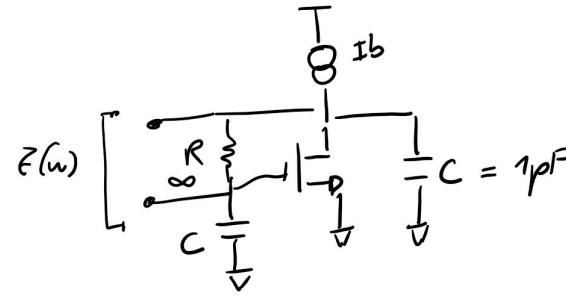


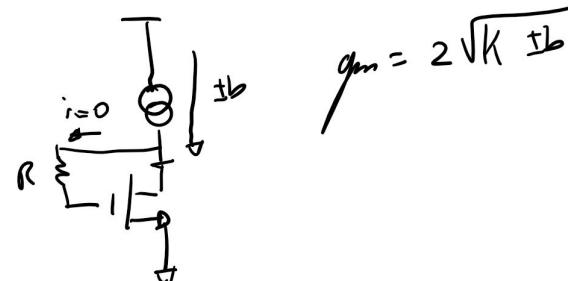
EXERCISE T 7.1

$$V_T = 0.5V$$

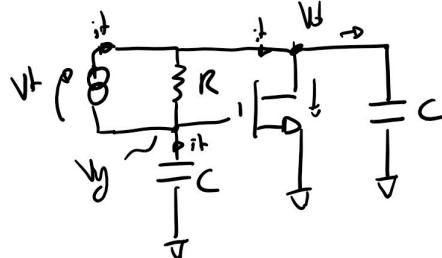
$$K = \frac{1}{2} f \omega C \alpha \left(\frac{V_T}{L} \right) = 2 \frac{\mu A}{V^2}$$



- BIAS POINT OF THE MOS



a) DERIVE $Z(w)$



$$Z(s) = \frac{V_L}{i_L} = \frac{(V_d - V_g)}{i_L} = \frac{V_d}{i_L} - \frac{V_g}{i_L}$$

$$V_g = -\frac{i_L}{nC}$$

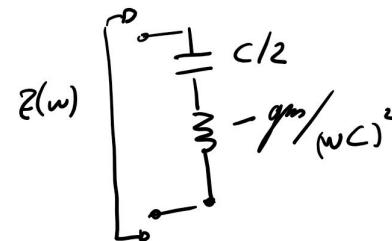
$$i_L = V_g g_m + V_d nC = -\frac{i_L}{nC} g_m + V_d nC$$

$$V_d = \frac{i_L}{nC} \left[1 + \frac{g_m}{nC} \right]$$

$$Z(s) = \frac{1}{nC} \left[1 + \frac{g_m}{nC} \right] + \frac{1}{nC} = \frac{2}{nC} + \frac{g_m}{(nC)^2}$$

$$\omega = j\omega$$

$$Z(w) = \frac{2}{j\omega C} + \frac{g_m}{(j\omega C)^2} = \frac{2}{j\omega C} - \frac{g_m}{(\omega C)^2}$$



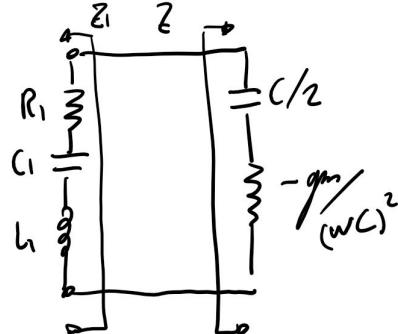
b) DERIVE OSCILLATION FREQUENCY WHEN Z_1 IS CONNECTED TO Z .

- From NEGATIVE RESISTOR MODE TO GUARANTEE POWER BALANCE @ ω_0

$$Z(j\omega_0) + Z_1(j\omega_0) = 0$$

$$\textcircled{1} \quad \text{Im}\{Z(j\omega)\} + \text{Im}\{Z_1(j\omega)\} = 0$$

$$\textcircled{2} \quad \text{Re}\{Z(j\omega)\} + \text{Re}\{Z_1(j\omega)\} = 0$$



$$\text{From } \textcircled{1}: \omega_0 L_1 - \frac{1}{\omega_0 C_1} - \frac{1}{\omega_0 C/2} = 0 \quad \leadsto \quad \omega_0 = \frac{1}{\sqrt{L_1 (C_1 \parallel C/2)}} \\ = 2\pi 40,13 \text{ MHz}$$

c) MINIMUM I_{bias} to guarantee start-up

@ steady state $\operatorname{Re}\{\bar{Z}_1(j\omega)\} + \operatorname{Re}\{\bar{Z}(j\omega)\} = 0$

@ start $\operatorname{Re}\{\bar{Z}_1(j\omega)\} + \operatorname{Re}\{\bar{Z}(j\omega)\} < 0$

$$R_1 - \frac{g_m}{\omega^2 C^2} < 0$$

$$\left\{ \begin{array}{l} g_m > R_1 \omega^2 C^2 \\ g_m = 2 \sqrt{K I_{bias}} \end{array} \right.$$

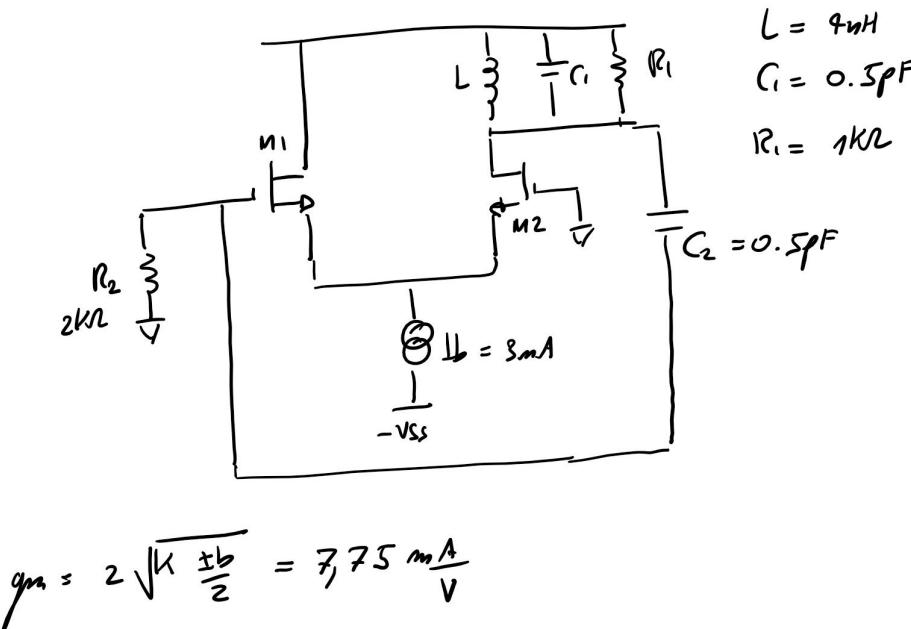
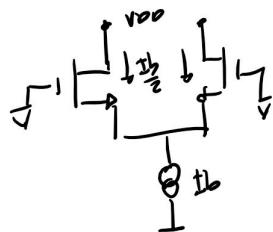
$$I_{bias} \geq \frac{1}{K} \left(R_1 \frac{\omega^2 C^2}{2} \right)^2 = 0.449 \mu A$$

EXERCISE 7.2

$$k = \frac{1}{2} \mu n C_0 \times \left(\frac{W}{L} \right) = 10 \frac{mA}{V^2}$$

$$V_T = 0.5V$$

MOS BIAS POINT



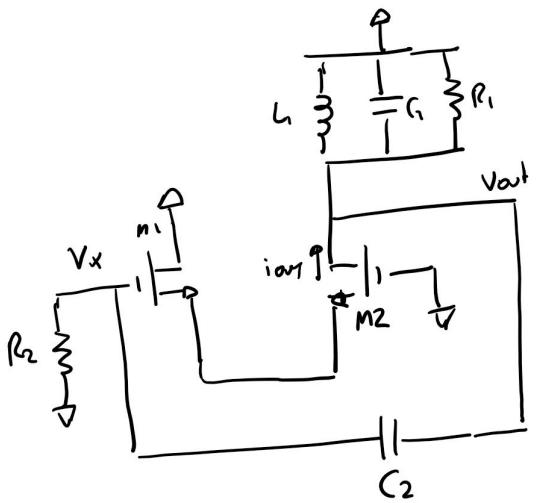
$$L = 4mH$$

$$C_1 = 0.5pF$$

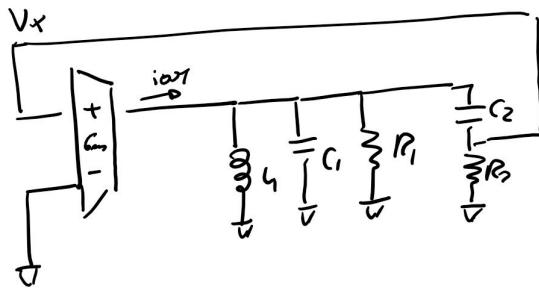
$$R_1 = 1k\Omega$$

$$g_m = 2 \sqrt{k \frac{I_D}{2}} = 7.75 \frac{mA}{V}$$

a) DERIVE ω_0 (WITH SIMPLIFICATIONS)



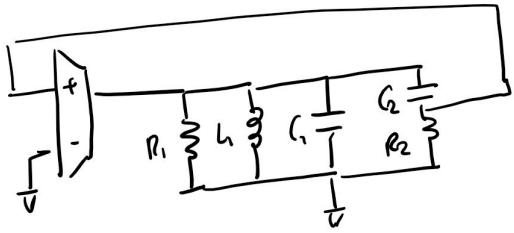
EQ. MODEL
TAN θ + TRANSCONCR



$$V_x = \frac{m_1}{t} i_{out}$$

$$i_{out} = g_m [V_x - \frac{i_{out}}{g_m}]$$

$$\frac{i_{out}}{V_x} = \frac{g_m}{2} = G_m$$



- SER - TO - PARALLEL TRANSFORMATION



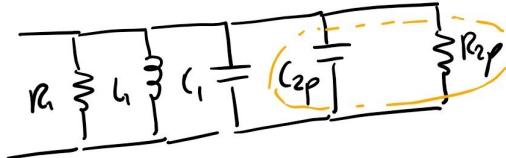
SER - TO - PAR Ω_{wo}

$$R_p = R_s (1 + \Omega_T^2)$$

$$X_p = X_s \left(1 + \frac{1}{\Omega_T^2}\right)$$

$$\Omega_T = \frac{X_s}{R_s} = \frac{R_p}{X_p}$$

PAR - TO - SER Ω_{wo}



$$R_{2p} = R_2 (1 + \Omega_T^2)$$

$$\frac{1}{w_0 C_{2p}} = \frac{1}{w_0 C_2} \left(1 + \frac{1}{\Omega_T^2}\right)$$

$$\Omega_T = \frac{1}{w_0 C_2 R_2} = w_0 C_{2p} R_{2p}$$

- we don't know w_0
make hypothesis of Ω_T

$$\text{if } \Omega_T \ll 1 \rightarrow w_0 \gg \frac{1}{C_2 R_2} = 2\pi 159 \text{ MHz}$$

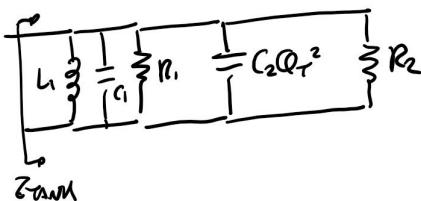
$$\text{if } \Omega_T \gg 1 \rightarrow w_0 \ll \frac{1}{C_2 R_2} = 2\pi 159 \text{ MHz}$$

- consider $\Omega_T \ll 1$

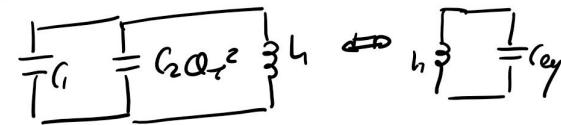
$$\begin{cases} R_{2p} = R_2 (1 + \Omega_T^2) \approx R_2 \\ \frac{1}{w_0 C_{2p}} = \frac{1}{w_0 C_2} (1 + \frac{1}{\Omega_T^2}) \approx \frac{1}{w_0 C_2 \Omega_T^2} \Rightarrow C_{2p} = C_2 \Omega_T^2 \end{cases}$$

Since
 $L \sim n^4 \rightarrow f \sim \text{GHz}$
 $C \sim \mu F$

• EQ TANK MODEL



$$\text{Im}\{Z_{\text{TANK}}\} = 0$$



$$w_0 L_1 - \frac{1}{w_0 [C_1 + (2Q_T)^2]} = 0 \quad Q_T = \frac{1}{w_0 R_2 C_2}$$

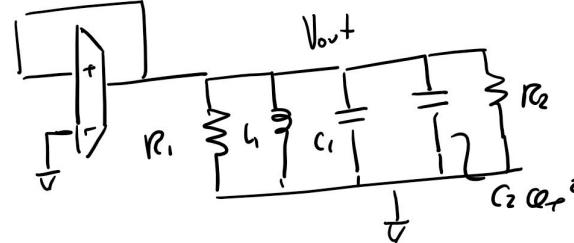
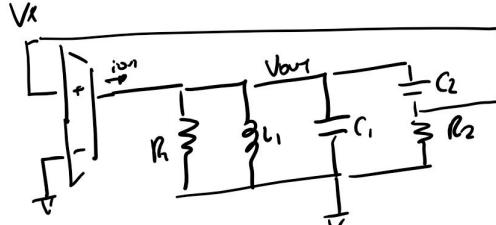
$$w_0 L_1 - \frac{1}{w_0 \left[C_1 + \frac{1}{(w_0 R_2)^2 C_2} \right]} = 0$$

$$w_0^2 L_1 C_1 + \frac{w_0^2 L_1}{w_0^2 R_2^2 C_2} = 1$$

$$\begin{aligned} w_0^2 &= \frac{1}{L_1 C_1} - \frac{1}{C_1 C_2 R_2^2} \\ &= (2\pi \cdot 3.55 \text{ GHz})^2 \end{aligned}$$

• CHECK THE ASSUMPTION

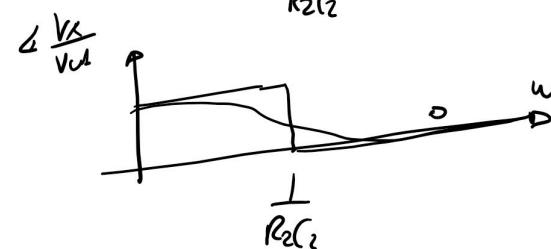
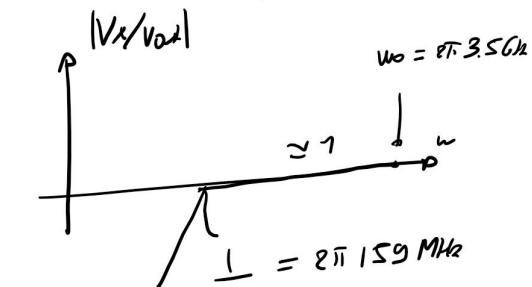
$$Q_T = \frac{1}{w_0 C_2 R_2} = 0.045 \ll 1$$



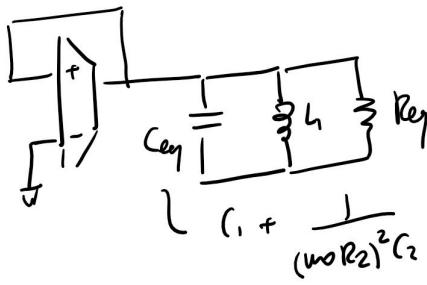
$$\frac{V_x}{R_1} \stackrel{\text{ion}}{\approx} V_{in}$$

$$\frac{V_x}{V_{in}} = \frac{R_2}{R_2 + \frac{1}{w_0 C_2}} = \frac{\omega C_2 R_2}{1 + \omega R_2 C_2}$$

$$V_x \approx V_{in} \text{ in AC}$$

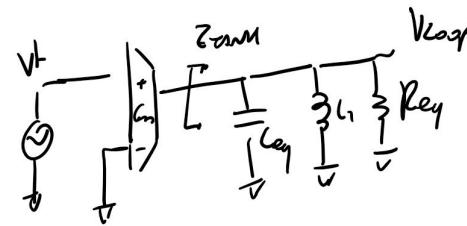


• CHECK THE STARTUP CONDITION



TO GUARANTEE STARTUP:

$$|G_{loop}(j\omega)| > 1$$

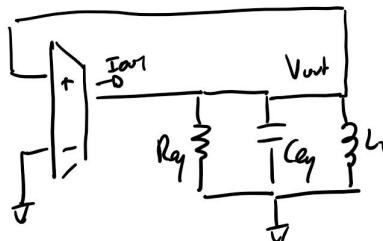
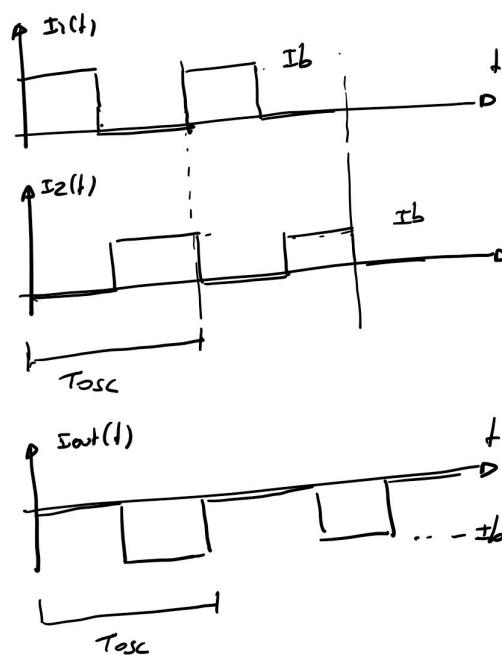
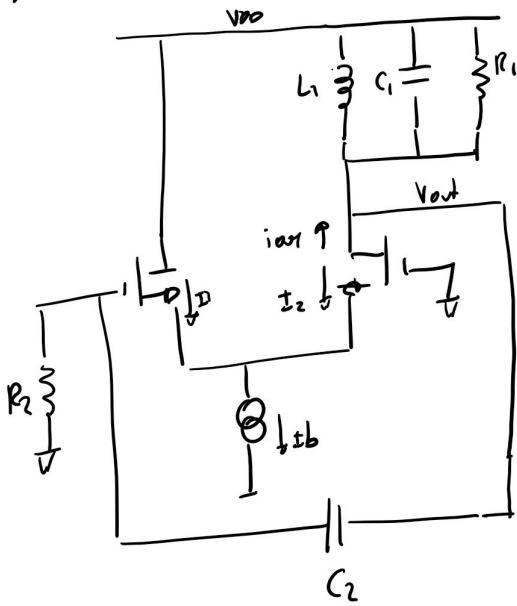


$$G_{loop}(j\omega) = G_m \cdot Z_{loop}(j\omega)$$

@ no resonance

$$|G_{loop}(j\omega)| = G_m \cdot R_{eq} = \frac{g_m}{2} (R_1 \parallel R_2) = 2.58 > 1$$

b) OSCILLATION AMPLIFIER CONSTRAINING MOS FULLY SWITCHING

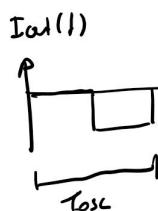


EVALUATE $V_{out}(t)$ WITH HARMONIC BALANCE

$$I_{a1}(t) \rightarrow I_{a1}(j\omega)$$

$$V_{out}(j\omega) = I(j\omega) Z_{loop}(j\omega)$$

$$V_{out}(j\omega) \rightarrow V_{out}(t)$$



SQUARE WAVE
 $D = 0.5$
 T_{osc}
 $A = -I_b$

$$I_{a1}(t) = y_0 + \sum_{n=1}^{+\infty} 2|y_n| \cos\left(\frac{2\pi}{T_{osc}} nt + \phi_n\right)$$

$$I_0 = y_0 = -\frac{I_b}{2}$$

$$y_n = \frac{1}{T_{osc}} \int_0^{T_{osc}} \{I_{a1}(t)\} dt = \frac{y_0}{T_{osc}}$$



Square wave

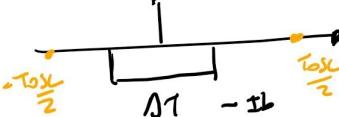
$$D = 0.5$$

Tosc

$$A = -Ib$$

$$I_{out\text{ truncated}}(t) = I_{out}(t) \text{ rect}\left(\frac{t}{T_{osc}}\right)$$

$$I_{out\text{ truncated}}(t) = -Ib \text{ rect}\left(\frac{t}{T_{osc}}\right)$$



$$I_{out}(t) = \gamma_0 + \sum_{n=1}^{+\infty} 2|\gamma_n| \cos\left(\frac{2\pi}{T_{osc}}nt + \phi\right)$$

$$\gamma_0 = \frac{1}{T_{osc}} \int_{-T_{osc}/2}^{T_{osc}/2} I_{out}(t) dt$$

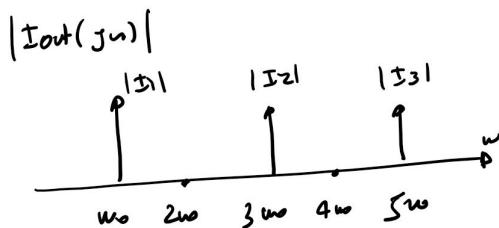
$$\gamma_0 = \frac{1}{T_{osc}} \int_{-T_{osc}/2}^{T_{osc}/2} -Ib \text{ rect}\left(\frac{t}{T_{osc}}\right) dt$$

$$= \frac{1}{T_{osc}} \Delta t (-Ib) \sin\left(\frac{n}{T_{osc}} \Delta t\right)$$

$$= -Ib D \frac{\sin(nD\pi)}{nD\pi}$$

$$\frac{\Delta t}{T_{osc}} = D$$

= avg - cycle

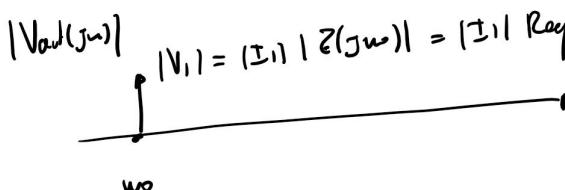
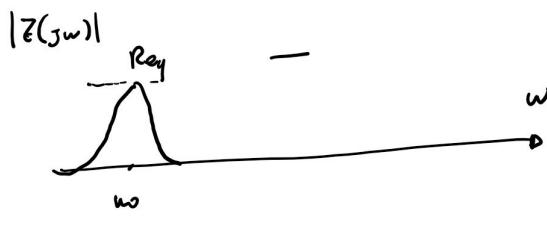


$$|I1| = 2|\gamma_1| = 2Ib \frac{1}{2} \frac{\sin(\pi/2)}{\pi/2} = \frac{2Ib}{\pi}$$

$$|I2| = 2|\gamma_2| = 2Ib \frac{1}{2} \frac{\sin(\pi)}{\pi} = 0$$

$$|I3| = 2|\gamma_3| = \frac{2}{3} \frac{Ib}{\pi}$$

$$|I4| = 0$$



$$V_{out}(t) = V_{out\text{dc}} + |V1| \cos(w_0 t)$$

|V1| = OSCILLATION AMPLITUDE

$$= \frac{2Ib}{\pi} \cdot Ray = \frac{2Ib}{\pi} (R_1 // R_2)$$

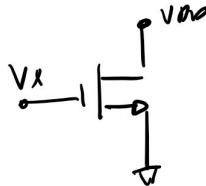
$$= 1,273 \text{ V}$$

c) OSCILATOR OPERATION REGION

M2

$$\begin{aligned} V_{out} &= V_{DD} + |V1| \cos(w_0 t) \\ V_d &> V_g - V_T \\ V_{DD} - V_1 &> -V_T \\ 1.25V - 1.273V &> -0.5V \\ -0.023V &> -0.5V \quad \text{OK} \end{aligned}$$

M1



$$V_d > V_g^{\text{max}} - V_T$$

$$V_g^{\text{max}} < V_{DD} + V_T$$

$$1.273V < 1.25V + 0.5V \quad \text{OK}$$

EXERCISE 7.3

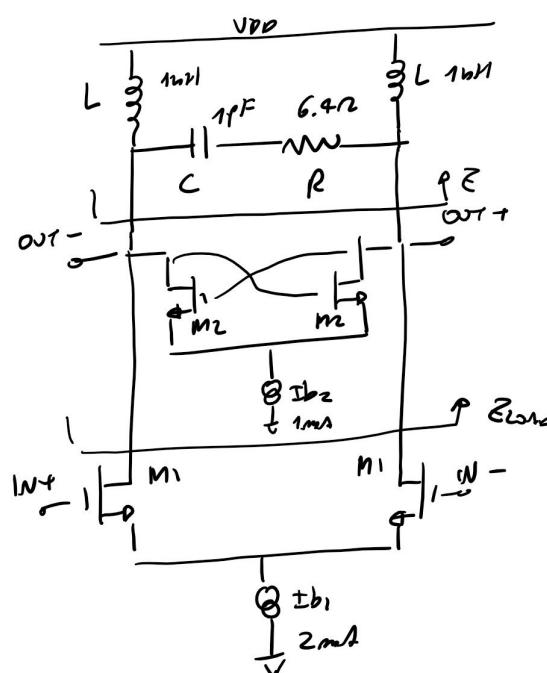
$$V_T = 0.5V$$

$$k = \frac{1}{2} \mu C_b = 0.1 \text{ mH/V}^2$$

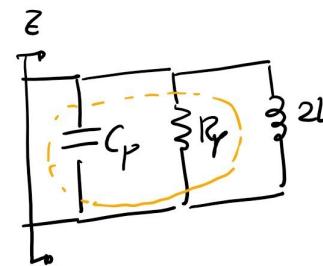
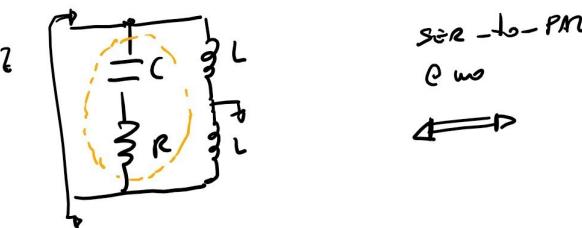
$$V_d = 2/3$$

$$g_m = 2\sqrt{k \left(\frac{V}{L}\right)_1 \frac{\pm b_1}{2}}$$

$$g_m = 2\sqrt{k \left(\frac{V}{L}\right)_2 \frac{\pm b_2}{2}}$$



a) RESONANT FREQUENCY OF Z_{LOAD}



$$\begin{cases} R_p = R(1 + Q_T^2) \\ \frac{1}{w_0 C_p} = \frac{1}{w_0 C} \left(1 + \frac{1}{Q_T^2}\right) \\ Q_T = \frac{1}{w_0 C R} = w_0 C_p R_p \end{cases}$$

WE DON'T KNOW w_0

HYPOTHESIS ON Q_T

$$\text{IF } Q_T \gg 1 \rightarrow w_0 \ll \frac{1}{CR} = 2\pi 24,87 \text{ GHz}$$

$$\text{IF } Q_T \ll 1 \rightarrow w_0 \gg \frac{1}{CR} = 2\pi 24,87 \text{ GHz}$$

Since $L \sim \mu H$ $C \sim pF$

$$f_0 \sim 6 \text{ GHz}$$

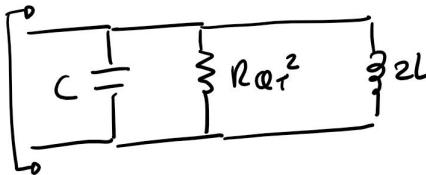
$$\text{so } Q_T \gg 1$$

if $Q_T \gg 1$

$$R_p = R(1 + Q_T^2) \approx RQ_T^2$$

$$\frac{1}{w_0 C_{2p}} = \frac{1}{w_0 C_2} \left(1 + \frac{1}{Q_T^2}\right) \approx \frac{1}{w_0 C_2} \Rightarrow C_{2p} \approx C_2$$

Z



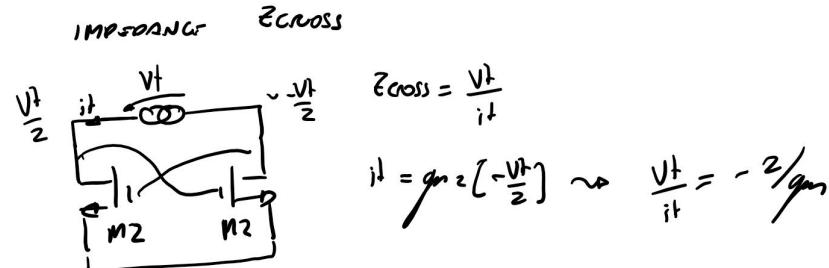
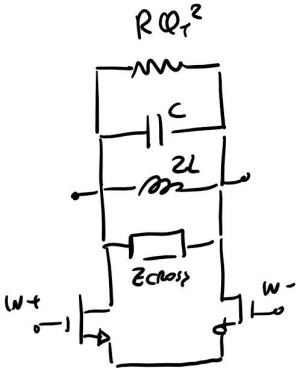
$$w_0 = \sqrt{\frac{1}{2LC}} \approx 2\pi 3,56 \text{ GHz}$$

• CHECK THE ASSUMPTION

$$Q_T = \frac{1}{w_0 R_C} = 7 > 7$$

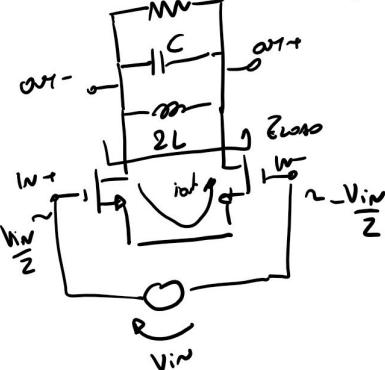
- SIZE $(W/L)_1$ AND $(W/L)_2$ (i) $\frac{V_{out,d}}{V_{in,d}} = 5 \text{ dB}$

(ii) $-3 \text{ dB BW} = 270 \text{ MHz}$

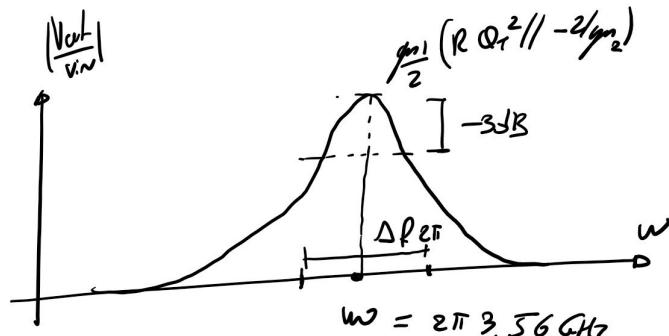


• DERIVE THE $\frac{V_{out,d}}{V_{in,d}}$

$$R_{eq} = RQ_T^2 // (-2/gm_2)$$



$$V_{out} = i_{out} Z_{load} \\ = Vt \frac{gm_1}{2} Z_{load}$$



$$\Delta f_BW = 270 \text{ MHz}$$

$$Q_{TMM} = \frac{f_0}{\Delta f} = 13,18$$

$$Q_{TMM} = w_0 C_{eq} \approx C_{eq} = 589,46 \text{ pF}$$

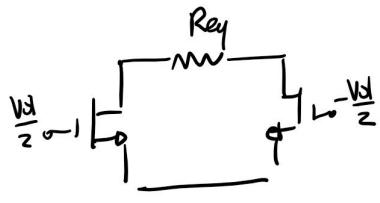
$$gm_2 = 2 \left[\frac{1}{RQ_T^2} - \frac{1}{R_{eq}} \right] = 2,98 \text{ mA/V} \\ = 2 \sqrt{\frac{1}{2} \mu C_{ox} \left(\frac{W}{L} \right)_2 \frac{I_b}{2}}$$

$$\left(\frac{W}{L}\right)_2 = 45$$

$$= RQ_T^2 // Z_{cross}$$

$$Z_{cross} = -2/gm_2$$

(i) Cmos 5dB gain

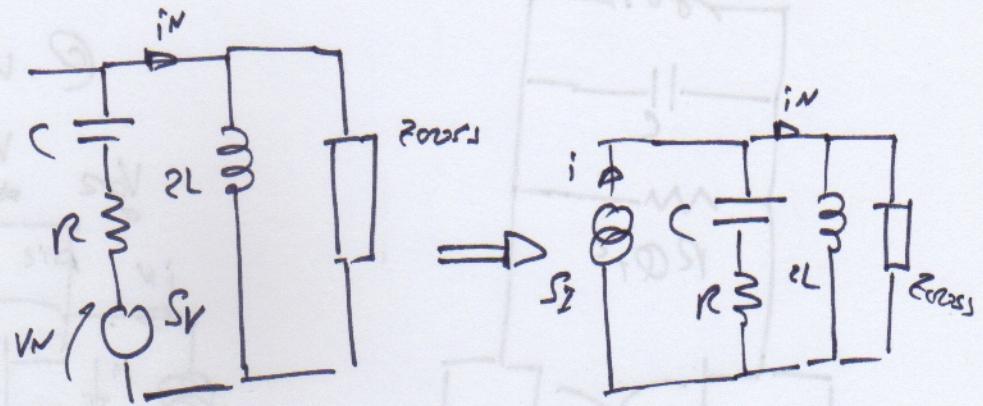
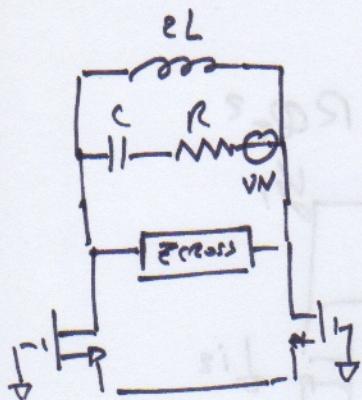


$$\frac{V_{out}}{V_{in}} = \frac{g_m}{2} R_{load} = 5 \text{ dB} = 10^{\frac{5}{20}}$$

∴ $g_m = 6.03 \text{ mS}$

$$= 2 \sqrt{\frac{1}{2} \mu_s C_{os} \left(\frac{V}{I} \right)_1 \frac{I_b}{2}}$$
$$\left(\frac{V}{I} \right)_1 = 91$$

b) TRANSISTOR AND RESISTOR NOISE @ w₀

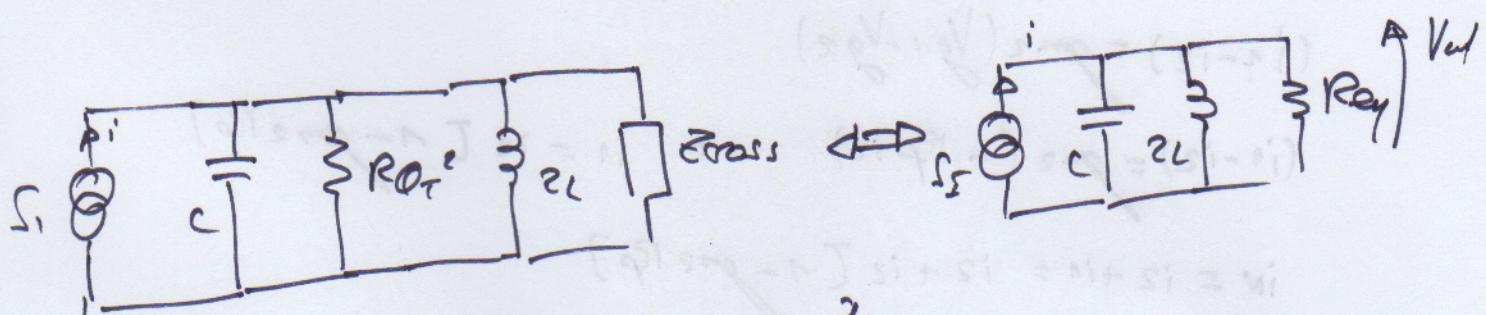


$$i = \frac{V_N}{R + \frac{1}{\omega C}} = \frac{V_N \omega C}{(1 + \omega C R)}$$

$\rho = j_{w_0}$

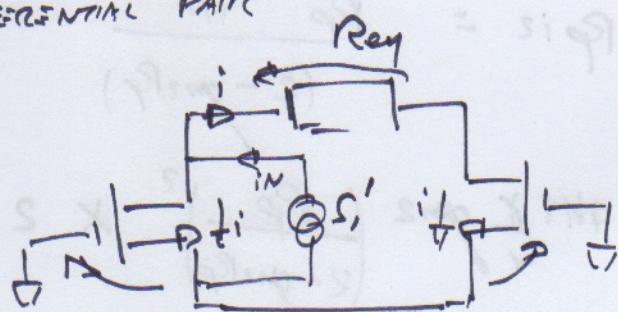
$$S_I = 4kT R \left| \frac{j \omega C}{1 + j \omega C R} \right|^2 = \frac{4kT R}{R} \frac{\omega^2 C^2}{1 + \omega^2 R^2 C^2} = \frac{4kT}{R} \frac{\frac{1}{Q_T^2}}{\left(1 + \frac{1}{Q_T^2}\right)^2}$$

$$S_I = \frac{4kT}{R(1 + Q_T^2)} = \frac{4kT}{R\rho}$$



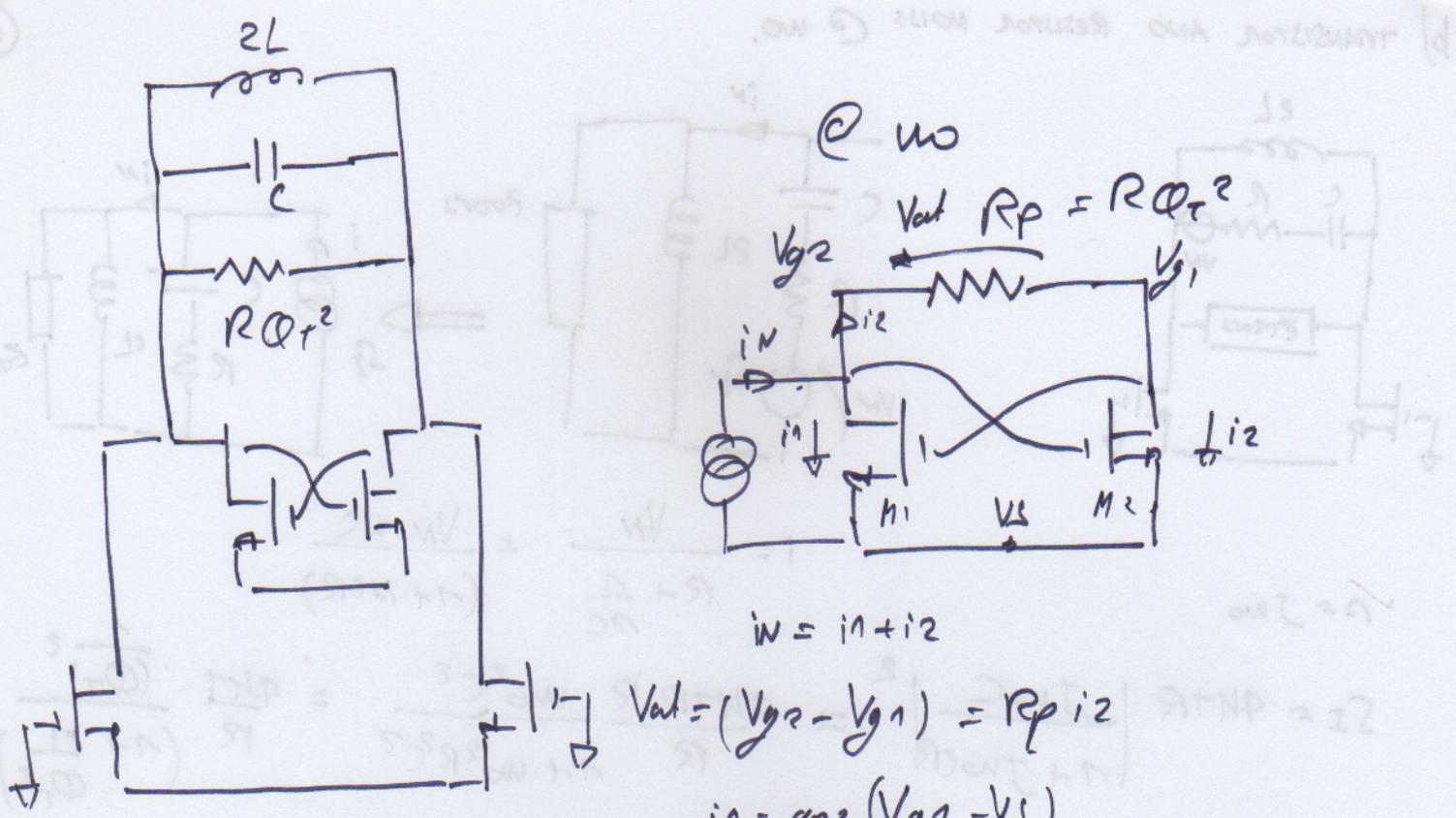
$$S_{V_{RAT}} = \frac{4kT}{R Q_T^2} (R_{eq})^2 = \left(4,284 \frac{nV}{\sqrt{Hz}}\right)^2 \quad T = 300K$$

• CONTRIBUTION OF MAIN DIFFERENTIAL PAIR



$$\begin{aligned} & @ w_0 \quad S_{V_{RAT}} = \frac{1}{4} \left[\frac{4kT}{\alpha} \mu_n \right] R_{eq}^2 \times 2 \\ & = \left(3,4022 \frac{nV}{\sqrt{Hz}}\right)^2 \quad \text{↳ BOTH TRANSISTORS} \end{aligned}$$

$$\left\{ \begin{array}{l} V_{out} = i R_{eq} \\ i_N = 2i \end{array} \right. \Rightarrow V_{out} = \frac{i_N}{2} R_{eq}$$



$$(i_1 - i_2) = g_m (V_{g1} - V_{g2})$$

$$(i_1 - i_2) = g_m (-R_P i_2) \rightarrow i_1 = i_2 [1 - g_m R_P]$$

$$i_N = i_2 + i_1 = i_2 + i_2 [1 - g_m R_P]$$

$$= i_2 [1 + 1 - g_m R_P]$$

$$V_{out} = R_P i_2 = \frac{R_P i_N}{(2 - g_m R_P)}$$

@no

$$S_{Vout} = qH1 \propto g_m^2 \left(\frac{R_P}{2 - g_m R_P} \right)^2 \times 2 \quad \leftarrow \text{Both transistors}$$

$$= k \cdot 55 \frac{V}{\sqrt{Hz}}^2$$

$$S_{Vout, tot} = S_{at}^{RES} + S_{at}^{MAIN} + S_{at}^{CROSS} = \left(6.036 \frac{V}{\sqrt{Hz}} \right)^2$$

c) $\left(\frac{K}{L}\right)_2$ LIMIT?

$R_{eq} > 0$ to prevent oscillations

$$R_{eq} = \frac{2R\Omega^2}{2 - R\Omega^2 g_m} > 0$$

$$2R\Omega^2 > 0$$

$$2 - R\Omega^2 g_m > 0$$

$$g_m < \frac{2}{R\Omega^2}$$

$$g_m = 2 \sqrt{\frac{1}{2} m C_{ox} \left(\frac{K}{L}\right)_2 \frac{I_b e}{2}}$$

$$\left(\frac{K}{L}\right)_2 < 204$$