

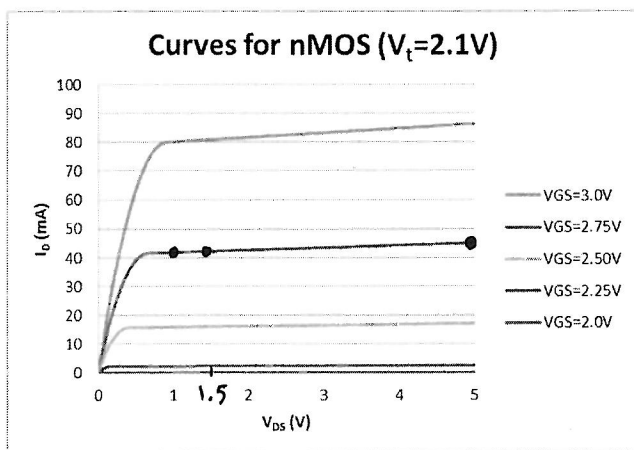
Name Solution

Section _____

1. (50 points) You wanted to characterize your BS170 nMOS transistor ($V_t=2.1V$) using the curve tracer and generated the family of curves and data points below. Using the trace corresponding to an overdrive voltage (V_{ov}) of 650 mV when $V_{DS}=1.5V$, determine (in no particular order) the

- a. transconductance parameter (k)
 b. channel length modulation (λ)
 c. channel length resistance (r_o)

- d. early voltage (V_A)
 e. transconductance gain (g_m)



$$V_{GS} - V_t = V_{ov} = 650 \text{ mV} \quad \therefore V_{GS} = 2.75 \text{ V}$$

$$V_{DS} = 1.50 \text{ V}$$

$V_{DS} \text{ (V)}$	$I_D \text{ (mA)}$				
	$V_{GS}=3.0V$	$V_{GS}=2.75V$	$V_{GS}=2.50V$	$V_{GS}=2.25V$	$V_{GS}=2.0V$
0	0	0	0	0	0
0.5	63.68	39.19	15.68	2.20	5.1E-10
1	80.14	41.80	15.83	2.23	1.01E-09
1.5	80.93	42.21	15.99	2.25	1.51E-09
2	81.71	42.62	16.14	2.27	2.01E-09
2.5	82.50	43.03	16.30	2.29	2.51E-09
3	83.28	43.44	16.45	2.31	3.01E-09
3.5	84.07	43.85	16.61	2.34	3.51E-09
4	84.86	44.26	16.76	2.36	4.01E-09
4.5	85.64	44.67	16.92	2.38	4.51E-09
5	86.43	45.08	17.07	2.40	5.01E-09

You may find the following equations useful. Write your final answers below.

$V_A = \frac{1}{\lambda}$	$g_m = kV_{OV} = 2 \frac{I_D}{V_{OV}}$	$r_o = \frac{1}{\lambda I_D}$	$g_m = \left. \frac{dI_D}{dV_{GS}} \right _{V_{DS}}$	$\text{slope} = \frac{1}{r_o} = \left. \frac{dI_D}{dV_{DS}} \right _{V_{GS}}$
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k 200 $\text{mA} \cdot \text{V}^{-2}$	λ 0.0197 V^{-1}	r_o 1.2 $\text{k}\Omega$	V_A 51.5 V	g_m 129.9 $\text{mA} \cdot \text{V}^{-1}$
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$$\frac{1}{r_o} = \left. \frac{dI_D}{dV_{DS}} \right|_{V_{GS}} = \frac{(45.08 - 41.8) \text{ mA}}{(5 - 1) \text{ V}} = 0.82 \text{ mS} \quad \therefore r_o = 1.2 \text{ k}\Omega$$

$$r_o = \frac{1}{\lambda I_D}, \quad \lambda = \frac{1}{r_o I_D} = \frac{1}{(1.2 \text{ k}\Omega)(42.21 \text{ mA})} = 0.0197 \approx 0.02 \text{ V}^{-1}$$

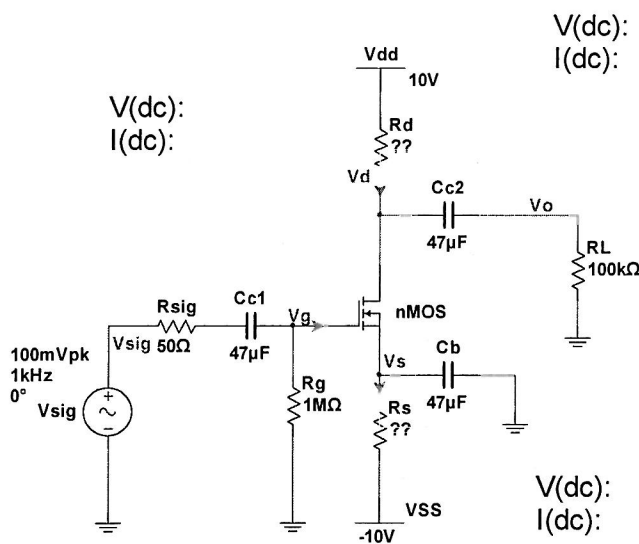
$$V_A = \frac{1}{\lambda} = \frac{1}{0.0197} = 51.5 \text{ V}$$

$$g_m = \frac{2 I_D}{V_{OV}} = \frac{2(42.21 \text{ mA})}{0.650 \text{ V}} = 129.9 \text{ mA/V}$$

$$k = g_m / V_{OV} = 200 \text{ mA/V}^2$$

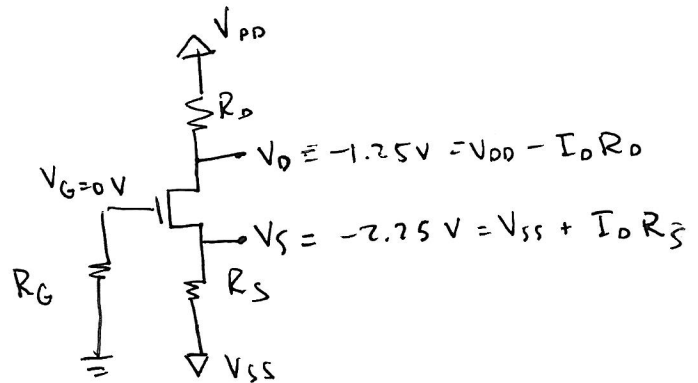
2. (50 points) Assume you found $\lambda=0.02 \text{ V}^{-1}$ and $k=200 \text{ mA}\cdot\text{V}^{-2}$ from part 1 where $V_t = 2.1 \text{ V}$. You then biased your transistor using the 2-supply biasing scheme below for $V_{OV} = 650 \text{ mV}$ and $V_{DS} = 1.5 \text{ V}$.

2a. (10 points) Draw the large signal model and determine the required R_d and R_s .



$V(\text{dc})$:
 $I(\text{dc})$:

Disconnect the capacitors



$V(\text{dc})$:
 $I(\text{dc})$:

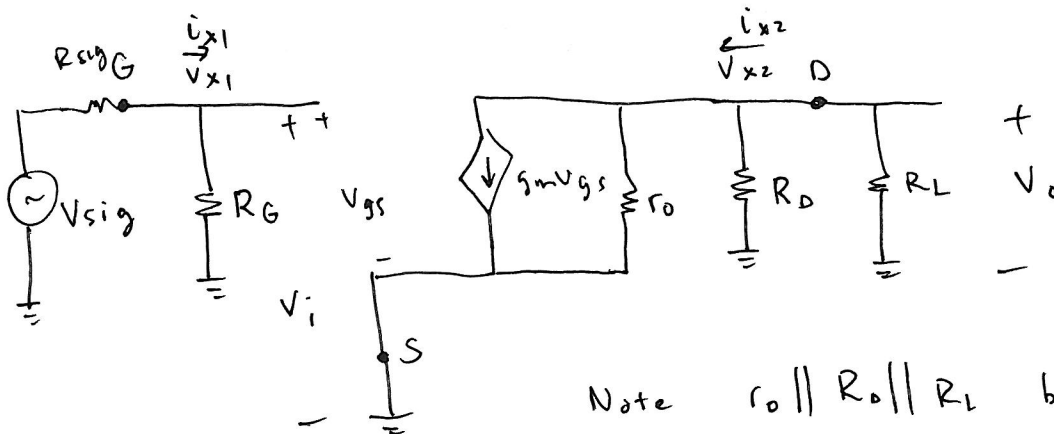
$$V_{GS} = 2.75 \text{ V} \Rightarrow V_S = -2.75 \text{ V}$$

$$V_{DS} = 1.5 \text{ V} \Rightarrow V_D = -2.75 + 1.5 = -1.25 \text{ V}$$

$$R_D = \frac{V_{DD} - V_D}{I_D} = \frac{10 - (-1.25)}{42.21 \text{ mA}} = \boxed{266.5 \Omega = R_D}$$

$$R_S = \frac{V_S - V_{SS}}{I_D} = \frac{-2.75 - (-10)}{42.21 \text{ mA}} = \boxed{171.8 \Omega = R_S}$$

2b. (10 points) Draw the hybrid- π small signal model. Ensure you account for r_o .



Note $r_o \parallel R_D \parallel R_L$ because source to ground

2c. (20 points) Determine relationships for A_{vo} , A_v , R_i , and R_o .

$$A_{vo} = \frac{v_o}{v_i} \bigg|_{R_L = \infty} = \frac{-g_m v_{gs} r_o \parallel R_D}{v_{gs}} = -g_m r_o \parallel R_D$$

$$A_v = \frac{v_o}{v_i} \bigg|_{R_L} = \frac{-g_m v_{gs} r_o \parallel R_D \parallel R_L}{v_{gs}} = -g_m r_o \parallel R_D \parallel R_L$$

$$R_i = \frac{v_{x1}}{i_{x1}} = R_G$$

$$R_o = \frac{v_{x2}}{i_{x2}} = r_o \parallel R_D$$

2d. (10 points) Calculate A_{vo} , A_v , R_i , and R_o from your relationships above.

$$A_{vo} = -(129.9 \text{ mA/V})(1.2 \text{ k} \parallel 266.5) = -28.32 \text{ V/V}$$

$$A_v = -(129.9 \text{ mA/V})(1.2 \text{ k} \parallel 266.5 \parallel 100 \text{ k}) = -28.26 \text{ V/V}$$

Note $A_v \approx A_{vo}$ because $R_L \gg R_D, r_o$

$$R_i = 1 \text{ M}\Omega$$

$$R_o = (1.2 \text{ k} \parallel 266.5) = 218.1 \Omega$$

ECE 321 – Quiz 4 (part 2)

MOSFET Characterization

Fall 2015

Name Solution

Section _____

ACADEMIC SECURITY. This Quiz is released at the beginning of the next lesson.

INTEGRITY: Your honor is extremely important. This academic security policy is designed to help you succeed in meeting academic requirements while practicing the honorable behavior our country rightfully demands of its military. Do not compromise your integrity by violating academic security or by taking unfair advantage of your classmates.

Collaboration Policy: No collaboration allowed. This is individual effort. You may not seek help from other cadets, nor from any faculty members. All help must be properly documented.

Permissible References: Any texts or online resources.

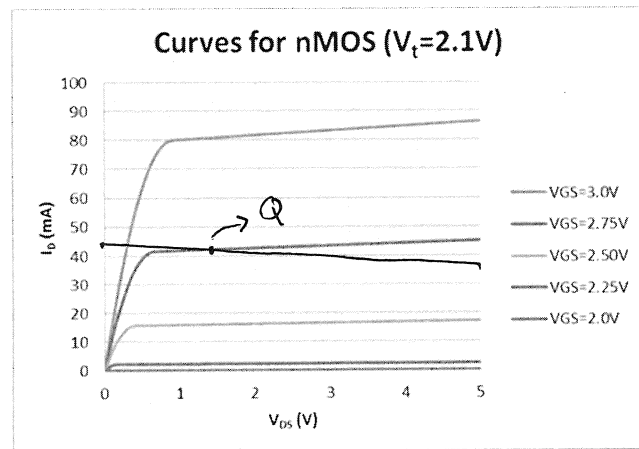
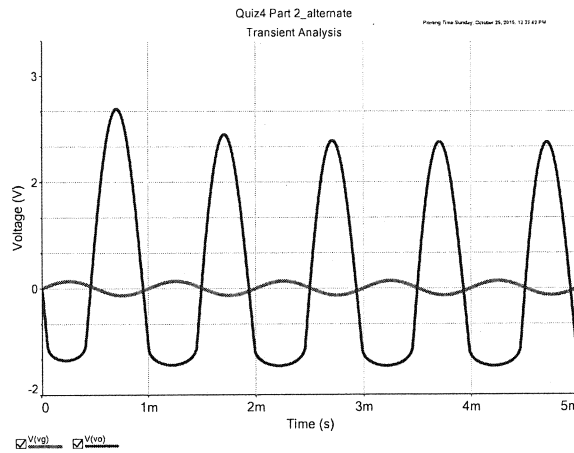
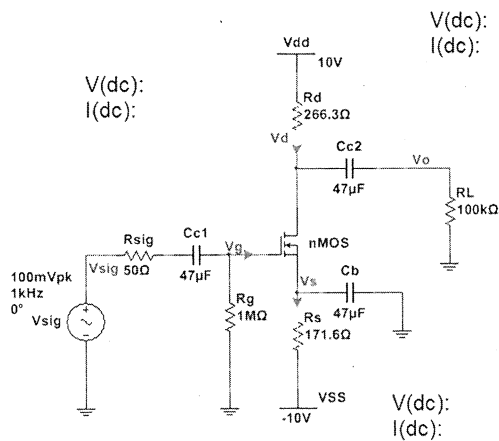
Grading: This is part 2 of the quiz.

Total _____

Grade _____

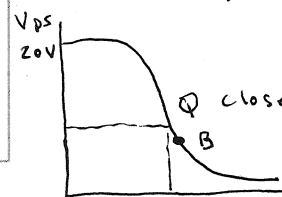
Documentation:

3. (50 points) Assume you found the correct bias point and simulated your circuit with below results showing v_o and v_i (i.e. v_g). Comment on the appropriateness of the bias point and the measures you would take to ensure undistorted voltage swing in both directions.



You may find the figure to the left helpful in your analysis and explanation.

the output is swinging too low into triode. See that $V_{DS} = 1.5V$ which is a small fraction of the total supply voltage ($V_{DD} - V_{SS} = 20V$).



So we need

to raise V_{DS} to allow our output

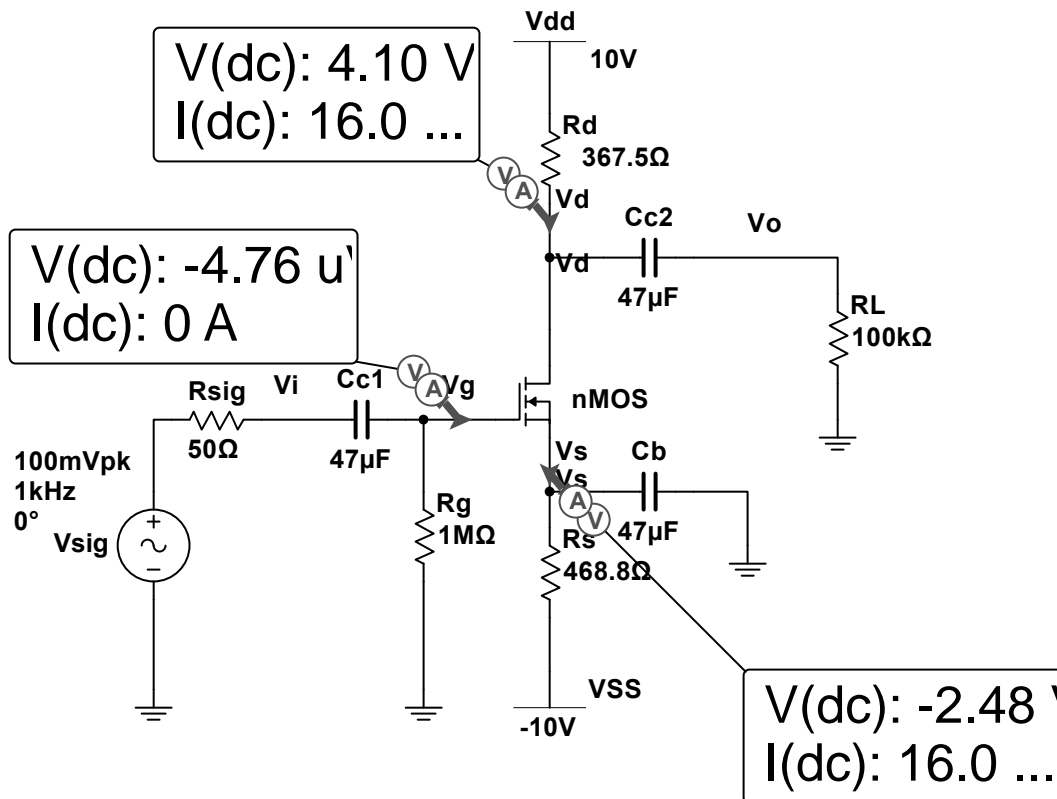
V_{GS} to swing both ways.

you could raise V_D by lowering R_D but this may adversely affect your gain remember $A_v = -g_m R_D \parallel R_L \parallel r_o = -28.3 \text{ V/V}$ from part 2d. Alternatively you could lower I_D so less voltage drop across R_D, R_S resulting in a bigger V_{DS} . Suppose we use the $V_{GS} = 2.5V$ curve. This is $V_{ov} = 0.4V$ giving $I_D = \frac{1}{2} (200)(0.4)^2 = 16mA$. Then $R_S = \frac{-2.5 - (-10)}{16mA} = 468.8\Omega$

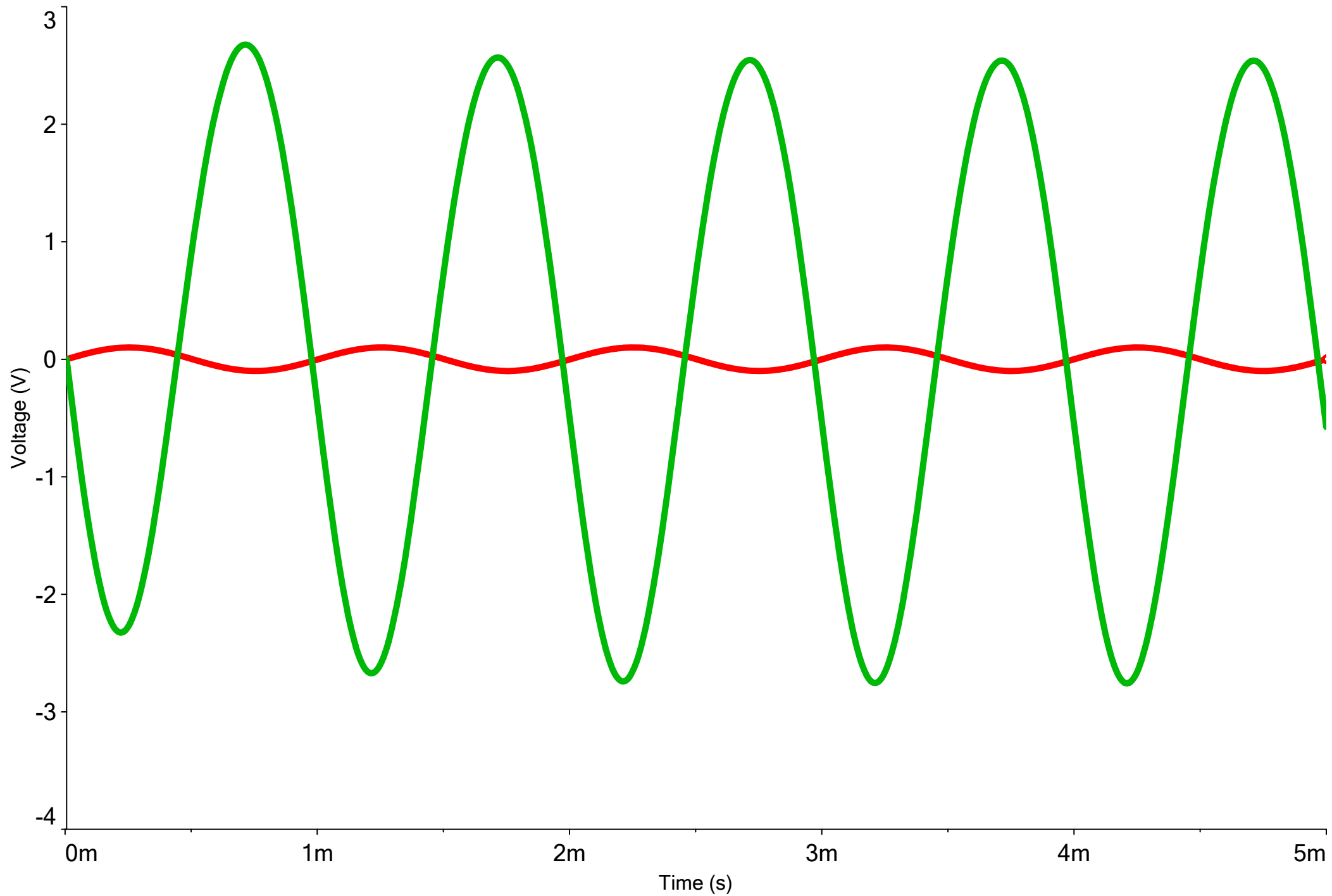
now $g_m = 200(0.4) = 80 \text{ mA/V}$ which is a 38% reduction so would anticipate A_v to also drop by ~38%. So we could now raise R_D by 38% giving $R_D = 266.3 \times 1.38 = 367.5\Omega$ so now $V_{DS} = 6.62V$

$$V_D = V_{DD} - I_D R_D = 4.12V \quad V_S = V_{SS} + I_D R_S = -2.5V$$

$V_{DS} = 6.62V$ which is now a bigger fraction of the total supply and we preserved our gain A_v



Transient

☒ $V(vi)$ ☒ $V(vo)$