#### 1. Incremental PID controller

The discrete PID controller is defined as follows:

$$u(n) = k_p e(n) + k_i \sum_{k=0}^{n} e(k) + k_d [e(n) - e(n-1)]$$
(1)

In order to switch to the incremental PID controller expression,

$$\Delta u(n) = u(n) - u(n-1) \tag{2}$$

is defined. Plugging in the controller results in,

$$\Delta u(n) = u(n) - u(n-1)$$

$$= k_p e(n) + k_i \sum_{k=0}^{n} e(k) + k_d [e(n) - e(n-1)] - k_p e(n-1)$$

$$- k_i \sum_{k=0}^{n-1} e(k) - k_d [e(n-1) - e(n-2)]$$

$$= k_p e(n) + k_i \sum_{k=0}^{n-1} e(k) + k_i e(n) + k_d e(n) - k_d e(n-1)$$

$$- k_p e(n-1) - k_i \sum_{k=0}^{n-1} e(k-1) - k_d e(n-1) + k_d e(n-2)$$

$$= k_p e(n) - k_p e(n-1) + k_i e(n) + k_d e(n) - 2k_d e(n-1) + k_d e(n-2)$$

$$= k_n [e(n) - e(n-1)] + k_i e(n) + k_d [e(n) - 2e(n-1) + e(n-2)]$$
(3)

## 2. Fuzzy PI controller

Let the inputs be e(n) and  $\Delta e(n)$  and the output be  $\Delta u(n)$  defined as error, change in error and change in control signal, respectively.

The membership functions Positive(P) and Negative(N) are defined as

$$\mu_{P}(e) \begin{cases} 0, & e < -L \\ \frac{e+L}{2L}, & -L \le e \le L \\ 1, & e > L \end{cases} \quad \mu_{N}(e) \begin{cases} 1, & e < -L \\ \frac{-e+L}{2L}, & -L \le e \le L \\ 0, & e > L \end{cases}$$
 (4)

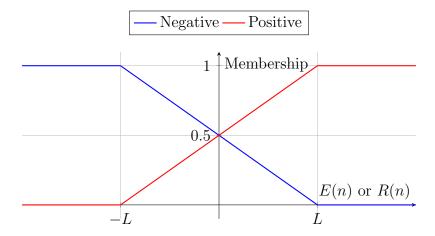


Figure 1: Input fuzzy sets: Negative and Positive

and

$$\mu_{P}(\Delta e) \begin{cases} 0, & \Delta e < -L \\ \frac{\Delta e + L}{2L}, & -L \le \Delta e \le L \\ 1, & \Delta e > L \end{cases} \quad \mu_{N}(\Delta e) \begin{cases} 1, & \Delta e < -L \\ \frac{-\Delta e + L}{2L}, & -L \le \Delta e \le L \\ 0, & \Delta e > L \end{cases}$$
 (5)

and depicted in Figure 1. The membership functions are chosen such that,

$$\mu_N(e) + \mu_P(e) = 1 \quad \mu_N(\Delta e) + \mu_P(\Delta e) = 1$$
 (6)

For the output variable the membership function is chosen as

$$\mu_N(\Delta u) \begin{cases} 1, & \Delta u = -H \\ 0, & \Delta u \neq -H \end{cases} \quad \mu_P(\Delta u) \begin{cases} 1, & \Delta u = H \\ 0, & \Delta u \neq H \end{cases} \quad \mu_Z(\Delta u) \begin{cases} 1, & \Delta u = 0 \\ 0, & \Delta u \neq 0 \end{cases}$$
 (7)

and is depicted in Figure 2.

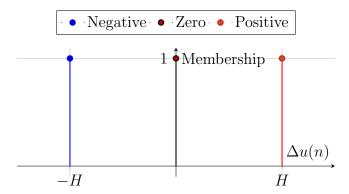


Figure 2: Output fuzzy sets as singleton values at  $-H,\ 0,$  and H

The following rule set is constructed.

Table 1: Fuzzy Rule Table: Control Output  $\Delta u(n)$  based on Error e(n) and Change of Error  $\Delta e(n)$ 

$e(n) \setminus \Delta e(n)$	N	$\mathbf{Z}$	P
${f N}$	N	N	Z
${f Z}$	N	Z	Р
P	Z		Р

The rule set is defined as follows:

IF e(n) is N AND  $\Delta e(n)$  is N THEN output is Negative

IF e(n) is N AND  $\Delta e(n)$  is P THEN output is Zero

IF e(n) is P AND  $\Delta e(n)$  is N THEN output is Zero

IF e(n) is P AND  $\Delta e(n)$  is P THEN output is Positive

The inference step results in:

$$\mu_N(e)\mu_N(\Delta e)$$
 for output  $-H$ 

$$\mu_N(e)\mu_P(\Delta e)$$
 for output 0

$$\mu_P(e)\mu_N(\Delta e)$$
 for output 0

$$\mu_P(e)\mu_P(\Delta e)$$
 for output H

and hence in,

$$\mu_N(e)\mu_N(\Delta e)$$
 for output  $-H$ 

$$\mu_P(e)\mu_P(\Delta e)$$
 for output H

The defuzzification step gives,

$$\Delta u(n) = \frac{\mu_P(e)\mu_P(\Delta e)H - \mu_N(e)\mu_N(\Delta e)H}{\mu_N(e)\mu_N(\Delta e) + \mu_N(e)\mu_P(\Delta e) + \mu_P(e)\mu_N(\Delta e) + \mu_P(e)\mu_P(\Delta e)}$$
(8)

The denominator is simplified as follows:

$$\mu_{N}(e)\mu_{N}(\Delta e) + \mu_{N}(e)\mu_{P}(\Delta e) + \mu_{P}(e)\mu_{N}(\Delta e) + \mu_{P}(e)\mu_{P}(\Delta e)$$

$$= (1 - \mu_{P}(e))\mu_{N}(\Delta e) + (1 - \mu_{P}(e))\mu_{P}(\Delta e) + \mu_{P}(e)\mu_{N}(\Delta e) + \mu_{P}(e)\mu_{P}(\Delta e)$$

$$= \mu_{N}(\Delta e) - \mu_{P}(e)\mu_{N}(\Delta e) + \mu_{P}(\Delta e) - \mu_{P}(e)\mu_{P}(\Delta e) + \mu_{P}(e)\mu_{N}(\Delta e) + \mu_{P}(e)\mu_{P}(\Delta e)$$

$$= \mu_{N}(\Delta e) + \mu_{P}(\Delta e)$$

$$= \mu_{N}(\Delta e) + \mu_{P}(\Delta e)$$

$$= 1$$

(9)

therefore,

$$\Delta u(n) = \mu_{P}(e)\mu_{P}(\Delta e)H - \mu_{N}(e)\mu_{N}(\Delta e)H$$

$$= H \begin{cases} -1, & e < -L \\ \frac{\Delta e + L}{2L} \frac{e + L}{2L} - \frac{-e + L}{2L} \frac{-\Delta e + L}{2L}, & -L \le e \le L \\ 1, & e > L \end{cases}$$

$$= \frac{\Delta e + L}{2L} \frac{e + L}{2L} - \frac{-e + L}{2L} \frac{-\Delta e + L}{2L}$$

$$= \frac{(\Delta e + L)(e + L) - (-e + L)(-\Delta e + L)}{4L^{2}}$$

$$= \frac{\Delta e(e + L) + L(e + L) - (-e)(-\Delta e + L) - L(-\Delta e + L)}{4L^{2}}$$

$$= \frac{e\Delta e + L\Delta e + Le + L^{2} - e\Delta e + Le + L\Delta e - L^{2}}{4L^{2}}$$

$$= \frac{e\Delta e + Le + Le + L\Delta e}{4L^{2}}$$

$$= \frac{L\Delta e + Le + Le + L\Delta e}{4L^{2}}$$

$$= \frac{2L\Delta e + 2Le}{4L^{2}}$$

$$= \frac{2L\Delta e + 2Le}{4L}$$

$$= \frac{\Delta e + e}{2L}$$

$$= \frac{1}{2L}\Delta e + \frac{1}{2L}e$$

$$= \frac{1}{2L}\Delta e + \frac{1}{2L}e$$

$$\Delta u(n) = \begin{cases} -H, & e < -L \\ \frac{H}{2L}\Delta e + \frac{H}{2L}e, & -L \le e \le L \\ H, & e > L \end{cases}$$

$$(10)$$

Since the control signal is a linear combination of  $\Delta e$  and e it is an incremental PI controller.

### 3. Tables

# 3.1. Aggressive Tracking

Advantage: Fast response and quick convergence to reference.

Disadvantage: May cause overshoot or oscillation in sensitive systems.

Table 2: Rule Table A: Aggressive Tracking

$e \backslash \Delta e$	N	Z	Р
N	N	N	N
Z	N	Р	Р
P	Р	Р	Р

# 3.2. Conservative Tracking

Table 3: Rule Table B: Conservative Tracking

$e \backslash \Delta e$	N	Z	Р
N	N	Z	Z
Z	Z	Z	Z
Р	Z	Z	Р

Advantage: Very stable with low overshoot.

Disadvantage: Slow response; possible steady-state error if not tuned well.

## 3.3. PD-like(Symmetric)

Table 4: Rule Table C: PD-like

$e \backslash \Delta e$	N	Z	Р
N	N	Z	Р
Z	N	Z	Р
Р	N	Z	Р

Advantage: Symmetric, general-purpose behavior; good if derivative effect dominates.

Disadvantage: No integral action  $\rightarrow$  steady-state error possible.