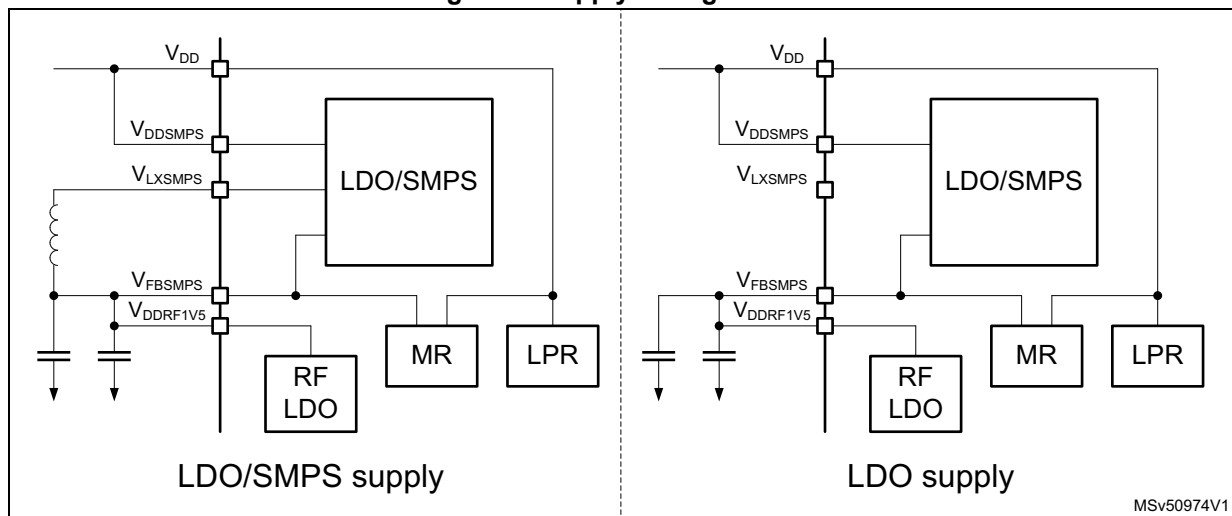
An embedded linear voltage regulator is used to supply the internal digital power  $V_{CORE}$ .  $V_{CORE}$  is the power supply for digital peripherals, SRAM1 and SRAM2. The Flash memory is supplied by  $V_{CORE}$  and  $V_{DD}$ .  $V_{CORE}$  is split in two parts:  $V_{DDO}$  part and an interruptible part  $V_{DDI}$ .

**Figure 5. Power supply overview**



The different supply configurations are shown in the figure below.

**Figure 6. Supply configurations**



The LDO or SMPS step-down converter operating mode can be configured by one of the following:

- by the MCU using the SMPSEN setting in PWR control register 5 (PWR\_CR5), that depends upon the MCU system operating mode (Run, Stop, Standby or Shutdown).
- by the sub-GHz radio using `SetRegulatorMode()` command and the sub-GHz radio operating mode (Sleep, Calibrate, Standby, Standby with HSE32 or Active).

After any POR and NRST reset, the LDO mode is selected. The SMPS selection has priority over LDO selection.

While the sub-GHz radio is in Standby with HSE32 or in Active mode, the supply mode is not altered until the sub-GHz radio enters Standby or Sleep mode. The sub-GHz radio activity may add a delay for entering the MCU software requested supply mode.

The LDO or SMPS supply mode can be checked with the SMPSRDY flag in Power status register 2 (PWR\_SR2).

**Note:** When the radio is active, the supply mode is not changed until after the radio activity is finished.

During Stop 1, Stop 2 and Standby modes, when the sub-GHz radio is not active, the LDO or SMPS step-down converter is switched off. When exiting low-power modes (except Shutdown), the SMPS step-down converter is set by hardware to the mode selected by the SMPSEN bit in PWR control register 5 (PWR\_CR5). SMPSEN is retained in Stop and Standby modes.

Independently from the MCU software selected supply operating mode, the sub-GHz radio allows the supply mode selection while the sub-GHz radio is active (thanks to the sub-GHz radio `SetRegulatorMode()` command).

The maximum load current delivered by the SMPS can be selected by the sub-GHz radio SUBGHZ\_SMPSC2R register.

The inrush current of the LDO and SMPS step-down converter can be controlled via the sub-GHz radio SUBGHZ\_PCR register. This information is retained in all but the sub-GHz radio Deep-sleep mode.



The SMPS needs a clock to be functional. If for any reason this clock stops, the device may be destroyed. To avoid this situation, a clock detection is used to, in case of a clock failure, switch off the SMPS and enable the LDO. The SMPS clock detection is enabled by the sub-GHz radio SUBGHZ\_SMPSC0R.CLKDE. By default, the SMPS clock detection is disabled and must be enabled before enabling the SMPS.

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**Danger:** Before enabling the SMPS, the SMPS clock detection must be enabled in the sub-GHz radio SUBGHZ\_SMPSC0R.CLKDE.

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### 3.9.2 Power supply supervisor

The devices integrate a power-on reset/power-down reset, coupled with a Brownout reset (BOR) circuitry.

BOR0 level cannot be disabled. Other BOR levels can be enabled by user option. When enabled, BOR is active in all power modes except in Shutdown

Five BOR thresholds can be selected through option bytes.

During power-on, BOR keeps the device under reset until the supply voltage  $V_{DD}$  reaches the specified  $V_{BORx}$  threshold:

- When  $V_{DD}$  drops below the selected threshold, a device reset is generated.
- When  $V_{DD}$  is above the  $V_{BORx}$  upper limit, the device reset is released and the system can start.

The devices feature an embedded PVD (programmable voltage detector) that monitors the  $V_{DD}$  power supply and compares it with the  $V_{PVD}$  threshold. An interrupt can be generated when  $V_{DD}$  drops below the  $V_{PVD}$  threshold and/or when  $V_{DD}$  is higher than the  $V_{PVD}$  threshold. The interrupt service routine can then generate a warning message and/or put the MCU into a safe state.

The PVD is enabled by software and can be configured to monitor the  $V_{DD}$  supply level needed for the sub-GHz radio operation. For this, the PVD must select its lowest threshold, and the PVD and the wakeup must be enabled by the EWPVD bit in PWR\_CR3 register. Only a voltage drop below the PVD level generates a wakeup event.

In addition, the devices embed a PVM (peripheral voltage monitor) that compares the independent supply voltage  $V_{DDA}$  with a fixed threshold to ensure that the peripheral is in its functional supply range.

Finally, a radio end-of-life monitor provides information on the  $V_{DD}$  supply when  $V_{DD}$  is too low to operate the sub-GHz radio. When reaching the EOL level, the software must stop all radio activity in a safe way.

### 3.9.3 Linear voltage regulator

Two embedded linear voltage regulators supply all the digital circuitries, except for the Standby circuitry and the Backup domain. The main regulator (MR) output voltage (VCORE) can be programmed by software to two different power ranges (range 1 and range 2), to optimize the consumption depending on the system maximum operating frequency.

The voltage regulators are always enabled after a reset. Depending on the application modes, the  $V_{CORE}$  supply is provided either by the main regulator or by the low-power regulator (LPR).

When MR is used, a dynamic voltage scaling is proposed to optimize power as follows:

- **range 1: high-performance range**  
The system clock frequency can be up to 48 MHz. The Flash memory access time for read access is minimum. Write and erase operations are possible.
- **range 2: low-power range**  
The system clock frequency can be up to 16 MHz. The Flash memory access time for a read access is increased as compared to range 1. Write and erase operations are possible.

### 3.9.4 VBAT operation

The VBAT pin is used to power the device VBAT domain (RTC, LSE and backup registers) from an external battery, an external super-capacitor, or from  $V_{DD}$  when no external battery nor an external super-capacitor are present. Three anti-tamper detection pins are available in VBAT mode.

VBAT operation is automatically activated when  $V_{DD}$  is not present.

An internal  $V_{BAT}$  battery charging circuit is embedded and can be activated when  $V_{DD}$  is present.

*Note:* When the microcontroller is supplied only from  $V_{BAT}$ , external interrupts and RTC alarm/events do not exit it from VBAT operation.

## 3.10 Low-power modes

The devices support several low-power modes to achieve the best compromise between low-power consumption, short startup time, available peripherals and available wakeup sources.

By default, the microcontroller is in Run mode, range 1, after a system or a power-on reset. It is up to the user to select one of the low-power modes described below:

- **Sleep mode:** CPU clock off, all peripherals including CPU core peripherals (among them NVIC, SysTick) can run and wake up the CPU when an interrupt or an event occurs.
- **Low-power run mode (LPRun):** when the system clock frequency is reduced below 2 MHz. The code is executed from the SRAM or from the Flash memory. The regulator is in low-power mode to minimize the operating current.
- **Low-power sleep mode (LPSleep):** entered from the LPRun mode.
- **Stop 0 and Stop 1 modes:** the content of SRAM1, SRAM2 and of all registers is retained. All clocks in the VCORE domain are stopped. PLL, MSI, HSI16 and HSE32 are disabled. LSI and LSE can be kept running.

RTC can remain active (Stop mode with RTC, Stop mode without RTC). The sub-GHz radio may remain active independently from the CPU.

Some peripherals with the wakeup capability can enable HSI16 RC during the Stop mode to detect their wakeup condition.

Stop 1 offers the largest number of active peripherals and wakeup sources, a smaller

wakeup time but a higher consumption compared with Stop 2.

In Stop 0 mode, the main regulator remains on, resulting in the fastest wakeup time but with much higher consumption. The active peripherals and wakeup sources are the same as in Stop 1 mode that uses the low-power regulator.

The system clock, when exiting Stop 0 or Stop 1 mode, can be either MSI up to 48 MHz or HSI16, depending on the software configuration.

- **Stop 2 mode:** part of the  $V_{CORE}$  domain is powered off. Only SRAM1, SRAM2, CPU and some peripherals preserve their contents (see [Table 7](#)).

All clocks in the  $V_{CORE}$  domain are stopped. PLL, MSI, HSI16 and HSE32 are disabled. LSI and LSE can be kept running.

RTC can remain active (Stop 2 mode with RTC, Stop 2 mode without RTC). The sub-GHz radio may also remain active independent from the CPU.

Some peripherals with the wakeup capability can enable HSI16 RC during the Stop 2 mode to detect their wakeup condition (see [Table 7](#)).

The system clock when exiting from Stop 2 mode, can be either MSI up to 48 MHz or HSI16, depending on the software configuration.

- **Standby mode:**  $V_{CORE}$  domain is powered off. However, it is possible to preserve the SRAM2 content as detailed below:
  - Standby mode with SRAM2 retention when the RRS bit is set in the PWR control register 3 (PWR\_CR3). In this case, SRAM2 is supplied by the low-power regulator.
  - Standby mode when the RRS bit is cleared in the PWR control register 3 (PWR\_CR3). In this case the main regulator and the low-power regulator are powered off.

All clocks in the  $V_{CORE}$  domain are stopped. PLL, MSI, HSI16 and HSE32 are disabled. LSI and LSE can be kept running.

The RTC can remain active (Standby mode with RTC, Standby mode without RTC). The sub-GHz radio and the PVD may also remain active when enabled independent from the CPU. In Standby mode, the PVD selects its lowest level.

The system clock, when exiting Standby modes, is MSI at 4 MHz.

- **Shutdown mode:**  $V_{CORE}$  domain is powered off. All clocks in the  $V_{CORE}$  domain are stopped. PLL, MSI, HSI16, LSI and HSE32 are disabled. LSE can be kept running. The system clock when exiting the Shutdown mode, is MSI at 4 MHz. In this mode, the supply voltage monitoring is disabled and the product behavior is not guaranteed in case of a power voltage drop.

The table below summarizes the peripheral features over all available modes. Wakeup capability is detailed in gray cells.

**Table 7. Functionalities depending on system operating mode<sup>(1)</sup>**

Peripheral	Run	Sleep	LPRun	LPSleep	Stop 0		Stop 1		Stop 2		Standby		Shutdown		VBAT
					- Wakeup capability		- Wakeup capability		- Wakeup capability		- Wakeup capability		- Wakeup capability		
CPU	Y	-	Y	-	R	-	R	-	R	-	-	-	-	-	-
Sub-GHz radio system	O	O	O	O	O	O	O	O	O	O	O	O	-	-	-
Flash memory (up to 256 Kbytes)	Y	O <sup>(2)</sup>	O <sup>(3)</sup>	O <sup>(2)</sup> <sub>(3)</sub>	R	-	R	-	R	-	R	-	R	-	R
Flash memory interface	Y	Y	Y	Y	R	-	R	-	R	-	-	-	-	-	-
SRAM1	Y	O <sup>(2)</sup>	Y	O <sup>(2)</sup>	R	-	R	-	R	-	-	-	-	-	-
SRAM2	Y	O <sup>(2)</sup>	Y	O <sup>(2)</sup>	R	-	R	-	R	-	O <sup>(4)</sup>	-	-	-	-
Backup registers	Y	Y	Y	Y	R	-	R	-	R	-	R	-	R	-	R
Brownout reset (BOR)	Y	Y	Y	Y	Y	Y	Y	Y	Y	Y	Y	Y	-	-	-
Programmable voltage detector (PVD)	O	O	O	O	O	O	O	O	O	O	O <sup>(5)</sup>	O <sup>(5)</sup>	-	-	-
Peripheral voltage monitor (PVM3)	O	O	O	O	O	O	O	O	O	O	-	-	-	-	-
DMAX (x = 1, 2)	O	O	O	O	R	-	R	-	-	-	-	-	-	-	-
DMAMUX1	O	O	O	O	R	-	R	-	-	-	-	-	-	-	-
High-speed internal (HSI16)	O	O	O	O	O <sup>(6)</sup>	-	O <sup>(6)</sup>	-	O <sup>(6)</sup>	-	-	-	-	-	-
High-speed external (HSE32)	O	O	O <sup>(7)</sup>	O <sup>(7)</sup>	O <sup>(7)</sup>	-	O <sup>(7)</sup>	-	O <sup>(7)</sup>	-	O <sup>(7)</sup>	-	-	-	-
Low-speed internal (LSI)	O	O	O	O	O	-	O	-	O	-	O	-	-	-	-
Low-speed external (LSE)	O	O	O	O	O	-	O	-	O	-	O	-	O	-	O
Multi-speed internal (MSI)	O	O	O	O	O	-	O	-	O	-	-	-	-	-	-
Clock security system (CSS)	O	O	O	O	R	-	R	-	-	-	-	-	-	-	-
Clock security system on LSE	O	O	O	O	O	O	O	O	O	O	O	O	-	-	-
RTC/auto wakeup	O	O	O	O	O	O	O	O	O	O	O	O	O	O	O
Number of tamper pins	3	3	3	3	3	O	3	O	3	O	3	O	3	O	3
USARTx (x = 1, 2)	O	O	O	O	O <sup>(8)</sup>	O <sup>(8)</sup>	O <sup>(8)</sup>	O <sup>(8)</sup>	-	-	-	-	-	-	-
Low-power UART (LPUART1)	O	O	O	O	O <sup>(8)</sup>	O <sup>(8)</sup>	O <sup>(8)</sup>	O <sup>(8)</sup>	O <sup>(8)</sup>	O <sup>(8)</sup>	-	-	-	-	-
I2Cx (x = 1, 2)	O	O	O	O	O <sup>(9)</sup>	O <sup>(9)</sup>	O <sup>(9)</sup>	O <sup>(9)</sup>	-	-	-	-	-	-	-
I2C3	O	O	O	O	O <sup>(9)</sup>	O <sup>(9)</sup>	O <sup>(9)</sup>	O <sup>(9)</sup>	O <sup>(9)</sup>	O <sup>(9)</sup>	-	-	-	-	-
SPI1	O	O	O	O	R	-	R	-	-	-	-	-	-	-	-

Table 7. Functionalities depending on system operating mode<sup>(1)</sup> (continued)

Peripheral	Run	Sleep	LPRun	LPSleep	Stop 0		Stop 1		Stop 2		Standby		Shutdown		VBAT
					-	Wakeup capability	-	Wakeup capability	-	Wakeup capability	-	Wakeup capability	-	Wakeup capability	
SUBGHZSPI	O	O	O	O	R	-	R	-	-	-	-	-	-	-	-
SPI2S2	O	O	O	O	R	-	R	-	-	-	-	-	-	-	-
ADC	O	O	O	O	R	-	R	-	-	-	-	-	-	-	-
DAC	O	O	O	O	R	-	R	-	-	-	-	-	-	-	-
VREFBUF	O	O	O	O	O	-	O	-	R	-	-	-	-	-	-
COMPx (x = 1, 2)	O	O	O	O	O	O	O	O	O	O	-	-	-	-	-
Temperature sensor	O	O	O	O	R	-	R	-	-	-	-	-	-	-	-
TIMx (x = 1, 2, 16, 17)	O	O	O	O	R	-	R	-	-	-	-	-	-	-	-
LPTIM1	O	O	O	O	O	O	O	O	O	O	-	-	-	-	-
LPTIMx (x = 2, 3)	O	O	O	O	O	O	O	O	-	-	-	-	-	-	-
Independent watchdog (IWDG)	O	O	O	O	O	O	O	O	O	O	O	O	-	-	-
Window watchdog (WWDG)	O	O	O	O	R	-	R	-	R	-	-	-	-	-	-
SysTick timer	O	O	O	O	R	-	R	-	R	-	-	-	-	-	-
True random number generator (RNG)	O <sub>(10)</sub>	O <sub>(10)</sub> <sup>(1)</sup>	R	R	R	-	R	-	-	-	-	-	-	-	-
AES hardware accelerator	O	O	O	O	R	-	R	-	-	-	-	-	-	-	-
PKA hardware accelerator	O	O	O	O	R	-	R	-	-	-	-	-	-	-	-
CRC calculation unit	O	O	O	O	R	-	R	-	R	-	-	-	-	-	-
HSEM	O	R	O	R	R	-	R	-	-	-	-	-	-	-	-
EXTI	O	O	O	O	R	O	R	O	R	O	-	-	-	-	-
GPIOs	O	O	O	O	O	O	O	O	O	O	R <sub>(11)</sub>	3 pins <sub>(12)</sub>	<sup>(13)</sup>	3 pins <sub>(12)</sub> <sup>(1)</sup>	-

1. Legend: Y = Yes (enabled). O = Optional (disabled by default and can be enabled by software). R = data retained.  
 - = Not available. Gray cells indicate wakeup capability.

2. The SRAM clock can be gated on or off.

3. Flash memory can be placed in power-down mode.

4. The SRAM2 content can optionally be retained when the PWR\_CR3.RRS bit is set.

5. Only when the sub-GHz radio is active.

6. Some peripherals with wakeup from Stop capability can request HSI16 to be enabled. In this case, HSI16 is woken up by the peripheral, and only feeds the peripheral that requested it. HSI16 is automatically put off when the peripheral does not need it anymore.

7. HSE32 can be used by sub-GHz radio system.
8. USART reception is functional in Stop 0 and Stop 1 modes. LPUART1 reception is functional in Stop 0, Stop 1, and Stop 2 modes. LPUART1 generates a wakeup interrupt on Start address match or received frame event.
9. I2Cx (x= 1, 2) address detection is functional in Stop 0 and Stop 1 modes. I2C3 address detection is functional in Stop 0, Stop 1 and Stop 2 modes. I2C3 generates a wakeup interrupt in case of address match.
10. Voltage scaling range 1 only.
11. I/Os can be configured with internal pull-up, pull-down or floating in Standby mode.
12. The I/Os with wakeup from Standby/Shutdown capability are PA0, PC13 and PB3.
13. I/Os can be configured with internal pull-up, pull-down or floating in Shutdown mode, but the configuration is lost when exiting the Shutdown mode.

Table 8. Low-power mode summary

Mode name	Entry	Wakeup source <sup>(1)</sup>	Wakeup system clock	Effect on clocks	Voltage regulators	
					MR	LPR
Sleep (Sleep-now or Sleep-on-exit)	WFI or return from ISR	Any interrupt	Same as before entering Sleep mode	CPU clock OFF No effect on other clocks or analog clock sources	ON	ON
	WFE	Wakeup event				
LPRun	Set LPR bit	Clear LPR bit	Same as LPRun clock	None	OFF	ON
LPSleep	Set LPR bit + WFI or return from ISR	Any interrupt	Same as before entering LPSleep mode	CPU clock OFF No effect on other clocks or analog clock sources	OFF	ON
	Set LPR bit + WFE	Wakeup event			OFF	ON
Stop 0	LPMS = 0b000 + SLEEPDEEP bit + WFI or return from ISR or WFE	Any EXTI line (configured in the EXTI registers). Specific peripherals events	HSI16 when STOPWUCK = 1 in RCC_CFGR. MSI with the frequency before entering the Stop mode when STOPWUCK = 0.	All clocks OFF except HSI16, LSI and LSE	ON	ON
Stop 1	LPMS = 0b001 + SLEEPDEEP bit + WFI or return from ISR or WFE				OFF	
Stop 2 (with I2C3, LPUART1, LPTIM1, SRAM1, SRAM2)	LPMS = 0b010+ SLEEPDEEP bit + WFI or return from ISR or WFE					

Table 8. Low-power mode summary (continued)

Mode name	Entry	Wakeup source <sup>(1)</sup>	Wakeup system clock	Effect on clocks	Voltage regulators	
					MR	LPR
Standby (with SRAM2)	LPMS = 0b011 + Set RRS bit + SLEEPDEEP bit + WFI or return from ISR or WFE	Wakeup PVD, RFIRQ, wakeup RFBUSY, WKUP pin edge, RTC and TAMP event, LSECSS, external reset in NRST pin, IWDG reset	MSI 4 MHz	All clocks OFF except LSI and LSE	OFF	ON
Standby	LPMS = 0b011 + Clear RRS bit + SLEEPDEEP bit + WFI or return from ISR or WFE				OFF	OFF
Shutdown	LPMS = 0b1xx + SLEEPDEEP bit + WFI or return from ISR or WFE	WKUP pin edge, RTC and TAMP event, external reset in NRST pin	MSI 4 MHz	All clocks OFF except LSE	OFF	OFF

1. Refer to [Table 7: Functionalities depending on system operating mode](#).

### Relation between MCU and sub-GHz radio operating modes

The CPU and sub-GHz radio have their own operating modes, as defined in the table below.

Table 9. MCU and sub-GHz radio operating modes

CPU operating mode	Sub-GHz radio operating mode	Description
Run, Sleep	Sleep, Calibration, Standby, Active (FS, TX, RX) <sup>(1)</sup>	LDO or SMPS regulator active, MCU running in main regulator (MR) mode.
LPRun, LPSleep	Deep-sleep	LDO or SMPS regulator off, MCU running in low power regulator (LPR) mode.
	Sleep, Calibration, Standby, Active (FS, TX, RX)	LDO or SMPS regulator active, MCU running in low power regulator (LPR) mode.
Stop 0	Sleep, Calibration, Standby, Active (FS, TX, RX) <sup>(1)</sup>	LDO or SMPS regulator active, MCU running in main regulator (MR) mode.
Stop 1 and Stop 2	Deep-sleep	LDO or SMPS regulator off, MCU using low power regulator (LPR) mode.
	Sleep, Calibration, Standby, Active (FS, TX, RX)	LDO or SMPS regulator active, MCU using low power regulator (LPR) mode.
Standby	Deep-sleep	LDO or SMPS regulator off, MCU regulator off or on in low power (LPR) mode <sup>(2)</sup> .
	Sleep, Calibration, Standby, Active (FS, TX, RX)	LDO or SMPS regulator active, MCU regulator off or on in low power (LPR) mode <sup>(2)</sup> .
Shutdown	Deep-sleep <sup>(3)</sup>	LDO or SMPS regulator off, MCU regulator off

1. In the MCU Run, Sleep and Stop 0 modes, the sub-GHz radio is prevented from entering Deep-sleep mode.

2. When retaining SRAM2 in Standby mode, the MCU uses the low-power regulator (LPR) mode.
3. When the CPU is in Shutdown mode, the sub-GHz radio cannot be activated and is forced in Deep-sleep mode.

### 3.10.1 Reset mode

In order to improve the consumption under reset, the I/Os state under and after reset is "analog state" (the I/O Schmitt trigger is disabled). In addition, the internal reset pull-up is deactivated when the reset source is internal.

## 3.11 Peripheral interconnect matrix

Several peripherals have direct connections between them. This allows autonomous communication between peripherals, saving CPU resources and, consequently, reducing power-supply consumption. In addition, these hardware connections allow fast and predictable latency.

Depending on peripherals, these interconnections can operate in Run, Sleep, LPRun, LPSleep, Stop 0, Stop 1 and Stop 2 modes.

**Table 10. Peripherals interconnect matrix<sup>(1) (2)</sup>**

Source	Destination													
	TIM1	TIM2	TIM16	TIM17	LPTIM1	LPTIM2	LPTIM3	ADC	DAC	COMP1	COMP2	DMAMUX1	IRTIM	SUBGHZSPI
TIM1	-	X	-	-	-	-	-	X	X	X	X	-	-	-
TIM2	X	-	-	-	-	-	-	X	X	X	X	-	-	-
TIM16	-	-	-	-	-	-	-	-	-	-	-	-	X	-
TIM17	X	-	-	-	-	-	-	-	-	-	-	-	X	-
LPTIM1	-	-	-	-	-	-	X	-	X	-	-	X	-	-
LPTIM2	-	-	-	-	-	-	X	-	X	-	-	X	-	-
LPTIM3	-	-	-	-	-	-	-	-	-	-	-	X	-	X
ADC	X	-	-	-	-	-	-	-	-	-	-	-	-	-
Temperature sensor	-	-	-	-	-	-	-	X	-	-	-	-	-	-
VBAT	-	-	-	-	-	-	-	X	-	-	-	-	-	-
VREFINT	-	-	-	-	-	-	-	X	-	-	-	-	-	-
HSE32	-	-	-	X	-	-	-	-	-	-	-	-	-	-
LSE	-	X	X	-	-	-	-	-	-	-	-	-	-	-
MSI	-	-	-	X	-	-	-	-	-	-	-	-	-	-
LSI	-	-	X	-	-	-	-	-	-	-	-	-	-	-
MCO	-	-	-	X	-	-	-	-	-	-	-	-	-	-



Table 10. Peripherals interconnect matrix<sup>(1) (2)</sup> (continued)

Source	Destination													
	TIM1	TIM2	TIM16	TIM17	LPTIM1	LPTIM2	LPTIM3	ADC	DAC	COMP1	COMP2	DMAMUX1	IRTIM	SUBGHZSPI
<b>GPIO EXTI</b>	-	-	-	-	-	-	-	X	X	-	-	X	-	-
<b>RTC</b>	-	-	X	-	X	X	-	-	-	-	-	-	-	-
<b>TAMP</b>	-	-	-	-	X	X	-	-	-	-	-	-	-	-
<b>COMP1</b>	X	X	X	X	X	X	-	-	-	-	-	-	-	-
<b>COMP2</b>	X	X	X	X	X	X	-	-	-	-	-	-	-	-
<b>SYST ERR</b>	X	-	X	X	-	-	-	-	-	-	-	-	-	-

1. For more details, refer to section "Interconnection details" of the reference manual.

2. The "-" symbol in grayed cells means no interconnect.

### 3.12 Reset and clock controller (RCC)

The following different clock sources can be used to drive the system clock (SYSCLK):

- HSI16 (high-speed internal) 16 MHz RC oscillator clock
- MSI (multi-speed internal) RC oscillator clock from 100 kHz to 48 MHz
- HSE32 (high-speed external) 32 MHz oscillator clock, with trimming capacitors.
- PLL clock

The MSI is used as system clock source after startup from reset, configured at 4 MHz.

The devices have the following additional clock sources:

- LSI: 32 kHz low-speed internal RC that may drive the independent watchdog and optionally the RTC used for auto-wakeup from Stop and Standby modes.
- LSE: 32.768 kHz low-speed external crystal that optionally drives the RTC used for auto-wakeup from Stop, Standby and Shutdown modes, or the real-time clock (RTCCLK).

Each clock source can be switched on or off independently when it is not used, to optimize power consumption.

Several prescalers can be used to configure the AHB frequencies (HCLK3/PCLK3, HCLK1), the high-speed APB2 (PCLK2) and the low-speed APB1 (PCLK1) domains. The maximum frequency of the AHB (HCLK3, HCLK1), the PCLK1 and the PCLK2 domains is 48 MHz.

Most peripheral clocks are derived from their bus clock (HCLK, PCLK) except the following:

- The clock used for true RNG, is derived (selected by software) from one of the following sources:
  - PLL VCO (PLLQCLK) (only available in Run mode)
  - MSI (only available in Run mode)
  - LSI clock
  - LSE clock
- The ADC clock is derived (selected by software) from one of the following sources:
  - system clock (SYSCLK) (only available in Run mode)
  - HSI16 clock (only available in Run mode)
  - PLL VCO (PLLCLK) (only available in Run mode)
- The DAC uses the LSI clock in sample and hold mode
- The (LP)U(S)ARTs clocks are derived (selected by software) from one of the following sources:
  - system clock (SYSCLK) (only available in Run mode)
  - HSI16 clock (available in Run and Stop modes)
  - LSE clock (available in Run and Stop modes)
  - APB clock (PCLK depending on which APB the U(S)ART is mapped) (available in CRun and CSleep when also enabled in (LP)U(S)ARTxSMEN.)

The wakeup from Stop mode is supported only when the clock is HSI16 or LSE.

- The I2Cs clocks are derived (selected by software) from one of the following sources:
  - system clock (SYSCLK) (only available in Run mode)
  - HSI16 clock (available in Run and Stop modes)
  - APB clock (PCLK depending on which APB the I2C is mapped) (available in CRun and CSleep when also enabled in I2CxSMEN.)

The wakeup from Stop mode is supported only when the clock is HSI16.

- The SPI2S2 I2S clock is derived (selected by software) from one of the following sources:
  - HSI16 clock (only available in Run mode)
  - PLL VCO (PLLQCLK) (only available in Run mode)
  - external input I2S\_CK (available in Run and Stop modes)
- The low-power timers (LPTIMx) clock is derived (selected by software) from one of the following sources:
  - LSI clock (available in Run and Stop modes)
  - LSE clock (available in Run and Stop modes)
  - HSI16 clock (only available in Run mode)
  - APB clock (PCLK depending on which APB the LPTIMx is mapped) (available in Run and CStop when enabled in LPTIMxSMEN.)
  - external clock mapped on LPTIMx\_IN1 (available in Run and Stop modes)

The functionality in Stop mode (including wakeup) is supported only when the clock is LSI or LSE, or in external clock mode.

- The RTC clock is derived (selected by software) from one of the following sources:
  - LSE clock
  - LSI clock
  - HSE32 clock divided by 32

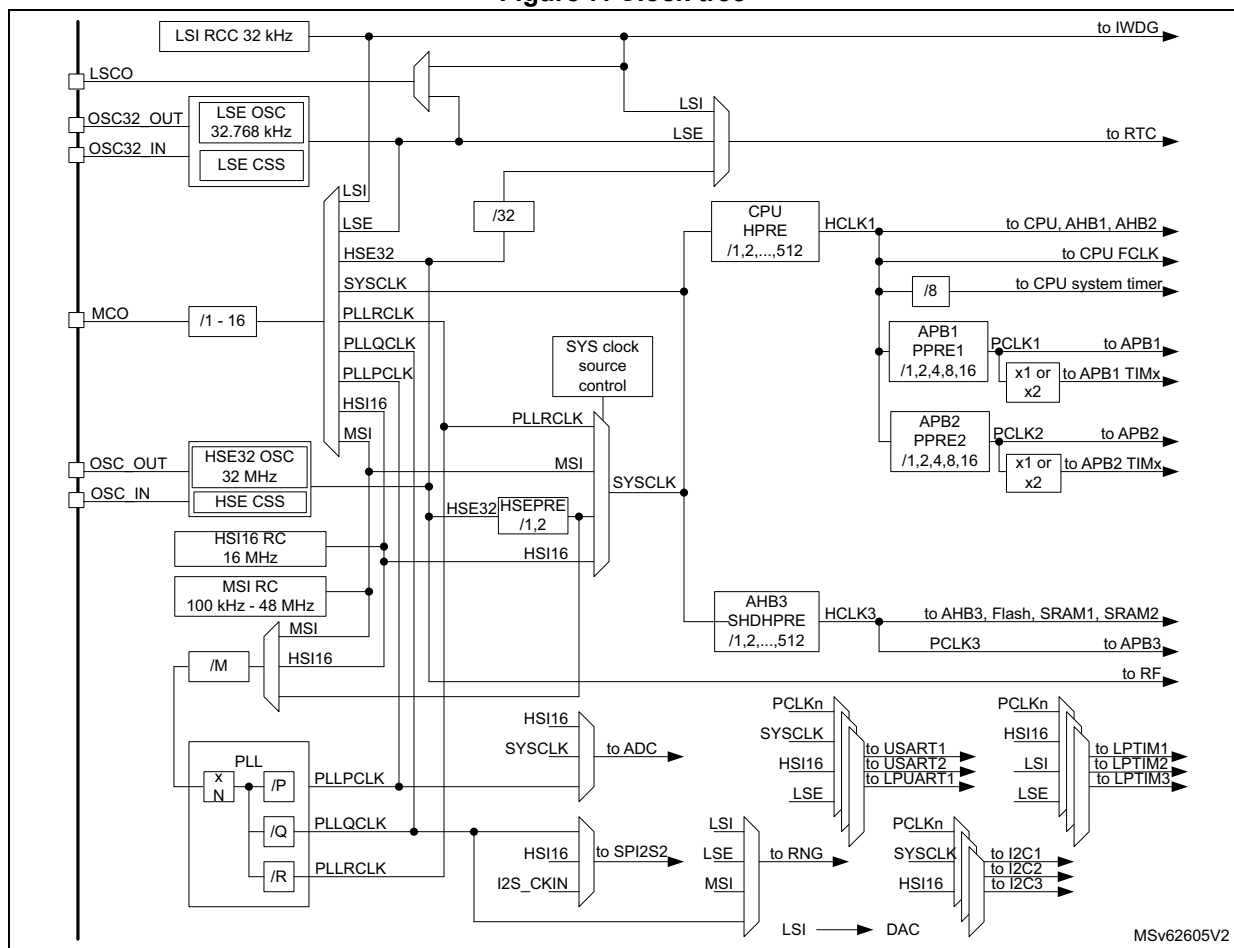
The functionality in Stop mode (including wakeup) is supported only when the clock is LSI or LSE.

- The IWDG clock is always the LSI clock.

The RCC feeds the CPU system timer (SysTick) external clock with the AHB clock (HCLK1) divided by eight. The SysTick can work either with this clock or directly with the CPU clock (HCLK1), configurable in the SysTick control and status register.

FCLK1 acts as CPU free-running clock. For more details, refer to the programming manual *STM32 Cortex®-M4 MCUs and MPUs programming manual (PM0214)*.

Figure 7. Clock tree



- For full details about the internal and external clock source characteristics, refer to the electrical characteristics section in the device datasheet.
- The ADC clock can additionally be derived from the AHB clock of the ADC bus interface, divided by a programmable factor (1, 2 or 4). When the programmable factor is 1, the AHB prescaler must be equal to 1.

### 3.13 General-purpose inputs/outputs (GPIOs)

Each of the GPIO pins can be configured by software as output (push-pull or open-drain), as input (with or without pull-up or pull-down) or as peripheral alternate function. Most of the GPIO pins are shared with digital or analog alternate functions. Fast I/O toggling can be achieved thanks to their mapping on the AHB2 bus.

The I/Os alternate function configuration can be locked if needed following a specific sequence in order to avoid spurious writing to the I/Os registers.

### 3.14 Directly memory access controller (DMA)

The DMA (direct memory access) is used to provide high-speed data transfer between peripherals and memory, as well as memory to memory. Data can be quickly moved by DMA without any CPU actions. This keeps CPU resources free for other operations.

The DMA controller has 14 channels in total. A full cross matrix allows the peripherals, with DMA support, to be mapped on any of the available DMA channels. Each DMA channel has an arbiter for handling the priority between DMA requests.

The DMA main features are listed below:

- 14 independently configurable channels (requests)
- a full cross matrix between peripherals and all 14 channels and an hardware trigger possibility through the DMAMUX1
- software programmable priorities between requests from channels of one DMA (four levels: very-high, high, medium, low), plus hardware priorities management in case of equality (example: request 1 has priority over request 2)
- independent source and destination transfer size (byte, half-word, word), emulating packing and unpacking. Source/destination addresses must be aligned on the data size.
- support for circular buffer management
- three event flags (DMA half-transfer, DMA transfer complete and DMA transfer error), logically ORed together in a single interrupt request for each channel
- memory-to-memory transfer
- peripheral-to-memory, memory-to-peripheral, and peripheral-to-peripheral transfers
- access to Flash memory, SRAM, APB and AHB peripherals, as source and destination
- programmable number of data to be transferred (up to 65536)

**Table 11. DMA1 and DMA2 implementation**

Feature	DMA1	DMA2
Number of channels	7	7

DMAMUX1 is used to route the peripherals with DMA source support, to any DMA channel.

## 3.15 Interrupts and events

### 3.15.1 Nested vectored interrupt controller (NVIC)

The devices embed an NVIC able to manage 16 priority levels, and to handle up to 62 maskable interrupt channels plus the 16 interrupt lines of the Cortex-M4.

The NVIC benefits are the following:

- low-latency interrupt processing
- interrupt entry vector table address passed directly to the core
- early processing of interrupts
- processing of late-arriving higher-priority interrupts
- support for tail chaining
- processor state automatically saved
- interrupt entry restored on interrupt exit, with no instruction overhead

The NVIC hardware block provides flexible interrupt management features with minimal interrupt latency.

### 3.15.2 Extended interrupt/event controller (EXTI)

The EXTI manages wakeup through configurable and direct event inputs. It provides wake-up requests to the power control, and generates interrupt requests to the CPU NVIC and events to the CPU event input.

Configurable events/interrupts come from peripherals that are able to generate a pulse and allow the selection between the event/interrupt trigger edge and a software trigger.

Direct events/interrupts come from peripherals having their own clearing mechanism.

## 3.16 Analog-to-digital converter (ADC)

A native 12-bit ADC is embedded into the devices. It can be extended to 16-bit resolution through hardware oversampling. The ADC has up to 12 external channels and four internal channels (temperature sensor, voltage reference, VBAT monitoring, DAC output). The ADC performs conversions in single-shot or scan mode. In scan mode, automatic conversion is performed on a selected group of analog inputs.

The ADC frequency is independent from the CPU frequency, allowing maximum sampling rate of ~2 Msps even with a low CPU speed. An auto-shutdown function guarantees that the ADC is powered off except during the active conversion phase.

The ADC can be served by the DMA controller. It can operate in the whole  $V_{DD}$  supply range.

The ADC features a hardware oversampler up to 256 samples, improving the resolution to 16 bits. Refer to the application note *Improving STM32 Series ADC resolution by oversampling* (AN2668).

An analog watchdog feature allows very precise monitoring of the converted voltage of one, some or all scanned channels. An interrupt is generated when the converted voltage is outside the programmed thresholds.

The events generated by the general-purpose timers (TIMx) can be internally connected to the ADC start triggers, to allow the application to synchronize A/D conversions with timers.

### 3.16.1 Temperature sensor

The temperature sensor (TS) generates a  $V_{TS}$  voltage that varies linearly with temperature.

The temperature sensor is internally connected to the ADC VIN[12] input channel, to convert the sensor output voltage into a digital value.

The sensor provides good linearity but it has to be calibrated to obtain good overall accuracy of the temperature measurement. As the offset of the temperature sensor may vary from part to part due to process variation, the uncalibrated internal temperature sensor is suitable only for relative temperature measurements.

To improve the accuracy of the temperature sensor, each part is individually factory-calibrated by ST. The resulting calibration data is stored in the device engineering bytes, accessible in read-only mode.

**Table 12. Temperature sensor calibration values**

Calibration value name	Description	Memory address
TS-CAL1	TS ADC raw data acquired at 30 °C ( $\pm 5$ °C), $V_{DDA} = V_{REF+} = 3.3$ V ( $\pm 10$ mV)	0x1FFF 75A8 - 0x1FFF 75A9
TS_CAL2	TS ADC raw data acquired at 130 °C ( $\pm 5$ °C), $V_{DDA} = V_{REF+} = 3.3$ V ( $\pm 10$ mV)	0x1FFF 75C8 - 0x1FFF 75C9

### 3.16.2 Internal voltage reference ( $V_{REFINT}$ )

$V_{REFINT}$  provides a stable (bandgap) voltage output for the ADC and comparators.  $V_{REFINT}$  is internally connected to the ADC VIN[13] input channel.

$V_{REFINT}$  is individually and precisely measured, for each part, by ST, during production test and stored in the device engineering bytes. It is accessible in read-only mode.

**Table 13. Internal voltage reference calibration values**

Calibration value name	Description	Memory address
VREFINT	Raw data acquired at 30 °C ( $\pm 5$ °C), $V_{DDA} = V_{REF+} = 3.0$ V ( $\pm 10$ mV)	0x1FFF 75AA - 0x1FFF 75AB

### 3.16.3 $V_{BAT}$ battery voltage monitoring

This embedded hardware feature allows the application to measure the  $V_{BAT}$  battery voltage using the ADC VIN[14] input channel. As  $V_{BAT}$  may be higher than  $V_{DDA}$ , and thus outside the ADC input range, the VBAT pin is internally connected to a bridge divider by three. As a consequence, the converted digital value is one third the  $V_{BAT}$  voltage.

### 3.17 Voltage reference buffer (VREFBUF)

The devices embed a voltage reference buffer that can be used as voltage reference for ADC, and also as voltage reference for external components through the VREF+ pin.

VREFBUF supports two voltages: 2.048 V and 2.5 V.

An external voltage reference can be provided through the VREF+ pin when VREFBUF is off.

### 3.18 Digital-to-analog converter (DAC)

The 1-channel 12-bit buffered DAC converts a digital value into an analog voltage available on the channel output. The architecture of each channel is based on an integrated resistor string and an inverting amplifier. The digital circuitry is common for both channels.

DAC main features:

- 1 DAC output channel
- 8-bit or 12-bit output mode
- buffer offset calibration (factory and user trimming)
- left or right data alignment in 12-bit mode
- synchronized update capability
- noise-wave generation
- triangular-wave generation
- independent or simultaneous conversion for DAC channels
- DMA capability for either DAC channel
- triggering with timer events, synchronized with DMA
- triggering with external events

Sample-and-hold low-power mode, with internal or external capacitor

### 3.19 Comparator (COMP)

The devices embed two rail-to-rail comparators with programmable reference voltage (internal or external), hysteresis and speed (low speed for low-power) and with selectable output polarity.

The reference voltage can be one of the following:

- external I/O
- internal reference voltage or submultiple (1/4, 1/2, 3/4)

All comparators can wake up from Stop mode, generate interrupts and breaks for the timers, and can be also combined into a window comparator.

### 3.20 True random number generator (RNG)

The devices embed a true RNG that delivers 32-bit random numbers generated by an integrated analog circuitry.

### 3.21 Advanced encryption standard hardware accelerator (AES)

The AES encrypts or decrypts data, using an algorithm and implementation fully compliant with the advanced encryption standard (AES) defined in FIPS (federal information processing standards) publication 197.

Multiple chaining modes are supported (ECB, CBC, CTR, GCM, GMAC, CCM), for key sizes of 128 or 256 bits. The AES supports DMA single transfers for incoming and outgoing data (two DMA channels required).

### 3.22 Public key accelerator (PKA)

The PKA is used to compute cryptographic public key primitives, specifically those related to RSA (Rivest, Shamir and Adleman), Diffie-Hellmann or ECC (elliptic curve cryptography) over GF(p) (Galois fields). These operations are executed in the Montgomery domain.

### 3.23 Timer and watchdog

The devices include one advanced 16-bit timer, one general-purpose 32-bit timer, two 16-bit basic timers, three low-power timers, two watchdog timers and a SysTick timer.

The table below compares the features of the advanced control, general purpose and basic timers.

**Table 14. Timer features**

Timer type	Timer name	Counter resolution (bits)	Counter type	Prescaler factor	DMA request generation	Capture/compare channels	Complementary outputs
Advanced control	TIM1	16	Up, down and up/down	Any integer between 1 and 65536	Yes	4	3
General purpose	TIM2	32					NA
	TIM16	16	Up			2	1
	TIM17						
Low power	LPTIM1 LPTIM2 LPTIM3						

#### 3.23.1 Advanced-control timer (TIM1)

The advanced-control timer TIM1 can be seen as a three-phase PWM multiplexed on six channels. Each channel has complementary PWM outputs with programmable inserted dead-times. Each channel can also be seen as complete general-purpose timers.



The four independent channels can be used for:

- input capture
- output compare
- PWM generation (edge or center-aligned modes) with full modulation capability (0 - 100 %)
- one-pulse mode output

In debug mode, the TIM1 counter can be frozen and the PWM outputs disabled to turn off any power switches driven by these outputs.

Many features are shared with those of the general-purpose timers (described in the next section) using the same architecture. TIM1 can then work together with TIM2 via the peripheral interconnect matrix, for synchronization or event chaining.

### 3.23.2 General-purpose timers (TIM2, TIM16, TIM17)

Each general-purpose timer can be used to generate PWM outputs, or act as a simple time base.

TIM2 main features:

- full-featured general-purpose timer
- four independent channels for input capture/output compare, PWM or one-pulse mode output
- counter that can be frozen in debug mode
- independent DMA request generation, support of quadrature encoders

TIM16 and TIM17 main features:

- general-purpose timers with mid-range features
- 16-bit auto-reload upcounters and 16-bit prescalers
- 1 channel and 1 complementary channel
- channels that can all be used for input capture/output compare, PWM or one-pulse mode output
- counter that can be frozen in debug mode
- independent DMA request generation

### 3.23.3 Low-power timers (LPTIM1, LPTIM2 and LPTIM3)

These low-power timers have an independent clock and run in Stop mode if they are clocked by LSE, LSI, or by an external clock. They are able to wake up the system from Stop mode.

LPTIM1 is active in Stop 0, Stop 1 and Stop 2 modes.

LPTIM2 and LPTIM3 are active in Stop 0 and Stop 1 modes.

LPTIM1/2/3 main features:

- 16-bit up counter with 16-bit autoreload register
- 16-bit compare register
- configurable output: pulse, PWM
- continuous/one-shot mode
- selectable software/hardware input trigger
- selectable clock source
- internal clock sources: LSE, either LSI, HSI16 or APB clock
- external clock source over LPTIM input (works even with no internal clock source running, used by pulse counter application)
- programmable digital glitch filter
- encoder mode (LPTIM1 only)

#### 3.23.4 Independent watchdog (IWDG)

The independent watchdog is based on a 12-bit downcounter and a 8-bit prescaler. The IWDG is clocked from an independent 32 kHz internal RC (LSI). As the IWDG operates independently from the main clock, it can operate in Stop and Standby modes.

The IWDG can be used either as a watchdog to reset the device when a problem occurs, or as a free running timer for application timeout management. The IWDG is hardware or software configurable through the option bytes. The counter can be frozen in debug mode.

#### 3.23.5 System window watchdog (WWDG)

The window watchdog is based on a 7-bit downcounter that can be set as free running. The WWDG can be used as a watchdog to reset the device when a problem occurs.

The WWDG is clocked from the main clock and has an early warning interrupt capability. The counter can be frozen in debug mode.

#### 3.23.6 SysTick timer

This timer is dedicated to real-time operating systems, but can also be used as a standard down counter.

SysTick timer main features:

- 24-bit down counter
- autoreload capability
- maskable system interrupt generation when the counter reaches 0
- programmable clock source

### 3.24 Real-time clock (RTC), tamper and backup registers

The RTC is an independent BCD timer/counter. The RTC provides a time-of-day clock/calendar with programmable alarm interrupts.

As long as the supply voltage remains in the operating range, the RTC never stops, regardless of the device status (Run mode, low-power mode or under reset).

The RTC provides an automatic wakeup to manage all low-power modes.

The RTC is functional in VBAT mode.

Twenty 32-bit backup registers are retained in all low-power modes and also in VBAT mode. These registers can be used to store sensitive data as their content is protected by a tamper detection circuit.

Three tamper pins and four internal tampers are available for anti-tamper detection. The external tamper pins can be configured for edge or level detection with or without filtering.

### 3.25 Inter-integrated circuit interface (I2C)

The device embeds three I2Cs, with features implementation listed in the table below.

The I<sup>2</sup>C bus interface handles communications between the microcontroller and the serial I<sup>2</sup>C bus. It controls all I<sup>2</sup>C bus-specific sequencing, protocol, arbitration and timing.

The I2C peripheral supports:

- I<sup>2</sup>C bus specification and user manual rev. 5 compatibility:
  - slave and master modes, multimaster capability
  - Standard-mode (Sm), with a bitrate up to 100 Kbit/s
  - Fast-mode (Fm), with a bitrate up to 400 Kbit/s
  - Fast-mode plus (Fm+), with a bitrate up to 1 Mbit/s and 20 mA output drive I/Os
  - 7-bit and 10-bit addressing mode, multiple 7-bit slave addresses
  - programmable setup and hold times
  - clock stretching (optional)
- SMBus (system management bus) specification rev 2.0 compatibility:
  - hardware PEC (packet error checking) generation and verification with ACK control
  - address resolution protocol (ARP) support
  - SMBus alert
- PMBus (power system management protocol) specification rev 1.1 compatibility
- independent clock: a choice of independent clock sources allowing the I<sup>2</sup>C communication speed to be independent from the PCLK reprogramming (see [Figure 7](#))
- wakeup from Stop mode on address match
- programmable analog and digital noise filters
- 1-byte buffer with DMA capability

**Table 15. I2C implementation**

I2C features <sup>(1)</sup>	I2C1 <sup>(2)</sup>	I2C2 <sup>(2)</sup>	I2C3
7-bit addressing mode	X	X	X
10-bit addressing mode	X	X	X
Standard-mode (up to 100 Kbit/s)	X	X	X
Fast-mode (up to 400 Kbit/s)	X	X	X
Fast-mode Plus with 20mA output drive I/Os (up to 1 Mbit/s)	X	X	X
Independent clock	X	X	X

Table 15. I2C implementation (continued)

I2C features <sup>(1)</sup>	I2C1 <sup>(2)</sup>	I2C2 <sup>(2)</sup>	I2C3
Wakeup from Stop mode	X <sup>(3)</sup>	X <sup>(3)</sup>	X <sup>(4)</sup>
SMBus/PMBus	X	X	X

1. X = supported.
2. The register content is lost in Stop 2 mode.
3. Wakeup supported from Stop 0 and Stop 1 modes.
4. Wakeup supported from Stop 0, Stop 1 and Stop 2 modes.

### 3.26 Universal synchronous/asynchronous receiver transmitter (USART/UART)

The devices embed two universal synchronous receiver transmitters, USART1 and USART2 (see [Table 16](#) for the implementation details).

Each USART provides asynchronous communication, IrDA SIR ENDEC support, multiprocessor communication mode, single-wire half-duplex communication mode. Each USART has LIN Master/Slave capability and provides hardware management of the CTS and RTS signals, and RS485 driver enable.

The USART is able to communicate at speeds of up to 4 Mbit/s, and also provides Smart Card mode (ISO 7816 compliant) and SPI-like communication capability.

The USART supports synchronous operation (SPI mode), and can be used as an SPI master.

The USART has a clock domain independent from the CPU clock, allowing the USART to wake up the MCU from Stop mode, using baudrates up to 200 kbaud.

The wakeup events from Stop mode are programmable and can be one of the following:

- start bit detection
- any received data frame
- a specific programmed data frame

The USART interface can be served by the DMA controller.

### 3.27 Low-power universal asynchronous receiver transmitter (LPUART)

The devices embed one low-power UART (LPUART1) that enables asynchronous serial communication with minimum power consumption. The LPUART supports half-duplex single-wire communication and modem operations (CTS/RTS), allowing multiprocessor communication.

The LPUART has a clock domain independent from the CPU clock, and can wake up the system from Stop mode using baudrates up to 220 Kbaud. The wakeup events from Stop mode are programmable and can be one of the following:

- start bit detection
- any received data frame
- a specific programmed data frame

Only a 32.768 kHz clock (LSE) is needed to allow LPUART communication up to 9600 baud. Therefore, even in Stop mode, the LPUART can wait for an incoming frame while having an extremely low-energy consumption. Higher speed clock can be used to reach higher baudrates.

The LPUART interface can be served by the DMA controller.

**Table 16. USART/LPUART features**

USART modes/features <sup>(1)</sup>	USART1/2	LPUART1
Hardware flow control for modem	X	X
Continuous communication using DMA	X	X
Multiprocessor communication	X	X
Synchronous mode (Master/Slave)	X	-
Smartcard mode	X	-
Single-wire half-duplex communication	X	X
IrDA SIR ENDEC block	X	-
LIN mode	X	-
Dual clock domain and wakeup from low-power mode	X	X
Receiver timeout interrupt	X	-
Modbus communication	X	-
Auto baud rate detection	X	-
Driver enable	X	X
USART data length	7, 8 and 9 bits	
Tx/Rx FIFO	X	X
Tx/Rx FIFO size	8	

1. X = supported.

### 3.28 Serial peripheral interface (SPI)/integrated-interchip sound interface (I2S)

The SPI/I2S interface can be used to communicate with external devices using the SPI protocol or the I<sup>2</sup>S audio protocol. SPI or I2S mode is selectable by software. SPI Motorola<sup>®</sup> mode is selected by default after a device reset.

The SPI protocol supports half-duplex, full-duplex and simplex synchronous, serial communication with external devices. The SPI interface can be configured as master and, in

this case, it provides the communication clock (SCK) to the external slave device. The SPI interface can also operate in multimaster configuration.

The I<sup>2</sup>S protocol is also a synchronous serial communication interface. It can operate in slave or master mode with half-duplex communication. It can address four different audio standards including the Philips I<sup>2</sup>S standard, the MSB- and LSB-justified standards and the PCM standard.

**Table 17. SPI and SPI/I2S implementation<sup>(1)</sup>**

Features	SPI1	SPI2S2	SUBGHZSPI
Enhanced NSSP and TI modes	Yes		
Hardware CRC calculation	Yes	Yes	No
I <sup>2</sup> S support	No	Yes	No
Data size configurable (bits)	from 4 to 16		
Rx/Tx FIFO size (bits)	32		
Wakeup capability from LPSleep	Yes		

1. The SPI1 and SPI2S2 instances are general purpose type while the SUBGHZSPI instance is dedicated for Sub-GHz radio control exclusively. Radio is controlled internally through SUBGHZSPI and, for debug purpose only, from the external.

## 3.29 Development support

### Serial-wire JTAG debug port (SWJ-DP)

The Arm SWJ-DP interface is embedded, and is a combined JTAG and serial-wire debug port, that enables either a serial-wire debug or a JTAG probe to be connected to the target.

The debug is performed using only two pins instead of the five required by the JTAG (JTAG pins can then be reused as GPIOs with alternate function). The JTAG TMS and TCK pins are shared with SWDIO and SWCLK, respectively, and a specific sequence on the TMS pin is used to switch between JTAG-DP and SW-DP.

## 4 Pinouts, pin description and alternate functions

Figure 8. UFBGA73 pinout

	1	2	3	4	5	6	7	8	9
A	VSSSMPS	VDDSMPS	-	PA14	VDDA	-	VDD	VBAT	PA12
B	VLXSMPS	VFBSMPS	PA15	PB15	VREF+	PC14- OSC32_IN	VSS	PA13	PA11
C	PB3	PB4	PB7	PB9	PC15- OSC32_OUT	PB14	PC13	PA10	-
D	-	PB5	PB8	PC2	PC3	PA0	PB13	PB2	VSS
E	PB6	VDD	VSS	PC5	PA9	PB12	PB1	VDDRF	VDD
F	PC1	PC0	PC4	PA6	NRST	PB0- VDD_TCXO	VDDRF1V55	OSC_OUT	-
G	-	PC6	PA1	PB11	VSS	VSSRF	VSSRF	VSSRF	OSC_IN
H	PA3	PA2	PA7	PB10	VDD	VSSRF	RFL_N	VDDPA	VR_PA
J	PA4	PA5	PA8	-	PH3	RFL_P	-	RFO_LP	RFO_HP

MSv48145V2

1. The above figure shows the package top view.

Table 18. Legend/abbreviations used in the pinout table

Name		Abbreviation	Definition
Pin name		Unless otherwise specified in brackets below the pin name, the pin function during and after reset is the same as the actual pin name	
Pin type		S	Supply pin
		I	Input only pin
		I/O	Input / output pin
		O	Output only pin
I/O structure		FT	5 V tolerant I/O
		RF	Radio RF pin
		Option for FT I/Os	
		_f	I/O, Fm+ capable
		_a	I/O, with Analog switch function supplied by V <sub>DDA</sub>
Notes		Unless otherwise specified by a note, all I/Os are set as analog inputs during and after reset.	
Pin functions	Alternate functions	Functions selected through GPIOx_AFR registers	
	Additional functions	Functions directly selected/enabled through peripheral registers	

Table 19. STM32WLE5J8/JB/JC pin definition

Pin number	Pin name (function after reset)	Pin type	I/O structure	Notes	Alternate functions	Additional functions
UFBGA73						
C1	PB3	I/O	FT_a	-	JTDO/TRACESWO, TIM2_CH2, SPI1_SCK, RF_IRQ0, USART1_RTS, CM4_EVENTOUT	COMP1_INM, COMP2_INM, ADC1_IN2, TAMP_IN3/WKUP3
C2	PB4	I/O	FT_fa	-	NJTRST, I2C3_SDA, SPI1_MISO, USART1_CTS, DEBUG_RF_LDORDY, TIM17_BKIN, CM4_EVENTOUT	COMP1_INP, COMP2_INP, ADC1_IN3
D2	PB5	I/O	FT_a	-	LPTIM1_IN1, I2C1_SMBA, SPI1_MOSI, RF_IRQ1, USART1_CK, COMP2_OUT, TIM16_BKIN, CM4_EVENTOUT	-
E3	VSS	S	-	-	-	-
E2	VDD	S	-	-	-	-
E1	PB6	I/O	FT_f	-	LPTIM1_ETR, I2C1_SCL, USART1_TX, TIM16_CH1N, CM4_EVENTOUT	-
C3	PB7	I/O	FT_f	-	LPTIM1_IN2, TIM1_BKIN, I2C1_SDA, USART1_RX, TIM17_CH1N, CM4_EVENTOUT	-
D3	PB8	I/O	FT_f	-	TIM1_CH2N, I2C1_SCL, RF_IRQ2, TIM16_CH1, CM4_EVENTOUT	-
C4	PB9	I/O	FT_f	-	TIM1_CH3N, I2C1_SDA, SPI2_NSS/I2S2_WS, IR_OUT, TIM17_CH1, CM4_EVENTOUT	-
F2	PC0	I/O	FT_f	-	LPTIM1_IN1, I2C3_SCL, LPUART1_RX, LPTIM2_IN1, CM4_EVENTOUT	-
F1	PC1	I/O	FT_f	-	LPTIM1_OUT, SPI2_MOSI/I2S2_SD, I2C3_SDA, LPUART1_TX, CM4_EVENTOUT	-
D4	PC2	I/O	FT	-	LPTIM1_IN2, SPI2_MISO, CM4_EVENTOUT	-
D5	PC3	I/O	FT	-	LPTIM1_ETR, SPI2_MOSI/I2S2_SD, LPTIM2_ETR, CM4_EVENTOUT	-
F3	PC4	I/O	FT	-	CM4_EVENTOUT	-
E4	PC5	I/O	FT	-	CM4_EVENTOUT	-



Table 19. STM32WLE5J8/JB/JC pin definition (continued)

Pin number	Pin name (function after reset)	Pin type	I/O structure	Notes	Alternate functions	Additional functions
UFBGA73						
G2	PC6	I/O	FT	-	I2S2_MCK, CM4_EVENTOUT	-
D6	PA0	I/O	FT_a	-	TIM2_CH1, I2C3_SMBA, I2S_CKIN, USART2_CTS, COMP1_OUT, TIM2_ETR, CM4_EVENTOUT	TAMP_IN2/WKUP1
G3	PA1	I/O	FT_a	-	TIM2_CH2, LPTIM3_OUT, I2C1_SMBA, SPI1_SCK, USART2_RTS, LPUART1_RTS, CM4_EVENTOUT	-
H2	PA2	I/O	FT_a	-	LSCO, TIM2_CH3, USART2_TX, LPUART1_TX, COMP2_OUT, DEBUG_PWR_LDORDY, CM4_EVENTOUT	LSCO
H1	PA3	I/O	FT_a	-	TIM2_CH4, I2S2_MCK, USART2_RX, LPUART1_RX, CM4_EVENTOUT	-
G5	VSS	S	-	-	-	-
H5	VDD	S	-	-	-	-
J1	PA4	I/O	FT	-	RTC_OUT2, LPTIM1_OUT, SPI1_NSS, USART2_CK, DEBUG_SUBGHZSPI_ NSSOUT, LPTIM2_OUT, CM4_EVENTOUT	-
J2	PA5	I/O	FT	-	TIM2_CH1, TIM2_ETR, SPI2_MISO, SPI1_SCK, DEBUG_SUBGHZSPI_ SCKOUT, LPTIM2_ETR, CM4_EVENTOUT	-
F4	PA6	I/O	FT	-	TIM1_BKIN, I2C2_SMBA, SPI1_MISO, LPUART1_CTS, DEBUG_SUBGHZSPI_ MISOOUT, TIM16_CH1, CM4_EVENTOUT	-
H3	PA7	I/O	FT_fa	-	TIM1_CH1N, I2C3_SCL, SPI1_MOSI, COMP2_OUT, DEBUG_SUBGHZSPI_ MOSIOUT, TIM17_CH1, CM4_EVENTOUT	-
J3	PA8	I/O	FT_a	-	MCO, TIM1_CH1, SPI2_SCK/I2S2_CK, USART1_CK, LPTIM2_OUT, CM4_EVENTOUT	-

Table 19. STM32WLE5J8/JB/JC pin definition (continued)

Pin number	Pin name (function after reset)	Pin type	I/O structure	Notes	Alternate functions	Additional functions
UFBGA73						
E5	PA9	I/O	FT_fa	-	TIM1_CH2, SPI2_NSS/I2S2_WS, I2C1_SCL, SPI2_SCK/I2S2_CK, USART1_TX, CM4_EVENTOUT	-
H4	PB10	I/O	FT_f	-	TIM2_CH3, I2C3_SCL, SPI2_SCK/I2S2_CK, LPUART1_RX, COMP1_OUT, CM4_EVENTOUT	-
G4	PB11	I/O	FT_f	-	TIM2_CH4, I2C3_SDA, LPUART1_TX, COMP2_OUT, CM4_EVENTOUT	-
F5	NRST	I/O	FT	-	-	-
J5	PH3	I/O	FT	-	CM4_EVENTOUT	BOOT0
H6	VSSRF	S	-	-	-	-
G6	VSSRF	S	-	-	-	-
J6	RFI_P	I	RF	-	-	-
H7	RFI_N	I	RF	-	-	-
G7	VSSRF	S	-	-	-	-
-	VSSRF	S	-	-	-	-
J8	RFO_LP	O	RF	-	-	-
G8	VSSRF	S	-	-	-	-
J9	RFO_HP	O	RF	-	-	-
-	VSSRF	S	-	-	-	-
H9	VR_PA	S	-	-	-	-
H8	VDDPA	S	-	-	-	-
-	VSSRF	S	-	-	-	-
G9	OSC_IN	I	RF	-	-	-
F8	OSC_OUT	O	RF	-	-	-
-	VSSRF	S	-	-	-	-
E8	VDDRF	S	-	-	-	-
F7	VDDRF1V55	S	-	-	-	-
D9	VSS	S	-	-	-	-
E9	VDD	S	-	-	-	-
F6	PB0-VDD_TCXO	I/O	FT_a	-	COMP1_OUT, CM4_EVENTOUT	-

Table 19. STM32WLE5J8/JB/JC pin definition (continued)

Pin number	Pin name (function after reset)	Pin type	I/O structure	Notes	Alternate functions	Additional functions
UFBGA73						
E7	PB1	I/O	FT_a	-	LPUART1_RTS_DE, LPTIM2_IN1, CM4_EVENTOUT	COMP2_INP, ADC1_IN5
D8	PB2	I/O	FT_a	-	LPTIM1_OUT, I2C3_SMBA, SPI1_NSS, DEBUG_RF_SMPSTRDY, CM4_EVENTOUT	COMP1_INP, COMP2_INM, ADC1_IN4
E6	PB12	I/O	FT	-	TIM1_BKIN, I2C3_SMBA, SPI2_NSS/I2S2_WS, LPUART1_RTS, CM4_EVENTOUT	-
D7	PB13	I/O	FT_fa	-	TIM1_CH1N, I2C3_SCL, SPI2_SCK/I2S2_CK, LPUART1_CTS, CM4_EVENTOUT	ADC1_IN0
C6	PB14	I/O	FT_fa	-	TIM1_CH2N, I2S2_MCK, I2C3_SDA, SPI2_MISO, CM4_EVENTOUT	ADC1_IN1
C8	PA10	I/O	FT_fa	-	RTC_REFIN, TIM1_CH3, I2C1_SDA, SPI2_MOSI/I2S2_SD, USART1_RX, DEBUG_RF_HSE32RDY, TIM17_BKIN, CM4_EVENTOUT	COMP1_INM, COMP2_INM, DAC1_OUT, ADC1_IN6
B9	PA11	I/O	FT_fa	-	TIM1_CH4, TIM1_BKIN2, LPTIM3_ETR, I2C2_SDA, SPI1_MISO, USART1_CTS, DEBUG_RF_NRESET, CM4_EVENTOUT	COMP1_INM, COMP2_INM, ADC1_IN7
A9	PA12	I/O	FT_fa	-	TIM1_ETR, LPTIM3_IN1, I2C2_SCL, SPI1_MOSI, RF_BUSY, USART1_RTS, CM4_EVENTOUT	ADC1_IN8
B8	PA13	I/O	FT_a	-	JTMS-SWDIO, I2C2_SMBA, IR_OUT, CM4_EVENTOUT	ADC1_IN9
B7	VSS	S	-	-	-	-
A7	VDD	S	-	-	-	-
A8	VBAT	S	-	-	-	-
C7	PC13	I/O	FT	-	CM4_EVENTOUT	TAMP_IN1/ RTC_OUT1/RTC_TS/W KUP2
B6	PC14-OSC32_IN	I/O	FT	-	CM4_EVENTOUT	OSC32_IN

Table 19. STM32WLE5J8/JB/JC pin definition (continued)

Pin number	Pin name (function after reset)	Pin type	I/O structure	Notes	Alternate functions	Additional functions
UFBGA73						
C5	PC15-OSC32_OUT	I/O	FT	-	CM4_EVENTOUT	OSC32_OUT
B5	VREF+	S	-	-	-	-
A5	VDDA	S	-	-	-	-
A4	PA14	I/O	FT_a	-	JTCK-SWCLK, LPTIM1_OUT, I2C1_SMBA, CM4_EVENTOUT	ADC1_IN10
B3	PA15	I/O	FT_fa	-	JTDI, TIM2_CH1, TIM2_ETR, I2C2_SDA, SPI1_NSS, CM4_EVENTOUT	COMP1_INM, COMP2_INP, ADC1_IN11
B4	PB15	I/O	FT_f		TIM1_CH3N, I2C2_SCL, SPI2_MOSI/I2S2_SD, CM4_EVENTOUT	-
B2	VFBSMPS	S	-	-	-	-
A2	VDDSMPS	S	-	-	-	-
B1	VLXSMPS	S	-	-	-	-
A1	VSSSMPS	S	-	-	-	-

Table 20. Alternate functions

Port		AF0	AF1	AF2	AF3	AF4	AF5	AF6	AF7	AF8	AF9	AF10	AF11	AF12	AF13	AF14	AF15
		SYS_AF	TIM1/ TIM2/ LPTIM1	TIM1/ TIM2	SPI2S2 TIM1/ LPTIM3	I2C1/ I2C2/ I2C3	SPI1/ SPI2S2	RF	USART1 / USART2	LPUART1	-	-	-	COMP1/ COMP2/ TIM1	DEBUG	TIM2/ TIM16/ TIM17/ LPTIM2	EVENOUT
Port A	PA0	-	TIM2_CH1	-	-	I2C3_SMBA	I2S_CKIN	-	USART2_CTS	-	-	-	-	COMP1_OUT	-	TIM2_ETR	CM4_EVENTOUT
	PA1	-	TIM2_CH2	-	LPTIM3_OUT	I2C1_SMBA	SPI1_SCK	-	USART2_RTS	LPUART1_RTS	-	-	-	-	-	-	CM4_EVENTOUT
	PA2	LSCO	TIM2_CH3	-	-	-	-	-	USART2_TX	LPUART1_TX	-	-	-	COMP2_OUT	DEBUG_PWR_LDORDY	-	CM4_EVENTOUT
	PA3	-	TIM2_CH4	-	-	-	I2S2_MCK	-	USART2_RX	LPUART1_RX	-	-	-	-	-	-	CM4_EVENTOUT
	PA4	RTC_OUT2	LPTIM1_OUT	-	-	-	SPI1_NSS	-	USART2_CK	-	-	-	-	-	DEBUG_SUBGHZSPI_NSSOUT	LPTIM2_OUT	CM4_EVENTOUT
	PA5	-	TIM2_CH1	TIM2_ETR	SPI2_MISO	-	SPI1_SCK	-	-	-	-	-	-	-	DEBUG_SUBGHZSPI_SCKOUT	LPTIM2_ETR	CM4_EVENTOUT
	PA6	-	TIM1_BKIN	-	-	I2C2_SMBA	SPI1_MISO	-	-	LPUART1_CTS	-	-	-	TIM1_BKIN	DEBUG_SUBGHZSPI_MISOOUT	TIM16_CH1	CM4_EVENTOUT
	PA7	-	TIM1_CH1N	-	-	I2C3_SCL	SPI1_MOSI	-	-	-	-	-	-	COMP2_OUT	DEBUG_SUBGHZSPI_MOSIOUT	TIM17_CH1	CM4_EVENTOUT
	PA8	MCO	TIM1_CH1	-	-	-	SPI2_SCK/ I2S2_CK	-	USART1_CK	-	-	-	-	-	-	LPTIM2_OUT	CM4_EVENTOUT
	PA9	-	TIM1_CH2	-	SPI2_NSS/ I2S2_WS	I2C1_SCL	SPI2_SCK/ I2S2_CK	-	USART1_TX	-	-	-	-	-	-	-	CM4_EVENTOUT
	PA10	RTC_REFIN	TIM1_CH3	-	-	I2C1_SDA	SPI2_MOSI/ I2S2_SD	-	USART1_RX	-	-	-	-	-	DEBUG_RF_HSE32RDY	TIM17_BKIN	CM4_EVENTOUT
	PA11	-	TIM1_CH4	TIM1_BKIN2	LPTIM3_ETR	I2C2_SDA	SPI1_MISO	-	USART1_CTS	-	-	-	-	TIM1_BKIN2	DEBUG_RF_NRESET	-	CM4_EVENTOUT



Table 20. Alternate functions (continued)

Port		AF0	AF1	AF2	AF3	AF4	AF5	AF6	AF7	AF8	AF9	AF10	AF11	AF12	AF13	AF14	AF15
		SYS_AF	TIM1/ TIM2/ LPTIM1	TIM1/ TIM2	SPI2S2 TIM1/ LPTIM3	I2C1/ I2C2/ I2C3	SPI1/ SPI2S2	RF	USART1 / USART2	LPUART1	-	-	-	COMP1/ COMP2/ TIM1	DEBUG	TIM2/ TIM16/ TIM17/ LPTIM2	EVENOUT
Port A (continued)	PA12	-	TIM1_ETR	-	LPTIM3_IN1	I2C2_SCL	SPI1_MOSI	RF_BUSY	USART1_RTS	-	-	-	-	-	-	-	CM4_EVENTOUT
	PA13	JTMS-SWDIO	-	-	-	I2C2_SMBA	-	-	-	IR_OUT	-	-	-	-	-	-	CM4_EVENTOUT
	PA14	JTCK-SWCLK	LPTIM1_OUT	-	-	I2C1_SMBA	-	-	-	-	-	-	-	-	-	-	CM4_EVENTOUT
	PA15	JTDI	TIM2_CH1	TIM2_ETR	-	I2C2_SDA	SPI1_NSS	-	-	-	-	-	-	-	-	-	CM4_EVENTOUT
Port B	PB0	-	-	-	-	-	-	-	-	-	-	-	-	COMP1_OUT	-	-	CM4_EVENTOUT
	PB1	-	-	-	-	-	-	-	-	LPUART1_RTS_DE	-	-	-	-	-	LPTIM2_IN1	CM4_EVENTOUT
	PB2	-	LPTIM1_OUT	-	-	I2C3_SMBA	SPI1_NSS	-	-	-	-	-	-	-	DEBUG_RF_SMPSTRDY	-	CM4_EVENTOUT
	PB3	JTDO-TRACE SWO	TIM2_CH2	-	-	-	SPI1_SCK	-	USART1_RTS	-	-	-	-	-	-	-	CM4_EVENTOUT
	PB4	NJTRST	-	-	-	I2C3_SDA	SPI1_MISO	-	USART1_CTS	-	-	-	-	-	DEBUG_RF_LDORDY	TIM17_BKIN	CM4_EVENTOUT
	PB5	-	LPTIM1_IN1	-	-	I2C1_SMBA	SPI1_MOSI	-	USART1_CK	-	-	-	-	COMP2_OUT	-	TIM16_BKIN	CM4_EVENTOUT
	PB6	-	LPTIM1_ETR	-	-	I2C1_SCL	-	-	USART1_TX	-	-	-	-	-	-	TIM16_CH1N	CM4_EVENTOUT
	PB7	-	LPTIM1_IN2	-	TIM1_BKIN	I2C1_SDA	-	-	USART1_RX	-	-	-	-	-	-	TIM17_CH1N	CM4_EVENTOUT
	PB8	-	TIM1_CH2N	-	-	I2C1_SCL	-	-	-	-	-	-	-	-	-	TIM16_CH1	CM4_EVENTOUT
	PB9	-	TIM1_CH3N	-	-	I2C1_SDA	SPI2_NSS/ I2S2_WS	-	-	IR_OUT	-	-	-	-	-	TIM17_CH1	CM4_EVENTOUT
	PB10	-	TIM2_CH3	-	-	I2C3_SCL	SPI2_SCK/ I2S2_CK	-	-	LPUART1_RX	-	-	-	COMP1_OUT	-	-	CM4_EVENTOUT

Table 20. Alternate functions (continued)

Port		AF0	AF1	AF2	AF3	AF4	AF5	AF6	AF7	AF8	AF9	AF10	AF11	AF12	AF13	AF14	AF15
		SYS_AF	TIM1/ TIM2/ LPTIM1	TIM1/ TIM2	SPI2S2 TIM1/ LPTIM3	I2C1/ I2C2/ I2C3	SPI1/ SPI2S2	RF	USART1 / USART2	LPUART1	-	-	-	COMP1/ COMP2/ TIM1	DEBUG	TIM2/ TIM16/ TIM17/ LPTIM2	EVENOUT
Port B (continued)	PB11	-	TIM2_ CH4	-	-	I2C3_ SDA	-	-	-	LPUART1_ TX	-	-	-	COMP2_ OUT	-	-	CM4_ EVENTOUT
	PB12	-	TIM1_ BKIN	-	TIM1_ BKIN	I2C3_ SMBA	SPI2_ NSS/ I2S2_WS	-	-	LPUART1_ RTS	-	-	-	-	-	-	CM4_ EVENTOUT
	PB13	-	TIM1_ CH1N	-	-	I2C3_ SCL	SPI2_ SCK/ I2S2_CK	-	-	LPUART1_ CTS	-	-	-	-	-	-	CM4_ EVENTOUT
	PB14	-	TIM1_ CH2N	-	I2S2_MCK	I2C3_ SDA	SPI2_ MISO	-	-	-	-	-	-	-	-	-	CM4_ EVENTOUT
	PB15	-	TIM1_ CH3N	-	-	I2C2_ SCL	SPI2_ MOSI/ I2S2_SD	-	-	-	-	-	-	-	-	-	CM4_ EVENTOUT



Table 20. Alternate functions (continued)

Port		AF0	AF1	AF2	AF3	AF4	AF5	AF6	AF7	AF8	AF9	AF10	AF11	AF12	AF13	AF14	AF15
		SYS_AF	TIM1/ TIM2/ LPTIM1	TIM1/ TIM2	SPI2S2 TIM1/ LPTIM3	I2C1/ I2C2/ I2C3	SPI1/ SPI2S2	RF	USART1 / USART2	LPUART1	-	-	-	COMP1/ COMP2/ TIM1	DEBUG	TIM2/ TIM16/ TIM17/ LPTIM2	EVENOUT
Port C	PC0	-	LPTIM1_ IN1	-	-	I2C3_ SCL	-	-	-	LPUART1_ RX	-	-	-	-	-	LPTIM2_ IN1	CM4_ EVENTOUT
	PC1	-	LPTIM1_ OUT	-	SPI2_ MOSI/ I2S2_SD	I2C3_ SDA	-	-	-	LPUART1_ TX	-	-	-	-	-	-	CM4_ EVENTOUT
	PC2	-	LPTIM1_ IN2	-	-	-	SPI2_ MISO	-	-	-	-	-	-	-	-	-	CM4_ EVENTOUT
	PC3	-	LPTIM1_ ETR	-	-	-	SPI2_ MOSI/ I2S2_SD	-	-	-	-	-	-	-	-	LPTIM2_ ETR	CM4_ EVENTOUT
	PC4	-	-	-	-	-	-	-	-	-	-	-	-	-	-	-	CM4_ EVENTOUT
	PC5	-	-	-	-	-	-	-	-	-	-	-	-	-	-	-	CM4_ EVENTOUT
	PC6	-	-	-	-	-	I2S2_ MCK	-	-	-	-	-	-	-	-	-	CM4_ EVENTOUT
	PC13	RTC_ OUT1, RTC_TS	-	-	-	-	-	-	-	-	-	-	-	-	-	-	CM4_ EVENTOUT
	PC14	-	-	-	-	-	-	-	-	-	-	-	-	-	-	-	CM4_ EVENTOUT
	PC15	-	-	-	-	-	-	-	-	-	-	-	-	-	-	-	CM4_ EVENTOUT
Port H	PH3	-	-	-	-	-	-	-	-	-	-	-	-	-	-	-	CM4_ EVENTOUT



## 5 Electrical characteristics

### 5.1 Parameter conditions

Unless otherwise specified, all voltages are referenced to  $V_{SS}$ .

#### 5.1.1 Minimum and maximum values

Unless otherwise specified, the minimum and maximum values are guaranteed in the worst conditions of ambient temperature, supply voltage and frequencies, by tests in production on 100 % of the devices, with an ambient temperature at  $T_A = 25\text{ }^{\circ}\text{C}$  and  $T_A = T_{A\text{max}}$  (given by the selected temperature range).

Data based on characterization results, design simulation and/or technology characteristics are indicated in the table footnotes and are not tested in production. Based on characterization, the minimum and maximum values refer to sample tests and represent the mean value plus or minus three times the standard deviation (mean  $\pm 3\sigma$ ).

#### 5.1.2 Typical values

Unless otherwise specified, typical data are based on  $T_A = 25\text{ }^{\circ}\text{C}$ ,  $V_{DD} = V_{DDA} = V_{BAT} = 3\text{ V}$ . Typical values are given only as design guidelines and are not tested.

Typical ADC accuracy values are determined by characterization of a batch of samples from a standard diffusion lot over the full temperature range, where 95 % of the devices have an error less than or equal to the value indicated (mean  $\pm 2\sigma$ ).

#### 5.1.3 Typical curves

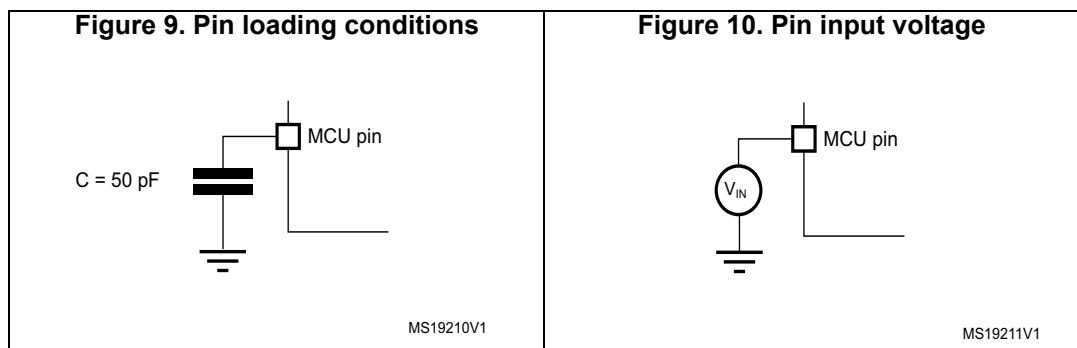
Unless otherwise specified, all typical curves are given only as design guidelines and are not tested.

#### 5.1.4 Loading capacitor

The loading conditions used for pin parameter measurement are shown in [Figure 9](#).

#### 5.1.5 Pin input voltage

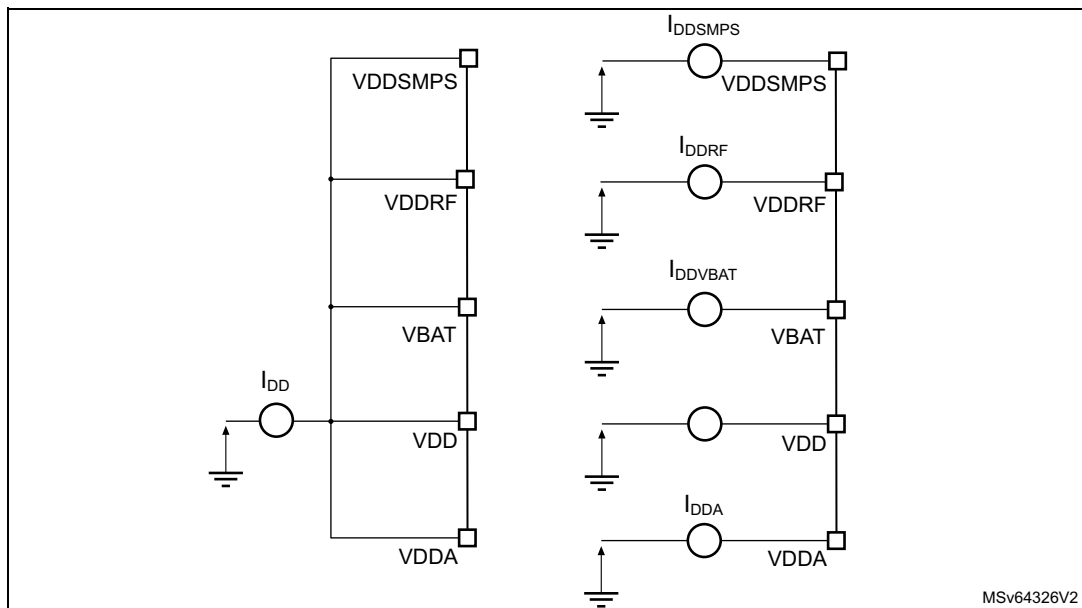
The input voltage measurement on a pin of the device is described in [Figure 10](#).





### 5.1.7 Current consumption measurement

Figure 12. Current consumption measurement scheme



## 5.2 Absolute maximum ratings

Stresses above the absolute maximum ratings listed in the tables below, may cause permanent damage to the device. These are stress ratings only and functional operation of the device at these conditions is not implied. Exposure to maximum rating conditions for extended periods may affect device reliability.

Device mission profile (application conditions) is compliant with JEDEC JESD47 Qualification Standard, extended mission profiles are available on demand.

Table 21. Voltage characteristics<sup>(1)</sup>

Symbol	Ratings	Min	Max	Unit
$V_{DDX} - V_{SS}$	External main supply voltage (including $V_{DD}$ , $V_{DDA}$ , $V_{DDRF}$ , $V_{DDSMPS}$ , $V_{BAT}$ , $V_{REF+}$ )	-0.3	3.9	V
$V_{IN}^{(2)}$	Input voltage on FT_XXX pins	$V_{SS} - 0.3$	$\min(V_{DD}, V_{DDA}, V_{DDRF}, V_{DDSMPS}) + 3.9^{(3)(4)}$	
	Input voltage on TT_XX pins		3.9	
	Input voltage on any other pin		3.9	
$ \Delta V_{DDX} $	Variations between different $V_{DDX}$ power pins of the same domain	-	50	mV
$ V_{SSx} - V_{SS} $	Variations between all the different ground pins <sup>(5)</sup>	-	50	
$V_{REF+} - V_{DDA}$	Allowed voltage difference for $V_{REF+} > V_{DDA}$	-	0.4	V

1. All main power ( $V_{DD}$ ,  $V_{DDRF}$ ,  $V_{DDA}$ ,  $V_{BAT}$ ) and ground ( $V_{SS}$ ,  $V_{SSA}$ ) pins must always be connected to the external power supply, in the permitted range.

2.  $V_{IN}$  maximum must always be respected. Refer to the next table for the maximum allowed injected current values.
3. This formula must be applied only on the power supplies related to the I/O structure described in [Table 19: STM32WLE5J8/JB/JC pin definition](#).
4. To sustain a voltage higher than 4 V, the internal pull-up/pull-down resistors must be disabled.
5. Include VREF- pin.

**Table 22. Current characteristics**

Symbol	Ratings	Max	Unit
$\Sigma I_{VDD}$	Total current into sum of all $V_{DD}$ power lines (source) <sup>(1)</sup>	130	mA
$\Sigma I_{VSS}$	Total current out of sum of all $V_{SS}$ ground lines (sink) <sup>(1)</sup>	130	
$I_{VDD(PIN)}$	Maximum current into each VDD power pin (source) <sup>(1)</sup>	100	
$I_{VSS(PIN)}$	Maximum current out of each VSS ground pin (sink) <sup>(1)</sup>	100	
$I_{IO(PIN)}$	Output current sunk by any I/O and control pin, except FT_f	20	
	Output current sunk by any FT_f pin	20	
	Output current sourced by any I/O and control pin	20	
$\Sigma I_{IO(PIN)}$	Total output current sunk by sum of all I/Os and control pins <sup>(2)</sup>	100	
	Total output current sourced by sum of all I/Os and control pins <sup>(2)</sup>	100	
$I_{INJ(PIN)}^{(3)}$	Injected current on FT_xxx, TT_xx, RST and B pins, except PB0 and PB1	-5 / +0 <sup>(4)</sup>	
	Injected current on PB0 and PB1	-5/0	
$\Sigma  I_{INJ(PIN)} $	Total injected current (sum of all I/Os and control pins) <sup>(5)</sup>	25	

1. All main power (VDD, VDDRF, VDDA, VBAT) and ground (VSS, VSSA) pins must always be connected to the external power supplies, in the permitted range.
2. This current consumption must be correctly distributed over all I/Os and control pins.
3. Positive injection (when  $V_{IN} > V_{DD}$ ) is not possible on these I/Os and does not occur for input voltages lower than the specified maximum value.
4. A negative injection is induced by  $V_{IN} < V_{SS}$ .  $I_{INJ(PIN)}$  must never be exceeded. Refer also to the previous table for the maximum allowed input voltage values.
5. When several inputs are submitted to a current injection, the maximum  $\Sigma |I_{INJ(PIN)}|$  is the absolute sum of the negative injected currents (instantaneous values).

**Table 23. Thermal characteristics**

Symbol	Ratings	Value	Unit
$T_{STG}$	Storage temperature range	-65 to +150	°C
$T_J$	Maximum junction temperature	105	

## 5.3 Operating conditions

### 5.3.1 Main performances

Table 24. Main performances at  $V_{DD} = 3\text{ V}$

Parameter		Test conditions	Typ	Unit
I <sub>CORE</sub>	Core current consumption	VBAT (V <sub>BAT</sub> = 3V, V <sub>DD</sub> = 0 V)	0.005	μA
		Shutdown	0.031	
		Standby (32-Kbyte RAM retention)	0.360	
		Stop 2, RTC enabled	1	
		Sleep (16 MHz)	770	
		LPRun (2 MHz)	220	
		Run, SMPS ON (48 MHz)	3450	
Rx boosted		LoRa 125 kHz, SMPS ON	4.82	mA
Tx low power	434 to 490 MHz, 14 dBm, 3.3 V	21		
	868 to 915 MHz, 14 dBm, 3.3 V	26		
Tx high power	434 to 490 MHz, 22 dBm, 3.3 V	120		
	868 to 915 MHz, 22 dBm, 3.3 V	107		

### 5.3.2 General operating conditions

Table 25. General operating conditions

Symbol	Parameter	Conditions	Min	Max	Unit
f <sub>HCLK</sub>	Internal AHB clock frequency	-	0	48	MHz
f <sub>PCLK1</sub>	Internal APB1 clock frequency	-			
f <sub>PCLK2</sub>	Internal APB2 clock frequency	-			
V <sub>DD</sub>	Standard operating voltage	-	1.8 <sup>(1)</sup>	3.6	V
V <sub>DDA</sub>	Analog supply voltage	ADC or COMP used	1.71	3.6	
		VREFBUF used	2.4		
		ADC, COMP, VREFBUF not used	0		
V <sub>BAT</sub>	Backup operating voltage	-	1.55	3.6	
V <sub>FBSMPS</sub>	SMPS feedback voltage	-	1.4	3.6	V
V <sub>DDRF</sub>	Minimum RF voltage	-	1.8	3.6	
V <sub>IN</sub>	I/O input voltage	TT_xx I/O	−0.3	V <sub>DD</sub> + 0.3	
		All I/O except TT_xx	−0.3	min between min (V <sub>DD</sub> , V <sub>DDA</sub> ) + 3.6 V and 5.5 V <sup>(2)(3)</sup>	

Table 25. General operating conditions (continued)

Symbol	Parameter	Conditions	Min	Max	Unit
$P_D$	Power dissipation at $T_A = 85\text{ }^\circ\text{C}$ for suffix 6 version <sup>(4)</sup>	UFBGA73	-	392.0	mW
$T_A$	Ambient temperature for suffix 6 version	Maximum power dissipation	-40	85	$^\circ\text{C}$
		Low-power dissipation <sup>(5)</sup>		105	
$T_J$	Junction temperature range	Suffix 6 version	-40	105	$^\circ\text{C}$

- When the reset is released, the functionality is guaranteed down to  $V_{BOR0}$  min.
- This formula has to be applied only on the power supplies related to the I/O structure described in [Table 19: STM32WLE5J8/JB/JC pin definition](#). Maximum I/O input voltage is the smallest value between  $\min(V_{DD}, V_{DDA}) + 3.6\text{ V}$  and  $5.5\text{ V}$ .
- For operation with voltage higher than  $\min(V_{DD}, V_{DDA}) + 0.3\text{ V}$ , the internal pull-up and pull-down resistors must be disabled.
- If  $T_A$  is lower, higher  $P_D$  values are allowed as long as  $T_J$  does not exceed  $T_J$  max (see [Table 92: Package thermal characteristics](#)).
- In low-power dissipation state,  $T_A$  can be extended to this range, as long as  $T_J$  does not exceed  $T_J$  max (see [Table 92: Package thermal characteristics](#)).

### 5.3.3 Sub-GHz radio characteristics

Electrical characteristics of the sub-GHz radio are given with the following conditions unless otherwise specified:

- $V_{DD} = 3.3\text{ V}$ . The current consumption is measured as described in [Figure 12](#).  $I_{DD}$  includes current consumption of all supplies ( $V_{DDRF}$ ,  $V_{DDSMPS}$ ,  $V_{DD}$ ,  $V_{DDA}$ ,  $V_{BAT}$ ). All peripherals except Sub-GHz radio are disabled and the system is in Standby mode.
- Temperature =  $25\text{ }^\circ\text{C}$
- HSE32 =  $32\text{ MHz}$
- $F_{RF} = 434/868/915\text{ MHz}$
- All RF impedances matched using reference design
- Reference design implementing a  $32\text{ MHz}$  crystal oscillator
- Transmit mode output power defined in  $50\text{ }\Omega$  load
- FSK BER (bit error rate) =  $0.1\text{ }\%$ , 2-level FSK modulation without pre-filtering, BR =  $4.8\text{ Kbit/s}$ , FDA =  $5\text{ kHz}$ , BW\_F =  $20\text{ kHz}$
- LoRa PER (packet error rate) =  $1\text{ }\%$ , packet of 64 bytes, preamble of 8 bytes, error correction code CR =  $4/5$ , CRC on payload enabled, no reduced encoding, no implicit header
- Sensitivities given using highest LNA gain step
- Power consumption measured with  $-140\text{ dBm}$  signal and AGC ON
- Blocking immunity, ACR and co-channel rejection, given for a single tone interferer and referenced to sensitivity  $+6\text{ dB}$ , blocking tests performed with unmodulated signal
- Bandwidth expressed on DSB (double-sided band)

Table 26. Sub-GHz radio power consumption

Symbol	Mode	Conditions		Min	Typ	Max	Unit
I <sub>DD</sub>	Deep-Sleep mode (Sleep with cold start) <sup>(1)(2)</sup>	All blocks off		-	50	-	nA
	Sleep mode (with warm start) <sup>(2)(3)</sup>	Configuration retained		-	140	-	
		Configuration retained + RC64k		-	810	-	
	Sleep, LDO mode <sup>(4)</sup>	LDO, band-gap, RC 13 MHz on	HSE32 off	-	414	-	μA
			HSE32 on	-	564	-	
	Sleep, SMPS mode <sup>(4)</sup>	Band-gap, RC 13 MHz on, SMPS 40 mA max	HSE32 off	-	700	-	
			HSE32 on	-	950	-	
	Standby mode (RC 13 MHz on)	RC 13 MHz on, HSE32 off		-	0.7	-	mA
	Standby mode (HSE32)	SMPS mode	40 mA max settings	-	1.05	-	
		LDO mode		-	0.99	-	
	Synthesizer mode	SMPS mode used with 40 mA drive capability		-	2.66	-	
		LDO mode		-	4.05	-	
	Receive mode, SMPS mode used	SMPS 40 mA max	FSK 4.8 Kbit/s	-	4.47	-	mA
			LoRa 125 kHz, FSK 4.8 Kbit/s	-	4.82	-	
			Rx boosted, FSK 4.8 Kbit/s	-	5.12	-	
			RX boosted, LoRa 125 kHz, FSK 4.8 Kbit/s	-	5.46	-	
	Receive mode, LDO mode used	FSK 4.8 Kbit/s		-	8.18	-	
		LoRa 125 kHz			8.90		
		RX boosted	FSK 4.8 Kbit/s		9.52		
			LoRa 125 kHz		10.22		

1. Cold start is equivalent to device at POR or when the device wakes up from Sleep mode with all blocks off.
2. Only Sub-GHz radio power consumption.
3. Warm start only happens when the device wakes up from Sleep mode with its configuration retained,
4. System in Stop 0 mode range 2.

Table 27. Sub-GHz radio power consumption in transmit mode (SMPS ON)

Symbol	Frequency band (MHz)	PA match (conditions)	Power output	Typ	Unit
$I_{DD}$	868 to 915	+14 dBm	+14 dBm, $V_{DDRF} = 3.3\text{ V}$	26	mA
			+10 dBm, $V_{DDRF} = 3.3\text{ V}$	20	
			+14 dBm, $V_{DDRF} = 1.8\text{ V}$	48	
			+10 dBm, $V_{DDRF} = 1.8\text{ V}$	38	
		+14 dBm (optimal settings)	+15 dBm, $V_{DDRF} = 3.3\text{ V}$	32.5	
			+10 dBm, $V_{DDRF} = 3.3\text{ V}$	15	
			+15 dBm, $V_{DDRF} = 1.8\text{ V}$	60	
			+10 dBm, $V_{DDRF} = 1.8\text{ V}$	29	
	434 to 490	+14 dBm	+15 dBm, $V_{DDRF} = 3.3\text{ V}$	25.5	mA
			+14 dBm, $V_{DDRF} = 3.3\text{ V}$	21	
			+10 dBm, $V_{DDRF} = 3.3\text{ V}$	15.5	
			+15 dBm, $V_{DDRF} = 1.8\text{ V}$	46.5	
			+14 dBm, $V_{DDRF} = 1.8\text{ V}$	39	
			+10 dBm, $V_{DDRF} = 1.8\text{ V}$	26	
	868 to 915	Low-power PA, SMPS OFF	+14 dBm	60	mA
	434 to 490			40	
	868 to 915	+22 dBm	+22 dBm	120	mA
			+20 dBm	106	
			+17 dBm	95	
			+14 dBm	88	
		+20 dBm (optimal settings)	+20 dBm	87	
		+17 dBm (optimal settings)	+17 dBm	58	
		+14 dBm (optimal settings)	+14 dBm	45	
	434 to 490	+22 dBm	+22 dBm	107	mA
			+20 dBm	92	
			+17 dBm	78	
			+14 dBm	65	
		+20 dBm (optimal settings)	+20 dBm	65	
		+17 dBm (optimal settings)	+17 dBm	42	
		+14 dBm (optimal settings)	+14 dBm	32	



Table 28. Sub-GHz radio general specifications

Symbol	Description	Conditions	Min	Typ	Max	Unit
FR	Frequency synthesizer range	Low-power PA	150	-	960	MHz
FSTEP	Frequency synthesizer step	High-resolution mode HSE / $2^{(2)(5)}$	-	095	-	Hz
PHN <sup>(1)(2)</sup>	Synthesizer phase noise (868 to 915 MHz)	100 kHz offset	-	-100	-	dBc/Hz
		1 MHz offset	-	-120	-	
		10 MHz offset	-	-135	-	
TS_FS	Synthesizer wakeup time	From Standby, HSE32 mode	-	40	-	μs
TS_HOP	Synthesizer hop time	10 MHz step	-	40	-	
TS_OSC	Crystal oscillator wakeup time	From Standby, RC <sup>(3)</sup> normal mode from HSE32 off	-	170	-	
OSC_TRM	Crystal oscillator trimming range for crystal frequency error compensation <sup>(4)</sup>	Min/max XTAL specifications	±15	±30	-	ppm
BR_F	Bitrate, FSK	Programmable (min modulation index is 0.5)	0.6	-	300 <sup>(5)</sup>	Kbit/s
FDA	Frequency deviation, FSK	Programmable (FDA + BR_F/2 ≤ 250 kHz)	0.6	-	200	kHz
BR_L	Bitrate, LoRa	Min for SF12, BW_L = 7.8 kHz Max for SF7, BW_L = 500 kHz	0.018	-	62.5 <sup>(6)</sup>	Kbit/s
BW_L	Signal BW, LoRa	Programmable	7.8	-	500 <sup>(6)</sup>	kHz
SF	Spreading factor for LoRa	Programmable, chips/symbol = $2^{SF}$	5	-	12	-

1. Phase Noise specifications are given for the recommended PLL bandwidth to be used for the specific modulation/BR, optimized settings may be used for specific applications.
2. Phase Noise is not constant over frequency, due to the topology of the PLL. For two frequencies close to each other, the phase noise may change significantly
3. Wakeup time till crystal oscillator frequency is within ±10 ppm.
4. OSC\_TRIM is the available trimming range to compensate for crystal initial frequency error and to allow crystal temperature compensation implementation. The total available trimming range is higher and allows the compensation for all device process variations
5. Maximum bit rate is assumed to scale with the RF frequency: for example 300 Kbit/s in the 869-to-915 MHz frequency band and only 50 Kbit/s at 150 MHz.
6. For RF frequencies below 400 MHz, there is a scaling between the frequency and supported bandwidth. Some bandwidths may not be available below 400 MHz.

Table 29. Sub-GHz radio receive mode specifications

Symbol	Description	Conditions	Min	Typ	Max	Unit
RXS_2FB	Sensitivity 2-FSK, RX boosted gain, split RF paths for RX and Tx, RF switch insertion loss excluded	BR = 0.6 Kbit/s, FDA = 0.8 kHz, BW = 4 kHz	-	-125	-	dBm
		BR = 1.2 Kbit/s, FDA = 5 kHz, BW = 20 kHz	-	-123	-	
		BR = 4.8 Kbit/s, FDA = 5 kHz, BW = 20 kHz	-	-117	-	
		BR = 38.4 Kbit/s, FDA = 40 kHz, BW = 160 kHz	-	-108	-	
		BR = 250 Kbit/s, FDA = 125 kHz, BW = 500 kHz	-	-103	-	
RXS_LB	Sensitivity LoRa, RX boosted gain, split RF paths for RX and Tx, RF switch insertion loss excluded	BW = 10.4 kHz, SF = 7	-	-135	-	
		BW = 10.4 kHz, SF = 12	-	-148	-	
		BW = 125 kHz, SF = 7	-	-125	-	
		BW = 125 kHz, SF = 12	-	-138	-	
		BW = 250 kHz, SF = 7	-	-122	-	
		BW = 250 kHz, SF = 12	-	-135	-	
		BW = 500 kHz, SF = 7	-	-118	-	
		BW = 500 kHz, SF = 12	-	-130	-	
RSX_2F	Sensitivity 2-FSK, RX power saving gain with direct tie connection between RX and Tx	BR = 4.8 Kbit/s, FDA = 5 kHz, BW = 20 kHz	-	-115	-	
RXS_L	Sensitivity LoRa, RX power saving gain with direct tie connection between RX and Tx	BW = 125 kHz, SF = 12	-	-135	-	
CCR_F	Co-channel rejection, FSK	-	-	-9	-	dB
CCR_L	Co-channel rejection, LoRa	SF = 7	-	7	-	
		SF = 12	-	19	-	
ACR_F	Adjacent channel rejection, FSK	Offset = $\pm 50$ kHz	-	44	-	
ACR_L	Adjacent channel rejection, LoRa	Offset = $\pm 1.5 \times BW\_L$ , BW = 125 kHz, SF = 7	-	60	-	
		Offset = $\pm 1.5 \times BW\_L$ , BW = 125 kHz, SF = 12	-	71	-	
BI_F	Blocking immunity, FSK	Offset = $\pm 1$ MHz, BR = 4.8 Kbit/s, FDA = 5 kHz, BW = 20 kHz	-	67	-	
		Offset = $\pm 2$ MHz, BR = 4.8 Kbit/s, FDA = 5 kHz, BW = 20 kHz	-	70	-	
		Offset = $\pm 10$ MHz, BR = 4.8 Kbit/s, FDA = 5 kHz, BW = 20 kHz	-	76	-	

Table 29. Sub-GHz radio receive mode specifications (continued)

Symbol	Description	Conditions	Min	Typ	Max	Unit
BI_L	Blocking immunity, LoRa	Offset = $\pm 1$ MHz, BW = 125 kHz, SF = 12	-	87	-	dB
		Offset = $\pm 2$ MHz, BW = 125 kHz, SF = 12	-	91	-	
		Offset = $\pm 10$ MHz, BW = 125 kHz, SF = 12	-	96	-	
IIP3	Third order input intercept point	Unwanted tones are 1 MHz and 1.96 MHz above LO. 868 to 915 MHz band	-	-9	-	dBm
		Unwanted tones are 1 MHz and 1.96 MHz above LO. 433 MHz band	-	-15	-	
IMA	Image attenuation	Without IQ calibration	-	30	-	dB
		With IQ calibration	-	54	-	
BW_F	DSB channel filter BW, FSK	Programmable, typical values	4.8	-	467	kHz
TS_RX	Receiver wakeup time	FS to RWX	-	41	-	$\mu$ s
FERR_L	Maximum tolerated frequency offset between transmitter and receiver, SF7 to SF12	All bandwidths, $\pm 25$ % of BW. The tighter limit between this line and the three lines below applies.	-	$\pm 25$	-	BW
	Maximum tolerated frequency offset between transmitter and receiver, SF10 to SF12	SF12	-50	-	50	ppm
		SF11	-100	-	100	
		SF10	-200	-	200	

Table 30. Sub-GHz radio transmit mode specifications

Symbol	Description	Conditions	Min	Typ	Max	Unit
TXOP	Max RF output power	Highest power step setting for low-power PA (LP PA)	-	+15 <sup>(1)</sup>	-	dBm
		Highest power step setting for high-power PA (HP PA)	-	+22	-	
TXDRP	RF output power drop versus supply voltage	LP PA, under SMPS or LDO VDDop range from 1.8 to 3.7 V	-	0.5	-	dB
		HP PA, +22 dBm, V <sub>DD</sub> = 2.7 V	-	2	-	
		HP PA, +22 dBm, V <sub>DD</sub> = 2.4 V	-	3	-	
		HP PA, +22 dBm, V <sub>DD</sub> = 1.8 V	-	6	-	
TXPRNG	RF output power range	Programmable in 31 steps, typical value	TXOP-31	-	TXOP	dBm
TXACC	RF output power step accuracy	-	-	$\pm 2$	-	dB

Table 30. Sub-GHz radio transmit mode specifications (continued)

Symbol	Description	Conditions	Min	Typ	Max	Unit
TXRMP	PA ramping time	Programmable	10	-	3400	μs
TS_TX	TX wakeup time	Frequency synthesizer enabled	-	36 + PA ramping	-	

1. For low-power PA, +15 dBm maximum RF output power can be reached with optimal settings.

Table 31. Sub-GHz radio power management specifications

Symbol	Description	Conditions	Frequency (MHz)			Unit
			470	490	868	
TRPOR	Required POR reset pulse duration	For $V_{DD} \geq 1.8$ V	50	100	-	μs
VEOLL	End-of-life low-threshold voltage	-	1.81	1.89	1.96	V
VEOLH	End-of-life high-threshold voltage	-	1.86	1.94	2.1	
VEOLD	End-of-life hysteresis voltage	VEOLH - VEOLL	50	53	56	mV
VREG	Main regulated supply	LDO or SMPS over process, voltage and temperature range	1.47	1.55	1.62	V
LDTRSMPS	Load transient for ILSMPS 100 μA to 100 mA in 10 μs LDO running	High BW mode	-	25	-	mV
		Low BW mode	-	47	-	
ILSMPS	SMPS load current	-	-	-	100	mA
IDDSMPS	SMPS quiescent current	SMPS high power, $V_{DD} = 3.3$ V	-	538	-	μA
		SMPS low power, $V_{DD} = 3.3$ V	-	460	-	
EFFSMPS	SMPS converter average efficiency $EFF = V_{REG} \times I_{LOAD} / V_{BAT} \times I_{DD}$	SMPS 100 mA max $V_{DD} = 3.3$ V, ILSMPS = 6 mA	-	71	-	%
		SMPS 100 mA max $V_{DD} = 3.3$ V, ILSMPS = 50 mA	-	89	-	
		SMPS 100 mA max $V_{DD} = 1.8$ V, ILSMPS = 6 mA	-	88	-	
		SMPS 100 mA max $V_{DD} = 2.0$ V, ILSMPS = 50 mA	-	91	-	
		SMPS 100 mA max $V_{DD} = 3.3$ V, ILSMPS = 100 mA	-	86	-	
Cout	Shared between LDO and SMPS	±20 % tolerance	-	470	-	nF
Lout	SMPS inductor	-	-	15	-	μH
TSSMPS	Sleep and Sleep, SMPS startup time	For ILIM = 50 mA	-	70	-	μs

Table 31. Sub-GHz radio power management specifications (continued)

Symbol	Description	Conditions	Frequency (MHz)			Unit
			470	490	868	
IDDLDO	LDO quiescent current	$V_{DD} = 3.3\text{ V}$ , ILOAD = 0 to 100 mA, current limiter off	-	95	-	$\mu\text{A}$
		$V_{DD} = 3.3\text{ V}$ , ILOAD = 100 mA, current limiter on	-	380	-	
		$V_{DD} = 3.3\text{ V}$ , ILOAD = 50 mA, current limiter on	-	280	-	
ILDO	LDO load current	-	-	100	-	mA
LDTRLDO	Load transient for ILDO 100 $\mu\text{A}$ to 100 mA in 10 $\mu\text{s}$	-	-	25	-	mV
TSLDO	Sleep and Sleep, LDO startup time	For ILIM = 50 mA	-	60	-	$\mu\text{s}$
VDIG	Digital regulator target voltage	-	1.14	1.2	1.26	V
ILM <sup>(1)</sup>	Current limiter max value	-	25	50	200	mA

1. The default current limiter value is set to 50 mA.

### 5.3.4 Operating conditions at power-up/power-down

Parameters given in the table below are derived from tests performed under the ambient temperature condition summarized in [Table 25: General operating conditions](#).

Table 32. Operating conditions at power-up/power-down

Symbol	Parameter	Min	Max	Unit
$t_{VDD}$	$V_{DD}$ rise time rate	-	$\infty$	$\mu\text{s/V}$
	$V_{DD}$ fall time rate	10	$\infty$	
$t_{VDDA}$	$V_{DDA}$ rise time rate	0	$\infty$	
	$V_{DDA}$ fall time rate	10	$\infty$	
$t_{VDDRF}$	$V_{DDRF}$ rise time rate	-	$\infty$	
	$V_{DDRF}$ fall time rate	-	$\infty$	

### 5.3.5 Embedded reset and power-control block characteristics

Parameters given in the table below are derived from tests performed under the ambient temperature conditions summarized in [Table 25: General operating conditions](#).

Table 33. Embedded reset and power-control block characteristics

Symbol	Parameter	Conditions <sup>(1)</sup>	Min	Typ	Max	Unit
$t_{RSTTEMPO}^{(2)}$	Reset temporization after BOR0 is detected	$V_{DD}$ rising	-	250	400	$\mu\text{s}$

Table 33. Embedded reset and power-control block characteristics (continued)

Symbol	Parameter	Conditions <sup>(1)</sup>	Min	Typ	Max	Unit
$V_{BOR0}^{(2)}$	Brownout reset threshold 0	Rising edge	1.62	1.66	1.70	V
		Falling edge	1.60	1.64	1.69	
$V_{BOR1}$	Brownout reset threshold 1	Rising edge	2.06	2.10	2.14	
		Falling edge	1.96	2.00	2.04	
$V_{BOR2}$	Brownout reset threshold 2	Rising edge	2.26	2.31	2.35	
		Falling edge	2.16	2.20	2.24	
$V_{BOR3}$	Brownout reset threshold 3	Rising edge	2.56	2.61	2.66	
		Falling edge	2.47	2.52	2.57	
$V_{BOR4}$	Brownout reset threshold 4	Rising edge	2.85	2.90	2.95	
		Falling edge	2.76	2.81	2.86	
$V_{PVD0}$	Programmable voltage detector threshold 0	Rising edge	2.10	2.15	2.19	
		Falling edge	2.00	2.05	2.10	
$V_{PVD1}$	PVD threshold 1	Rising edge	2.26	2.31	2.36	V
		Falling edge	2.15	2.20	2.25	
$V_{PVD2}$	PVD threshold 2	Rising edge	2.41	2.46	2.51	
		Falling edge	2.31	2.36	2.41	
$V_{PVD3}$	PVD threshold 3	Rising edge	2.56	2.61	2.66	
		Falling edge	2.47	2.52	2.57	
$V_{PVD4}$	PVD threshold 4	Rising edge	2.69	2.74	2.79	
		Falling edge	2.59	2.64	2.69	
$V_{PVD5}$	PVD threshold 5	Rising edge	2.85	2.91	2.96	
		Falling edge	2.75	2.81	2.86	
$V_{PVD6}$	PVD threshold 6	Rising edge	2.92	2.98	3.04	
		Falling edge	2.84	2.90	2.96	
$V_{hyst\_BORH0}$	Hysteresis voltage of BORH0	Hysteresis in continuous mode	-	20	-	mV
		Hysteresis in other mode	-	30	-	
$V_{hyst\_BOR\_PVD}$	Hysteresis voltage of BORH (except BORH0) and PVD	-	-	100	-	
$I_{DD} (BOR\_PVD)^{(2)}$	BOR <sup>(3)</sup> (except BOR0) and PVD consumption from $V_{DD}$	-	-	1.1	1.6	$\mu A$
$V_{PVM3}$	$V_{DDA}$ peripheral voltage monitoring	Rising edge	1.61	1.65	1.69	V
		Falling edge	1.6	1.64	1.68	
$V_{hyst\_PVM3}$	PVM3 hysteresis	-	-	10	-	mV
$I_{DD} (PVM3)^{(2)}$	PVM3 consumption from $V_{DD}$	-	-	2	-	$\mu A$

1. Continuous mode means Run and Sleep modes, or temperature sensor enable in LPRun and LPSleep modes.
2. Guaranteed by design.
3. BOR0 is enabled in all modes (except Shutdown) and its consumption is therefore included in the supply current characteristics tables.

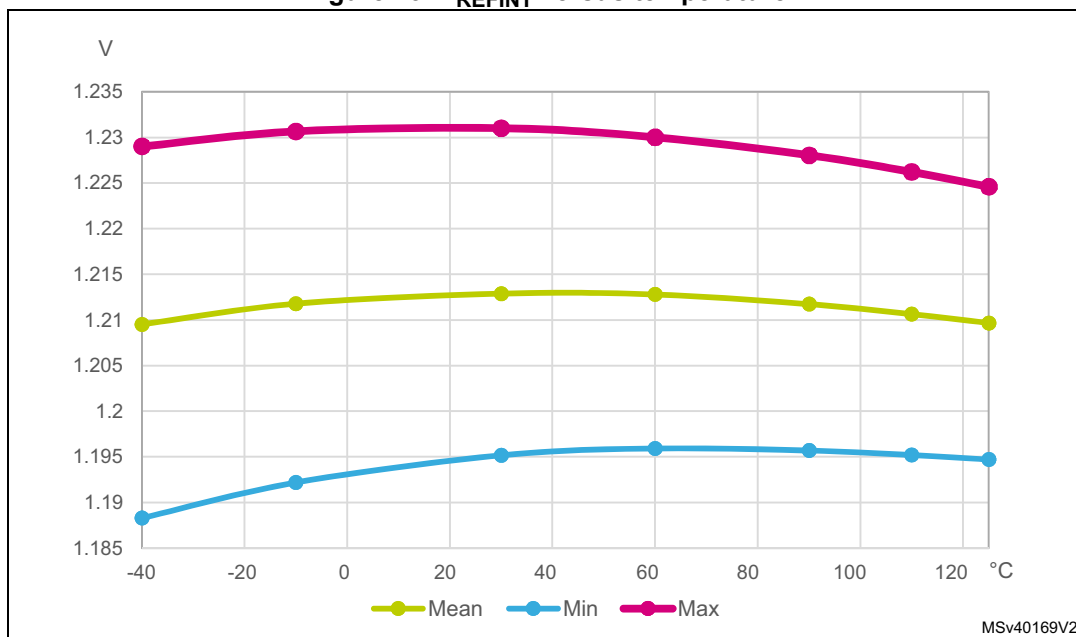
### 5.3.6 Embedded voltage reference

Parameters given in the table below are derived from tests performed under the ambient temperature and supply voltage conditions summarized in [Table 25: General operating conditions](#).

**Table 34. Embedded internal voltage reference**

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$V_{REFINT}$	Internal reference voltage	$-40\text{ }^{\circ}\text{C} < T_J < +105\text{ }^{\circ}\text{C}$	1.182	1.212	1.232	V
$t_{S\_vrefint}^{(1)}$	ADC sampling time when reading the internal reference voltage	-	4 <sup>(2)</sup>	-	-	$\mu\text{s}$
$t_{start\_vrefint}$	Start time of reference voltage buffer when ADC is enable	-	-	8	12 <sup>(2)</sup>	
$I_{DD}(V_{REFINTBUF})$	$V_{REFINT}$ buffer consumption from $V_{DD}$ when converted by ADC	-	-	12.5	20 <sup>(2)</sup>	$\mu\text{A}$
$\Delta V_{REFINT}$	Internal reference voltage spread over the temperature range	$V_{DD} = 3.3\text{ V}$	-	5	7.5 <sup>(2)</sup>	mV
$T_{Coeff}$	Temperature coefficient	$-40\text{ }^{\circ}\text{C} < T_J < +105\text{ }^{\circ}\text{C}$	-	30	50 <sup>(2)</sup>	ppm/ $^{\circ}\text{C}$
$A_{Coeff}$	Long term stability	1000 hours, $T = 25\text{ }^{\circ}\text{C}$	-	300	1000 <sup>(2)</sup>	ppm
$V_{DDCoeff}$	Voltage coefficient	$3.0\text{ V} < V_{DD} < 3.6\text{ V}$	-	250	1200 <sup>(2)</sup>	ppm/V
$V_{REFINT\_DIV1}$	1/4 reference voltage	-	24	25	26	% $V_{REFINT}$
$V_{REFINT\_DIV2}$	1/2 reference voltage		49	50	51	
$V_{REFINT\_DIV3}$	3/4 reference voltage		74	75	76	

1. The shortest sampling time can be determined in the application by multiple iterations.
2. Guaranteed by design.

Figure 13.  $V_{REFINT}$  versus temperature

### 5.3.7 Supply current characteristics

The current consumption is a function of several parameters and factors such as operating voltage, ambient temperature, I/O pin loading, device software configuration, operating frequencies, I/O pin switching rate, program location in memory and executed binary code.

The current consumption is measured as described in [Figure 12](#).

#### Typical and maximum current consumption

The device is put under the following conditions:

- All I/O pins are in analog input mode.
- All peripherals are disabled, except when explicitly mentioned.
- The Flash memory access time is adjusted with the minimum wait-states number, depending on the  $f_{HCLK}$  frequency. Refer to the table 'Number of wait states according to Flash clock (HCLK3) frequency' in the reference manual (RM0461).
- $f_{PCLK} = f_{HCLK}$  when the peripherals are enabled.
- $f_{PCLK} = f_{HCLK} = f_{HCLKS}$  for the Flash memory and shared peripherals.

Parameters given in the tables below ([Table 35](#) to [Table 47](#)) are derived from tests performed under ambient temperature and supply voltage conditions summarized in [Table 25: General operating conditions](#).



**Table 35. Current consumption in Run and LPRun modes, CoreMark code with data running from Flash memory, ART enable (cache ON, prefetch OFF)**

Symbol	Parameter	Conditions			Typ			Max <sup>(1)</sup>		Unit
		-	Voltage scaling	f <sub>HCLK</sub> (MHz)	25 °C	55 °C	85 °C	25 °C	85 °C	
I <sub>DD</sub> (Run)	Supply current in Run mode	f <sub>HCLK</sub> = f <sub>MSI</sub> All peripherals disabled	Range 2	16	1.85	1.90	1.95	2.20	2.40	mA
				8	1.10	1.15	1.20	1.40	1.60	
				2	0.585	0.610	0.670	TBD	TBD	
			SMPS Range 2	16	1.60	1.60	1.65	TBD	TBD	
				8	1.00	1.05	1.05	TBD	TBD	
				2	0.730	0.750	0.780	TBD	TBD	
			Range 1	48	5.55	5.65	5.80	7.40	11.0	
				32	3.85	3.95	4.05	5.60	8.40	
				16	2.15	2.20	2.30	3.70	6.60	
			SMPS Range 1	48	3.40	3.45	3.55	TBD	TBD	
				32	2.50	2.55	2.60	TBD	TBD	
				16	1.60	1.60	1.65	TBD	TBD	
I <sub>DD</sub> (LPRun)	Supply current in LPRun mode	f <sub>HCLK</sub> = f <sub>MSI</sub> All peripherals disabled		2	0.220	0.235	0.290	0.270	0.490	
				1	0.120	0.135	0.185	0.150	0.390	
				0.4	0.058	0.0715	0.120	0.084	0.330	

1. Guaranteed by characterization results, unless otherwise specified.



**Table 36. Current consumption in Run and LPRun modes, CoreMark code with data running from SRAM1**

Symbol	Parameter	Conditions			Typ			Max <sup>(1)</sup>		Unit
		-	Voltage scaling	f <sub>HCLK</sub> (MHz)	25 °C	55 °C	85 °C	25 °C	85 °C	
I <sub>DD</sub> (Run)	Supply current in Run mode	f <sub>HCLK</sub> = f <sub>MSI</sub> All peripherals disabled	Range 2	16	1.90	1.90	2.00	2.20	2.40	mA
				8	1.10	1.15	1.20	1.40	1.60	
				2	TBD	TBD	TBD	TBD	TBD	
			SMPS Range 2	16	1.45	1.45	1.50	TBD	TBD	
				8	1.00	1.05	1.05	TBD	TBD	
				2	0.730	0.750	0.780	TBD	TBD	
			Range 1	48	5.65	5.75	5.90	6.50	6.70	
				32	3.90	4.00	4.10	4.60	4.80	
				16	2.20	2.25	2.30	2.50	2.80	
			SMPS Range 1	48	3.45	3.50	3.60	TBD	TBD	
				32	2.50	2.55	2.60	TBD	TBD	
				16	1.60	1.60	1.65	TBD	TBD	
I <sub>DD</sub> (LPRun)	Supply current in LPRun mode	f <sub>HCLK</sub> = f <sub>MSI</sub> All peripherals disabled		2	0.220	0.230	0.285	0.240	0.480	
				1	0.120	0.130	0.180	0.140	0.380	
				0.4	0.052	0.064	0.115	0.077	0.320	

1. Guaranteed by characterization results, unless otherwise specified.

**Table 37. Typical current consumption in Run and LPRun modes, with different codes running from Flash memory, ART enable (cache ON, prefetch OFF)**

Symbol	Parameter	Conditions			Typ	Unit	Typ	Unit
		-	Voltage scaling	Code	25 °C		25 °C	
$I_{DD}(\text{Run})$	Supply current in Run mode	$f_{HCLK} = f_{MSI}$ All peripherals disabled	Range 2 $f_{HCLK} = 16 \text{ MHz}$	Reduced code	1.90	mA	118.75	$\mu\text{A/MHz}$
				CoreMark <sup>(1)</sup>	1.85		115.63	
				Dhrystone 2.1	1.85		115.63	
				Fibonacci	1.80		112.50	
				While(1)	1.60		100.00	
			SMPS Range 2 $f_{HCLK} = 16 \text{ MHz}$	Reduced code	1.45		90.63	
				CoreMark <sup>(1)</sup>	1.40		87.50	
				Dhrystone 2.1	1.40		87.50	
				Fibonacci	1.40		87.50	
				While(1)	1.30		81.25	
			Range 1 $f_{HCLK} = 48 \text{ MHz}$	Reduced code	5.70		118.75	
				CoreMark <sup>(1)</sup>	5.55		115.63	
				Dhrystone 2.1	5.50		114.58	
				Fibonacci	5.40		112.50	
				While(1)	4.65		96.88	
			SMPS Range 1 $f_{HCLK} = 48 \text{ MHz}$	Reduced code	3.50		72.92	
				CoreMark <sup>(1)</sup>	3.40		70.83	
				Dhrystone 2.1	3.40		70.83	
				Fibonacci	3.30		68.75	
				While(1)	2.90		60.42	



**Table 37. Typical current consumption in Run and LPRun modes, with different codes running from Flash memory, ART enable (cache ON, prefetch OFF) (continued)**

Symbol	Parameter	Conditions			Typ	Unit	Typ	Unit
		-	Voltage scaling	Code	25 °C		25 °C	
I <sub>DD</sub> (LPRun)	Supply current in LPRun mode	f <sub>HCLK</sub> = f <sub>MSI</sub> = 2 MHz All peripherals disabled		Reduced code	0.225	μA	112.50	μA/MHz
				CoreMark <sup>(1)</sup>	0.220		110.00	
				Dhrystone 2.1	0.220		110.00	
				Fibonacci	0.240		120.00	
				While(1)	0.175		87.50	

1. CoreMark used for characterization results provided in [Table 35](#) and [Table 37](#).

**Table 38. Typical current consumption in Run and LPRun modes,  
with different codes running from SRAM1**

Symbol	Parameter	Conditions			Typ	Unit	Typ	Unit
		-	Voltage scaling	Code	25 °C		25 °C	
$I_{DD}(\text{Run})$	Supply current in Run mode	$f_{HCLK} = f_{MSI}$ All peripherals disabled	Range 2 $f_{HCLK} = 16 \text{ MHz}$	Reduced code	1.95	mA	121.88	$\mu\text{A/MHz}$
				CoreMark <sup>(1)</sup>	1.90		118.75	
				Dhrystone 2.1	1.90		118.75	
				Fibonacci	1.90		118.75	
				While(1)	1.75		109.38	
			Range 2 SMPS ON $f_{HCLK} = 16 \text{ MHz}$	Reduced code	1.45		90.63	
				CoreMark <sup>(1)</sup>	1.45		90.63	
				Dhrystone 2.1	1.45		90.63	
				Fibonacci	1.45		90.63	
				While(1)	1.35		84.38	
			Range 1 $f_{HCLK} = 48 \text{ MHz}$	Reduced code	5.90		122.92	
				CoreMark <sup>(1)</sup>	5.65		117.71	
				Dhrystone 2.1	5.70		118.75	
				Fibonacci	5.65		117.71	
				While(1)	5.10		106.25	
			Range 1 SMPS ON $f_{HCLK} = 48 \text{ MHz}$	Reduced code	3.60		75.00	
				CoreMark <sup>(1)</sup>	3.45		71.88	
				Dhrystone 2.1	3.50		72.92	
				Fibonacci	3.45		71.88	
				While(1)	3.15		65.63	



**Table 38. Typical current consumption in Run and LPRun modes,  
with different codes running from SRAM1 (continued)**

Symbol	Parameter	Conditions			Typ	Unit	Typ	Unit
		-	Voltage scaling	Code	25 °C		25 °C	
$I_{DD}(LPRun)^{(2)}$	Supply current in LPRun mode	$f_{HCLK} = f_{MSI} = 2 \text{ MHz}$ All peripherals disabled			Reduced code	$\mu\text{A}$	112.50	$\mu\text{A/MHz}$
					CoreMark <sup>(1)</sup>		110.00	
					Dhrystone 2.1		112.50	
					Fibonacci		112.50	
					While(1)		97.50	

1. CoreMark used for characterization results provided in [Table 35](#) and [Table 37](#).
2. Flash memory in power-down mode.

**Table 39. Current consumption in Sleep and LPSleep modes, Flash memory ON**

Symbol	Parameter	Conditions			Typ			Max <sup>(1)</sup>		Unit
		-	Voltage scaling	f <sub>HCLK</sub> (MHz)	25 °C	55 °C	85 °C	25 °C	85 °C	
I <sub>DD</sub> (Sleep)	Supply current in Sleep mode	f <sub>HCLK</sub> = f <sub>MSI</sub> All peripherals disabled	Range 2	16	0.770	0.800	0.860	1.00	1.30	mA
				8	0.570	0.600	0.655	0.780	0.990	
				2	0.445	0.470	0.525	0.650	0.860	
			Range 1	48	1.70	1.70	1.80	2.10	2.30	
				32	1.25	1.30	1.40	1.60	1.90	
				16	0.845	0.875	0.945	1.10	1.40	
			SMPS Range 1	48	1.35	1.40	1.45	TBD	TBD	
				32	1.15	1.15	1.20	TBD	TBD	
				16	0.895	0.915	0.950	TBD	TBD	
I <sub>DD</sub> (LPSleep)	Supply current in LPSleep mode	f <sub>HCLK</sub> = f <sub>MSI</sub> All peripherals disabled		2	0.068	0.0805	0.130	0.095	0.330	
				1	0.044	0.0565	0.105	0.069	0.310	
				0.4	0.0225	0.040	0.0885	0.052	0.290	
				0.1	0.018	0.032	0.081	0.045	0.280	

1. Guaranteed by characterization results, unless otherwise specified.

**Table 40. Current consumption in LPSleep mode, Flash memory in power-down**

Symbol	Parameter	Conditions		Typ			Max <sup>(1)</sup>		Unit
		-	f <sub>HCLK</sub> (MHz)	25 °C	55 °C	85 °C	25 °C	85 °C	
I <sub>DD</sub> (LPSleep)	Supply current in LPSleep mode	f <sub>HCLK</sub> = f <sub>MS</sub> All peripherals disabled	2	58.0	74.5	125	86.0	330	μA
			1	35.5	50.5	99.0	60.0	300	
			0.4	18.5	33.5	81.5	41.0	280	
			0.1	11.0	26.5	74.5	36.0	280	

1. Guaranteed by characterization results, unless otherwise specified.

**Table 41. Current consumption in Stop 2 mode**

Symbol	Parameter	Conditions	Typ				Max <sup>(1)</sup>			Unit
		V <sub>DD</sub> (V)	0 °C	25 °C	55 °C	85 °C	0 °C	25 °C	85 °C	
I <sub>DD</sub> (Stop 2)	Supply current in Stop 2 mode RTC disabled	1.8	0.545	0.830	2.45	8.45	1.20	2.20	24.0	μA
		2.4	0.525	0.850	2.60	8.80	TBD	TBD	TBD	
		3.0	0.605	0.885	2.80	9.25	1.10	2.60	26.0	
		3.6	0.630	0.935	3.10	9.75	1.40	2.80	26.0	
I <sub>DD</sub> (Stop 2 with RTC)	Supply current in Stop 2 mode RTC enabled, clocked by LSI <sup>(2)</sup>	1.8	0.650	0.880	2.55	8.25	1.30	2.30	24.0	
		2.4	0.630	0.945	2.70	8.85	TBD	TBD	TBD	
		3.0	0.715	1.00	2.90	9.70	1.40	2.80	26.0	
		3.6	0.750	1.10	3.15	10.5	1.50	3.00	26.0	

1. Guaranteed based on test during characterization, unless otherwise specified.

2. LSI using LSIPRE = 1 configuration.



**Table 42. Current consumption in Stop 1 mode**

Symbol	Parameter	Conditions	Typ				Max <sup>(1)</sup>			Unit
		V <sub>DD</sub> (V)	0 °C	25 °C	55 °C	85 °C	0 °C	25 °C	85 °C	
I <sub>DD</sub> (Stop 1)	Supply current in Stop 1 mode RTC disabled	1.8	2.05	4.00	14.0	47.0	6.10	20.0	200	μA
		2.4	2.15	3.95	14.0	47.0	TBD	TBD	TBD	
		3.0	2.15	4.15	14.0	47.5	5.90	20.0	200	
		3.6	2.25	4.20	14.0	48.0	6.20	20.0	200	
I <sub>DD</sub> (Stop 1 with RTC)	Supply current in Stop 1 mode RTC enabled, clocked by LSI <sup>(2)</sup>	1.8	2.30	4.35	14.0	47.0	6.30	20.0	200	
		2.4	2.50	4.35	14.5	47.5	TBD	TBD	TBD	
		3.0	2.55	4.55	14.5	48.0	6.40	21.0	200	
		3.6	2.65	4.70	15.0	48.5	6.70	21.0	200	

1. Guaranteed based on test during characterization, unless otherwise specified.
2. LSI using LSIPRE = 1 configuration.

**Table 43. Current consumption in Stop 0 mode**

Symbol	Parameter	Conditions	Typ				Max <sup>(1)</sup>			Unit
		V <sub>DD</sub> (V)	0 °C	25 °C	55 °C	85 °C	0 °C	25 °C	85 °C	
I <sub>DD</sub> (Stop 0)	Supply current in Stop 0 mode RTC disabled	1.8	335	345	365	415	480	500	740	μA
		2.4	360	370	395	445	TBD	TBD	TBD	
		3.0	390	400	425	475	540	570	800	
		3.6	425	435	460	515	580	600	840	

1. Guaranteed based on test during characterization, unless otherwise specified.



Table 44. Current consumption in Standby mode

Symbol	Parameter	Conditions		Typ				Max <sup>(1)</sup>			Unit
		-	V <sub>DD</sub> (V)	0 °C	25 °C	55 °C	85 °C	0 °C	25 °C	85 °C	
I <sub>DD</sub> (Standby)	Supply current in Standby mode RTC disabled Backup registers retained	No retention	1.8	0.009	0.027	0.245	1.00	TBD	TBD	TBD	μA
			2.4	0.022	0.051	0.340	1.35	TBD	TBD	TBD	
			3.0	0.046	0.071	0.470	1.75	TBD	TBD	TBD	
			3.6	0.075	0.125	0.650	2.30	TBD	TBD	TBD	
		SRAM2 retained	1.8	0.130	0.205	0.820	2.90	0.200	0.550	8.20	
			2.4	0.140	0.225	0.915	3.25	TBD	TBD	TBD	
			3.0	0.165	0.255	1.05	3.70	0.280	0.710	9.40	
			3.6	0.190	0.300	1.20	4.25	0.330	0.770	10.0	
I <sub>DD</sub> (Standby with RTC)	Supply current in Standby mode (backup registers and SRAM2 retained) RTC enabled	RTC clocked by LSI (PREDIV = 1)	1.8	0.215	0.295	0.895	3.10	TBD	TBD	TBD	μA
			2.4	0.230	0.325	0.990	3.45	TBD	TBD	TBD	
			3.0	0.260	0.360	1.15	3.95	TBD	TBD	TBD	
			3.6	0.305	0.425	1.30	4.55	TBD	TBD	TBD	
		RTC clocked by LSE quartz <sup>(2)</sup> in low drive mode	1.8	0.270	0.350	0.975	3.15	TBD	TBD	TBD	
			2.4	0.295	0.390	1.10	3.50	TBD	TBD	TBD	
			3.0	0.345	0.445	1.25	4.00	TBD	TBD	TBD	
			3.6	0.415	0.535	1.45	4.60	TBD	TBD	TBD	

1. Guaranteed by characterization results, unless otherwise specified.

2. Based on characterization done with a 32.768 kHz crystal (MC306-G-06Q-32.768, manufacturer JFVNY) with two 6.8 pF loading capacitors.

**Table 45. Current consumption in Shutdown mode**

Symbol	Parameter	Conditions		Typ				Max <sup>(1)</sup>			Unit
		-	V <sub>DD</sub> (V)	0 °C	25 °C	55 °C	85 °C	0 °C	25 °C	85 °C	
I <sub>DD</sub> (Shutdown)	Supply current in Shutdown mode RTC disabled Backup registers retained		1.8	0.001	0.008	0.105	0.380	0.001	0.043	1.70	μA
			2.4	0.008	0.018	0.135	0.445	TBD	TBD	TBD	
			3.0	0.018	0.031	0.180	0.545	0.078	0.150	2.40	
			3.6	0.041	0.062	0.260	0.690	0.110	0.190	2.90	
I <sub>DD</sub> (Shutdown with RTC)	Supply current in Shutdown mode (backup registers retained) RTC enabled	RTC clocked by an external clock	1.8	0.054	0.065	0.145	0.545	TBD	TBD	TBD	
			2.4	0.090	0.105	0.200	0.665	TBD	TBD	TBD	
			3.0	0.160	0.175	0.295	0.860	TBD	TBD	TBD	
			3.6	0.250	0.280	0.440	1.15	TBD	TBD	TBD	
		RTC clocked by LSE quartz <sup>(2)</sup> in low drive mode	1.8	0.140	0.155	0.270	0.605	TBD	TBD	TBD	
			2.4	0.165	0.185	0.315	0.705	TBD	TBD	TBD	
			3.0	0.205	0.225	0.380	0.855	TBD	TBD	TBD	
			3.6	0.265	0.295	0.500	1.10	TBD	TBD	TBD	

1. Guaranteed by characterization results, unless otherwise specified.

2. Based on characterization done with a 32.768 kHz crystal (MC306-G-06Q-32.768, manufacturer JFVNY) with two 6.8 pF loading capacitors.

**Table 46. Current consumption in VBAT mode**

Symbol	Parameter	Conditions		Typ				Unit
		-	V <sub>BAT</sub> (V)	0 °C	25 °C	55 °C	85 °C	
I <sub>DD</sub> (VBAT)	Backup domain supply current	RTC disabled	1.8	1.00	3.00	19.0	95.0	nA
			2.4	1.00	3.00	22.0	110	
			3.0	1.00	5.00	31.0	150	
			3.6	3.00	11.0	50.0	220	
		RTC enabled and clocked by LSE quartz <sup>(1)</sup>	1.8	140	150	180	275	
			2.4	155	170	200	310	
			3.0	185	200	235	375	
			3.6	230	245	295	485	

1. Based on characterization done with a 32.768 kHz crystal (MC306-G-06Q-32.768, manufacturer JFVNY) with two 6.8 pF loading capacitors.

**Table 47. Current under Reset condition**

Symbol	Conditions	Typ	Max <sup>(1)</sup>	Unit
	V <sub>DD</sub> (V)	25 °C	25 °C	
I <sub>DD</sub> (RST)	1.8 V	600	TBD	μA
	2.4 V	650	TBD	
	3.0 V	700	TBD	
	3.6 V	780	TBD	

1. Guaranteed by characterization results, unless otherwise specified.

### I/O system current consumption

The current consumption of the I/O system has two components: a static and a dynamic.

#### I/O static current consumption

All the I/Os used as inputs with pull-up generate current consumption when the pin is externally held low. The value of this current consumption can be simply computed by using the pull-up/pull-down resistors values given in [Table 67: I/O static characteristics](#).

For the output pins, any external pull-down or external load must also be considered to estimate the current consumption.

Additional I/O current consumption is due to I/Os configured as inputs if an intermediate voltage level is externally applied. This current consumption is caused by the input Schmitt trigger circuits used to discriminate the input value. Unless this specific configuration is required by the application, this supply current consumption can be avoided by configuring these I/Os in analog mode. This is notably the case of ADC input pins which should be configured as analog inputs.

**Caution:** Any floating input pin can also settle to an intermediate voltage level or switch inadvertently, as a result of external electromagnetic noise. To avoid current consumption related to floating pins, these pins must either be configured in analog mode, or forced internally to a definite digital value. This can be done either by using pull-up/down resistors or by configuring the pins in output mode.

#### I/O dynamic current consumption

In addition to the internal peripheral current consumption measured previously (see [Table 48: Peripheral current consumption](#)), the I/Os used by an application also contribute to the current consumption. When an I/O pin switches, it uses the current from the I/O supply voltage to supply the I/O pin circuitry and to charge/discharge the capacitive load (internal or external) connected to the pin:

$$I_{SW} = V_{DD} \times f_{SW} \times C$$

where

- $I_{SW}$  is the current sunk by a switching I/O to charge/discharge the capacitive load.
- $V_{DD}$  is the I/O supply voltage.
- $f_{SW}$  is the I/O switching frequency.
- $C$  is the total capacitance seen by the I/O pin:  $C = C_{Io} + C_{EXT}$ .
- $C_{EXT}$  is the PCB board capacitance plus any connected external device pin capacitance.

The test pin is configured in push-pull output mode and is toggled by software at a fixed frequency.

### On-chip peripheral current consumption

The current consumption of the on-chip peripherals is given in the table below. The device is placed under the following conditions:

- All I/O pins are in analog mode.
- The given value is calculated by measuring the difference of the current consumptions:
  - when the peripheral is clocked on
  - when the peripheral is clocked off
- Ambient operating temperature and supply voltage conditions summarized in [Table 21: Voltage characteristics](#).
- The power consumption of the digital part of the on-chip peripherals is given in the table below. The power consumption of the analog part of the peripherals (where applicable) is indicated in each related section of the datasheet.

**Table 48. Peripheral current consumption**

Peripheral		Range 1	Range 2	LPRun and LPSleep	Unit
AHB1	CRC1	0.417	0.375	1.00	μA/MHz
	DMA1	2.29	1.88	1.45	
	DMA2	2.50	1.94	1.50	
	DMAMUX1	3.96	3.38	2.50	
	All AHB1 peripherals	9.17	7.50	9.30	
AHB2	GPIOA	0.000	0.125	0.200	μA/MHz
	GPIOB	0.000	0.125	0.150	
	GPIOC	0.000	0.125	0.150	
	GPIOH	0.000	0.0625	0.100	
	All AHB2 peripherals	0.625	0.563	0.400	
AHB3	AES1	2.50	2.13	1.80	μA/MHz
	FLASH	7.92	6.56	11.3	
	PKA	3.33	2.75	2.15	
	RNG1	1.04	N/A	N/A	
	RNG1 independent clock domain	0.625	N/A	N/A	
	SRAM1	0.625	0.375	0.550	
	SRAM2	0.417	0.375	0.500	
	All AHB3 peripherals <sup>(1)</sup>	16.0	13.4	16.0	
APB1	DAC	0.833	0.688	0.500	μA/MHz
	I2C1	1.67	1.38	1.05	
	I2C1 independent clock domain	2.29	1.94	1.40	
	I2C2	1.67	1.38	1.05	
	I2C2 independent clock domain	2.50	2.00	1.60	
	I2C3	1.67	1.38	0.900	

Table 48. Peripheral current consumption (continued)

Peripheral		Range 1	Range 2	LPRun and LPSleep	Unit
APB1	I2C3 independent clock domain	2.29	1.88	1.30	μA/MHz
	LPTIM1	1.67	1.44	1.50	
	LPTIM1 independent clock domain	2.50	2.19	1.45	
	LPTIM2	1.67	1.38	0.900	
	LPTIM2 independent clock domain	2.50	2.13	1.55	
	LPTIM3	0.833	0.688	0.650	
	LPTIM3 independent clock domain	2.29	1.94	0.650	
	LPUART1	20.8	1.81	3.55	
	LPUART1 independent clock domain	2.50	2.06	1.35	
	RTCAPB	2.08	1.81	1.50	
	SPI2	1.46	1.19	0.900	
	TIM2	4.58	3.81	2.95	
	USART2	1.88	1.56	1.35	
	USART2 independent clock domain	4.58	3.75	3.05	
	WWDG1	0.417	0.313	0.050	
	All APB1 peripherals <sup>(1)</sup>	19.6	16.1	20.2	
APB2	ADC	2.29	1.00	0.700	μA/MHz
	ADC independent clock domain	0.208	0.125	0.300	
	SPI1	1.25	1.06	0.900	
	TIM1	6.25	5.19	8.30	
	TIM16	2.29	1.94	1.35	
	TIM17	2.29	1.88	1.25	
	USART1	1.67	1.38	1.00	
	USART1 independent clock domain	4.17	3.38	2.90	
	All APB2 peripherals <sup>(1)</sup>	15.8	13.0	15.8	
APB3	SUBGHZSPI	1.46	1.25	1.10	μA/MHz
	All APB3 peripherals	1.46	1.25	1.10	
All peripherals <sup>(1)</sup>		62.9	52.3	59.7	

1. Without independent clocks.

### 5.3.8 Wakeup time from low-power modes and voltage scaling transition times

The wakeup times given in the table below, are the latency between the event and the execution of the first user instruction.

The device goes in low-power mode after the WFE (wait for event) instruction.

Table 49. Low-power mode wakeup timings<sup>(1)</sup>

Symbol	Parameter	Conditions		Typ	Max	Unit
$t_{WUSLEEP}$	Wakeup time from Sleep to Run mode	-		0.188	0.222	N° of CPU cycles
$t_{WULPSLEEP}$	Wakeup time from LPSleep to LPRun mode	Wakeup in Flash with memory in power-down during LPSleep mode (FPDS = 1 in PWR_CR1) and with clock MSI = 2 MHz		3.81	4.38	
$t_{WUSTOP0}$	Wakeup time from Stop 0 mode in Flash memory <sup>(2)</sup>	To Run mode (Range 1)	Wakeup clock MSI = 48 MHz	2.14	2.90	$\mu s$
			Wakeup clock MSI = 16 MHz	2.78	3.58	
			Wakeup clock HSI16 = 16 MHz	1.99	TBD	
			Wakeup clock HSI16 = 16 MHz with HSIKERON enabled	1.01	1.13	
			Wakeup clock MSI = 4 MHz	6.79	8.21	
			Wakeup clock MSI = 2 MHz	10.4	12.2	
		To LPRun mode	Wakeup clock MSI = 2 MHz	10.5	12.3	
$t_{WUSTOP1}$	Wakeup time from Stop 1 mode in Flash memory <sup>(2)</sup>	To Run mode (Range 1)	Wakeup clock MSI = 48 MHz	5.15	6.55	$\mu s$
			Wakeup clock MSI = 16 MHz	5.73	7.14	
			Wakeup clock HSI16 = 16 MHz	5.71	7.10	
			Wakeup clock HSI16 = 16 MHz with HSIKERON enabled	4.57	6.52	
			Wakeup clock MSI = 4 MHz	8.43	9.93	
			Wakeup clock MSI = 2 MHz	11.9	13.7	
		To LPRun mode	Wakeup clock MSI = 2 MHz	10.6	13.9	
$t_{WUSTOP2}$	Wakeup time from Stop 2 mode in Flash memory <sup>(2)</sup>	To Run mode (Range 1)	Wakeup clock MSI = 48 MHz	5.56	6.85	$\mu s$
			Wakeup clock MSI = 16 MHz	6.32	7.59	
			Wakeup clock HSI16 = 16 MHz	6.28	7.51	
			Wakeup clock HSI16 = 16 MHz with HSIKERON enabled	6.26	7.53	
			Wakeup clock MSI = 4 MHz	9.69	10.9	
			Wakeup clock MSI = 2 MHz	14.0	15.4	
		To LPRun mode	Wakeup clock MSI = 2 MHz	10.5	13.8	
$t_{WUSTBY}$	Wakeup time from Standby to Run mode	Range 1	Wakeup clock MSI = 4 MHz	34.3	39.2	$\mu s$
			Wakeup clock MSI = 8 MHz	22.4	25.6	
$t_{WUSHUTD}$	Wakeup time from Shutdown to Run mode	Range 1	Wakeup clock MSI = 4 MHz	264	316	

1. Guaranteed by characterization results ( $V_{DD} = 3\text{ V}$ ,  $T = 25\text{ }^{\circ}\text{C}$ ).

2. Wakeup time is equivalent when code is executed from SRAM1 compared to Flash memory. It is also equivalent when going to Range 2 rather than Range 1.



Table 50. Regulator modes transition times<sup>(1)</sup>

Symbol	Parameter	Conditions	Typ	Max	Unit
$t_{WULPRUN}$	Wakeup time from LPRun to Run mode <sup>(2)</sup>	Code run with MSI = 2 MHz	19.6	TBD	$\mu s$
$t_{VOST}$	Regulator transition time from Range 2 to Range 1 <sup>(3)</sup>	Code run with HSI16	21.9	32.2	
	Regulator transition time from Range 1 to Range 2 <sup>(3)</sup>		23.1	33.9	

1. Guaranteed by characterization results ( $V_{DD} = 3\text{ V}$ ,  $T = 25\text{ }^{\circ}\text{C}$ ).

2. Time until REGLPF flag is cleared in PWR\_SR2.

3. Time until VOSF flag is cleared in PWR\_SR2.

### 5.3.9 External clock source characteristics

#### High-speed external user clock generated from an external source

The high-speed external (HSE32) clock can be supplied with a 32 MHz crystal oscillator or by a TCXO (temperature controlled crystal oscillator).

#### Crystal oscillator

The devices include internal programmable capacitances that can be used to tune the crystal frequency in order to compensate the PCB parasitic one.

Characteristics in the tables below, are measured over recommended operating conditions, unless otherwise specified. Typical values are referred to  $T_A = 25\text{ }^{\circ}\text{C}$  and  $V_{DD} = 3\text{ V}$ .

Table 51. HSE32 crystal requirements<sup>(1)</sup>

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$f_{NOM}$	Oscillator frequency	-	-	32	-	MHz
$f_{TOL}$	Frequency tolerance	Includes initial accuracy, stability over temperature, aging and frequency pulling due to incorrect load capacitance.	-	-	20	ppm
$C_L$	Load capacitance	-	6	-	8	pF
ESR	Equivalent series resistance	-	-	-	100	$\Omega$
$P_D$	Drive level	-	-	-	100	$\mu\text{W}$

1. 32 MHz XTAL is specified for two specific references: NX2016SA and NX1612SA.

Table 52. HSE32 oscillator characteristics

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$t_{SUA(HSE)}$	Startup time for 80% amplitude stabilization	$V_{DDRF}$ stabilized, SUBGHZ_HSEINTRIMR = 0x12, -40 to +125 $^{\circ}\text{C}$ temperature range	-	1000	-	$\mu s$
$t_{SUR(HSE)}$	Startup time for HSEReady signal	$V_{DDRF}$ stabilized, SUBGHZ_HSEINTRIMR = 0x12, -40 to +125 $^{\circ}\text{C}$ temperature range	-	180	-	

Table 52. HSE32 oscillator characteristics (continued)

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$I_{DDRF(HSE)}$	HSE32 current consumption	HSEGMCR = 000, SUBGHZ_HSEINTRIMR = 0x12	-	50	-	$\mu A$
$XOT_{g(HSE)}$	SUBGHZ_HSEINTRIMR granularity	Capacitor bank	-	1	5	ppm
$XOT_{fp(HSE)}$	SUBGHZ_HSEINTRIMR frequency pulling		$\pm 15$	$\pm 30$	-	
$XOT_{nb(HSE)}$	SUBGHZ_HSEINTRIMR number of tuning bits		-	6	-	bit
$XOT_{st(HSE)}$	SUBGHZ_HSEINTRIMR setting time		-	-	0.1	ms

For more information about the trimming methodology of the oscillator, refer to application note *HSE trimming for STM32 wireless MCUs* (AN5042).

## TCXO

Table 53. HSE32 TCXO characteristics

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
VTCXO	Regulated voltage range for TCXO voltage supply	$VDDOP > VTCXO + 200 \text{ mV}$	1.6	1.7	3.3	V
ILTCXO	Load current for TCXO	-	-	1.5	4	mA
TSVTCXO	Startup time for TCXO	From enable to regulated voltage within 25 mV from target	-	-	50	$\mu s$
IDDTXO	Current consumption for TCXO	Quiescent current	-	-	70	$\mu A$
		Relative to load current	-	1.6	2	%
ATCXO	Amplitude voltage for external TCXO applied to OSC_IN pin	Provided through a 220 $\Omega$ resistor in series with a 10 pF capacitance.	0.4	0.6	1.2	V <sub>pk-pk</sub>

## Low-speed external user clock generated from an external source

The low-speed external (LSE) clock can be supplied with a 32.768 kHz crystal resonator oscillator. The information provided in this section is based on design simulation results obtained with typical external components specified in the table below. In the application,

the resonator and the load capacitors have to be placed as close as possible to the oscillator pins to minimize output distortion and startup stabilization time.

Refer to the crystal resonator manufacturer for more details on the resonator characteristics (frequency, package, accuracy).

**Table 54. Low-speed external user clock characteristics<sup>(1)</sup>**

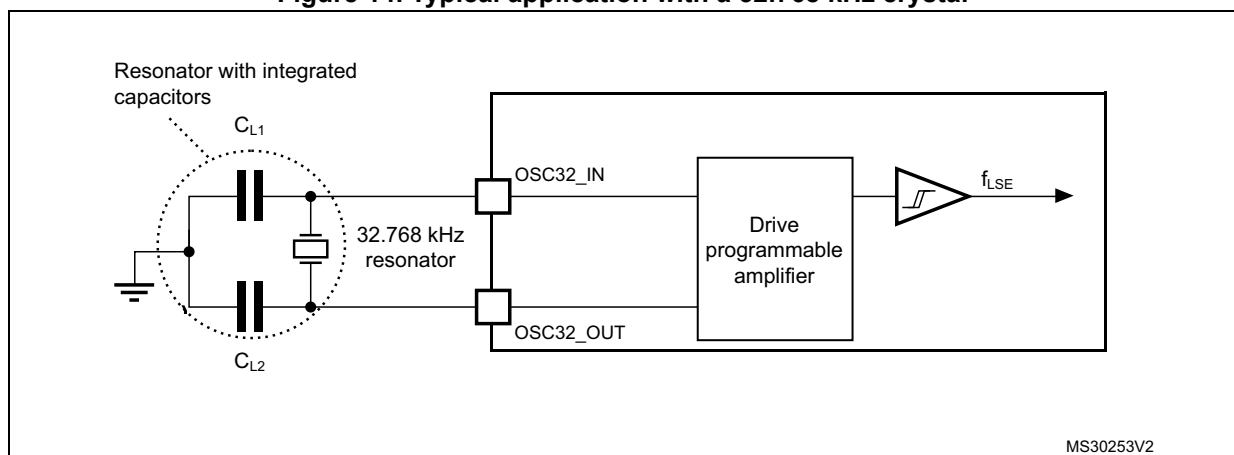
Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$I_{DD(LSE)}$	LSE current consumption	LSEDRV[1:0] = 00 - Low drive capability	-	250	-	nA
		LSEDRV[1:0] = 01 - Medium-low drive capability	-	315	-	
		LSEDRV[1:0] = 10 - Medium-high drive capability	-	500	-	
		LSEDRV[1:0] = 11 - High drive capability	-	630	-	
$G_{mcritmax}$	Maximum critical crystal $g_m$	LSEDRV[1:0] = 00 - Low drive capability	-	-	0.50	$\mu A/V$
		LSEDRV[1:0] = 01 - Medium-low drive capability	-	-	0.75	
		LSEDRV[1:0] = 10 - Medium-high drive capability	-	-	1.70	
		LSEDRV[1:0] = 11 - High drive capability	-	-	2.70	
$t_{SU(LSE)}^{(2)}$	Startup time	$V_{DD}$ stabilized	-	2	-	s

1. Guaranteed by design.

2.  $t_{SU(LSE)}$  is the startup time measured from the moment it is enabled (by software) until a stable 32 MHz oscillation is reached. This value is measured for a standard crystal and it can vary significantly with the crystal manufacturer.

For more information on the crystal selection, refer to application note *Oscillator design guide for STM8AF/AL/S and STM32 microcontrollers* (AN2867).

**Figure 14. Typical application with a 32.768 kHz crystal**

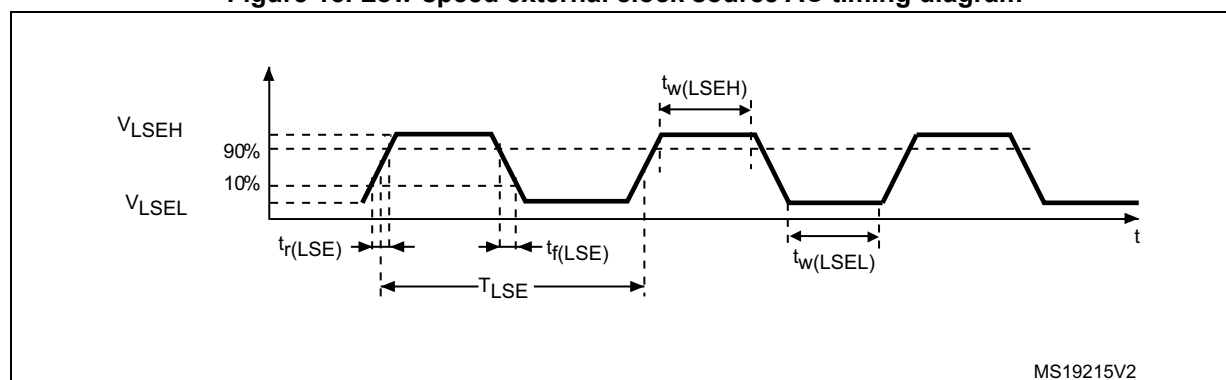


**Note:** No external resistors are required between OSC32\_IN and OSC32\_OUT, and it is forbidden to add one.

In bypass mode, the LSE oscillator is switched off and the input pin is a standard GPIO.

The external clock signal has to respect the I/O characteristics detailed in [Section 5.3.16: I/O port characteristics](#). The recommend clock input waveform is shown in the figure below.

Figure 15. Low-speed external clock source AC timing diagram



MS19215V2

Table 55. Low-speed external user clock characteristics<sup>(1)</sup> – Bypass mode

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$f_{LSE\_ext}$	User external clock source frequency	-	21.2	32.768	44.4	kHz
$V_{LSEH}$	OSC32_IN input pin high-level voltage	-	$0.7 \times V_{DDx}$	-	$V_{DDx}$	V
$V_{LSEL}$	OSC32_IN input pin low-level voltage	-	$V_{SS}$	-	$0.3 \times V_{DDx}$	
$t_{w(LSEH)}$ $t_{w(LSEL)}$	OSC32_IN high or low time	-	250	-	-	ns
$f_{tolLSE}$	Frequency tolerance	Includes initial accuracy, stability over temperature, aging and frequency pulling	-500	-	+500	ppm

1. Guaranteed by design.

### 5.3.10 Internal clock source characteristics

Parameters given in the table below are derived from tests performed under ambient temperature and supply voltage conditions summarized in [Table 25: General operating conditions](#). The provided curves are characterization results, not tested in production.

#### High-speed internal (HSI16) RC oscillator

Table 56. HSI16 oscillator characteristics<sup>(1)</sup>

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$f_{HSI16}$	HSI16 frequency	$V_{DD} = 3.0\text{ V}$ , $T_A = 30\text{ }^{\circ}\text{C}$	15.88	-	16.08	MHz

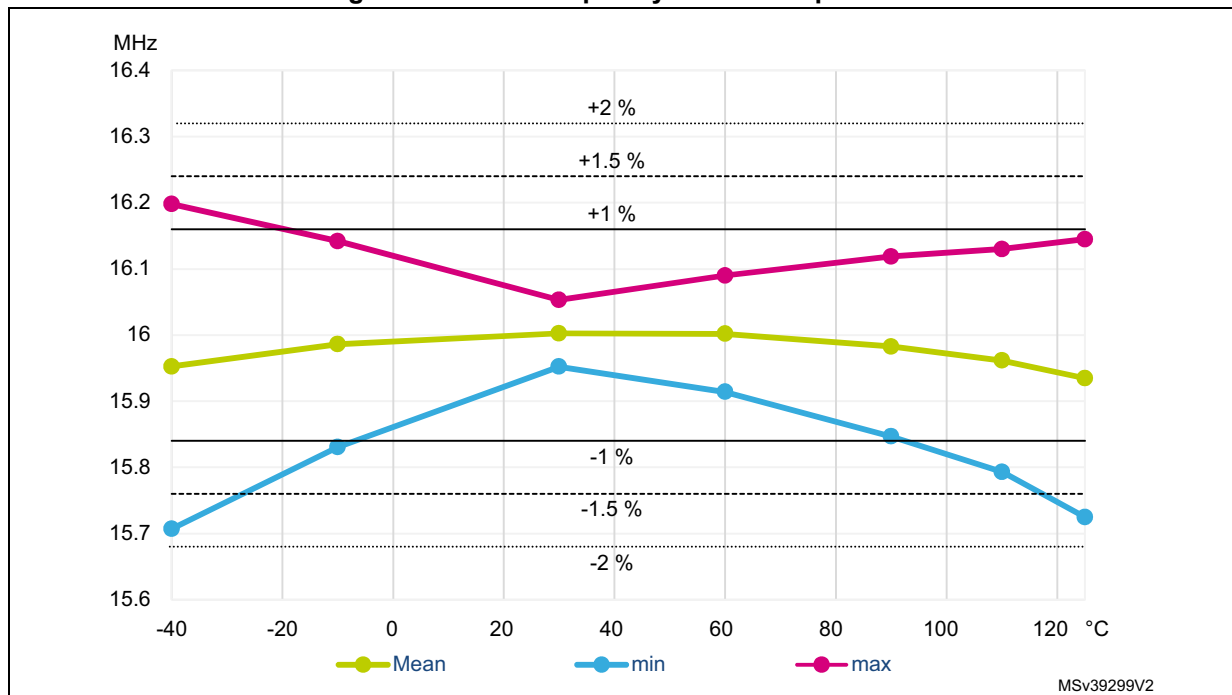
Table 56. HSI16 oscillator characteristics<sup>(1)</sup> (continued)

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
TRIM	HSI16 user trimming step	Trimming code is not a multiple of 48	0.2	0.3	0.4	%
		Trimming code is a multiple of 48	-4	-6	-8	
DuCy(HSI16) <sup>(2)</sup>	Duty cycle	-	45	-	55	
$\Delta_{Temp}(HSI16)$	HSI16 oscillator frequency drift over temperature	$T_A = 0 \text{ to } 85 \text{ }^\circ\text{C}$	-1	-	1	
		$T_A = -40 \text{ to } 125 \text{ }^\circ\text{C}$	-2	-	1.5	
$\Delta_{VDD}(HSI16)$	HSI16 oscillator frequency drift over $V_{DD}$	$V_{DD} = 1.8 \text{ V to } 3.6 \text{ V}$	-0.1	-	0.05	$\mu\text{s}$
$t_{su}(HSI16)^{(2)}$	HSI16 oscillator start-up time	-	-	0.8	1.2	
$t_{stab}(HSI16)^{(2)}$	HSI16 oscillator stabilization time	-	-	3	5	
$I_{DD}(HSI16)^{(2)}$	HSI16 oscillator power consumption	-	-	155	190	

1. Guaranteed by characterization results.

2. Guaranteed by design.

Figure 16. HSI16 frequency versus temperature



## Multi-speed internal (MSI) RC oscillator

Table 57. MSI oscillator characteristics<sup>(1)</sup>

Symbol	Parameter	Conditions		Min	Typ	Max	Unit
$f_{\text{MSI}}$	MSI frequency after factory calibration, done at $V_{\text{DD}} = 3 \text{ V}$ and $T_{\text{A}} = 30 \text{ }^{\circ}\text{C}$	MSI mode	Range 0	98.7	100	101.3	kHz
			Range 1	197.4	200	202.6	
			Range 2	394.8	400	405.2	
			Range 3	789.6	800	810.4	
			Range 4	0.987	1	1.013	MHz
			Range 5	1.974	2	2.026	
			Range 6	3.948	4	4.052	
			Range 7	7.896	8	8.104	
			Range 8	15.79	16	16.21	
			Range 9	23.69	24	24.31	
			Range 10	31.58	32	32.42	
			Range 11	47.38	48	48.62	
		PLL mode XTAL = 32.768 kHz	Range 0	-	98.304	-	kHz
			Range 1	-	196.608	-	
			Range 2	-	393.216	-	
			Range 3	-	786.432	-	
			Range 4	-	1.016	-	MHz
			Range 5	-	1.999	-	
			Range 6	-	3.998	-	
			Range 7	-	7.995	-	
			Range 8	-	15.991	-	
			Range 9	-	23.986	-	
			Range 10	-	32.014	-	
			Range 11	-	48.005	-	
$\Delta_{\text{TEMP}}(\text{MSI})^{(2)}$	MSI oscillator frequency drift over temperature	MSI mode	$T_{\text{A}} = 0 \text{ to } 85 \text{ }^{\circ}\text{C}$	-3.5	-	3	%
			$T_{\text{A}} = -40 \text{ to } 125 \text{ }^{\circ}\text{C}$	-8	-	6	

Table 57. MSI oscillator characteristics<sup>(1)</sup> (continued)

Symbol	Parameter	Conditions			Min	Typ	Max	Unit
$\Delta V_{DD}(MSI)^{(2)}$	MSI oscillator frequency drift over $V_{DD}$ (reference is 3 V)	MSI mode	Range 0 to 3	$V_{DD} = 1.8$ to 3.6 V	-1.2	-	0.5	%
				$V_{DD} = 2.4$ to 3.6 V	-0.5	-		
			Range 4 to 7	$V_{DD} = 1.8$ to 3.6 V	-2.5	-	0.7	
				$V_{DD} = 2.4$ to 3.6 V	-0.8	-		
			Range 8 to 11	$V_{DD} = 1.8$ to 3.6 V	-5	-	1	
				$V_{DD} = 2.4$ to 3.6 V	-1.6	-		
$\Delta F_{SAMPLING}(MSI)^{(2)(4)}$	Frequency variation in sampling mode <sup>(3)</sup>	MSI mode	$T_A = -40$ to 85 °C		-	1	2	
			$T_A = -40$ to 125 °C		-	2	4	
CC jitter(MSI) <sup>(4)</sup>	RMS cycle-to-cycle jitter	PLL mode Range 11		-	-	60	-	ps
P jitter(MSI) <sup>(4)</sup>	RMS period jitter	PLL mode Range 11		-	-	50	-	
$t_{SU}(MSI)^{(4)}$	MSI oscillator start-up time	Range 0		-	-	10	20	$\mu s$
		Range 1		-	-	5	10	
		Range 2		-	-	4	8	
		Range 3		-	-	3	7	
		Range 4 to 7		-	-	3	6	
		Range 8 to 11		-	-	2.5	6	
$t_{STAB}(MSI)^{(4)}$	MSI oscillator stabilization time	PLL mode Range 11	10 % of final frequency	-	-	0.25	0.5	ms
			5 % of final frequency	-	-	0.5	1.25	
			1 % of final frequency	-	-	-	2.5	

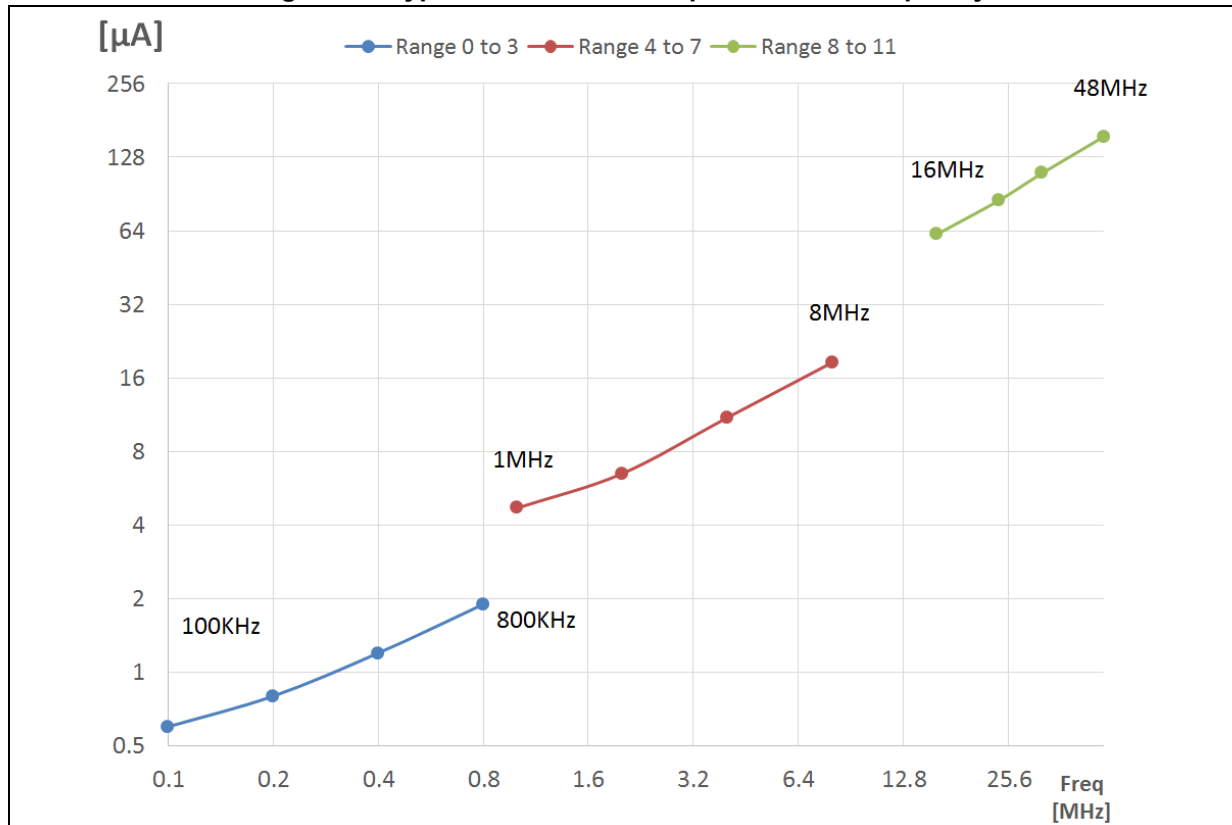
Table 57. MSI oscillator characteristics<sup>(1)</sup> (continued)

Symbol	Parameter	Conditions		Min	Typ	Max	Unit
$I_{DD}(MSI)^{(4)}$	MSI oscillator power consumption	MSI and PLL mode	Range 0	-	-	0.6	1
			Range 1	-	-	0.8	1.2
			Range 2	-	-	1.2	1.7
			Range 3	-	-	1.9	2.5
			Range 4	-	-	4.7	6
			Range 5	-	-	6.5	9
			Range 6	-	-	11	15
			Range 7	-	-	18.5	25
			Range 8	-	-	62	80
			Range 9	-	-	85	110
			Range 10	-	-	110	130
			Range 11	-	-	155	190

1. Guaranteed by characterization results.
2. This is a deviation for an individual part once the initial frequency has been measured.
3. Sampling mode means LPRun and LPSleep modes with temperature sensor disabled.
4. Guaranteed by design.



Figure 17. Typical current consumption vs. MSI frequency



### Low-speed internal (LSI) RC oscillator

Table 58. LSI oscillator characteristics<sup>(1)</sup>

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$f_{LSI}$	LSI frequency	$V_{DD} = 3\text{ V}$ , $T_A = 30\text{ }^{\circ}\text{C}$	31.04	-	32.96	kHz
		$V_{DD} = 1.8\text{ to }3.6\text{ V}$ , $T_A = -40\text{ to }125\text{ }^{\circ}\text{C}$	29.5	-	34	
$t_{SU(LSI)}^{(2)}$	LSI oscillator startup time	-	-	80	130	$\mu\text{s}$
$t_{STAB(LSI)}^{(2)}$	LSI oscillator stabilization time	5 % of final frequency	-	125	180	
$I_{DD(LSI)}^{(2)}$	LSI oscillator power consumption	-	-	110	180	nA

1. Guaranteed by characterization results.

2. Guaranteed by design.

### 5.3.11 PLL characteristics

Parameters given in the table below are derived from tests performed under temperature and  $V_{DD}$  supply voltage conditions summarized in [Table 25: General operating conditions](#).

**Table 59. PLL characteristics<sup>(1)</sup>**

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$f_{PLL\_IN}$	PLL input clock <sup>(2)</sup>	-	2.66	-	16	MHz
	PLL input clock duty cycle	-	45	-	55	%
$f_{PLL\_P\_OUT}$	PLL multiplier output clock P	Voltage scaling Range 1	3	-	48	MHz
		Voltage scaling Range 2	3	-	16	
$f_{PLL\_Q\_OUT}$	PLL multiplier output clock Q	Voltage scaling Range 1	12	-	48	
		Voltage scaling Range 2	12	-	16	
$f_{PLL\_R\_OUT}$	PLL multiplier output clock R	Voltage scaling Range 1	12	-	48	
		Voltage scaling Range 2	12	-	16	
$f_{VCO\_OUT}$	PLL VCO output	Voltage scaling Range 1	96	-	344	
		Voltage scaling Range 2	96	-	128	
$t_{LOCK}$	PLL lock time	-	-	15	40	μs
Jitter	RMS cycle-to-cycle jitter	System clock 48 MHz	-	40	-	ps
	RMS period jitter		-	30	-	
$I_{DD}(PLL)$	PLL power consumption on $V_{DD}$ <sup>(1)</sup>	VCO freq = 96 MHz	-	200	260	μA
		VCO freq = 192 MHz	-	300	380	
		VCO freq = 344 MHz	-	520	650	

1. Guaranteed by design.

2. Take care of using the appropriate division factor M to obtain the specified PLL input clock values. The M factor is shared between the two PLLs.

### 5.3.12 Flash memory characteristics

**Table 60. Flash memory characteristics<sup>(1)</sup>**

Symbol	Parameter	Conditions	Typ	Max	Unit
$t_{prog}$	64-bit programming time	-	81.7	90.8	μs
$t_{prog\_row}$	One row (64 double-words) programming time	Normal programming	5.2	5.5	ms
		Fast programming	3.8	4.0	
$t_{prog\_page}$	One 2-Kbyte page programming time	Normal programming	41.8	43.0	
		Fast programming	30.4	31.0	
$t_{ERASE}$	2-Kbyte page erase time	-	22.0	24.5	
$t_{ME}$	Mass erase time	-	22.1	25.0	

Table 60. Flash memory characteristics<sup>(1)</sup> (continued)

Symbol	Parameter	Conditions	Typ	Max	Unit
I <sub>DD</sub>	Average consumption from V <sub>DD</sub>	Write mode	3.4	-	mA
		Erase mode	3.4	-	
	Maximum current (peak)	Write mode	7 (for 6 μs)	-	
		Erase mode	7 (for 67 μs)	-	

1. Guaranteed by design.

Table 61. Flash memory endurance and data retention

Symbol	Parameter	Conditions	Min <sup>(1)</sup>	Unit
N <sub>END</sub>	Endurance	T <sub>A</sub> = -40 to +105 °C	10	kcycles
t <sub>RET</sub>	Data retention	1 kcycle <sup>(2)</sup> at T <sub>A</sub> = 85 °C	30	Years
		1 kcycle <sup>(2)</sup> at T <sub>A</sub> = 105 °C	15	
		1 kcycle <sup>(2)</sup> at T <sub>A</sub> = 125 °C	7	
		10 kcycles <sup>(2)</sup> at T <sub>A</sub> = 55 °C	30	
		10 kcycles <sup>(2)</sup> at T <sub>A</sub> = 85 °C	15	
		10 kcycles <sup>(2)</sup> at T <sub>A</sub> = 105 °C	10	

1. Guaranteed by characterization results.

2. Cycling performed over the whole temperature range.

### 5.3.13 EMC characteristics

Susceptibility tests are performed on a sample basis during device characterization.

#### Functional EMS (electromagnetic susceptibility)

While a simple application is executed on the device (toggling two LEDs through I/O ports), the device is stressed by the following electromagnetic events until a failure occurs (failure indicated by the LEDs):

- **ESD** (electrostatic discharge, positive and negative) applied to all device pins until a functional disturbance occurs (test compliant with IEC 61000-4-2 standard)
- **FTB** (burst of fast transient voltage, positive and negative) applied to VDD and VSS pins, through a 100 pF capacitor, until a functional disturbance occurs (test compliant with IEC 61000-4-4 standard)

A device reset allows normal operations to be resumed.

The test results given in the table below, are based on the EMS levels and classes defined in application note *EMC design guide for STM8, STM32 and Legacy MCUs* (AN1709).

Table 62. EMS characteristics

Symbol	Parameter	Conditions	Level/Class
$V_{FESD}$	Voltage limits to be applied on any I/O pin to induce a functional disturbance	$V_{DD} = 3.3\text{ V}$ , $T_A = +25\text{ }^{\circ}\text{C}$ , $f_{HCLK} = 48\text{ MHz}$ , conforming to IEC 61000-4-2	2B
$V_{EFTB}$	Fast transient voltage burst limits to be applied through 100 pF on VDD and VSS pins to induce a functional disturbance	$V_{DD} = 3.3\text{ V}$ , $T_A = +25\text{ }^{\circ}\text{C}$ , $f_{HCLK} = 48\text{ MHz}$ , conforming to IEC 61000-4-4	5A

### Designing hardened software to avoid noise problems

EMC characterization and optimization are performed at component level with a typical application environment and simplified MCU software.

*Note: Good EMC performance is highly dependent on the user application and the software in particular. It is then recommended that the user applies EMC software optimization and prequalification tests in relation with the EMC level requested for the application.*

#### Software recommendations

The software flow must include the management of runaway conditions such as:

- corrupted program counter
- unexpected reset
- critical data corruption (control registers)

#### Prequalification trials

Most of the common failures (unexpected reset and program counter corruption) can be reproduced by manually forcing a low state on the NRST pin or the oscillator pins for 1 s.

To complete these trials, ESD stress can be applied directly on the device, over the range of specification values. When unexpected behavior is detected, the software can be hardened to prevent unrecoverable errors occurring. For more details, refer to the application note *Software techniques for improving microcontrollers EMC performance* (AN1015).

### Electromagnetic interference (EMI)

The electromagnetic field emitted by the device is monitored while a simple application is executed (toggling two LEDs through the I/O ports). This emission test is compliant with the IEC 61967-2 standard, that specifies the test board and the pin loading.

Table 63. EMI characteristics

Symbol	Parameter	Conditions	Monitored frequency band	Peripheral ON SMPS OFF $f_{HSE} = /f_{CPUM4}, f_{CPUM0}]$	Unit
				$f_{HSE} = 32 \text{ MHz}$ $f_{CPU} = 48 \text{ MHz}$	
$S_{EMI}$	Peak level	$V_{DD} = 3.6 \text{ V}$ , $T_A = 25 \text{ }^{\circ}\text{C}$ , UFBGA73 package compliant with IEC 61967-2	0.1 MHz to 30 MHz	1	dB $\mu$ V
			30 MHz to 130 MHz	4	
			130 MHz to 1 GHz	0	
			1 GHz to 2 GHz	7	
			EMI level	2	-

### 5.3.14 Electrical sensitivity characteristics

Based on three different tests (ESD, LU) using specific measurement methods, the device is stressed to determine its performance in terms of electrical sensitivity.

#### Electrostatic discharge (ESD)

Electrostatic discharges (a positive then a negative pulse separated by 1 s) are applied to the pins of each sample according to each pin combination. The sample size depends on the number of supply pins in the device (3 parts  $\times$  (n+1) supply pins). This test conforms to the ANSI/JEDEC standard.

Table 64. ESD absolute maximum ratings

Symbol	Ratings	Conditions	Class	Maximum value <sup>(1)</sup>	Unit
$V_{ESD(HBM)}$	Electrostatic discharge voltage (human body model)	$T_A = +25 \text{ }^{\circ}\text{C}$ , conforming to ANSI/ESDA/JEDEC JS-001	2	2000	V
$V_{ESD(CDM)}$	Electrostatic discharge voltage (charge device model)	$T_A = +25 \text{ }^{\circ}\text{C}$ , conforming to ANSI/ESD STM5.3.1 JS-002	C2a	500	

1. Guaranteed by characterization results.

#### Static latch-up

The following complementary static tests are required on three parts to assess the latch-up performance:

- A supply overvoltage is applied to each power supply pin.
- A current injection is applied to each input, output and configurable I/O pin.

These tests are compliant with EIA/JESD 78A IC latch-up standard.

Table 65. Electrical sensitivities

Symbol	Parameter	Conditions	Class
LU	Static latch-up class	$T_A = +105 \text{ }^{\circ}\text{C}$ conforming to JESD78A	Level A

### 5.3.15 I/O current injection characteristics

As a general rule, current injection to the I/O pins, due to external voltage below  $V_{SS}$  or above  $V_{DD}$  (for standard, 3V-capable I/O pins), must be avoided during normal product operation. However, in order to give an indication of the robustness of the microcontroller in case abnormal injection accidentally happens, susceptibility tests are performed on a sample basis during device characterization.

#### Functional susceptibility to I/O current injection

While a simple application is executed on the device, the device is stressed by injecting current into the I/O pins programmed in floating-input mode. While current is injected into the I/O pin, one at a time, the device is checked for functional failures.

The failure is indicated by an out-of-range parameter: ADC error above a certain limit (higher than 5 LSB TUE), out of conventional limits of induced leakage current on adjacent pins (out of the  $-5\ \mu\text{A}/0\ \mu\text{A}$  range) or other functional failure (for example reset occurrence or oscillator frequency deviation).

The characterization results are given in the table below.

Negative induced leakage current is caused by negative injection and positive induced leakage current is caused by positive injection.

**Table 66. I/O current injection susceptibility<sup>(1)</sup>**

Symbol	Description	Functional susceptibility		Unit
		Negative injection	Positive injection	
$I_{INJ}$	Injected current on all pins except PB0, PB1	-5	N/A <sup>(2)</sup>	mA
	Injected current on PB0, PB1 pins	-5	0	

1. Guaranteed by characterization results.

2. Injection not possible.

### 5.3.16 I/O port characteristics

#### General input/output characteristics

Unless otherwise specified, the parameters given in the table below are derived from tests performed under the conditions summarized in [Table 25: General operating conditions](#). All I/Os are designed as CMOS- and TTL-compliant.

Table 67. I/O static characteristics

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$V_{IL}$	I/O input low-level voltage <sup>(1)</sup>	$1.8\text{ V} < V_{DD} < 3.6\text{ V}$	-	-	$0.3 \times V_{DD}$	V
	I/O input low-level voltage <sup>(2)</sup>				$0.39 \times V_{DD} - 0.06$	
$V_{IH}$	I/O input high-level voltage <sup>(1)</sup>		$0.7 \times V_{DD}$	-	-	
	I/O input high-level voltage <sup>(2)</sup>		$0.49 \times V_{DD} + 0.26$	-	-	
$V_{hys}$	TT_xx, FT_XXX and NRST I/O input hysteresis		-	200	-	mV
$I_{lkg}$	FT_xx input leakage current	$0 \leq V_{IN} \leq \text{Max}(V_{DDXXX})^{(3)}$	-	-	$\pm 100$	nA
		$\text{Max}(V_{DDXXX}) \leq V_{IN} \leq \text{Max}(V_{DDXXX}) + 1\text{ V}^{(2)(3)(4)}$	-	-	650	
		$\text{Max}(V_{DDXXX}) + 1\text{ V} < V_{IN} \leq 5.5\text{ V}^{(2)(3)(4)(5)(6)}$	-	-	$200^{(7)}$	
	FT_lu, FT_u and PC3 IO input leakage current	$0 \leq V_{IN} \leq \text{Max}(V_{DDXXX})^{(3)}$	-	-	$\pm 150$	
		$\text{Max}(V_{DDXXX}) \leq V_{IN} \leq \text{Max}(V_{DDXXX}) + 1\text{ V}^{(2)(3)}$	-	-	2500	
		$\text{Max}(V_{DDXXX}) + 1\text{ V} < V_{IN} \leq 5.5\text{ V}^{(1)(3)(4)(8)}$	-	-	250	
	TT_xx input leakage current	$V_{IN} \leq \text{Max}(V_{DDXXX})^{(3)}$	-	-	$\pm 150$	
		$\text{Max}(V_{DDXXX}) \leq V_{IN} < 3.6\text{ V}^{(3)}$	-	-	2000	
$R_{PU}$	Weak pull-up equivalent resistor <sup>(1)</sup>	$V_{IN} = V_{SS}$	25	40	55	k $\Omega$
$R_{PD}$	Weak pull-down equivalent resistor <sup>(1)</sup>	$V_{IN} = V_{DD}$	25	40	55	
$C_{IO}$	I/O pin capacitance	-	-	5	-	pF

1. Tested in production.

2. Guaranteed by design, not tested in production.

3. Represents the pad leakage of the I/O itself. The total product pad leakage is given by  $I_{Total\_leak\_max} = 10\text{ }\mu\text{A} + \text{number of I/Os where } V_{IN} \text{ is applied on the pad} \times I_{lkg}(\text{Max})$ .

4.  $\text{Max}(V_{DDXXX})$  is the maximum value among all the I/O supplies.

5.  $V_{IN}$  must be lower than  $[\text{Max}(V_{DDXXX}) + 3.6\text{ V}]$ .

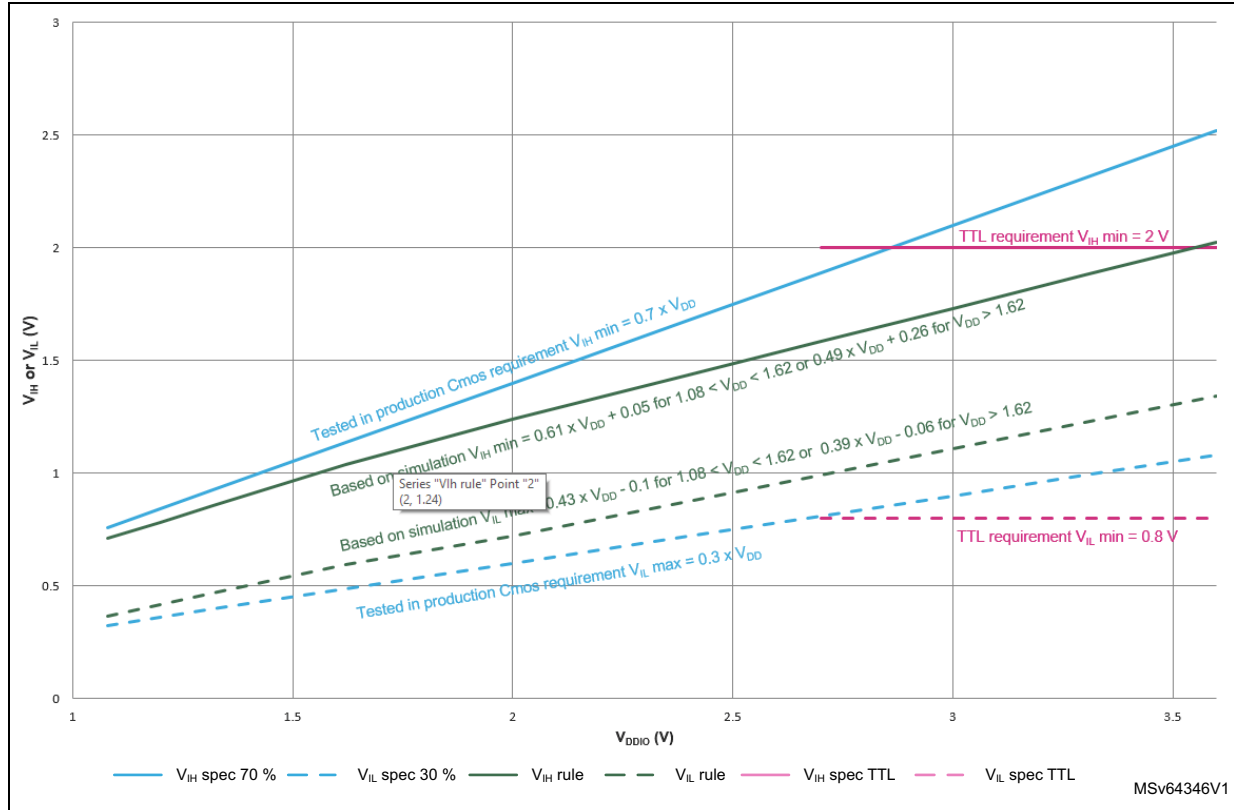
6. Refer to the figure below.

7. To sustain a voltage higher than  $[\text{Min}(V_{DD}, V_{DDA}) + 0.3\text{ V}]$ , the internal pull-up and pull-down resistors must be disabled on all FT\_xx I/O, except FT\_lu, FT\_u and PC3.

8. Pull-up and pull-down resistors are designed with a true resistance in series with a switchable PMOS/NMOS, whose contribution to the series resistance is minimal ( $\sim 10\%$ ).

All I/Os are CMOS- and TTL-compliant (no software configuration required). Their characteristics cover more than the strict CMOS-technology or TTL parameters, as shown in the figure below.

**Figure 18. I/O input characteristics -  $V_{IL}$  and  $V_{IH}$  on all I/Os except BOOT0**



### Output driving current

The GPIOs can sink or source up to  $\pm 8$  mA, and sink or source up to  $\pm 20$  mA (with a relaxed  $V_{OL}/V_{OH}$ ).

In the user application, the number of I/O pins that can drive current must be limited to respect the absolute maximum rating specified in [Section 5.2: Absolute maximum ratings](#).

The sum of the currents sourced by all the I/Os on  $V_{DD}$ , plus the maximum consumption of the MCU sourced on  $V_{DD}$ , cannot exceed the absolute maximum rating  $\Sigma I_{VDD}$  (see [Table 21: Voltage characteristics](#)).

The sum of the currents sunk by all the I/Os on  $V_{SS}$ , plus the maximum consumption of the MCU sunk on  $V_{SS}$ , cannot exceed the absolute maximum rating  $\Sigma I_{VSS}$  (see [Table 21: Voltage characteristics](#)).

### Output voltage levels

Unless otherwise specified, the parameters given in the table below are derived from tests performed under the ambient temperature and supply voltage conditions summarized in [Table 25: General operating conditions](#). All I/Os are CMOS- and TTL-compliant (FT or TT unless otherwise specified).



Table 68. Output voltage characteristics<sup>(1)</sup>

Symbol	Parameter	Conditions	Min	Max	Unit
$V_{OL}^{(2)}$	Output low-level voltage for an I/O pin	CMOS port <sup>(3)</sup> $ I_{IO}  = 8 \text{ mA}$ , $V_{DD} \geq 2.7 \text{ V}$	-	0.4	V
$V_{OH}^{(2)}$	Output high-level voltage for an I/O pin		$V_{DD} - 0.4$	-	
$V_{OL}^{(2)}$	Output low-level voltage for an I/O pin	TTL port <sup>(3)</sup> $ I_{IO}  = 8 \text{ mA}$ , $V_{DD} \geq 2.7 \text{ V}$	-	0.4	
$V_{OH}^{(2)}$	Output high-level voltage for an I/O pin		2.4	-	
$V_{OL}^{(2)}$	Output low-level voltage for an I/O pin	$ I_{IO}  = 20 \text{ mA}$ , $V_{DD} \geq 2.7 \text{ V}$	-	1.3	
$V_{OH}^{(2)}$	Output high-level voltage for an I/O pin		$V_{DD} - 1.3$	-	
$V_{OL}^{(2)}$	Output low-level voltage for an I/O pin	$ I_{IO}  = 4 \text{ mA}$ , $V_{DD} \geq 1.8 \text{ V}$	-	0.4	
$V_{OH}^{(2)}$	Output high-level voltage for an I/O pin		$V_{DD} - 0.45$	-	
$V_{OLFM+}^{(2)}$	Output low-level voltage for an FT I/O pin in FM+ mode (FT I/O with "f" option)	$ I_{IO}  = 20 \text{ mA}$ , $V_{DD} \geq 2.7 \text{ V}$	-	0.4	
		$ I_{IO}  = 10 \text{ mA}$ , $V_{DD} \geq 1.8 \text{ V}$	-	0.4	
		$ I_{IO}  = 2 \text{ mA}$ , $1.8 \text{ V} \geq V_{DD} \geq 1.08 \text{ V}$	-	0.4	

1. The  $I_{IO}$  current sourced or sunk by the device must always respect the absolute maximum rating specified in [Table 21: Voltage characteristics](#). The sum of the currents sourced or sunk by all the I/Os (I/O ports and control pins) must always respect the absolute maximum ratings  $\Sigma I_{IO}$ .
2. Guaranteed by design.
3. TTL and CMOS outputs are compatible with JEDEC standards JESD36 and JESD52.

### Input/output AC characteristics

Unless otherwise specified, the parameters given in the table below are derived from tests performed under the ambient temperature and supply voltage conditions summarized in [Table 25: General operating conditions](#).

Table 69. I/O AC characteristics<sup>(1)(2)</sup>

OSPEEDx[1:0] <sup>(3)</sup>	Symbol	Parameter	Conditions	Min	Max	Unit
0b00	Fmax	Maximum frequency	$C = 50 \text{ pF}$ , $2.7 \text{ V} \leq V_{DD} \leq 3.6 \text{ V}$	-	5	MHz
			$C = 50 \text{ pF}$ , $1.8 \text{ V} \leq V_{DD} \leq 2.7 \text{ V}$	-	1	
			$C = 10 \text{ pF}$ , $2.7 \text{ V} \leq V_{DD} \leq 3.6 \text{ V}$	-	10	
			$C = 10 \text{ pF}$ , $1.8 \text{ V} \leq V_{DD} \leq 2.7 \text{ V}$	-	1.5	
	Tr/Tf	Output rise and fall time	$C = 50 \text{ pF}$ , $2.7 \text{ V} \leq V_{DD} \leq 3.6 \text{ V}$	-	25	ns
			$C = 50 \text{ pF}$ , $1.8 \text{ V} \leq V_{DD} \leq 2.7 \text{ V}$	-	52	
			$C = 10 \text{ pF}$ , $2.7 \text{ V} \leq V_{DD} \leq 3.6 \text{ V}$	-	17	
			$C = 10 \text{ pF}$ , $1.8 \text{ V} \leq V_{DD} \leq 2.7 \text{ V}$	-	37	

Table 69. I/O AC characteristics<sup>(1)(2)</sup> (continued)

OSPEEDx[1:0] <sup>(3)</sup>	Symbol	Parameter	Conditions	Min	Max	Unit
0b01	Fmax	Maximum frequency	C = 50 pF, 2.7 V ≤ V <sub>DD</sub> ≤ 3.6 V	-	25	MHz
			C = 50 pF, 1.8 V ≤ V <sub>DD</sub> ≤ 2.7 V	-	10	
			C = 10 pF, 2.7 V ≤ V <sub>DD</sub> ≤ 3.6 V	-	50	
			C = 10 pF, 1.8 V ≤ V <sub>DD</sub> ≤ 2.7 V	-	15	
	Tr/Tf	Output rise and fall time	C = 50 pF, 2.7 V ≤ V <sub>DD</sub> ≤ 3.6 V	-	9	ns
			C = 50 pF, 1.8 V ≤ V <sub>DD</sub> ≤ 2.7 V	-	16	
			C = 10 pF, 2.7 V ≤ V <sub>DD</sub> ≤ 3.6 V	-	4.5	
			C = 10 pF, 1.8 V ≤ V <sub>DD</sub> ≤ 2.7 V	-	9	
0b10	Fmax	Maximum frequency	C = 50 pF, 2.7 V ≤ V <sub>DD</sub> ≤ 3.6 V	-	50	MHz
			C = 50 pF, 1.8 V ≤ V <sub>DD</sub> ≤ 2.7 V	-	25	
			C = 10 pF, 2.7 V ≤ V <sub>DD</sub> ≤ 3.6 V	-	100 <sup>(4)</sup>	
			C = 10 pF, 1.8 V ≤ V <sub>DD</sub> ≤ 2.7 V	-	37.5	
	Tr/Tf	Output rise and fall time	C = 50 pF, 2.7 V ≤ V <sub>DD</sub> ≤ 3.6 V	-	5.8	ns
			C = 50 pF, 1.8 V ≤ V <sub>DD</sub> ≤ 2.7 V	-	11	
			C = 10 pF, 2.7 V ≤ V <sub>DD</sub> ≤ 3.6 V	-	2.5	
			C = 10 pF, 1.8 V ≤ V <sub>DD</sub> ≤ 2.7 V	-	5	
0b11	Fmax	Maximum frequency	C = 30 pF, 2.7 V ≤ V <sub>DD</sub> ≤ 3.6 V	-	120 <sup>(4)</sup>	MHz
			C = 30 pF, 1.8 V ≤ V <sub>DD</sub> ≤ 2.7 V	-	50	
			C = 10 pF, 2.7 V ≤ V <sub>DD</sub> ≤ 3.6 V	-	180 <sup>(4)</sup>	
			C = 10 pF, 1.8 V ≤ V <sub>DD</sub> ≤ 2.7 V	-	75 <sup>(4)</sup>	
	Tr/Tf	Output rise and fall time	C = 30 pF, 2.7 V ≤ V <sub>DD</sub> ≤ 3.6 V	-	3.3	ns
			C = 30 pF, 1.8 V ≤ V <sub>DD</sub> ≤ 2.7 V	-	6	
			C = 10 pF, 2.7 V ≤ V <sub>DD</sub> ≤ 3.6 V	-	1.7	
			C = 10 pF, 1.8 V ≤ V <sub>DD</sub> ≤ 2.7 V	-	3.3	

1. The maximum frequency is defined with (T<sub>r</sub> + T<sub>f</sub>) ≤ 2/3 T, and duty cycle comprised between 45 and 55 %.

2. The fall and rise time are defined, respectively, between 90 and 10 %, and between 10 and 90 % of the output waveform.

3. OSPEED0[1:0] in GPIOA\_OSPEEDR, GPIOB\_OSPEEDR and GPIOC\_OSPEEDR. OSPEED3[1:0] in GPIOH\_OSPEEDR

4. This value represents the I/O capability but the maximum system frequency is limited to 48 MHz.

### 5.3.17 NRST pin characteristics

The NRST pin input driver uses the CMOS technology. It is connected to a permanent pull-up resistor, R<sub>PJ</sub>.

Unless otherwise specified, the parameters given in the table below are derived from tests performed under the ambient temperature and supply voltage conditions summarized in [Table 25: General operating conditions](#).

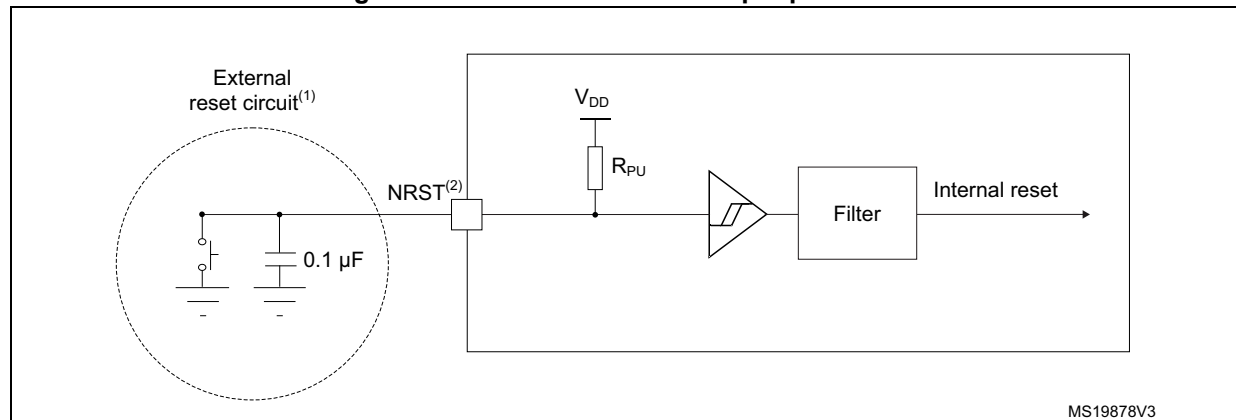
Table 70. NRST pin characteristics<sup>(1)</sup>

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$V_{IL(NRST)}$	NRST input low level voltage	-	-	-	$0.3 \times V_{DD}$	V
$V_{IH(NRST)}$	NRST input high level voltage	-	$0.7 \times V_{DD}$	-	-	
$V_{hys(NRST)}$	NRST Schmitt trigger voltage hysteresis	-	-	200	-	mV
$R_{PU}$	Weak pull-up equivalent resistor <sup>(2)</sup>	$V_{IN} = V_{SS}$	25	40	55	k $\Omega$
$V_{F(NRST)}$	NRST input, filtered pulse	-	-	-	70	ns
$V_{NF(NRST)}$	NRST input, not filtered pulse	$1.8 \text{ V} \leq V_{DD} \leq 3.6 \text{ V}$	350	-	-	

1. Guaranteed by design.

2. The pull-up is designed with a true resistance in series with a switchable PMOS. This PMOS contribution to the series resistance is minimal (~10 %).

Figure 19. Recommended NRST pin protection



1. The reset network protects the device against parasitic resets.

2. The user must ensure that the level on the NRST pin can go below the  $V_{IL(NRST)}$  max level specified in the above table. Otherwise the reset is not taken into account by the device.

3. The external capacitor on NRST must be placed as close as possible to the device.

### 5.3.18 Analog switches booster

Table 71. Analog switches booster characteristics<sup>(1)</sup>

Symbol	Parameter	Min	Typ	Max	Unit
$V_{DD}$	Supply voltage	1.8	-	3.6	V
$t_{SU(BOOST)}$	Booster startup time	-	-	240	$\mu\text{s}$
$I_{DD(BOOST)}$	Booster consumption for $1.8 \text{ V} \leq V_{DD} \leq 2.0 \text{ V}$	-	-	250	$\mu\text{A}$
	Booster consumption for $2.0 \text{ V} \leq V_{DD} \leq 2.7 \text{ V}$	-	-	500	
	Booster consumption for $2.7 \text{ V} \leq V_{DD} \leq 3.6 \text{ V}$	-	-	900	

1. Guaranteed by design.

### 5.3.19 Analog-to-digital converter characteristics

Unless otherwise specified, the parameters given in the table below are preliminary values derived from tests performed under ambient temperature,  $f_{\text{PCLK}}$  frequency and  $V_{\text{DDA}}$  supply voltage conditions summarized in [Table 25: General operating conditions](#).

**Note:** *It is recommended to perform a calibration after each power-up.*

**Table 72. ADC characteristics<sup>(1)</sup>**

Symbol	Parameter	Conditions <sup>(2)</sup>	Min	Typ	Max	Unit
$V_{\text{DDA}}$	Analog supply voltage	-	1.62	-	3.6	V
$V_{\text{REF+}}$	Positive reference voltage	$V_{\text{DDA}} \geq 2 \text{ V}$	2	-	$V_{\text{DDA}}$	
		$V_{\text{DDA}} < 2 \text{ V}$	$V_{\text{DDA}}$			
$f_{\text{ADC}}$	ADC clock frequency	Range 1	0.14	-	35	MHz
		Range 2	0.14	-	16	
$f_{\text{s}}$	Sampling rate	12 bits, $V_{\text{DDA}} > 2 \text{ V}$	-	-	2.50	MSPS
		10 bits, $V_{\text{DDA}} > 2 \text{ V}$	-	-	2.92	
		8 bits, $V_{\text{DDA}} > 2 \text{ V}$	-	-	3.50	
		6 bits, $V_{\text{DDA}} > 2 \text{ V}$	-	-	4.38	
		12 bits, $V_{\text{DDA}} \leq 2 \text{ V}$	-	-	2.18	
		10 bits, $V_{\text{DDA}} \leq 2 \text{ V}$	-	-	2.50	
		8 bits, $V_{\text{DDA}} \leq 2 \text{ V}$	-	-	2.92	
		6 bits, $V_{\text{DDA}} \leq 2 \text{ V}$	-	-	3.50	
$f_{\text{TRIG}}$	External trigger frequency	$f_{\text{ADC}} = 35 \text{ MHz}$ , 12 bits, $V_{\text{DDA}} > 2 \text{ V}$	-	-	2.35	MHz
		$f_{\text{ADC}} = 35 \text{ MHz}$ , 12 bits, $V_{\text{DDA}} \leq 2 \text{ V}$	-	-	2.18	
		12 bits, $V_{\text{DDA}} > 2 \text{ V}$	-	-	$f_{\text{ADC}}/15$	
		12 bits, $V_{\text{DDA}} \leq 2 \text{ V}$	-	-	$f_{\text{ADC}}/16$	
$V_{\text{AIN}}$	Conversion voltage range	-	$V_{\text{SSA}}$	-	$V_{\text{REF+}}$	V
$R_{\text{AIN}}$	External input impedance	-	-	-	50	k $\Omega$
$C_{\text{ADC}}$	Internal sample and hold capacitor	-	-	5	-	pF
$t_{\text{STAB}}$	ADC power-up time	-	2			Conversion cycle
$t_{\text{CAL}}$	Calibration time	$f_{\text{ADC}} = 35 \text{ MHz}$	2.35			$\mu\text{s}$
		-	82			$1/f_{\text{ADC}}$

Table 72. ADC characteristics<sup>(1)</sup> (continued)

Symbol	Parameter	Conditions <sup>(2)</sup>	Min	Typ	Max	Unit
$t_{LATR}$	Trigger conversion latency; Regular and injected channels without conversion abort	CKMODE = 00	2	-	3	$1/f_{ADC}$
		CKMODE = 01	-	-	2.75	
		CKMODE = 10	-	-	2.63	
		CKMODE = 11	-	-	3	
$t_s$	Sampling time	$f_{ADC} = 35\text{ MHz}$	0.043	-	4.59	$\mu s$
		-	1.5	-	160.5	$1/f_{ADC}$
$t_{ADCVREG\_STUP}$	ADC voltage regulator start-up time	-	-	-	20	$\mu s$
$t_{CONV}$	Total conversion time (including sampling time)	$f_{ADC} = 35\text{ MHz}$ Resolution = 12 bits	0.40	-	4.95	$\mu s$
		Resolution = 12 bits	$t_s + 12.5$ cycles for successive approximation = 14 to 173			$1/f_{ADC}$
$I_{DDA(ADC)}$	ADC consumption from $V_{DDA}$	$f_s = 2.5\text{ Msps}$	-	272	-	$\mu A$
		$f_s = 1\text{ Msps}$	-	118	-	
		$f_s = 10\text{ ksp/s}$	-	15	-	
$I_{DDV(ADC)}$	ADC consumption from $V_{REF+}$ single ended mode	$f_s = 2.5\text{ Msps}$	-	65	-	$\mu A$
		$f_s = 1\text{ Msps}$	-	26	-	
		$f_s = 10\text{ ksp/s}$	-	0.26	-	

1. Guaranteed by design

2. I/O analog switch voltage booster must be enabled (BOOSTEN = 1 in the SYSCFG\_CFGR1) when  $V_{DDA} < 2.4\text{ V}$  and disabled when  $V_{DDA} \geq 2.4\text{ V}$ .Table 73. Maximum ADC  $R_{AIN}$  values

Resolution	Sampling cycle at 35 MHz (ns)	Sampling time at 35 MHz (ns)	Max. $R_{AIN}^{(1)(2)}(\Omega)$
12 bits	1.5 <sup>(3)</sup>	43	50
	3.5	100	680
	7.5	214	2200
	12.5	357	4700
	19.5	557	8200
	39.5	1129	15000
	79.5	2271	33000
	160.5	4586	50000

Table 73. Maximum ADC  $R_{AIN}$  values (continued)

Resolution	Sampling cycle at 35 MHz (ns)	Sampling time at 35 MHz (ns)	Max. $R_{AIN}^{(1)(2)}(\Omega)$
10 bits	1.5 <sup>(3)</sup>	43	68
	3.5	100	820
	7.5	214	3300
	12.5	357	5600
	19.5	557	10000
	39.5	1129	22000
	79.5	2271	39000
	160.5	4586	50000
8 bits	1.5 <sup>(3)</sup>	43	82
	3.5	100	1500
	7.5	214	3900
	12.5	357	6800
	19.5	557	12000
	39.5	1129	27000
	79.5	2271	50000
	160.5	4586	50000
6 bits	1.5 <sup>(3)</sup>	43	390
	3.5	100	2200
	7.5	214	5600
	12.5	357	10000
	19.5	557	15000
	39.5	1129	33000
	79.5	2271	50000
	160.5	4586	50000

1. Guaranteed by design.
2. I/O analog switch voltage booster must be enabled (BOOSTEN = 1 in the SYSCFG\_CFGR1) when  $V_{DDA} < 2.4$  V and disabled when  $V_{DDA} \geq 2.4$  V.
3. Only allowed with  $V_{DDA} > 2$  V

Table 74. ADC accuracy<sup>(1)(2)(3)</sup>

Symbol	Parameter	Conditions <sup>(4)</sup>	Min	Typ	Max	Unit
ET	Total unadjusted error	$V_{DDA} = V_{REF+} = 3\text{ V}$ , $f_{ADC} = 35\text{ MHz}$ , $f_s \leq 2.5\text{ Msps}$ , $T_A = 25\text{ }^{\circ}\text{C}$	-	3	4	LSB
		$2\text{ V} < V_{DDA} = V_{REF+} < 3.6\text{ V}$ , $f_{ADC} = 35\text{ MHz}$ , $f_s \leq 2.5\text{ Msps}$ , $T_A = \text{entire range}$	-	3	6.5	
		$1.65\text{ V} < V_{DDA} = V_{REF+} < 3.6\text{ V}$ , $T_A = \text{entire range}$ , Range 1: $f_{ADC} = 35\text{ MHz}$ , $f_s \leq 2.2\text{ Msps}$ Range 2: $f_{ADC} = 16\text{ MHz}$ , $f_s \leq 1.1\text{ Msps}$	-	3	7.5	
EO	Offset error	$V_{DDA} = V_{REF+} = 3\text{ V}$ , $f_{ADC} = 35\text{ MHz}$ , $f_s \leq 2.5\text{ Msps}$ , $T_A = 25\text{ }^{\circ}\text{C}$	-	1.5	2	LSB
		$2\text{ V} < V_{DDA} = V_{REF+} < 3.6\text{ V}$ , $f_{ADC} = 35\text{ MHz}$ , $f_s \leq 2.5\text{ Msps}$ , $T_A = \text{entire range}$	-	1.5	4.5	
		$1.65\text{ V} < V_{DDA} = V_{REF+} < 3.6\text{ V}$ , $T_A = \text{entire range}$ , Range 1: $f_{ADC} = 35\text{ MHz}$ , $f_s \leq 2.2\text{ Msps}$ Range 2: $f_{ADC} = 16\text{ MHz}$ , $f_s \leq 1.1\text{ Msps}$	-	1.5	5.5	
EG	Gain error	$V_{DDA} = V_{REF+} = 3\text{ V}$ , $f_{ADC} = 35\text{ MHz}$ , $f_s \leq 2.5\text{ Msps}$ , $T_A = 25\text{ }^{\circ}\text{C}$	-	3	3.5	LSB
		$2\text{ V} < V_{DDA} = V_{REF+} < 3.6\text{ V}$ , $f_{ADC} = 35\text{ MHz}$ , $f_s \leq 2.5\text{ Msps}$ , $T_A = \text{entire range}$	-	3	5	
		$1.65\text{ V} < V_{DDA} = V_{REF+} < 3.6\text{ V}$ , $T_A = \text{entire range}$ , Range 1: $f_{ADC} = 35\text{ MHz}$ , $f_s \leq 2.2\text{ Msps}$ Range 2: $f_{ADC} = 16\text{ MHz}$ , $f_s \leq 1.1\text{ Msps}$	-	3	6.5	
ED	Differential linearity error	$V_{DDA} = V_{REF+} = 3\text{ V}$ , $f_{ADC} = 35\text{ MHz}$ , $f_s \leq 2.5\text{ Msps}$ , $T_A = 25\text{ }^{\circ}\text{C}$	-	1.2	1.5	LSB
		$2\text{ V} < V_{DDA} = V_{REF+} < 3.6\text{ V}$ , $f_{ADC} = 35\text{ MHz}$ , $f_s \leq 2.5\text{ Msps}$ , $T_A = \text{entire range}$	-	1.2	1.5	
		$1.65\text{ V} < V_{DDA} = V_{REF+} < 3.6\text{ V}$ , $T_A = \text{entire range}$ , Range 1: $f_{ADC} = 35\text{ MHz}$ , $f_s \leq 2.2\text{ Msps}$ Range 2: $f_{ADC} = 16\text{ MHz}$ , $f_s \leq 1.1\text{ Msps}$	-	1.2	1.5	
EL	Integral linearity error	$V_{DDA} = V_{REF+} = 3\text{ V}$ , $f_{ADC} = 35\text{ MHz}$ , $f_s \leq 2.5\text{ Msps}$ , $T_A = 25\text{ }^{\circ}\text{C}$	-	2.5	3	LSB
		$2\text{ V} < V_{DDA} = V_{REF+} < 3.6\text{ V}$ , $f_{ADC} = 35\text{ MHz}$ , $f_s \leq 2.5\text{ Msps}$ , $T_A = \text{entire range}$	-	2.5	3	
		$1.65\text{ V} < V_{DDA} = V_{REF+} < 3.6\text{ V}$ , $T_A = \text{entire range}$ , Range 1: $f_{ADC} = 35\text{ MHz}$ , $f_s \leq 2.2\text{ Msps}$ Range 2: $f_{ADC} = 16\text{ MHz}$ , $f_s \leq 1.1\text{ Msps}$	-	2.5	3.5	
ENOB	Effective number of bits	$V_{DDA} = V_{REF+} = 3\text{ V}$ , $f_{ADC} = 35\text{ MHz}$ , $f_s \leq 2.5\text{ Msps}$ , $T_A = 25\text{ }^{\circ}\text{C}$	10.1	10.2	-	bit
		$2\text{ V} < V_{DDA} = V_{REF+} < 3.6\text{ V}$ , $f_{ADC} = 35\text{ MHz}$ , $f_s \leq 2.5\text{ Msps}$ , $T_A = \text{entire range}$	9.6	10.2	-	
		$1.65\text{ V} < V_{DDA} = V_{REF+} < 3.6\text{ V}$ , $T_A = \text{entire range}$ , Range 1: $f_{ADC} = 35\text{ MHz}$ , $f_s \leq 2.2\text{ Msps}$ Range 2: $f_{ADC} = 16\text{ MHz}$ , $f_s \leq 1.1\text{ Msps}$	9.5	10.2	-	

Table 74. ADC accuracy<sup>(1)(2)(3)</sup> (continued)

Symbol	Parameter	Conditions <sup>(4)</sup>	Min	Typ	Max	Unit
SINAD	Signal-to-noise and distortion ratio	$V_{DDA} = V_{REF+} = 3\text{ V}$ , $f_{ADC} = 35\text{ MHz}$ , $f_s \leq 2.5\text{ Msps}$ , $T_A = 25\text{ }^{\circ}\text{C}$	62.5	63	-	dB
		$2\text{ V} < V_{DDA} = V_{REF+} < 3.6\text{ V}$ , $f_{ADC} = 35\text{ MHz}$ , $f_s \leq 2.5\text{ Msps}$ , $T_A = \text{entire range}$	59.5	63	-	
		$1.65\text{ V} < V_{DDA} = V_{REF+} < 3.6\text{ V}$ , $T_A = \text{entire range}$ , Range 1: $f_{ADC} = 35\text{ MHz}$ , $f_s \leq 2.2\text{ Msps}$ Range 2: $f_{ADC} = 16\text{ MHz}$ , $f_s \leq 1.1\text{ Msps}$	59	63	-	
SNR	Signal-to-noise ratio	$V_{DDA} = V_{REF+} = 3\text{ V}$ , $f_{ADC} = 35\text{ MHz}$ , $f_s \leq 2.5\text{ Msps}$ , $T_A = 25\text{ }^{\circ}\text{C}$	63	64	-	dB
		$2\text{ V} < V_{DDA} = V_{REF+} < 3.6\text{ V}$ , $f_{ADC} = 35\text{ MHz}$ , $f_s \leq 2.5\text{ Msps}$ , $T_A = \text{entire range}$	60	64	-	
		$1.65\text{ V} < V_{DDA} = V_{REF+} < 3.6\text{ V}$ , $T_A = \text{entire range}$ , Range 1: $f_{ADC} = 35\text{ MHz}$ , $f_s \leq 2.2\text{ Msps}$ Range 2: $f_{ADC} = 16\text{ MHz}$ , $f_s \leq 1.1\text{ Msps}$	60	64	-	
THD	Total harmonic distortion	$V_{DDA} = V_{REF+} = 3\text{ V}$ , $f_{ADC} = 35\text{ MHz}$ , $f_s \leq 2.5\text{ Msps}$ , $T_A = 25\text{ }^{\circ}\text{C}$	-	-74	-73	dB
		$2\text{ V} < V_{DDA} = V_{REF+} < 3.6\text{ V}$ , $f_{ADC} = 35\text{ MHz}$ , $f_s \leq 2.5\text{ Msps}$ , $T_A = \text{entire range}$	-	-74	-70	
		$1.65\text{ V} < V_{DDA} = V_{REF+} < 3.6\text{ V}$ , $T_A = \text{entire range}$ , Range 1: $f_{ADC} = 35\text{ MHz}$ , $f_s \leq 2.2\text{ Msps}$ Range 2: $f_{ADC} = 16\text{ MHz}$ , $f_s \leq 1.1\text{ Msps}$	-	-74	-70	

1. Based on characterization results, not tested in production.
2. ADC DC accuracy values are measured after internal calibration.
3. Injecting negative current on any analog input pin significantly reduces the accuracy of A-to-D conversion of signal on another analog input. It is recommended to add a Schottky diode (pin to ground) to analog pins susceptible to receive negative current.
4. I/O analog switch voltage booster enabled (BOOSTEN = 1 in the SYSCFG\_CFGR1) when  $V_{DDA} < 2.4\text{ V}$  and disabled when  $V_{DDA} \geq 2.4\text{ V}$ .



Figure 20. ADC accuracy characteristics

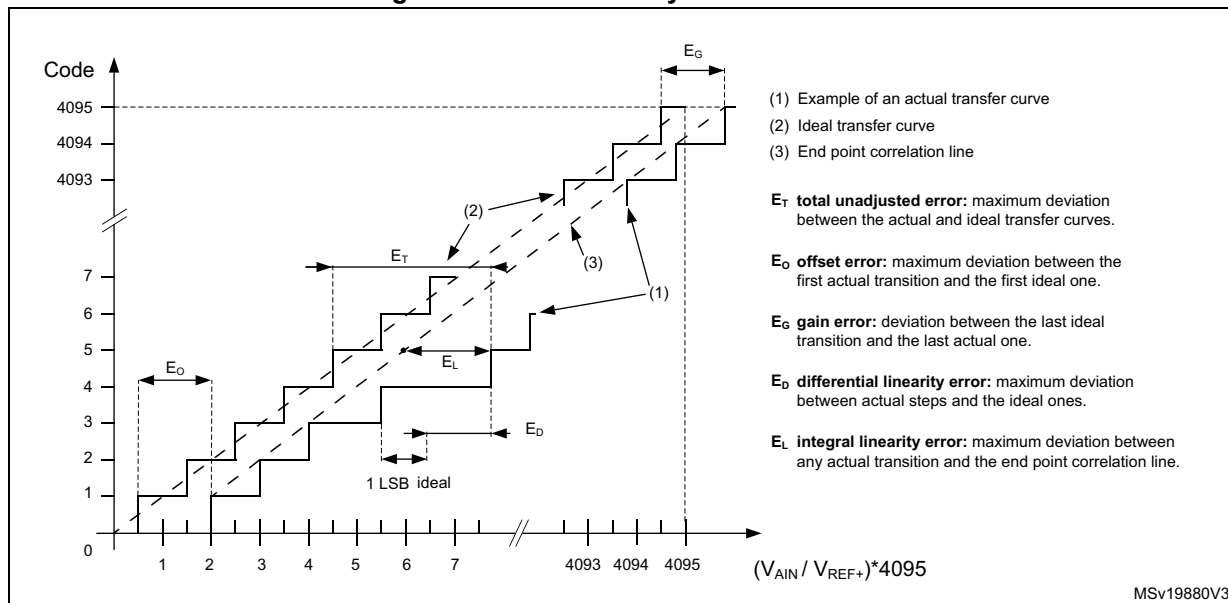
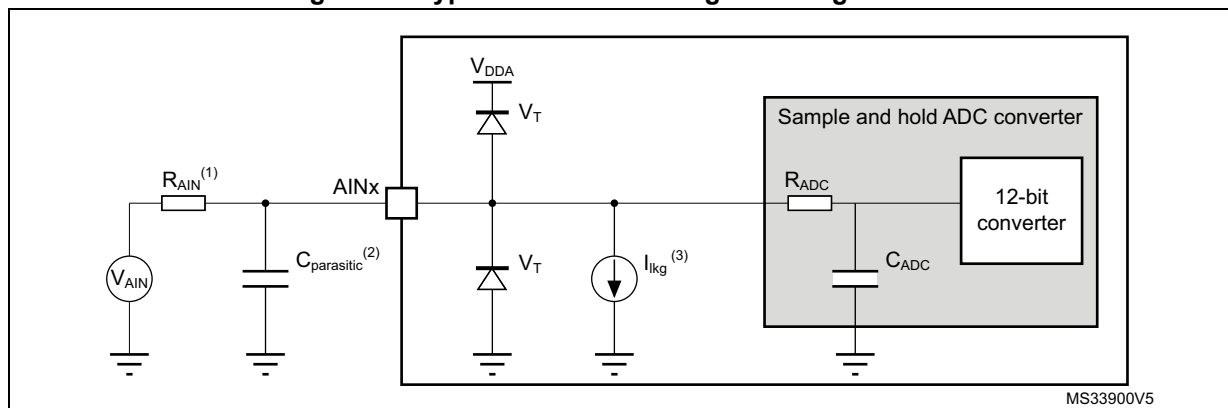


Figure 21. Typical connection diagram using the ADC



1. Refer to [Table 72: ADC characteristics](#) for the values of  $R_{AIN}$ ,  $R_{ADC}$  and  $C_{ADC}$ .
2.  $C_{parasitic}$  represents the capacitance of the PCB (dependent on soldering and PCB layout quality) plus the pad capacitance (refer to [Table 67: I/O static characteristics](#) for the value of the pad capacitance). A high  $C_{parasitic}$  value downgrades the conversion accuracy. To remedy this,  $f_{ADC}$  must be reduced.
3. Refer to [Table 67: I/O static characteristics](#) for the values of  $I_{lkg}$ .

### General PCB design guidelines

Power supply decoupling must be performed as shown in [Figure 11: Power supply scheme](#). The 100 nF capacitor must be ceramic (good quality) and must be placed as close as possible to the chip.

### 5.3.20 Temperature sensor characteristics

Table 75. TS characteristics

Symbol	Parameter	Min	Typ	Max	Unit
$T_L^{(1)}$	$V_{TS}$ linearity with temperature	-	$\pm 1$	$\pm 2$	$^{\circ}\text{C}$
Avg_Slope <sup>(2)</sup>	Average slope	2.3	2.5	2.7	mV / $^{\circ}\text{C}$
$V_{30}$	Voltage at 30 $^{\circ}\text{C}$ ( $\pm 5$ $^{\circ}\text{C}$ ) <sup>(3)</sup>	0.742	0.76	0.785	V
$t_{\text{START}}^{(1)}$ (TS_BUF) <sup>(1)</sup>	Sensor buffer startup time in continuous mode <sup>(4)</sup>	-	8	15	$\mu\text{s}$
$t_{\text{START}}^{(1)}$	Startup time when entering in continuous mode <sup>(4)</sup>	-	70	120	$\mu\text{s}$
$t_{S\_temp}^{(1)}$	ADC sampling time when reading the temperature	5	-	-	$\mu\text{s}$
$I_{DD}(\text{TS})^{(1)}$	Temperature sensor consumption from $V_{DD}$ , when selected by the ADC	-	4.7	7	$\mu\text{A}$

1. Guaranteed by design.
2. Guaranteed by characterization results.
3. Measured at  $V_{DDA} = 3.3 \text{ V} \pm 10 \text{ mV}$ . The  $V_{30}$  ADC conversion result is stored in the TS\_CAL1 byte. Refer to [Table 12: Temperature sensor calibration values](#).
4. Continuous mode means Run and Sleep modes, or temperature sensor enable in LPRun and LPSleep modes.

### 5.3.21 $V_{BAT}$ monitoring characteristics

Table 76.  $V_{BAT}$  monitoring characteristics<sup>(1)</sup>

Symbol	Parameter	Min	Typ	Max	Unit
R	Resistor bridge for $V_{BAT}$	-	39	-	k $\Omega$
Q	Ratio on $V_{BAT}$ measurement	-	3	-	-
$E_r^{(2)}$	Error on Q	-10	-	10	%
$t_{S\_vbat}^{(2)}$	ADC sampling time when reading $V_{BAT}$	12	-	-	$\mu\text{s}$

1.  $1.55 \text{ V} < V_{BAT} < 3.6 \text{ V}$ .
2. Guaranteed by design.

Table 77.  $V_{BAT}$  charging characteristics

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$R_{BC}$	Battery charging resistor	VBRS = 0	-	5	-	k $\Omega$
		VBRS = 1	-	1.5	-	

## 5.3.22 Voltage reference buffer characteristics

Table 78. VREFBUF characteristics<sup>(1)</sup>

Symbol	Parameter	Conditions		Min	Typ	Max	Unit
$V_{DDA}$	Analog supply voltage	Normal mode	$V_{RS} = 0$	2.4	-	3.6	V
			$V_{RS} = 1$	2.8	-	3.6	
		Degraded mode <sup>(2)</sup>	$V_{RS} = 0$	1.65	-	2.4	
			$V_{RS} = 1$	1.65	-	2.8	
$V_{REFBUF\_OUT}$	Voltage reference output	Normal mode	$V_{RS} = 0$	2.046 <sup>(3)</sup>	2.048	2.049 <sup>(3)</sup>	
			$V_{RS} = 1$	2.498 <sup>(3)</sup>	2.5	2.502 <sup>(3)</sup>	
		Degraded mode <sup>(2)</sup>	$V_{RS} = 0$	$V_{DDA} - 150 \text{ mV}$	-	$V_{DDA}$	
			$V_{RS} = 1$	$V_{DDA} - 150 \text{ mV}$	-	$V_{DDA}$	
TRIM	Trim step resolution	-	-	-	$\pm 0.05$	$\pm 0.1$	%
CL	Load capacitor	-	-	0.5	1	1.5	$\mu\text{F}$
esr	Equivalent series resistor of $C_{load}$	-	-	-	-	2	$\Omega$
$I_{load}$	Static load current	-	-	-	-	4	mA
$I_{line\_reg}$	Line regulation	$2.8 \text{ V} \leq V_{DDA} \leq 3.6 \text{ V}$	$I_{load} = 500 \mu\text{A}$	-	200	1000	ppm/V
			$I_{load} = 4 \text{ mA}$	-	100	500	
$I_{load\_reg}$	Load regulation	$500 \mu\text{A} \leq I_{load} \leq 4 \text{ mA}$	Normal mode	-	50	500	ppm/mA
$T_{coeff}$	Temperature coefficient	$-40^\circ\text{C} < T_J < +105^\circ\text{C}$		-	-	$T_{coeff\_vrefint} + 50$	ppm/ $^\circ\text{C}$
		$0^\circ\text{C} < T_J < +50^\circ\text{C}$		-	-	$T_{coeff\_vrefint} + 50$	
PSRR	Power supply rejection	DC		40	60	-	dB
		100 kHz		25	40	-	
$t_{START}$	Startup time	$CL = 0.5 \mu\text{F}^{(4)}$		-	300	350	$\mu\text{s}$
		$CL = 1.1 \mu\text{F}^{(4)}$		-	500	650	
		$CL = 1.5 \mu\text{F}^{(4)}$		-	650	800	
$I_{INRUSH}$	Control of maximum DC current drive on VREFBUF_OUT during start-up phase <sup>(5)</sup>	-	-	-	8	-	mA

Table 78. VREFBUF characteristics<sup>(1)</sup> (continued)

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$I_{DDA}$ (VREFBUF)	VREFBUF consumption from $V_{DDA}$	$I_{load} = 0 \mu A$	-	16	25	$\mu A$
		$I_{load} = 500 \mu A$	-	18	30	
		$I_{load} = 4 mA$	-	35	50	

1. Guaranteed by design, unless otherwise specified.
2. In degraded mode, VREFBUF cannot maintain accurately the output voltage that follows ( $V_{DDA}$  - drop voltage).
3. Guaranteed by test in production.
4. The capacitive load must include a 100 nF capacitor in order to cut-off the high-frequency noise.
5. To correctly control the VREFBUF in-rush current during start-up phase and scaling change, the  $V_{DDA}$  voltage must be in the range [2.4 V to 3.6 V] and [2.8 V to 3.6 V] respectively for  $V_{RS} = 0$  and  $V_{RS} = 1$ .

### 5.3.23 Digital-to-analog converter characteristics

Table 79. DAC characteristics<sup>(1)</sup>

Symbol	Parameter	Conditions		Min	Typ	Max	Unit
$V_{DDA}$	Analog supply voltage for DAC ON	DAC output buffer OFF, DAC_OUT pin not connected (internal connection only)		1.71	-	3.6	V
		Other modes		1.80	-		
$V_{REF+}$	Positive reference voltage	DAC output buffer OFF, DAC_OUT pin not connected (internal connection only)		1.71	-	$V_{DDA}$	V
		Other modes		1.80	-		
$R_L$	Resistive load	DAC output buffer ON	Connected to $V_{SSA}$	5	-	-	k $\Omega$
			Connected to $V_{DDA}$	25	-	-	
$R_O$	Output impedance	DAC output buffer OFF		9.6	11.7	13.8	
$R_{BON}$	Output impedance sample-and-hold mode, output buffer ON	$V_{DD} = 2.7 V$		-	-	2	
		$V_{DD} = 2.0 V$		-	-	3.5	
$R_{BOFF}$	Output impedance sample and hold mode, output buffer OFF	$V_{DD} = 2.7 V$		-	-	16.5	
		$V_{DD} = 2.0 V$		-	-	18.0	
$C_L$	Capacitive load	DAC output buffer ON		-	-	50	pF
$C_{SH}$		Sample-and-hold mode		-	0.1	1	$\mu F$
$V_{DAC\_OUT}$	Voltage on DAC_OUT output	DAC output buffer ON		0.2	-	$V_{REF+} - 0.2$	V
		DAC output buffer OFF		0	-	$V_{REF+}$	

Table 79. DAC characteristics<sup>(1)</sup> (continued)

Symbol	Parameter	Conditions		Min	Typ	Max	Unit
$t_{\text{SETTLING}}$	Settling time (full scale: for a 12-bit code transition between the lowest and the highest input codes, when DAC_OUT reaches final value $\pm 0.5$ LSB, $\pm 1$ LSB, $\pm 2$ LSB, $\pm 4$ LSB, $\pm 8$ LSB)	Normal mode DAC output buffer ON CL $\leq$ 50 pF RL $\geq$ 5 k $\Omega$	$\pm 0.5$ LSB	-	1.7	3	$\mu\text{s}$
			$\pm 1$ LSB	-	1.6	2.9	
			$\pm 2$ LSB	-	1.55	2.85	
			$\pm 4$ LSB	-	1.48	2.8	
			$\pm 8$ LSB	-	1.4	2.75	
		Normal mode DAC output buffer OFF, $\pm 1$ LSB, CL = 10 pF		-	2	2.5	
$t_{\text{WAKEUP}}^{(2)}$	Wakeup time from off state (setting the ENx bit in the DAC control register) until final value $\pm 1$ LSB	Normal mode DAC output buffer ON CL $\leq$ 50 pF, RL $\geq$ 5 k $\Omega$		-	4.2	7.5	$\mu\text{s}$
		Normal mode DAC output buffer OFF, CL $\leq$ 10 pF		-	2	5	
PSRR	$V_{\text{DDA}}$ supply rejection ratio	Normal mode DAC output buffer ON CL $\leq$ 50 pF, RL $\geq$ 5 k $\Omega$ , DC		-	-80	-28	dB
$T_{\text{W\_to\_W}}$	Minimum time between two consecutive writes into the DAC_DORx register to guarantee a correct DAC_OUT for a small variation of the input code (1 LSB)	DAC_MCR:MODEx[2:0] = 000 or 001 CL $\leq$ 50 pF, RL $\geq$ 5 k $\Omega$		1	-	-	$\mu\text{s}$
		DAC_MCR:MODEx[2:0] = 010 or 011 CL $\leq$ 10 pF		1.4	-	-	
$t_{\text{SAMP}}$	Sampling time in sample and hold mode (code transition between the lowest input code and the highest input code when DACOUT reaches final value $\pm 1$ LSB)	DAC_OUT pin connected	DAC output buffer ON, $C_{\text{SH}}$ = 100 nF	-	0.7	3.5	ms
			DAC output buffer OFF, $C_{\text{SH}}$ = 100 nF	-	10.5	18	
		DAC_OUT pin not connected (internal connection only)	DAC output buffer OFF	-	2	3.5	$\mu\text{s}$
$I_{\text{leak}}$	Output leakage current	Sample and hold mode, DAC_OUT pin connected		-	-	_(3)	nA
$C_{\text{Iint}}$	Internal sample and hold capacitor	-		5.2	7	8.8	pF
$t_{\text{TRIM}}$	Middle code offset trim time	DAC output buffer ON		50	-	-	$\mu\text{s}$
$V_{\text{offset}}$	Middle code offset for 1 trim code step	$V_{\text{REF+}}$ = 3.6 V		-	1500	-	$\mu\text{V}$
		$V_{\text{REF+}}$ = 1.8 V		-	750	-	

Table 79. DAC characteristics<sup>(1)</sup> (continued)

Symbol	Parameter	Conditions		Min	Typ	Max	Unit
$I_{DDA(DAC)}$	DAC consumption from $V_{DDA}$	DAC output buffer ON	No load, middle code (0x800)	-	315	500	$\mu A$
			No load, worst code (0xF1C)	-	450	670	
		DAC output buffer OFF	No load, middle code (0x800)	-	-	0.2	
		Sample and hold mode, $C_{SH} = 100 \text{ nF}$		-	$315 \times \frac{T_{on}}{T_{on} + T_{off}}^{(4)}$	$670 \times \frac{T_{on}}{T_{on} + T_{off}}^{(4)}$	
$I_{DDV(DAC)}$	DAC consumption from $V_{REF+}$	DAC output buffer ON	No load, middle code (0x800)	-	185	240	$\mu A$
			No load, worst code (0xF1C)	-	340	400	
		DAC output buffer OFF	No load, middle code (0x800)	-	155	205	
		Sample and hold mode, buffer ON, $C_{SH} = 100 \text{ nF}$ , worst case		-	$185 \times \frac{T_{on}}{T_{on} + T_{off}}^{(4)}$	$400 \times \frac{T_{on}}{T_{on} + T_{off}}^{(4)}$	
		Sample and hold mode, buffer OFF, $C_{SH} = 100 \text{ nF}$ , worst case		-	$155 \times \frac{T_{on}}{T_{on} + T_{off}}^{(4)}$	$205 \times \frac{T_{on}}{T_{on} + T_{off}}^{(4)}$	

1. Guaranteed by design.
2. In buffered mode, the output can overshoot above the final value for low input code (starting from min value).
3. Refer to [Table 67: I/O static characteristics](#).
4.  $T_{on}$  is the Refresh phase duration.  $T_{off}$  is the Hold phase duration. Refer to the reference manual for more details.

Figure 22. 12-bit buffered/non-buffered DAC

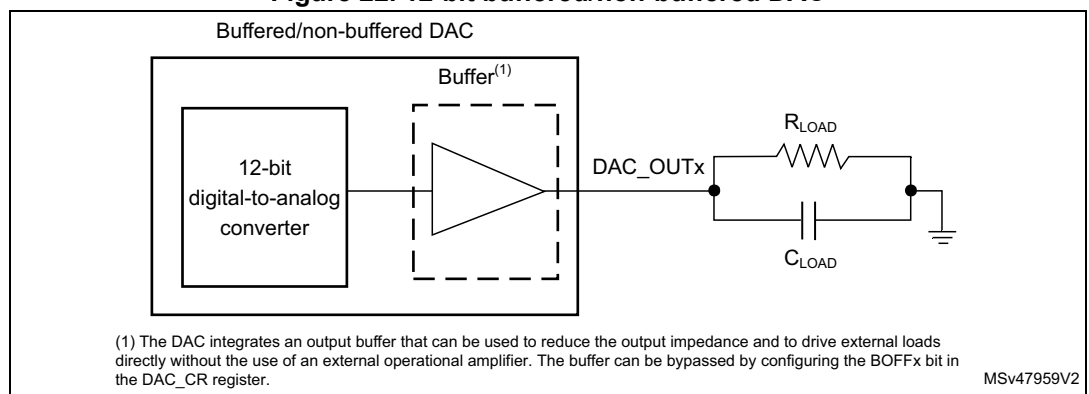


Table 80. DAC accuracy<sup>(1)</sup>

Symbol	Parameter	Conditions		Min	Typ	Max	Unit
DNL	Differential non linearity <sup>(2)</sup>	DAC output buffer ON		-	-	±2	LSB
		DAC output buffer OFF		-	-	±2	
-	Monotonicity	10 bits		Guaranteed			
INL	Integral non linearity <sup>(3)</sup>	DAC output buffer ON, CL ≤ 50 pF, RL ≥ 5 kΩ		-	-	±4	
		DAC output buffer OFF, CL ≤ 50 pF, no RL		-	-	±4	
Offset	Offset error at code 0x800 <sup>(3)</sup>	DAC output buffer ON CL ≤ 50 pF, RL ≥ 5 kΩ	V <sub>REF+</sub> = 3.6 V	-	-	±12	
			V <sub>REF+</sub> = 1.8 V	-	-	±25	
		DAC output buffer OFF, CL ≤ 50 pF, no RL		-	-	±8	
Offset1	Offset error at code 0x001 <sup>(4)</sup>	DAC output buffer OFF, CL ≤ 50 pF, no RL		-	-	±5	
OffsetCal	Offset Error at code 0x800 after calibration	DAC output buffer ON CL ≤ 50 pF, RL ≥ 5 kΩ	V <sub>REF+</sub> = 3.6 V	-	-	±5	
			V <sub>REF+</sub> = 1.8 V	-	-	±7	
Gain	Gain error <sup>(5)</sup>	DAC output buffer ON, CL ≤ 50 pF, RL ≥ 5 kΩ		-	-	±0.5	%
		DAC output buffer OFF, CL ≤ 50 pF, no RL		-	-	±0.5	
TUE	Total unadjusted error	DAC output buffer ON, CL ≤ 50 pF, RL ≥ 5 kΩ		-	-	±30	LSB
		DAC output buffer OFF, CL ≤ 50 pF no RL		-	-	±12	
TUECal	Total unadjusted error after calibration	DAC output buffer ON, CL ≤ 50 pF, RL ≥ 5 kΩ		-	-	±23	LSB
SNR	Signal-to-noise ratio	DAC output buffer ON, CL ≤ 50 pF, RL ≥ 5 kΩ, 1 kHz, BW 500 kHz		-	71.2	-	dB
		DAC output buffer OFF, CL ≤ 50 pF, no RL, 1 kHz, BW 500 kHz		-	71.6	-	
THD	Total harmonic distortion	DAC output buffer ON, CL ≤ 50 pF, RL ≥ 5 kΩ, 1 kHz		-	-78	-	dB
		DAC output buffer OFF, CL ≤ 50 pF, no RL, 1 kHz		-	-79	-	
SINAD	Signal-to-noise and distortion ratio	DAC output buffer ON, CL ≤ 50 pF, RL ≥ 5 kΩ, 1 kHz		-	70.4	-	dB
		DAC output buffer OFF, CL ≤ 50 pF, no RL, 1 kHz		-	71	-	

Table 80. DAC accuracy<sup>(1)</sup> (continued)

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
ENOB	Effective number of bits	DAC output buffer ON, $CL \leq 50$ pF, $RL \geq 5$ k $\Omega$ , 1 kHz	-	11.4	-	bits
		DAC output buffer OFF, $CL \leq 50$ pF, no RL, 1 kHz	-	11.5	-	

1. Guaranteed by design.
2. Difference between two consecutive codes - 1 LSB.
3. Difference between measured value at code  $i$  and the value at code  $i$  on a line drawn between code 0 and last code 4095.
4. Difference between the value measured at code (0x001) and the ideal value.
5. Difference between the ideal slope of the transfer function and the measured slope computed from code 0x000 and 0xFFFF when buffer is OFF, and from code giving 0.2 V and ( $V_{REF+} - 0.2$ ) V when buffer is ON.

### 5.3.24 Comparator characteristics

Table 81. COMP characteristics<sup>(1)</sup>

Symbol	Parameter	Conditions		Min	Typ	Max	Unit
V <sub>DDA</sub>	Analog supply voltage	-		1.62	-	3.6	V
V <sub>IN</sub>	Comparator input voltage range	-		0	-	V <sub>DDA</sub>	
V <sub>BG</sub> <sup>(2)</sup>	Scaler input voltage	-		V <sub>REFINT</sub>			
V <sub>SC</sub>	Scaler offset voltage	-		-	±5	±10	mV
I <sub>DDA</sub> (SCALER)	Scaler static consumption from V <sub>DDA</sub>	BRG_EN = 0 (bridge disabled)		-	200	300	nA
		BRG_EN = 1 (bridge enabled)		-	0.8	1	µA
t <sub>START_SCALER</sub>	Scaler startup time	-		-	100	200	µs
t <sub>START</sub>	Comparator startup time to reach propagation delay specification	High-speed mode	V <sub>DDA</sub> ≥ 2.7 V	-	-	5	µs
			V <sub>DDA</sub> < 2.7 V	-	-	7	
		Medium mode	V <sub>DDA</sub> ≥ 2.7 V	-	-	15	
			V <sub>DDA</sub> < 2.7 V	-	-	25	
		Ultra-low-power mode		-	-	40	
t <sub>D</sub> <sup>(3)</sup>	Propagation delay with 100 mV overdrive	High-speed mode	V <sub>DDA</sub> ≥ 2.7 V	-	55	80	ns
			V <sub>DDA</sub> < 2.7 V	-	55	100	
		Medium mode		-	0.55	0.9	µs
		Ultra-low-power mode		-	4	7	
V <sub>offset</sub>	Comparator offset error	Full common mode range		-	±5	±20	mV
V <sub>hys</sub>	Comparator hysteresis	No hysteresis		-	0	-	mV
		Low hysteresis		-	8	-	
		Medium hysteresis		-	15	-	
		High hysteresis		-	27	-	



Table 81. COMP characteristics<sup>(1)</sup> (continued)

Symbol	Parameter	Conditions		Min	Typ	Max	Unit
I <sub>DDA</sub> (COMP)	Comparator consumption from V <sub>DDA</sub>	Ultra-low-power mode	Static	-	400	600	nA
			With 50 kHz ±100 mV overdrive square signal	-	1200	-	
		Medium mode	Static	-	5	7	μA
			With 50 kHz ±100 mV overdrive square signal	-	6	-	
		High-speed mode	Static	-	70	100	
			With 50 kHz ±100 mV overdrive square signal	-	75	-	

1. Guaranteed by design, unless otherwise specified.
2. Refer to [Table 34: Embedded internal voltage reference](#).
3. Guaranteed by characterization results.

### 5.3.25 Timers characteristics

Parameters given in the following tables are guaranteed by design. Refer to [Section 5.3.16: I/O port characteristics](#) for details on the input/output alternate function characteristics (output compare, input capture, external clock, PWM output).

Table 82. TIMx<sup>(1)</sup> characteristics

Symbol	Parameter	Conditions	Min	Max	Unit
t <sub>res</sub> (TIM)	Timer resolution time	-	1	-	t <sub>TIMxCLK</sub>
		f <sub>TIMxCLK</sub> = 48 MHz	15.625	-	ns
f <sub>EXT</sub>	Timer external clock frequency on CH1 to CH4	-	0	f <sub>TIMxCLK</sub> /2	MHz
		f <sub>TIMxCLK</sub> = 48 MHz	0	40	
Res <sub>TIM</sub>	Timer resolution	TIM1, TIM16, TIM17	-	16	bit
		TIM2	-	32	
t <sub>COUNTER</sub>	16-bit counter clock period	-	1	65536	t <sub>TIMxCLK</sub>
		f <sub>TIMxCLK</sub> = 48 MHz	0.015625	1024	μs
t <sub>MAX_COUNT</sub>	Maximum possible count with 32-bit counter	-	-	65536 × 65536	t <sub>TIMxCLK</sub>
		f <sub>TIMxCLK</sub> = 48 MHz	-	67.10	s

1. TIMx is used as a general term where x stands for 1, 2, 16 or 17.

Table 83. IWDG min/max timeout period at 32 kHz (LSI)<sup>(1)</sup>

Prescaler divider	PR[2:0] bits	Min timeout (RL[11:0] = 0x000)	Max timeout (RL[11:0] = 0xFFFF)	Unit
/4	0x0	0.125	512	ms
/8	0x1	0.250	1024	
/16	0x2	0.500	2048	
/32	0x3	1.0	4096	
/64	0x4	2.0	8192	
/128	0x5	4.0	16384	
/256	0x6 or 0x7	8.0	32768	

1. The exact timings still depend on the phasing of the APB interface clock versus the LSI clock, hence there is always a full RC period of uncertainty.

### 5.3.26 Communication interfaces characteristics

#### I<sup>2</sup>C interface characteristics

The I<sup>2</sup>C interface meets the timings requirements of the I<sup>2</sup>C-bus specification and user manual rev. 03 for:

- Standard-mode (Sm): bitrate up to 100 Kbit/s
- Fast-mode (Fm): bitrate up to 400 Kbit/s
- Fast-mode Plus (Fm+): bitrate up to 1 Mbit/s

The I<sup>2</sup>C timings requirements are guaranteed by design when the I2C peripheral is properly configured (refer to the reference manual) and when the I2CCLK frequency is greater than the minimum shown in the table below.

Table 84. Minimum I2CCLK frequency in all I2C modes

Symbol	Parameter	Conditions		Min	Unit
f <sub>(I2CCLK)</sub>	I2CCLK frequency	Standard-mode	-	2	MHz
		Fast-mode	Analog filter ON, DNF = 0	8	
			Analog filter OFF, DNF = 1	9	
		Fast-mode Plus	Analog filter ON, DNF = 0	18	
			Analog filter OFF, DNF = 1	16	

The SDA and SCL I/O requirements are met with the following restrictions:

- The SDA and SCL I/O pins are not “true” open-drain. When configured as open-drain, the PMOS connected between the I/O pin and V<sub>DDIOx</sub> is disabled, but is still present.
- The 20 mA output drive requirement in Fast-mode Plus is partially supported. This limits the maximum load C<sub>load</sub> supported in Fast-mode Plus, given by these formulas:
  - $t_r(\text{SDA/SCL}) = 0.8473 \times R_p \times C_{\text{load}}$
  - $R_p(\text{min}) = [V_{\text{DD}} - V_{\text{OL}}(\text{max})] / I_{\text{OL}}(\text{max})$
 where R<sub>p</sub> is the I2C lines pull-up. Refer to [Section 5.3.16: I/O port characteristics](#) for more details.

All I2C SDA and SCL I/Os embed an analog filter (refer to the table below for its characteristics).

**Table 85. I2C analog filter characteristics<sup>(1)</sup>**

Symbol	Parameter	Min	Max	Unit
$t_{AF}$	Maximum pulse width of spikes that are suppressed by the analog filter	50 <sup>(2)</sup>	100 <sup>(3)</sup>	ns

1. Guaranteed by characterization.
2. Spikes with widths below  $t_{AF(min)}$  filtered.
3. Spikes with widths above  $t_{AF(max)}$  not filtered.

## USART characteristics

Unless otherwise specified, the parameters given in the table below are derived from tests performed under the ambient temperature,  $f_{PCLKx}$  frequency and supply voltage conditions summarized in [Table 25: General operating conditions](#), with the following configuration:

- OSPEEDRy[1:0] set to 10 (output speed)
- capacitive load C = 30 pF
- measurement points at CMOS levels:  $0.5 \times V_{DD}$

Refer to [Section 5.3.16: I/O port characteristics](#) for more details on the input/output alternate function characteristics (NSS, CK, TX, and RX for USART).

**Table 86. USART characteristics**

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
f <sub>CK</sub>	USART clock frequency	Master mode	-	-	6	MHz
		Slave mode	-	-	16	
t <sub>su(NSS)</sub>	NSS setup time	Slave mode	t <sub>ker</sub> + 5	-	-	ns
t <sub>h(NSS)</sub>	NSS hold time	Slave mode	2	-	-	
t <sub>w(CKH)</sub>	CK high time	Master mode	1 / f <sub>CK</sub> / 2 - 1	1 / f <sub>CK</sub> / 2	1 / f <sub>CK</sub> / 2 + 1	
t <sub>w(CKL)</sub>	CK low time					
t <sub>su(RX)</sub>	Data input setup time	Master mode	22	-	-	
		Slave mode	3	-	-	
t <sub>h(RX)</sub>	Data input hold time	Master mode	0	-	-	
		Slave mode	1	-	-	
t <sub>v(TX)</sub>	Data output valid time	Master mode	-	13	22	
		Slave mode	-	0.5	1	
t <sub>h(TX)</sub>	Data output hold time	Master mode	10	-	-	
		Slave mode	0	-	-	

**SPI characteristics**

Unless otherwise specified, the parameters given in the table below are derived from tests performed under the ambient temperature,  $f_{PCLKx}$  frequency and supply voltage conditions summarized in [Table 25: General operating conditions](#), with the following configuration:

- output speed set to OSPEEDRy[1:0] = 11
- capacitive load C = 30 pF
- measurements done at CMOS levels:  $0.5 \times V_{DD}$

Refer to [Section 5.3.16: I/O port characteristics](#) for more details on the input/output alternate function characteristics (NSS, SCK, MOSI, MISO for SPI).

**Table 87. SPI characteristics<sup>(1)</sup>**

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$f_{SCK}$ $1/t_{c(SCK)}$	SPI clock frequency	Master mode $1.65 < V_{DD} < 3.6$ V, Range 1	-	-	24	MHz
		Master transmitter mode $1.65 < V_{DD} < 3.6$ V, Range 1			24	
		Slave receiver mode $1.65 < V_{DD} < 3.6$ V, Range 1			24	
		Slave mode transmitter/full duplex $2.7 < V_{DD} < 3.6$ V, Range 1			24 <sup>(2)</sup>	
		Slave mode transmitter/full duplex $1.65 < V_{DD} < 3.6$ V, Range 1			24 <sup>(2)</sup>	
$t_{su(NSS)}$	NSS setup time	Slave mode, SPI prescaler = 2	$3 \times T_{PCLK}$	-	-	-
$t_{h(NSS)}$	NSS hold time	Slave mode, SPI prescaler = 2	$2 \times T_{PCLK}$	-	-	
$t_{w(SCKH)}$ $t_{w(SCKL)}$	SCK high and low time	Master mode	$T_{PCLK} - 1$	$T_{PCLK}$	$T_{PCLK} + 1$	
$t_{su(MI)}$	Data input setup time	Master mode	1	-	-	ns
$t_{su(SI)}$		Slave mode	1	-	-	
$t_{h(MI)}$	Data input hold time	Master mode	6	-	-	
$t_{h(SI)}$		Slave mode	2	-	-	
$t_{a(SO)}$	Data output access time	Slave mode	9	12	34	
$t_{dis(SO)}$	Data output disable time		9	10	16	

Table 87. SPI characteristics<sup>(1)</sup> (continued)

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$t_{v(SO)}$	Data output valid time	Slave mode, $2.7 < V_{DD} < 3.6$ V Range 1	-	10	13.5	ns
		Slave mode, $2.7 < V_{DD} < 3.6$ V Range 2	-	17	18	
		Slave mode, $1.8 < V_{DD} < 3.6$ V Range 1	-	10	20	
		Slave mode, $1.8 < V_{DD} < 3.6$ V Range 2	-	17	24	
$t_{v(MO)}$		Master mode (after enable edge)	-	1	1.5	
$t_{h(SO)}$	Data output hold time	Slave mode (after enable edge)	8	-	-	
$t_{h(MO)}$		Master mode (after enable edge)	0	-	-	

1. Guaranteed by characterization results.
2. Maximum frequency in Slave transmitter mode is determined by the sum of  $t_{v(SO)}$  and  $t_{su(MI)}$ , that has to fit into SCK low or high phase preceding the SCK sampling edge. This value can be achieved when the SPI communicates with a master having  $t_{su(MI)} = 0$  while Duty(SCK) = 50 %.

Figure 23. SPI timing diagram - Slave mode and CPHA = 0

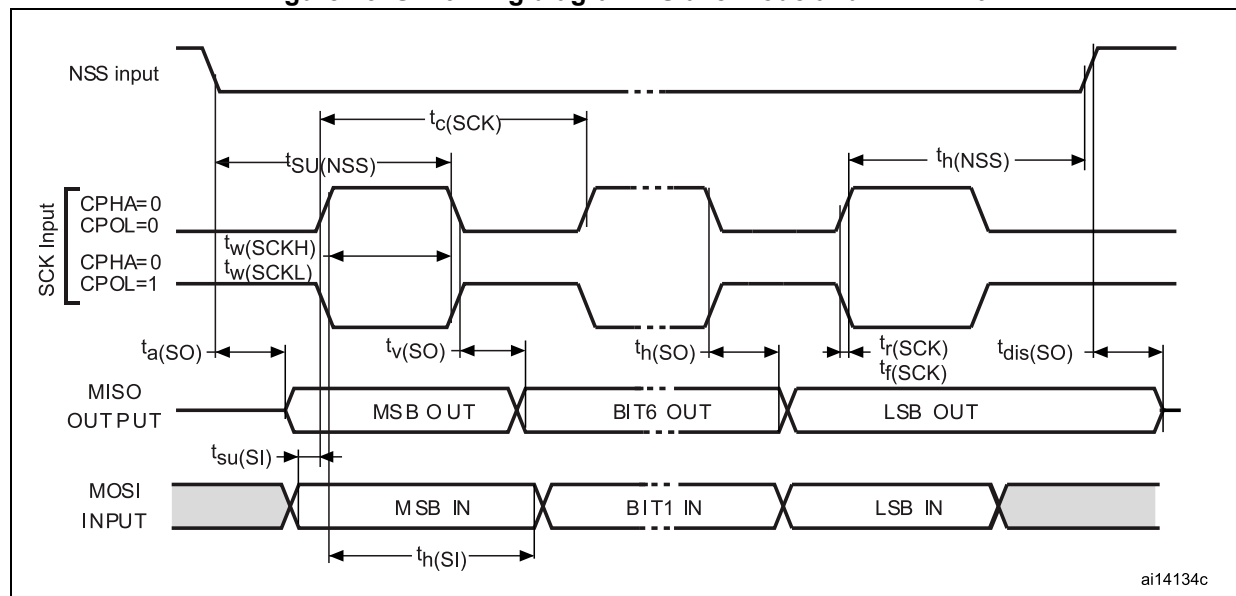
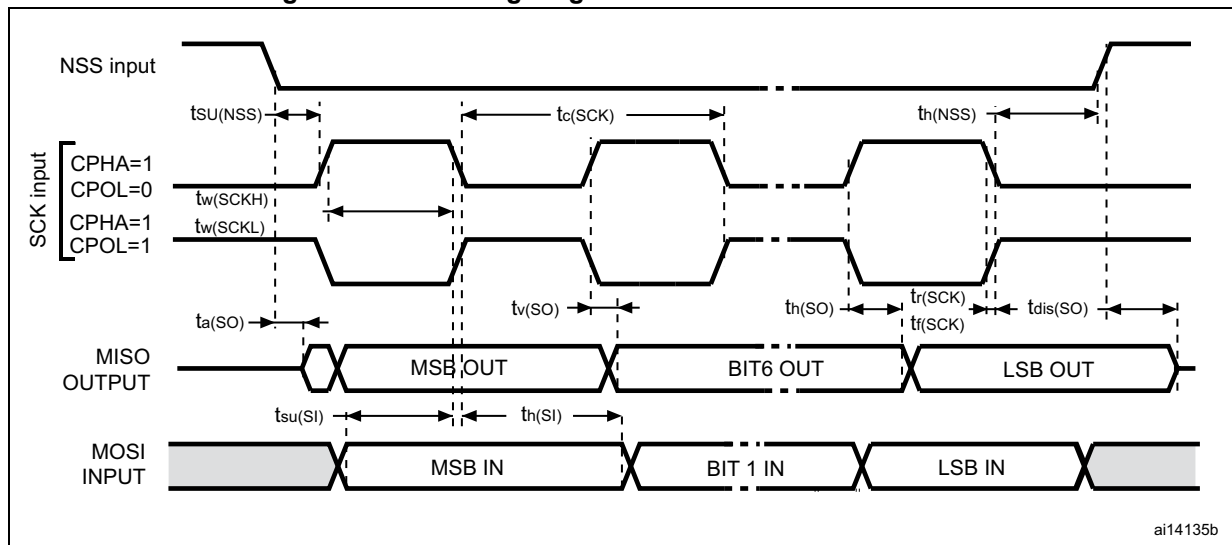
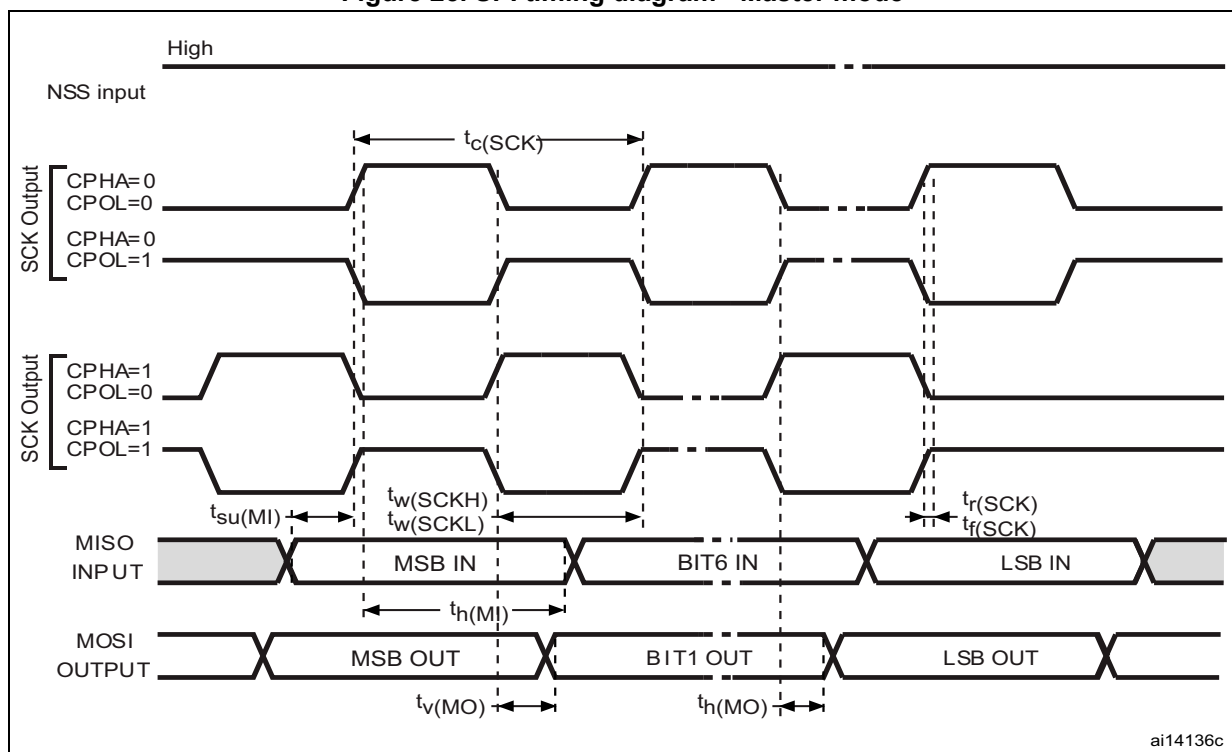


Figure 24. SPI timing diagram - Slave mode and CPHA = 1



1. Measurement points are set at CMOS levels:  $0.3 V_{DD}$  and  $0.7 V_{DD}$ .

Figure 25. SPI timing diagram - Master mode



1. Measurement points are set at CMOS levels:  $0.3 V_{DD}$  and  $0.7 V_{DD}$ .

**JTAG/SWD characteristics**

Unless otherwise specified, the parameters given in the table below are derived from tests performed under the ambient temperature,  $f_{PCLKx}$  frequency and supply voltage conditions summarized in [Table 25: General operating conditions](#), with the following configuration:

- capacitive load  $C = 30$  pF
- measurement done at CMOS levels:  $0.5 \times V_{DD}$ .

Refer to [Section 5.3.16: I/O port characteristics](#) for more details.

**Table 88. Dynamic JTAG characteristics**

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$f_{PP}$ $1/t_{c(TCK)}$	TCK clock frequency	$2.7\text{ V} < V_{DD} < 3.6\text{ V}$	-	-	33	MHz
		$1.8\text{ V} < V_{DD} < 3.6\text{ V}$	-	-	25	
$t_{i_{su}(TMS)}$	TMS input setup time	-	0.5	-	-	ns
$t_{i_{h}(TMS)}$	TMS input hold time	-	1	-	-	
$t_{i_{su}(TDI)}$	TDI input setup time	-	1	-	-	
$t_{i_{h}(TDI)}$	TDI input hold time	-	2.5	-	-	
$t_{ov}(TDO)$	TDO output valid time	$2.7\text{ V} < V_{DD} < 3.6\text{ V}$	-	12	15	
		$1.8\text{ V} < V_{DD} < 3.6\text{ V}$	-	12	20	
$t_{oh}(TDO)$	TDO output hold time	-	10	-	-	

**Table 89. Dynamic SWD characteristics**

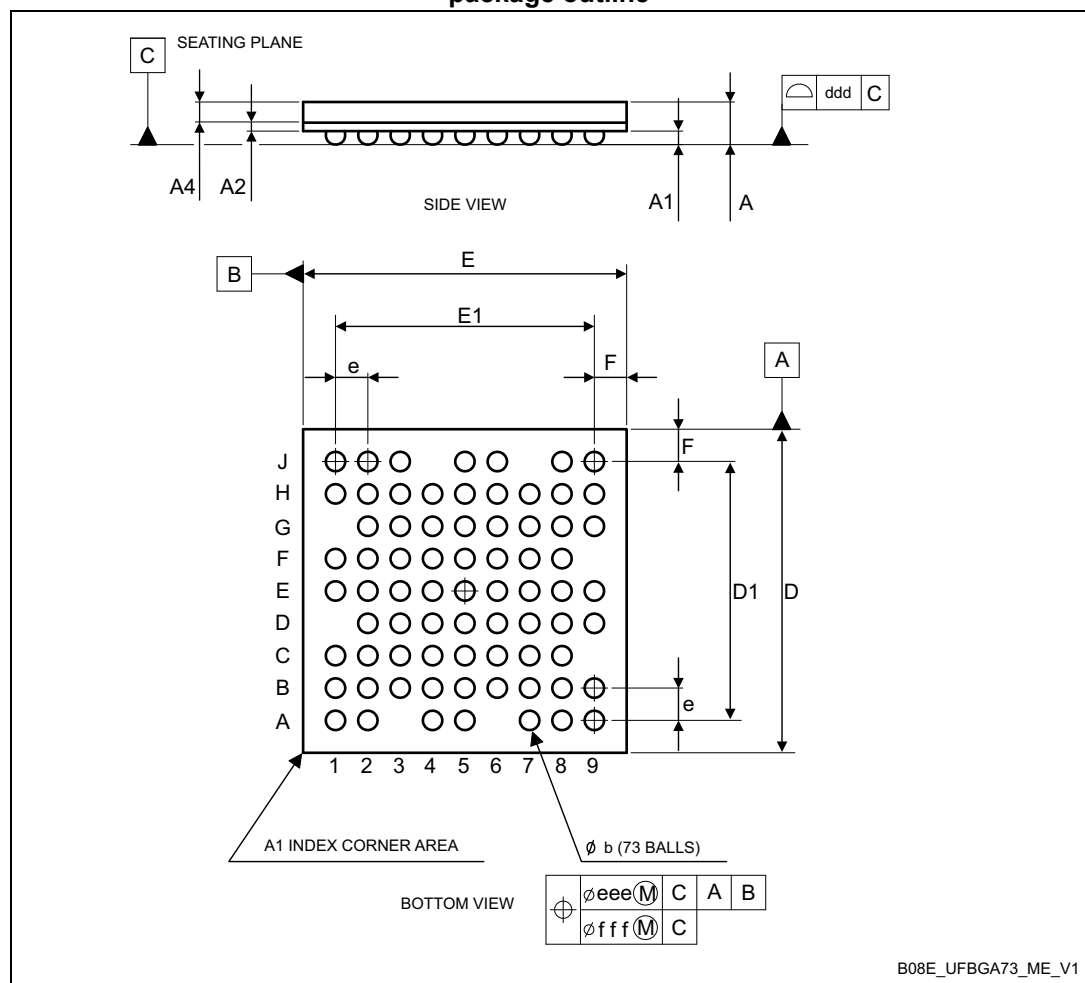
Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$f_{PP}$ $1/t_{c(SWCLK)}$	SWCLK clock frequency	$2.7\text{ V} < V_{DD} < 3.6\text{ V}$	-	-	58	MHz
		$1.8 < V_{DD} < 3.6\text{ V}$	-	-	41	
$t_{i_{su}(SWDIO)}$	SWDIO input setup time	-	1	-	-	ns
$t_{i_{h}(SWDIO)}$	SWDIO input hold time	-	2	-	-	
$t_{ov}(SWDIO)$	SWDIO output valid time	$2.7\text{ V} < V_{DD} < 3.6\text{ V}$	-	15	17	
		$1.8\text{ V} < V_{DD} < 3.6\text{ V}$	-	15	24	
$t_{oh}(SWDIO)$	SWDIO output hold time	-	9	-	-	

## 6 Package information

In order to meet environmental requirements, ST offers these devices in different grades of ECOPACK packages, depending on their level of environmental compliance. ECOPACK specifications, grade definitions and product status are available at [www.st.com](http://www.st.com). ECOPACK is an ST trademark.

### 6.1 UFBGA73 package information

Figure 26. UFBGA - 73 balls, 5 × 5 mm, ultra thin fine pitch ball grid array package outline



1. Drawing is not to scale.
2. - The terminal A1 corner must be identified on the top surface by using a corner chamfer, ink or metalized markings, or other feature of package body or integral heat slug.  
- A distinguishing feature is allowable on the bottom surface of the package to identify the terminal A1 corner. Exact shape of each corner is optional.

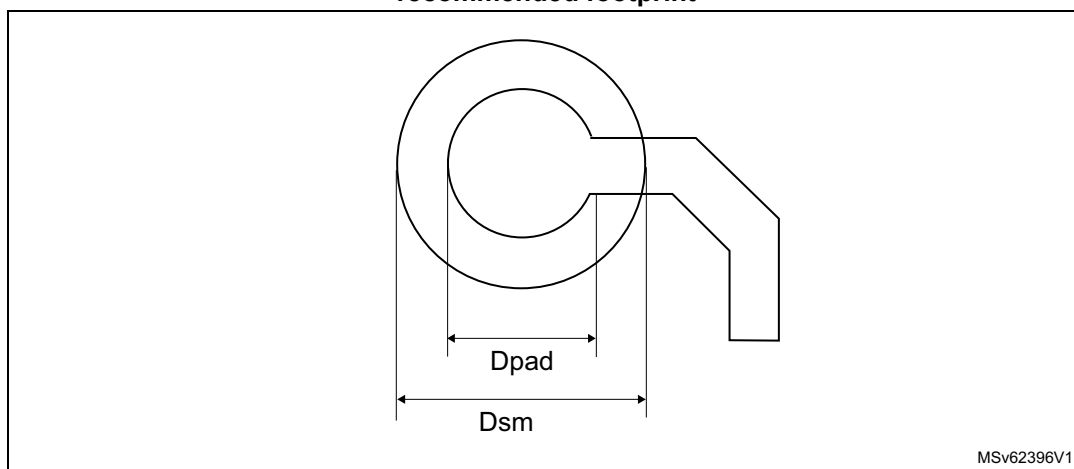


**Table 90. UFBGA - 73 balls, 5 × 5 mm, ultra thin fine pitch ball grid array mechanical data**

Symbol	millimeters			inches <sup>(1)</sup>		
	Min	Typ	Max	Min	Typ	Max
A <sup>(2)</sup>	-	-	0.60	-	-	0.236
A1	-	-	0.11	-	-	0.0043
A2	-	0.13	-	-	0.0051	-
A4	-	0.32	-	-	0.0126	-
b <sup>(3)</sup>	0.24	0.29	0.34	0.0094	0.0114	0.0134
D	4.85	5.00	5.15	0.1909	0.1969	0.2028
D1	-	4.00	-	-	0.1575	-
E	4.85	5.00	5.15	0.1909	0.1969	0.2028
E1	-	4.00	-	-	0.1575	-
e	-	0.50	-	-	0.0197	-
F	-	0.50	-	-	0.0197	-
ddd	-	-	0.08	-	-	0.0031
eee <sup>(4)</sup>	-	-	0.15	-	-	0.0059
fff <sup>(5)</sup>	-	-	0.05	-	-	0.0020

- Values in inches are converted from mm and rounded to 4 decimal digits.
- UFBGA stands for Ultra-Thin Profile Fine Pitch Ball Grid Array.
  - Ultra Thin profile:  $0.50 < A \leq 0.65\text{mm}$  / Fine pitch:  $e < 1.00\text{mm}$  pitch.
  - The total profile height (Dim A) is measured from the seating plane to the top of the component
  - The maximum total package height is calculated by the following methodology:  
 $A \text{ Max} = A1 \text{ Typ} + A2 \text{ Typ} + A4 \text{ Typ} + \sqrt{(A1^2 + A2^2 + A4^2 \text{ tolerance values})}$
- The typical balls diameters before mounting is 0.20 mm.
- The tolerance of position that controls the location of the pattern of balls with respect to datum A and B. For each ball there is a cylindrical tolerance zone eee perpendicular to datum C and located on true position with respect to datum A and B as defined by e. The axis perpendicular to datum C of each ball must lie within this tolerance zone.
- The tolerance of position that controls the location of the balls within the matrix with respect to each other. For each ball there is a cylindrical tolerance zone fff perpendicular to datum C and located on true position as defined by e. The axis perpendicular to datum C of each ball must lie within this tolerance zone. Each tolerance zone fff in the array is contained entirely in the respective zone eee above. The axis of each ball must lie simultaneously in both tolerance zones.

**Figure 27. UFBGA - 73 balls, 5 × 5 mm, ultra thin fine pitch ball grid array recommended footprint**



**Table 91. UFBGA recommended PCB design rules (0.5 mm pitch BGA)**

Dimension	Recommended values
Pitch	0.5 mm
$D_{pad}$	0.230 mm
$D_{sm}$	0.330 mm typ. (depends on the soldermask registration tolerance)
Stencil opening	0.280 mm
Stencil thickness	Between 0.100 mm and 0.125 mm
Pad trace width	0.100 mm
Ball diameter	0.280 mm

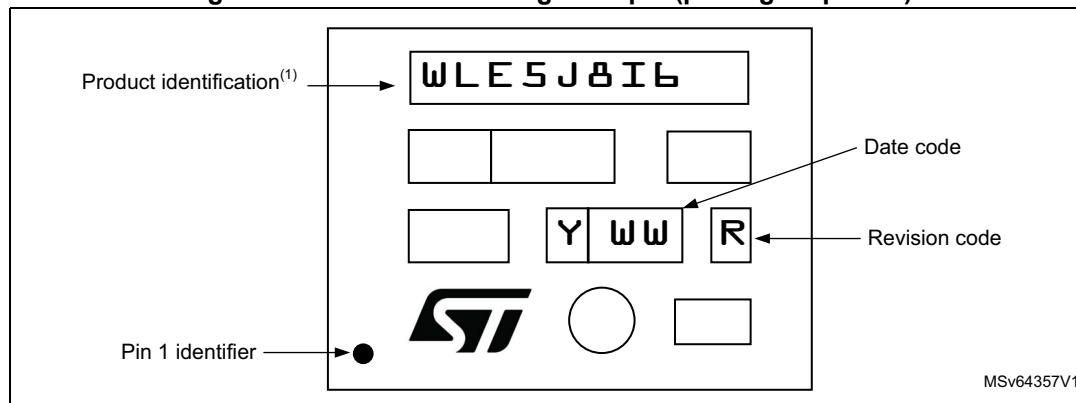
### Device marking for UFBGA73

The following figure gives an example of topside marking versus pin 1 position identifier location.

The printed markings may differ depending on the supply chain.

Other optional marking or inset/upset marks, which depend on supply chain operations, are not indicated below.

**Figure 28. UFBGA73 marking example (package top view)**



1. Parts marked as "ES", "E" or accompanied by an Engineering Sample notification letter, are not yet qualified and therefore not approved for use in production. ST is not responsible for any consequences resulting from such use. In no event will ST be liable for the customer using any of these engineering samples in production. ST's Quality department must be contacted prior to any decision to use these engineering samples to run a qualification activity.

## 6.2 Package thermal characteristics

The maximum chip junction temperature ( $T_J \text{ max}$ ) must never exceed the values given in [Table 25: General operating conditions](#).

The maximum chip-junction temperature,  $T_J \text{ max}$  (in °C), can be calculated using the equation:

$$T_J \text{ max} = T_A \text{ max} + (P_D \text{ max} \times \Theta_{JA})$$

where:

- $T_A \text{ max}$  is the maximum ambient temperature in °C.
- $\Theta_{JA}$  is the package junction-to-ambient thermal resistance, in °C/W.
- $P_D \text{ max}$  is the sum of  $P_{INT} \text{ max}$  and  $P_{I/O} \text{ max}$  ( $P_D \text{ max} = P_{INT} \text{ max} + P_{I/O} \text{ max}$ ).
- $P_{INT} \text{ max}$  is the product of  $I_{DD}$  and  $V_{DD}$ , expressed in Watt. This is the maximum chip internal power.

$P_{I/O} \text{ max}$  represents the maximum power dissipation on output pins:

$$P_{I/O} \text{ max} = \sum (V_{OL} \times I_{OL}) + \sum ((V_{DD} - V_{OH}) \times I_{OH})$$

taking into account the actual  $V_{OL} / I_{OL}$  and  $V_{OH} / I_{OH}$  of the I/Os at low and high level in the application.

When the SMPS is used, a portion of the power consumption is dissipated into the external inductor, therefore reducing the device power dissipation. This portion depends mainly on the inductor ESR characteristics.

As the radiated RF power is quite low (< 4 mW), it is not necessary to remove it from the device power consumption.

RF characteristics (such as sensitivity, Tx power consumption) are provided up to 85 °C.

**Table 92. Package thermal characteristics**

Symbol	Parameter	Value	Unit
$\Theta_{JA}$	<b>Thermal resistance junction-ambient</b> UFBGA73 - 5 mm x 5 mm	20.1	°C/W
$\Theta_{JB}$		20.2	
$\Theta_{JC}$		31.1	

# 7 Ordering information

Example:	STM32	WL	E5	J	8	I	6	TR
Device family								
STM32 = Arm® based 32-bit microcontroller								
Product type								
WL = wireless long range								
Device subfamily								
E5 = Cortex-M4, full set of modulations								
Pin count								
J = 73 pins								
Flash memory size								
8 = 64 Kbytes B = 128 Kbytes C = 256 Kbytes								
Package								
I = UFBGA (5 x 5mm)								
Temperature range								
6 = Industrial temperature range, -40 to 85 °C (105 °C junction)								
Packing								
TR = tape and reel xxx = programmed parts								

For a list of available options (such as speed or package) or for further information on any aspect of this device, contact the nearest ST sales office.

## 8 Revision history

Table 93. Document revision history

Date	Revision	Changes
20-Dec-2019	1	Initial release.
8-Jan-2020	2	Updated <i>Ultra-low-power platform</i> features.
25-Feb-2020	3	Updated: <ul style="list-style-type: none"> <li>– <math>T_J</math> in <a href="#">Table 2: Main features and peripheral counts</a> and in the whole doc</li> <li>– <a href="#">Section 3.9.1: Power supply schemes</a></li> <li>– <a href="#">Table 12: Temperature sensor calibration values</a></li> <li>– <a href="#">Table 13: Internal voltage reference calibration values</a></li> <li>– Ff renamed RF in <a href="#">Table 18: Legend/abbreviations used in the pinout table</a> and <a href="#">Table 19: STM32WLE5J8/JB/JC pin definition</a></li> <li>– new RF AF in <a href="#">Table 20: Alternate functions</a></li> <li>– <a href="#">Section 5.1.2: Typical values</a></li> <li>– <a href="#">Figure 12: Current consumption measurement scheme</a></li> <li>– <a href="#">Table 21: Voltage characteristics</a></li> <li>– <a href="#">Table 23: Thermal characteristics</a></li> <li>– <a href="#">Table 25: General operating conditions</a></li> <li>– <a href="#">Section 5.3.3: Sub-GHz radio characteristics</a></li> <li>– <a href="#">Table 34: Embedded internal voltage reference</a></li> <li>– <a href="#">Table 57: MSI oscillator characteristics</a></li> <li>– <a href="#">Table 58: LSI oscillator characteristics</a></li> <li>– <a href="#">Table 67: I/O static characteristics</a></li> <li>– <a href="#">Table 68: Output voltage characteristics</a></li> <li>– <a href="#">Table 69: I/O AC characteristics</a></li> <li>– <a href="#">Table 72: ADC characteristics</a></li> <li>– <a href="#">Table 75: TS characteristics</a></li> <li>– <a href="#">Table 76: <math>V_{BAT}</math> monitoring characteristics</a></li> </ul> Removed table LSI2 oscillator.
27-Feb-2020	4	Updated: <ul style="list-style-type: none"> <li>– <a href="#">Features</a></li> <li>– <a href="#">Table 24: Main performances at <math>VDD = 3\text{ V}</math></a></li> </ul>

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