

An Information-Theoretic Perspective on Successive-Cancellation List Decoding

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- Recent interest in **efficient** transmission of **short** information blocks¹
 - 5G New Radio (5G NR) and already 6G^{2,3}
 - Internet of things (IoT) and wireless sensor networks (WSN), etc.³
- **Successive-cancellation list (SCL) decoding** combined with outer cyclic redundancy checks (CRCs) makes polar codes **competitive for short blocks**
 - **Adopted** for uplink and downlink control information for the enhanced mobile broadband (eMBB) communication service⁴
 - **Candidate** for ultra-reliable low-latency communications (URLLCs) and massive machine-type communications (mMTCs)⁴
 - **Useful** for communicating over the fading channels with no CSI^{5,6}
 - **Part of possible solutions** for unsourced random access problem^{7,8}

¹[Polyanskiy et al., 2010], Channel coding rate in the finite blocklength regime... (Trans. Inf. Theory)

²[Durisi et al., 2016], Towards massive, ultra-reliable, and low-latency wireless communications with short packets... (IEEE Proc.)

³[Mahmood et al., 2020], White paper on critical and massive machine type communication towards 6G... (CoRR)

⁴[Bioglio et al., 2021], Design of polar codes in 5G new radio... (IEEE Commun. Surveys & Tutorials)

⁵[Xhemrishi et al., 2019], List decoding of short codes for communication over unknown fading channels... (Asilomar)

⁶[Yuan et. al., 2021], Polar-coded non-coherent communication... (Commun. Lett.)

⁷[Fengler et. al., 2022], Pilot-based unsourced RA with a massive MIMO receiver, interference cancellation, and power control... (J-SAC)

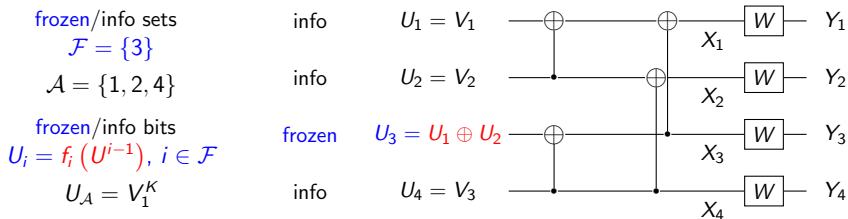
⁸[Gkagkos et al., 2022], FASURA: A scheme for quasi-static massive MIMO unsourced random access channels... (CoRR)

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Linear Codes based on Polar Transform

$$X^N = U^N G_N \quad \text{where} \quad G_N \triangleq B_N \begin{bmatrix} 1 & 0 \\ 1 & 1 \end{bmatrix}^{\otimes n} \quad \text{and} \quad N = 2^n$$



- Reed–Muller (RM) codes: optimize \mathcal{F} to maximize minimum distance^{9,10}
- Polar codes: optimize \mathcal{F} for successive cancellation (SC) decoding^{11,12}
 Requires estimation of $H(W_N^{(i)}) \triangleq H(U_i | Y^N, U^{i-1})$, for $i \in \{1, \dots, N\}$
- Any binary linear code is obtained with suitable \mathcal{F} and f_i ,¹³ $i \in \mathcal{F}$

⁹[Muller, 1954], Application of boolean algebra to switching circuit design and to error detection... (Trans. IRE Elect. Comp.)

¹⁰[Reed, 1954], A class of multiple-error-correcting codes and the decoding scheme... (Trans. IRE Inf. Theory)

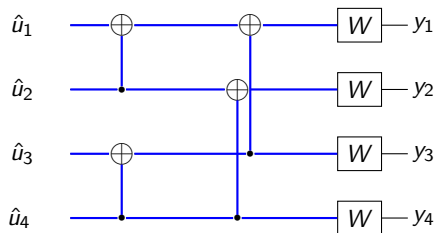
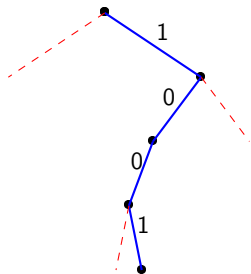
¹¹[Stolte, 2002], Rekursive Codes mit der Plotkin-Konstruktion und ihre Decodierung... (PhD Thesis, TU Darmstadt)

¹²[Arkan, 2009], Channel polarization: A method for constructing capacity-achieving codes for BMSCs... (Trans. Inf. Theory)

¹³[Trifonov and Miloslavskaya, 2016], Polar subcodes... (J-SAC)

Successive-Cancellation Decoding

Example: $u_3 = 0$ (frozen bit)



- Errors made by SC decoding **cannot be corrected** by later decisions
- Use SC list (SCL) decoding^{11,14,15} to **approach ML decoding performance**

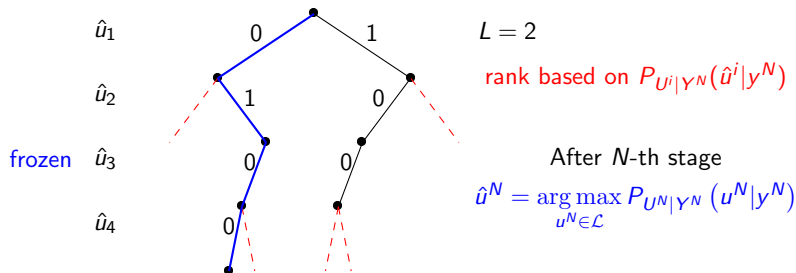
¹¹[Stolte, 2002], Rekursive Codes mit der Plotkin-Konstruktion und ihre Decodierung... (PhD Thesis, TU Darmstadt)

¹⁴[Dumer and Shabunov, 2006], Soft-decision decoding of Reed-Muller codes: recursive lists... (Trans. Inf. Theory)

¹⁵[Tal and Vardy, 2015], List decoding of polar codes... (Trans. Inf. Theory)

Successive-Cancellation List Decoding

For each i , both options are stored for decision on \hat{u}_i (i -th decoding stage,) which **doubles** the number of **paths** at each decoding stage



- Keep a list \mathcal{L} of past decisions for a set of L likely paths
Use an optimal rule to choose between final codewords on the list \mathcal{L}
- Recall that any binary linear code can be represented as a polar code
Can be used for **any binary linear block code**; may **not be efficient**

Motivating Question

- What list size is **sufficient to approach ML decoding** performance for a given code and channel?
 - Can be attacked via simulation but **quite complex for long codes and lists**
 - Simulation alone **unlikely to provide insight** into the question
 - A theoretical answer **might enable better code designs** for SCL decoding
 - A recent related work co-authored by Vardy¹⁶ provides an **upper bound on the sufficient list size** by generalizing a previous result co-authored by Urbanke¹⁷

¹⁶[Fazeli et al., 2021], List decoding of polar codes: How large should the list be to achieve ML decoding?... (ISIT)

¹⁷[Hashemi et al., 2018], Decoder partitioning: Towards practical list decoding of polar codes... (Trans. Commun.)

Summary of Results

- New bounds **related** to the list size required for near-ML performance
 - To avoid losing true codeword, **its rank must not be larger than list size**
 - The expected log-rank of correct codeword is **upper bounded by an entropy**
 - **Bounds on this entropy are derived** and relatively easy to compute
 - The log-rank of correct codeword **concentrates** around this mean
 - As an application, the analysis is used to modify RM codes
New codes **outperform 5G codes** under SCL decoding with practical list sizes
- For the binary erasure channel (BEC)
 - This entropy equals the dimension of an affine subspace
 - The random dimension sequence can be approximated by a Markov chain
 - For a fixed number of erasures, the **approximation is reasonably accurate**

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An Information Theoretic Perspective (1)

Basic Idea: After m steps, consider the conditional entropy $H(U^m|Y^N)$

The chain rule of entropy implies:

$$\begin{aligned} H(U^m|Y^N) &= \sum_{i=1}^m H(U_i|U^{i-1}, Y_1^N) \\ &= \sum_{i=1}^m H(W_N^{(i)}) \end{aligned}$$

But what about the frozen bits?

An Information-Theoretic Perspective (2)

- For the first m input bits, the information/frozen sets are denoted

$$\mathcal{A}^{(m)} \triangleq \mathcal{A} \cap \{1, \dots, m\} \quad \text{and} \quad \mathcal{F}^{(m)} \triangleq \mathcal{F} \cap \{1, \dots, m\}$$

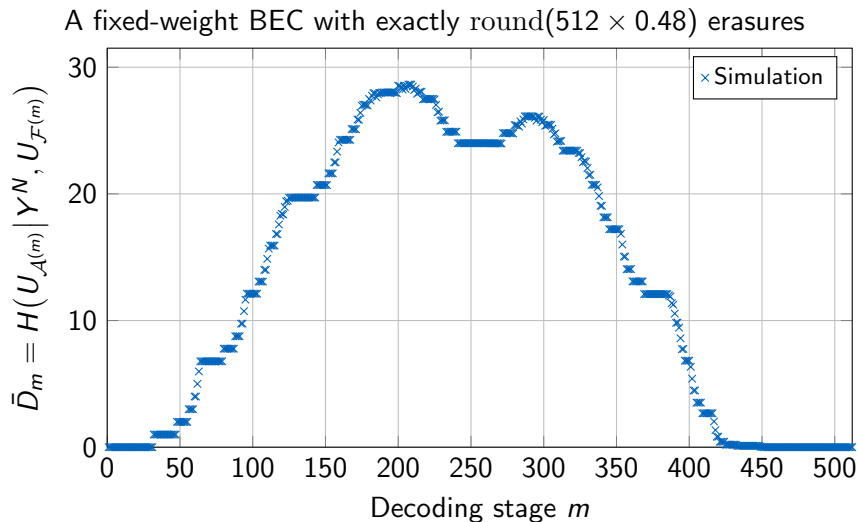
- Key Idea:** information entropy given frozen bits

$$d_m(y^N) \triangleq H(U_{\mathcal{A}^{(m)}} | Y^N = y^N, U_{\mathcal{F}^{(m)}}) \quad \text{and} \quad \bar{D}_m = H(U_{\mathcal{A}^{(m)}} | Y^N, U_{\mathcal{F}^{(m)}})$$

$$\sum_{i \in \mathcal{A}^{(m)}} H(W_N^{(i)}) - \sum_{i \in \mathcal{F}^{(m)}} (1 - H(W_N^{(i)})) \leq \bar{D}_m \leq \sum_{i \in \mathcal{A}^{(m)}} H(W_N^{(i)})$$

Q: Is the sequence $\{\bar{D}_m\}$ relevant for the list size for near-ML decoding on BMSCs?

An Exemplary (512, 256) Code



Theorem

Upon observing y^N when u^N is sent, we define the set (for $\alpha \in (0, 1]$)
 $\mathcal{S}_\alpha^{(m)}(u^m, y^N) \triangleq \{\tilde{u}^m : \mathbb{P}(\tilde{u}_{\mathcal{A}^{(m)}} | y^N, \tilde{u}_{\mathcal{F}^{(m)}}) \geq \alpha \mathbb{P}(u_{\mathcal{A}^{(m)}} | y_1^N, u_{\mathcal{F}^{(m)}})\}$. Then,

$$\mathbb{E} \left[\log_2 |\mathcal{S}_\alpha^{(m)}| \right] \leq \bar{D}_m + \log_2 \frac{1}{\alpha} = H(U_{\mathcal{A}^{(m)}} | Y^N, U_{\mathcal{F}^{(m)}}) + \log_2 \frac{1}{\alpha}$$

- Choosing $\alpha < 1$ (say 0.94) captures near misses and matches entropy better
 - Consider SCL decoding with max list size L_m after the m -th decoding step
 - It needs to satisfy $L_m \geq |\mathcal{S}_1^{(m)}|$ for the true u_1^m to stay on the list
- A first-order code design criterion: $\log_2 L_m \geq \bar{D}_m$
(reduce the peak $\bar{D}_{\max} \triangleq \max_m \bar{D}_m$)

A Few Remarks

- This approach currently has two weaknesses:
 - Entropy determined by typical events but coding cares about rare events
 - Sequence \bar{D}_m averaged over Y^N but decoder sees one realization $d_m(y^N)$

Theorem

For a wide range of BMS channels, the random variable D_m concentrates around its mean \bar{D}_m , i.e., for any $\beta > 0$, we have

$$\mathbb{P} \left\{ \frac{1}{N} |D_m - \bar{D}_m| > \beta \right\} \leq 2 \exp \left(-\frac{\beta^2}{c^2} N \right)$$

where c is a positive constant defined by the channel

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Successive Cancellation Inactivation Decoding

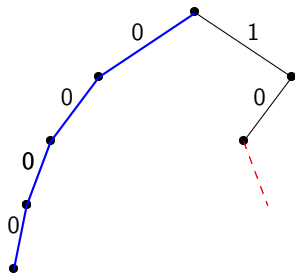
- Consider SCL decoding with **unbounded list size** on the BEC
 - Set of all valid paths after any decoding stage forms an **affine subspace**
 - SCL decoding tracks **all** valid paths defined by this subspace
- SC inactivation (SCI) decoding instead stores a **basis for space**¹⁸
 - If SC decoding step outputs erasure, **inactivate** the bit and add basis vector
 - Later messages in decoder are functions of inactivated bits (i.e., basis vectors)
 - If SC decoding of frozen bit is an unerased message, then resulting equation may allow one to **consolidate** the basis (i.e., remove a basis vector)

¹⁸[Coşkun, Neu, and Pfister, 2020], Successive cancellation inactivation decoding for modified Reed–Muller and eBCH codes... (ISIT)

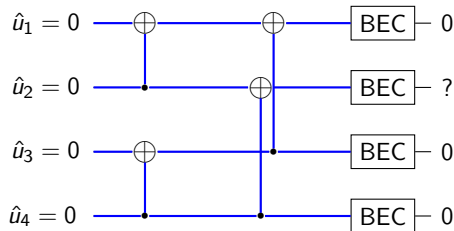
Successive Cancellation Inactivation Decoding: Algorithm

Example: $u_3 = 0$ (frozen bits)

$d_m(y_1^N)$ = subspace dimension after m -th decoding stage



$$d_4(y_1^4) = 0$$



- **Unique** solution only if $d_N(y_1^N) = 0$; otherwise declare an **error**
- Equivalent to SCL decoding **with unbounded list size** aka **ML decoding**

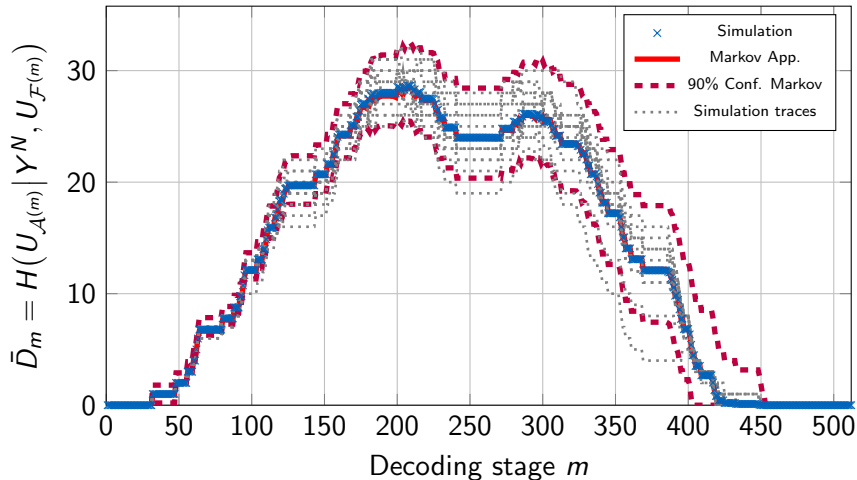
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- Polarization-adjusted convolutional (PAC) codes¹⁹:
 - Given set \mathcal{A} and a rate-1 convolutional code (CC) with memory ν
 - Encode information and frozen bit sequence with CC before applying polar transform
 - Decode using SCL or other methods, e.g., sequential decoding
 - For short lengths, RM frozen indices appear to be a good choice
- Dynamic RM (dRM) code ensemble¹⁸:
 - Let \mathcal{A} be the information indices of an RM code
 - Modified RM code where frozen bits are random linear function of past bits
 - Closely related to PAC codes

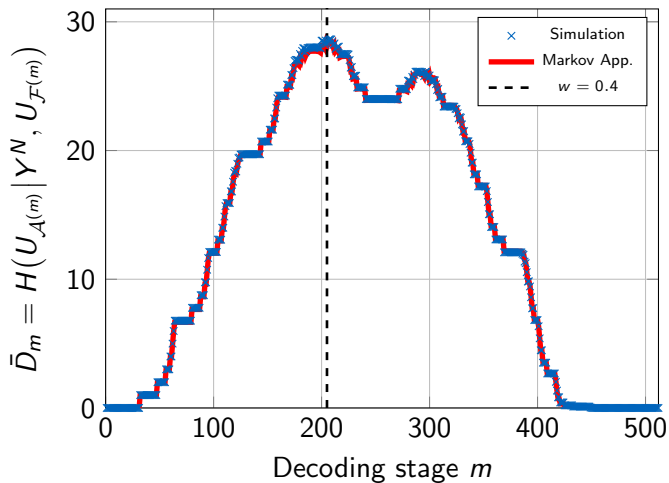
¹⁹[Arkan, 2019], From sequential decoding to channel polarization and back again... (CoRR)

¹⁸[Coşkun, Neu and Pfister, 2020] Successive cancellation inactivation decoding for modified Reed–Muller and eBCH codes... (ISIT)

A fixed-weight BEC with exactly $\text{round}(512 \times 0.48)$ erasures



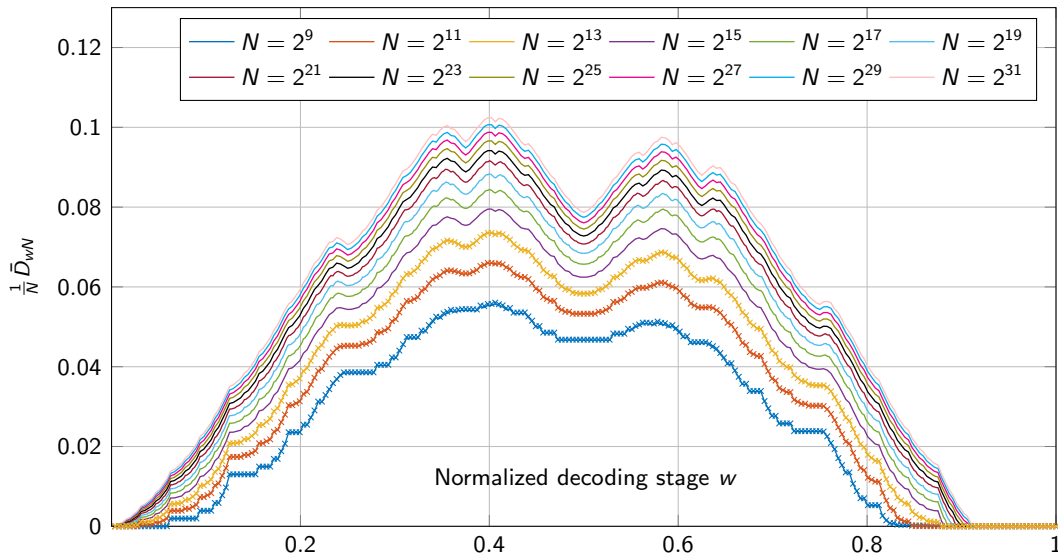
(512, 256) dRM Code



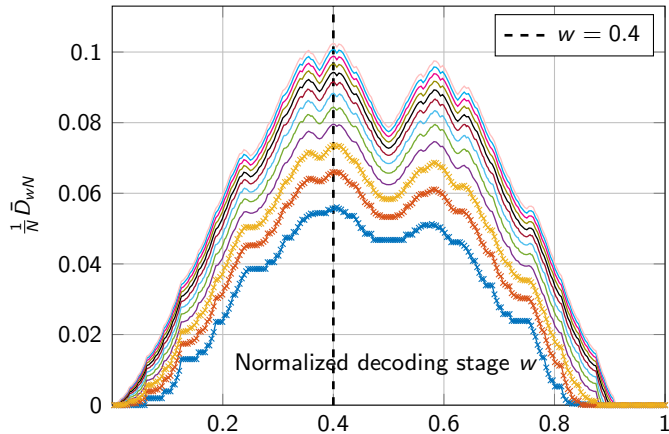
- How does \bar{D}_m behave as block length increases?
- Let $w \triangleq \frac{m}{N}$, $m \in \{1, \dots, N\}$, define the sequence $\frac{1}{N} \bar{D}_{wN}$
E.g.: $w = 0.4 \rightarrow \frac{1}{512} \bar{D}_{0.4N}$

Growth Rate of Subspace Dimension for dRM Codes with $R = 0.5$

A fixed-weight BEC with exactly $\text{round}(N \times 0.48)$ erasures



Growth Rate of Subspace Dimension for dRM Codes with $R = 0.5$



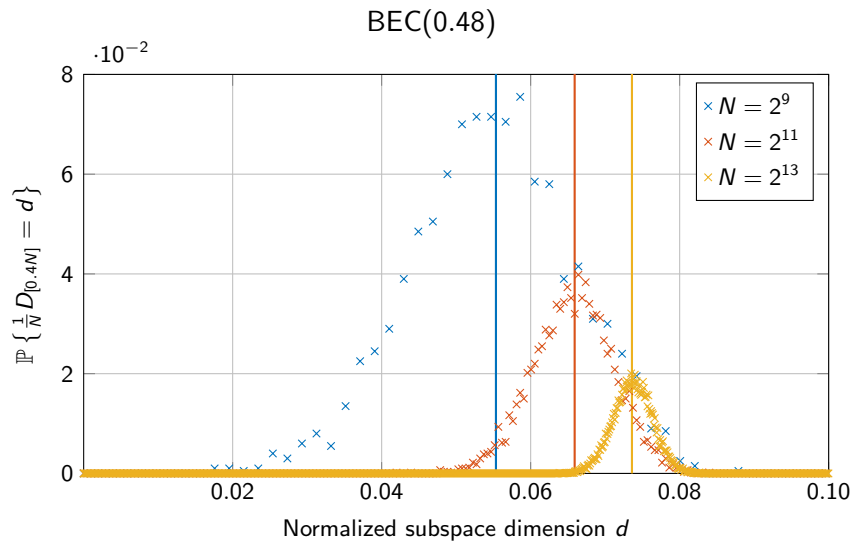
- How does the PMF of $\frac{1}{N}\bar{D}_{wN}$ behave as **block length increases** for a fixed w ?

- Set $w = 0.4$ and plot

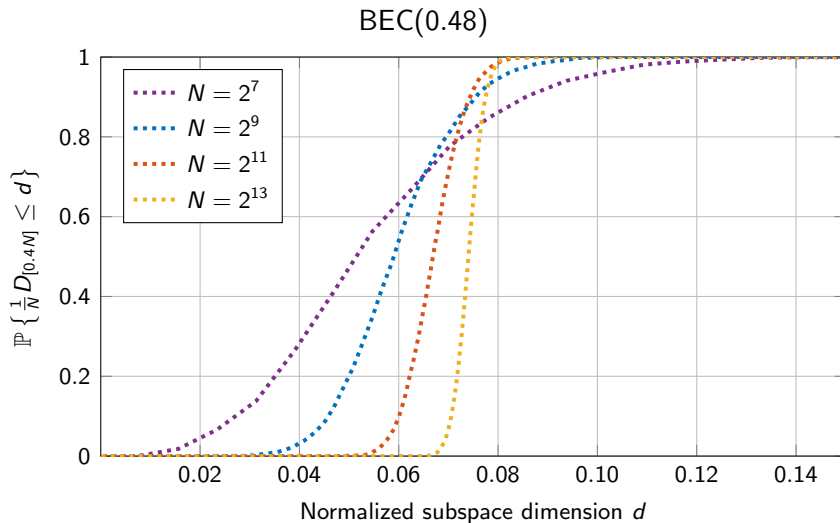
$$\mathbb{P}\left\{\frac{1}{N}\bar{D}_{[0.4N]} = d\right\}$$

where $d \in \mathcal{D}_N \subset [0, R]$

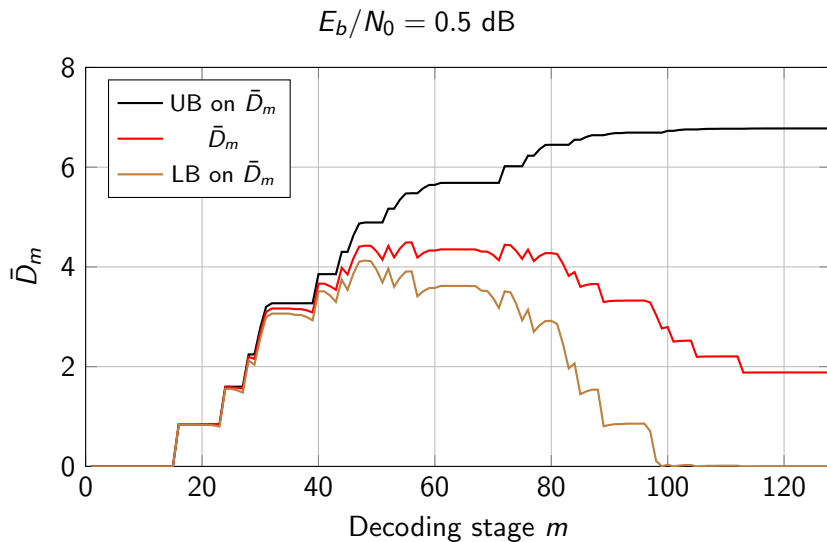
PMFs for $\frac{1}{N}D_{[0.4N]}$ for dRM Codes with $R = 0.5$



CDFs for $\frac{1}{N}D_{[0.4N]}$ for dRM Codes with $R = 0.5$



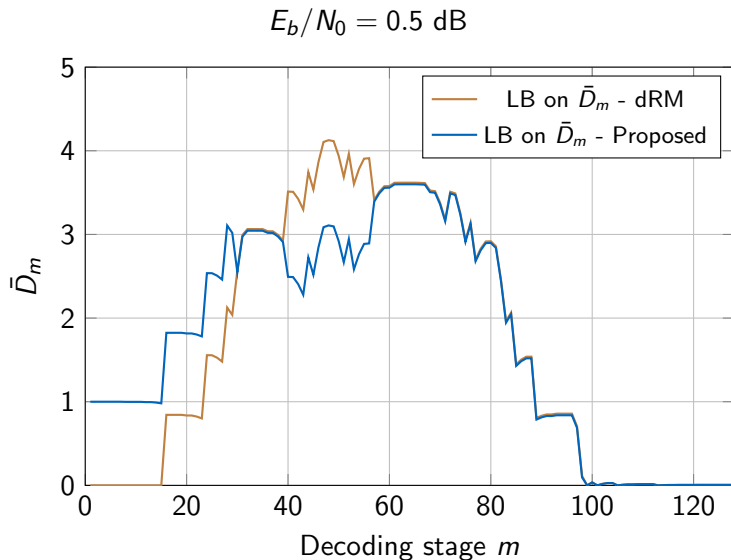
(128, 64) dRM Code over the BAWGNC



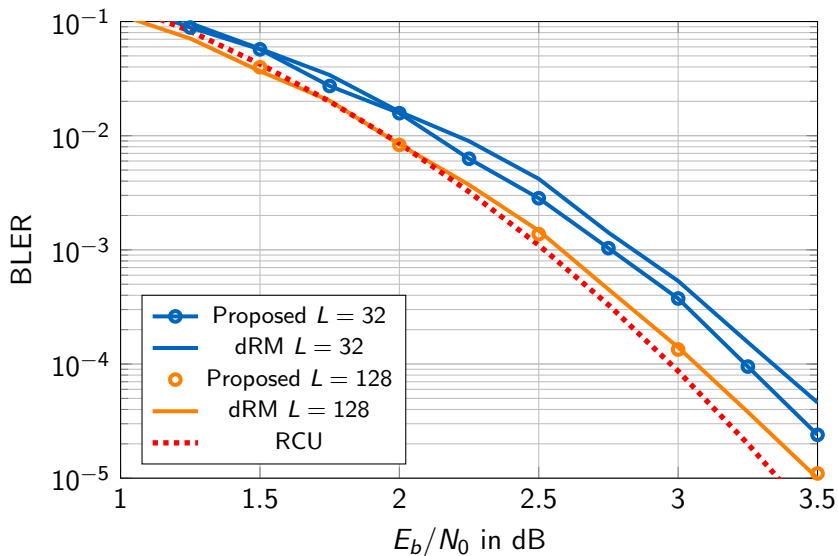
(128, 64) Proposed vs dRM Code over the BAWGNC

Proposed Code

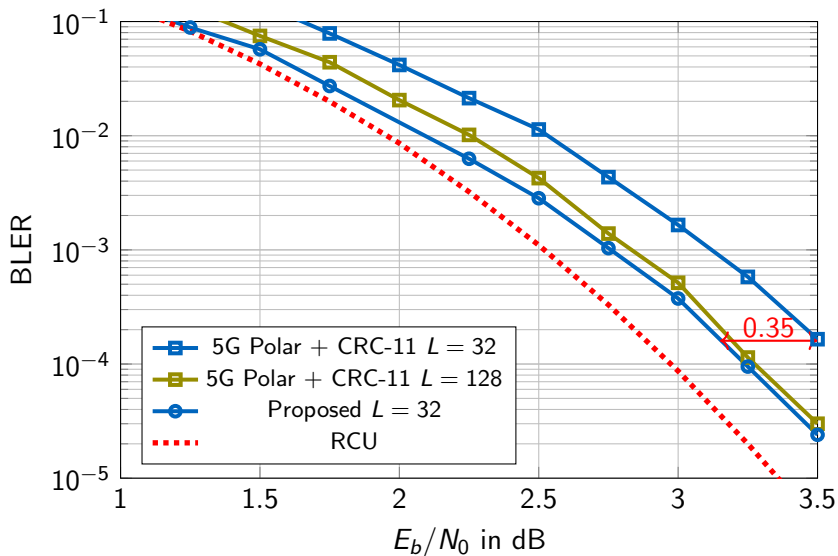
- $u_{\{30,40\}}$ dynamic frozen bits
- $u_{\{1,57\}}$ info. bits



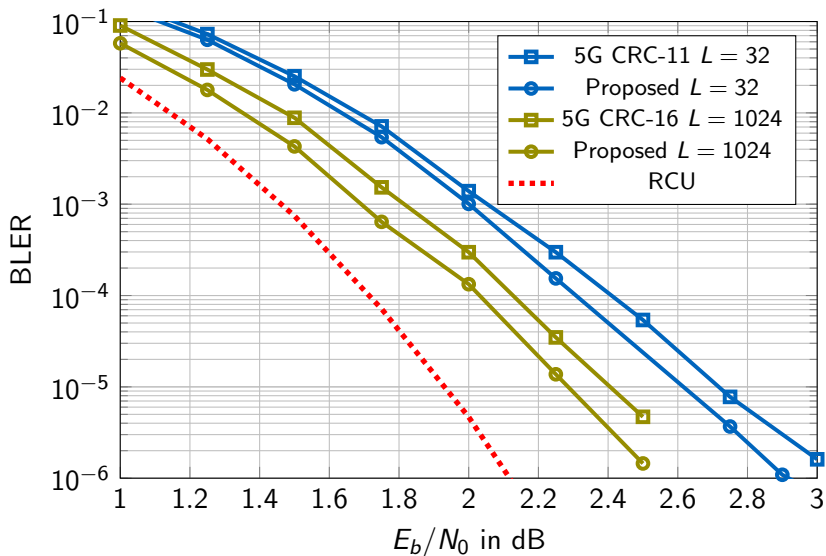
(128, 64) Codes over the BAWGNC



(128, 64) Codes over the BAWGNC



(512, 256) Codes over the BAWGNC



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- Recent advances in polar codes allow performance near random coding union bound for codes up to 512 bits of length with moderate complexity
 - Dynamic frozen bits act as outer code that improves minimum distance
 - SCL decoding can fully exploit outer code with large enough list size
- “What list size is sufficient to approach maximum-likelihood (ML) decoding performance under an SCL decoder?”
 - Information theory provides some estimates of required list size
 - For the BEC, the estimate is quite accurate and even relevant for optimum decoding
- Analysis leads to **improved designs** (in comparison with the PAC code and 5G polar codes) under SCL decoding with list sizes $L \in [8, 1024]$

Thanks for your attention