

1) Square law modulator

message = $m(t)$

carrier = $c(t) = A_c \cos(2\pi f_c t)$

input = $m(t) + c(t)$

non linear device

input \rightarrow [device] \rightarrow output = $T\{\text{input}\}$

$$\begin{aligned} \text{Output} = T\{\text{input}\} &= m(t) + c(t) + (m(t) + c(t))^2 \\ &= m(t) + c(t) + m(t)^2 + 2m(t)c(t) + c(t)^2 \\ &= m(t) + A_c \cos(2\pi f_c t) + m(t)^2 + 2m(t) A_c \cos(2\pi f_c t) + \frac{A_c^2}{2} (1 + \cos(4\pi f_c t)) \end{aligned}$$

Band pass filter

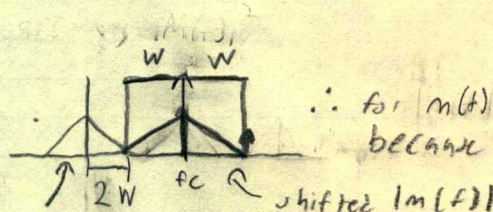
input = $m(t) + c(t) + m(t)^2 + 2c(t)m(t) + c(t)^2$

input \rightarrow [BPF] \rightarrow output

$$\begin{aligned} \text{Output} = T\{\text{input}\} &= m(t)^2 + c(t) + m(t)^2 + 2m(t)c(t) + \frac{A_c^2}{2} (1 + \cos(4\pi f_c t)) \\ &= (1 + 2m(t)) A_c \cos(2\pi f_c t) \therefore \text{DSB-PC} \checkmark \\ &\text{As long as } f_c > 3 \text{ Bandwidth} \end{aligned}$$

Reason:

Reasoning



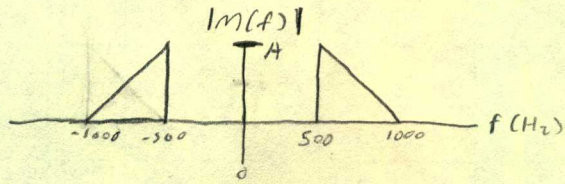
$$|m(f) * m(f)| = F\{m(t)^2\}$$

 \therefore for $m(t)^2$ to not be included, $f_c > 3W$

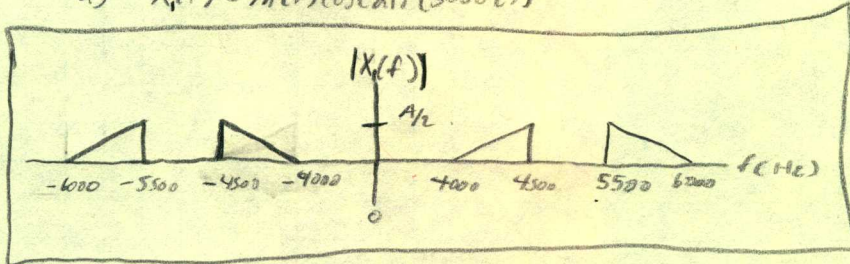
because when $m(t)$ is squared the bandwidth is doubled. A signal convolved with itself doubles the bandwidth.

With this bandwidth requirement, the $c(t)^2$ term will create an impulse far outside the BPF at or near $6W$. The $m(t)$ term has half the bandwidth of $m(t)^2$ and will also be removed by the BPF, leaving only the DSB-PC terms.

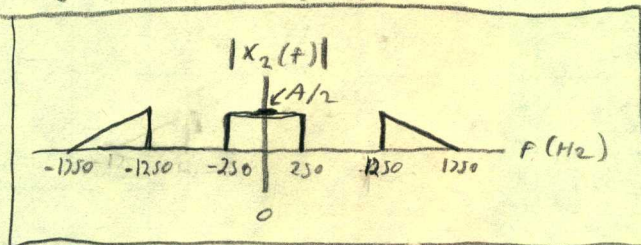
2)



a) $x_1(t) = m(t) \cos(2\pi(5000)t)$



b) $x_2(t) = m(t) \cos(2\pi(750)t)$



WCF, CW: there is overlap; creating the square wave