


RF Circuits Design

L11



Oscillators

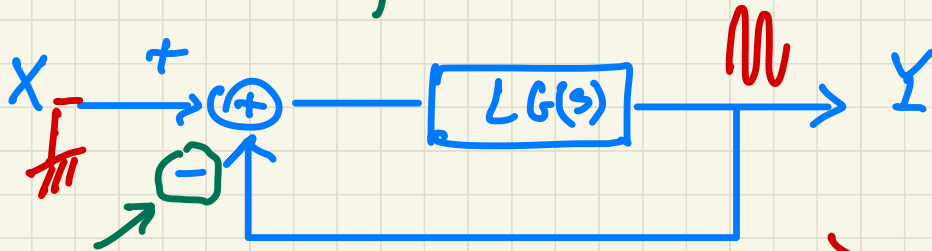
ex. VCO or CCO
XO

electrically - tuned oscillators
crystal oscillators

Mathematical models :

- 1) Feedback system
- 2) Negative resistance

Feedback system :

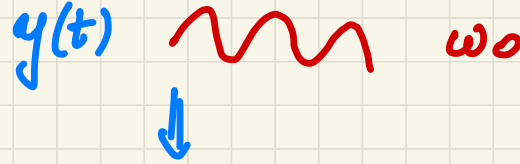


$$\frac{Y}{X} = \frac{LG(s)}{1 \oplus LG(s)}$$

a) negative feedback

→ autonomous system

$$\frac{Y}{X} = \frac{LG(s)}{1 + LG(s)}$$



Oscillation condition: $Y(j\omega_0) \neq 0$ with $X(j\omega_0) = 0$

$$0 \neq \frac{Y(j\omega_0)}{X(j\omega_0)} = \frac{LG(j\omega_0)}{1 + LG(j\omega_0)} \rightarrow \infty$$

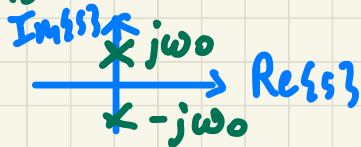
0

\Leftrightarrow

$$LG(j\omega_0) = -1$$

ω_0 is such that the loop at that freq. is equal to -1

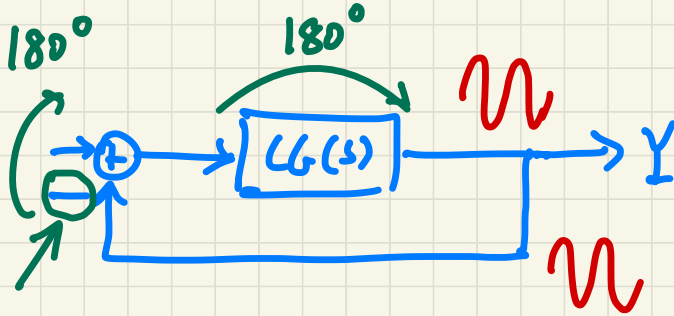
$\Leftrightarrow s = j\omega_0$ is a solution of $LG(s) = -1$
 ($j\omega_0$ is a pole of the closed-loop system)



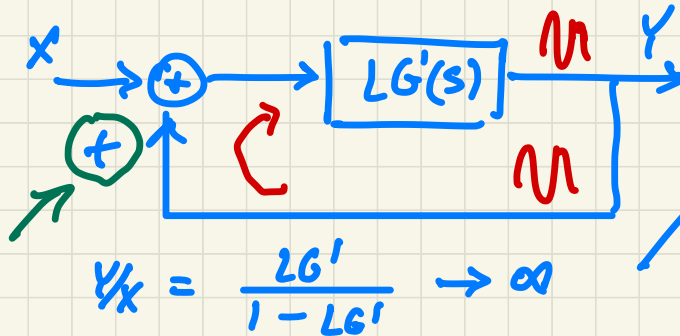
$$LG(j\omega_0) = -1 \quad \Leftrightarrow$$

$$\begin{cases} |LG(j\omega_0)| = 1 \\ \angle LG(j\omega_0) = 180^\circ \end{cases}$$

Barkhausen's conditions



b) positive feedback



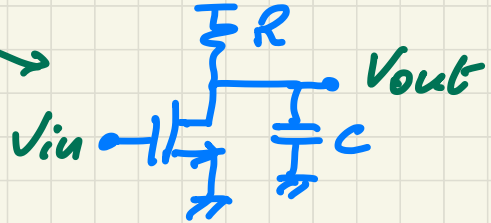
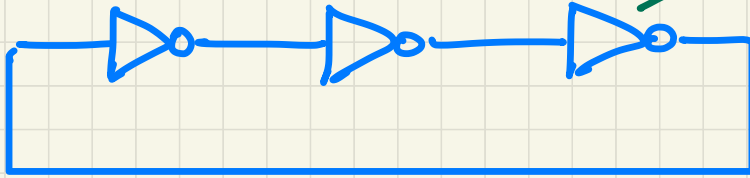
Oscillation condition

$$LG'(j\omega_0) = 1$$

$$\begin{cases} |LG'(j\omega_0)| = 1 \\ \angle LG'(j\omega_0) = 0^\circ \end{cases}$$

EXAMPLES

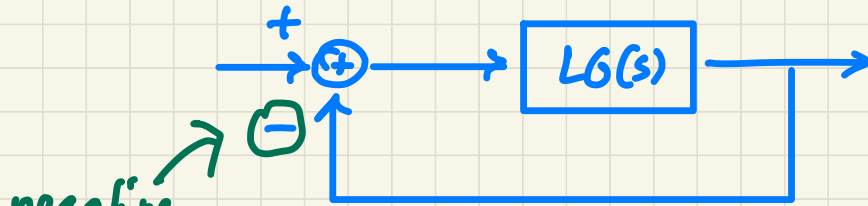
- RC oscillator (e.g. ring oscillator)



$$\frac{V_{out}}{V_{in}} = - \frac{G}{1 + s\tau}$$

$$G > 0$$

(simple linear model)



$$LG(s) = \frac{G^3}{(1 + s\tau)^3}$$

Oscillation condition : 1. $\angle LG(j\omega_0) = \pi$

$$4 \frac{G^3}{(1+j\omega_0\tau)^3} = \pi \quad ;$$

$$\underbrace{4G^3}_0 - 3 \cdot \arctan(\omega_0\tau) = \pi \quad ; \quad \arctan \omega_0\tau = -\frac{\pi}{3}$$

$$\omega_0\tau = \sqrt{3} \quad ; \quad \underline{\omega_0} = \frac{\sqrt{3}}{\tau} \quad \leftarrow \tau = RC$$

$$2. \quad |LG(j\omega_0)| = 1 \quad ;$$

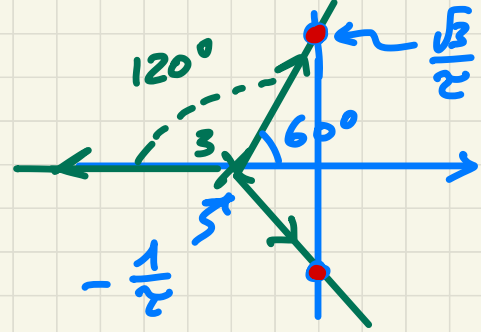
$$\frac{G^3}{[1 + \underbrace{(\omega_0\tau)^2}_{\sqrt{3}}]^{3/2}} = 1$$

$$\Rightarrow \frac{G^3}{(1+3)^{3/2}} = 1 \quad ; \quad G^3 = 2^3 \quad ; \quad \underline{G} = 2$$

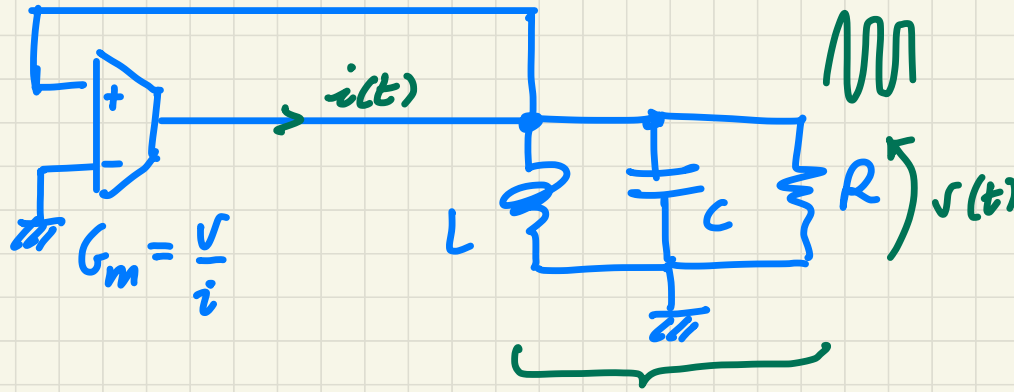
Root locus

$$LG(s) = \frac{G^3}{(1+s\tau)^3}$$

$$s = \pm j\omega_0 = \pm j\frac{\sqrt{3}}{\tau} \quad \leftarrow$$



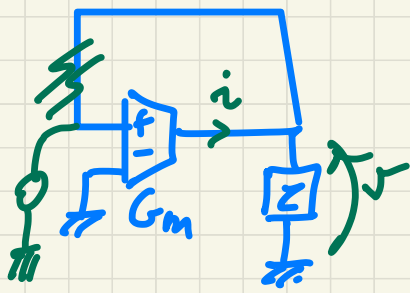
• LC oscillator



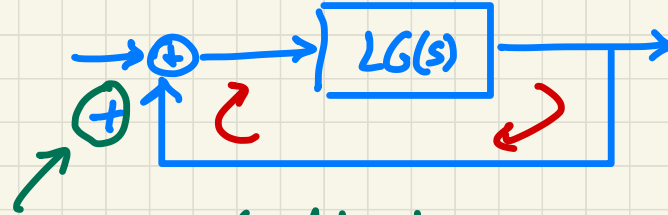
$$Z(s) = R \cdot \frac{1}{1^2 + s\omega_n/Q + \omega_n^2}$$

$$\omega_n = \frac{1}{\sqrt{LC}}$$

$$Q = \omega_n RC$$



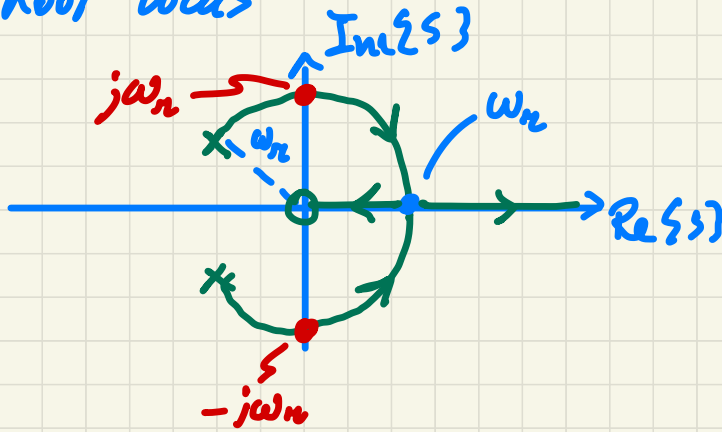
\Rightarrow



to highlight, it's positive feedback

$$LG(s) = G_m \cdot Z(s) = G_m R \cdot \frac{1 \omega_n / Q}{\underbrace{s^2 + 1 \omega_n / Q + \omega_n^2}_{2 \text{ complex poles}}}$$

Root locus



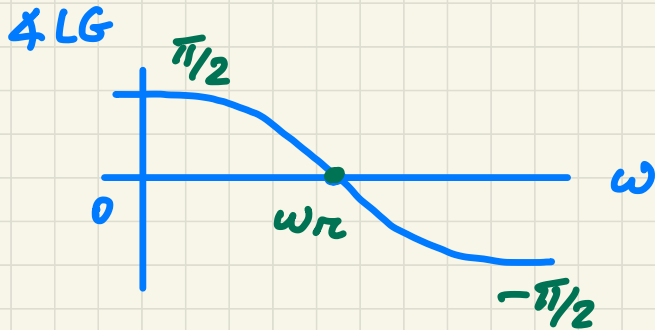
$\Rightarrow 1 = \pm j\omega_n = \pm j\omega_0$
frequency of oscillation
is identical to the resonance
frequency

$$1. \quad \angle LG(j\omega_0) = 0 ;$$

$$\angle \frac{j\omega_0\omega_r/Q}{-\omega_0^2 + j\omega_0\omega_r/Q + \omega_r^2} = \frac{\pi}{2} - \arctan \frac{\omega_0\omega_r/Q}{\omega_r^2 - \omega_0^2} = 0$$

\Downarrow

$$\underline{\omega_0 = \omega_r}$$



$$2. \quad |LG(j\omega_0)| = 1 ;$$

$$\Rightarrow \underline{G_m R = 1}$$

$$\frac{G_m R \cdot \cancel{\omega_0 \omega_r / Q}}{\sqrt{\underbrace{(\omega_r^2 - \omega_0^2)^2}_0 + \cancel{\left(\frac{\omega_0 \omega_r}{Q}\right)^2}}} = 1$$

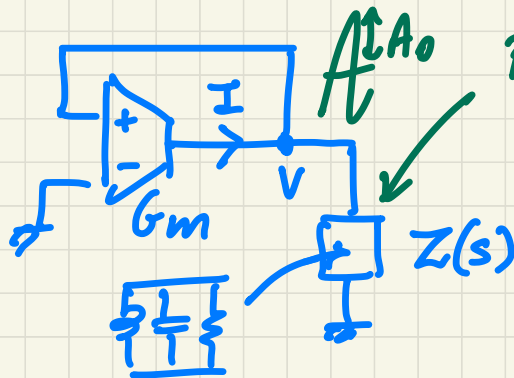
Negative Resistance Model

oscillation condition :

balance between dissipated power and the active power

$$\underline{Z_a(j\omega_0) + Z(j\omega_0) = 0}$$

example :



Power dissipated

$$\frac{1}{2} \cdot \frac{A_o^2}{R} =$$

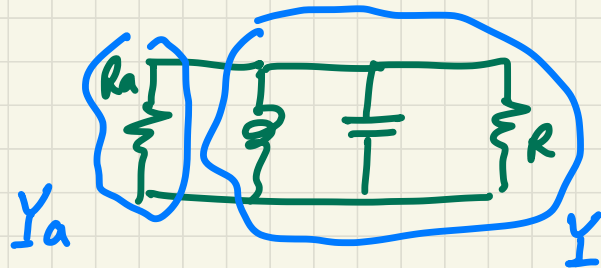
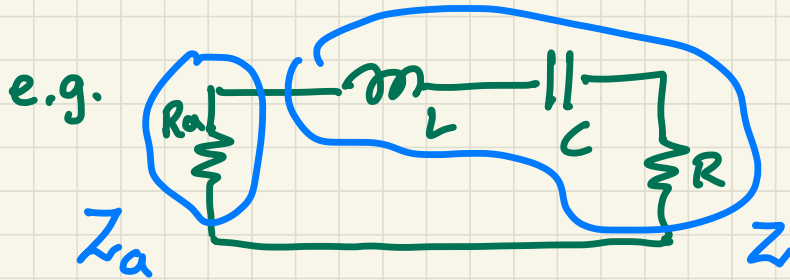
Active power

$$\frac{1}{2} G_m \cdot A_o^2$$

$$G_m = \frac{1}{R}; G_m R = 1$$

$$\underline{Z_a(j\omega_0) = -Z(j\omega_0)}$$

$$\begin{cases} \operatorname{Re} \{ Z_a(j\omega_0) \} = -\operatorname{Re} \{ Z(j\omega_0) \} \\ \operatorname{Im} \{ Z_a(j\omega_0) \} = -\operatorname{Im} \{ Z(j\omega_0) \} \end{cases}$$



$$\begin{cases} R_a = -R \\ \omega_0 L + \frac{1}{\omega_0 C} = 0 ; \quad \omega_0 = \frac{1}{\sqrt{LC}} \end{cases}$$