

## Control Systems

Continuous linear time-invariant systems (LTI)

"easy"  $\begin{cases} q'(t) = Aq(t) + Bf(t) \\ y(t) = Cq(t) + Df(t) \end{cases}$  ← inhomogeneous system of ODEs

Continuous linear time variant systems

"harder"  $\begin{cases} q'(t) = A(t)q(t) + B(t)f(t) \\ y(t) = C(t)q(t) + D(t)f(t) \end{cases}$

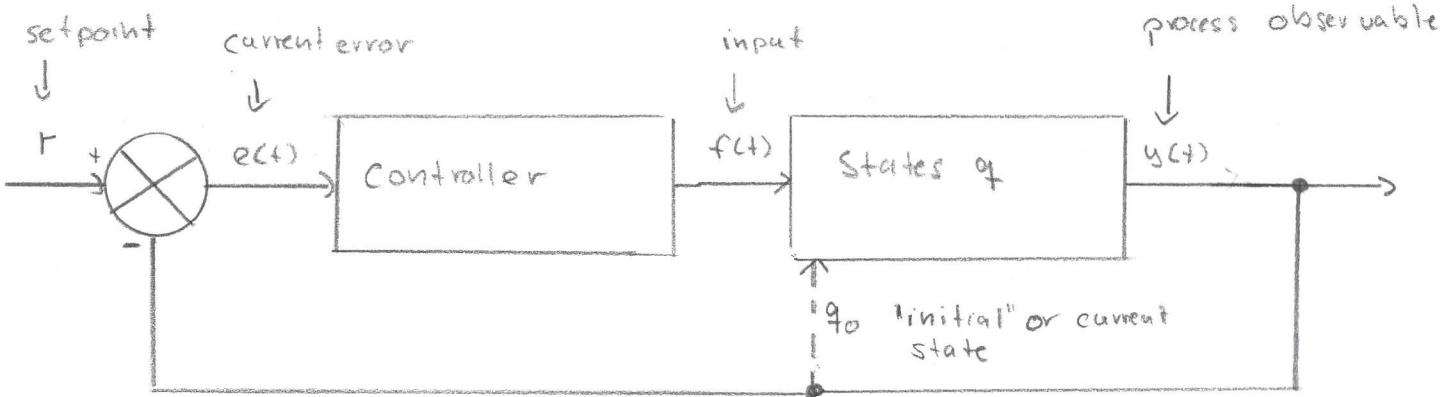
↑ linear but coefficients are time dependent

Non-linear systems

→ no matrix form

"hard"  $\begin{cases} q_1' = aq_1 + bq_2 + cq_1q_2 \\ q_2' = -dq_1^3 + e \end{cases}$

$$\begin{aligned} \dot{q} &= Aq + Bf \\ y &= Cq + Df \end{aligned} \quad \left. \begin{array}{l} \text{LTI system} \end{array} \right\}$$

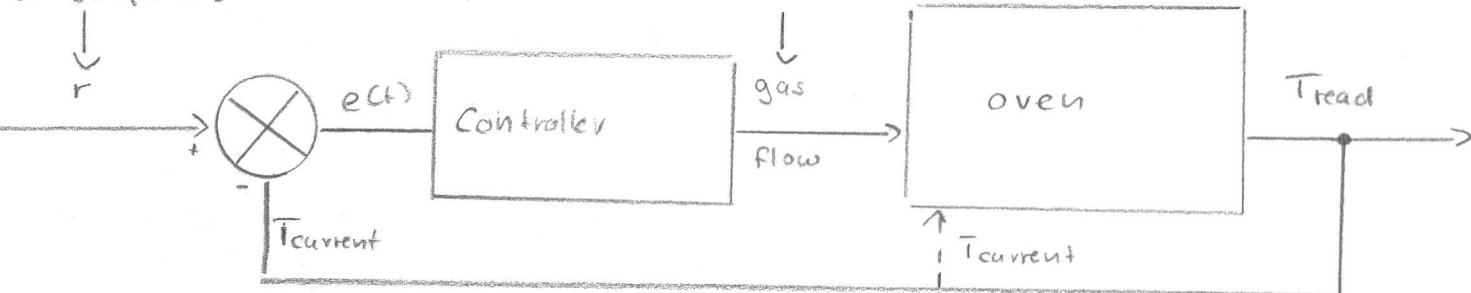


Control problem: how to set input  $f(t)$  to drive the system  
from current state  $q_0$  to desired observable  $y(t)$

Example : Oven

MP: Manipulated Variable (value pos/voltage)

SP: Set point



physical model requires knowledge  
of V vs. gas flow rate

Discrete Operation ( $r = 400$ )

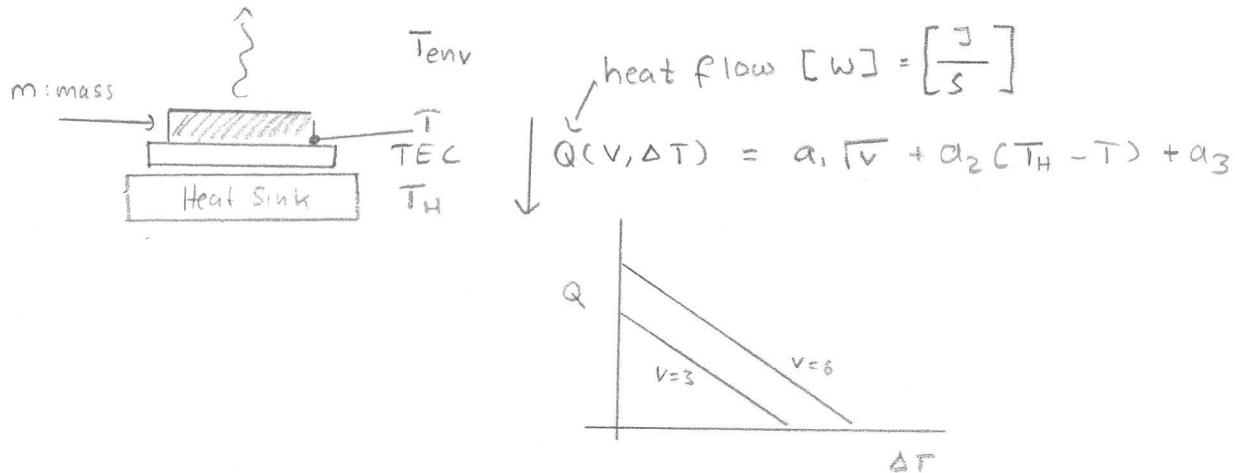
t	Tread	e(t)	MV
0	150	250	10 V
10 s	180	220	9 V
20 s	200	200	1 V
30 s	1	1	12 V

Model (LTI) if available or observe

- know e(t) current
- know e(t) past
- no knowledge of future

"Controller is mapping e(t) to MV"

## Concrete Example : Cold Stage



How does  $T$  change?

$$\frac{dT}{dt} = k(T_{env} - T) = \underbrace{\frac{Q}{mc_f}}_{\text{heat capacity}} \quad \left. \begin{array}{l} \text{Newton's law} \\ \text{of cooling} \end{array} \right\} \text{physical model} \quad \left. \begin{array}{l} \text{heat flux} \\ + \text{storage} \end{array} \right\} \text{of system}$$

Convert model to State Space Form

$$\frac{dT}{dt} = kT_{env} - kT - \frac{a_1}{mc} \bar{V} + \frac{a_2}{mc} T - \frac{a_3}{mc} + \frac{a_2}{mc} \bar{T}_H$$

$$\frac{dT}{dt} = (-k + \frac{a_2}{mc})T - \frac{a_1}{mc} \bar{V} - \frac{a_2}{mc} \bar{T}_H - \frac{a_3}{mc} + kT_{env}$$

state variable(s) :  $\vec{q} = \vec{T}$

inputs :  $\vec{f} = V, T_{env}, \bar{T}_H$

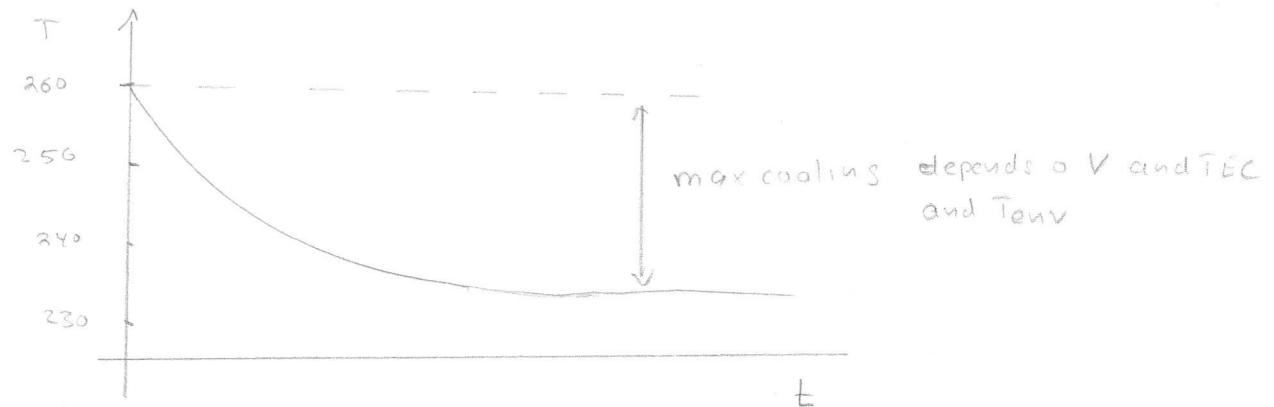
output(s) :  $\vec{y} = \vec{T}$

$$q = T \quad q' = \dot{T}$$

$$q' = \underbrace{\left[ -k + \frac{a_2}{mc} \right]}_A \vec{q} + \underbrace{\left[ -\frac{a_1}{mc} \quad -\frac{a_2}{mc} \quad -\frac{a_3}{mc} \quad k \right]}_B \begin{bmatrix} \bar{V} \\ \bar{T}_H \\ 1 \\ T_{env} \end{bmatrix} + f$$

$$y = \underbrace{\begin{bmatrix} 1 \end{bmatrix}}_C \vec{q} + \underbrace{\begin{bmatrix} 0 & 0 & 0 & 0 \end{bmatrix}}_D \vec{f}$$

## Simple Solution : ODE solver



```
function odes(q, p, t)
```

```
    A = p[1]
```

```
    B = p[2]
```

```
    f = p[3]
```

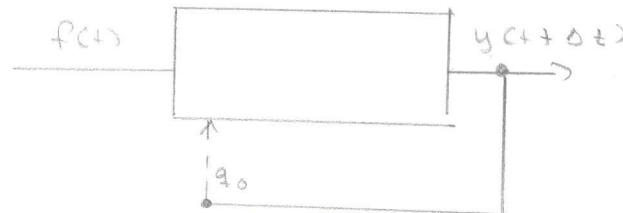
```
    return Aq + Bf
```

```
end
```

# Discrete Event Simulation

$$\dot{q} = Aq + Bf$$

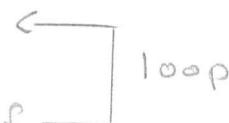
$$y = Cq + Df$$



- initialize ODE at  $t = t_0$  with IVP  $q_0, f$

- integrate to  $t + \Delta t$

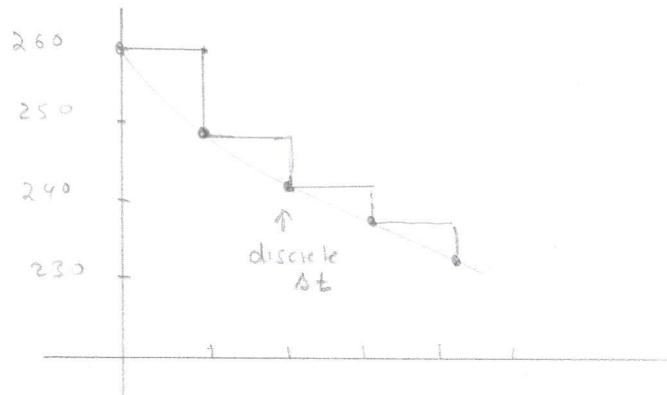
- evaluate solution  $\rightarrow$  build time array ; possibly update  $f$



<u><math>t</math></u>	<u><math>q</math></u>	<u><math>f</math></u>	$\Delta t = 20s$
0	1	1	
20	1	1	
40	1	1	
60	1	1	

$\left. \begin{array}{l} \\ \end{array} \right\} f \text{ may change each } \Delta t$

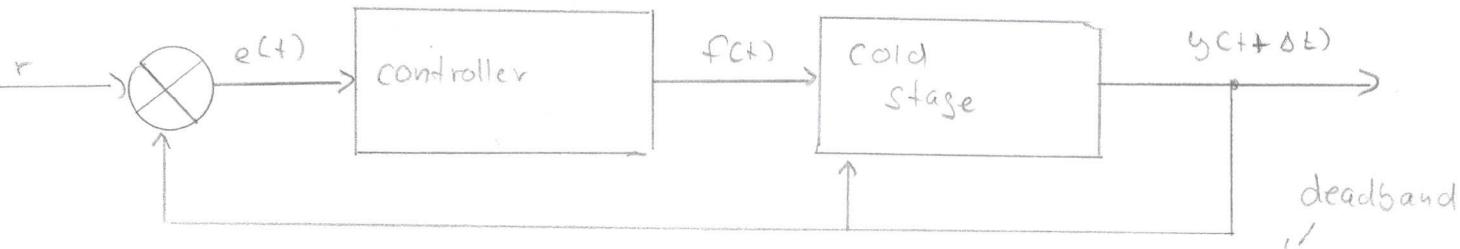
## Discrete Event Example



# On/Off Control

$r = 300$

$+1 = \text{cooling} \rightarrow \text{TEC} = +12V$   
 $-1 = \text{heating} \rightarrow -12V$



$$e(t) = r - y(t)$$

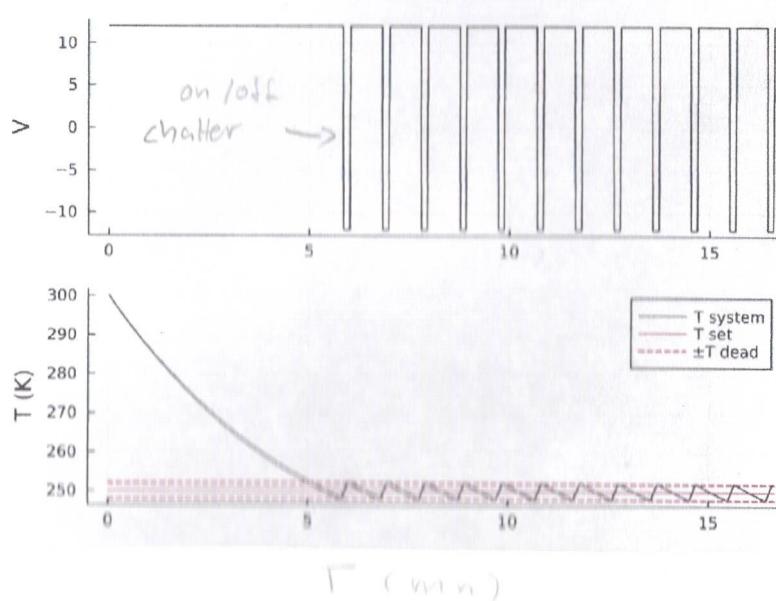
$$f(t) = \begin{cases} +1 & e(t) + d < 0 \wedge e(t + \Delta t) + d > 0 \\ -1 & e(t) - d > 0 \wedge e(t - \Delta t) - d < 0 \\ \text{no change} & \text{else} \end{cases}$$

deadband

past error

t	y(t)	e(t)	e(t)+d	e(t)-d	f(t)	
0	303	-3	-1	-5	+1	cooling
1	295	+5	+7	+3	-1	heating
2	299	+1	+3	-1	-1	(no change) "dead band"
3	304	-4	-2	-6	+1	cooling
4						
5						
6						

## Example Cold Stage



- solution oscillates
- control within  $\pm d$
- prone to chatter

dead band

## Proportional Control

$$g(t) = k_p \frac{\overbrace{r - y(t)}^{\text{error}}}{\overbrace{s}^{\text{span}}} + p_0$$

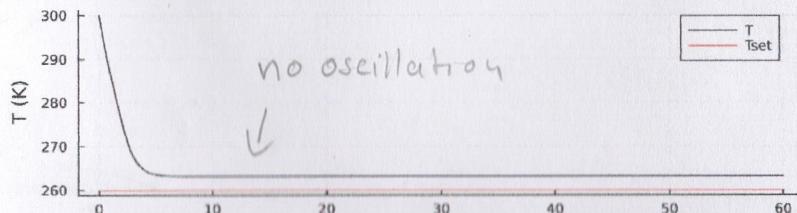
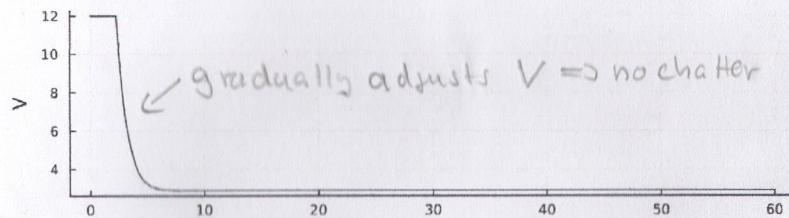
↑ proportional gain      ↑ span      ↑ offset

$$f(t) = \begin{cases} -1 & g(t) < -1 \\ 1 & g(t) > 1 \\ g(t) & \text{else} \end{cases}$$

$T_{set}$



$T_{span}$



Span: 100% output  
if not within  $\pm s$

} offset error can not be eliminated

## Proportional Integral Control (PI)

$$e(t) = \frac{r - y(t)}{g(t)}$$

$$g(t) = k_p e(t) + k_i$$

## Integral over past error

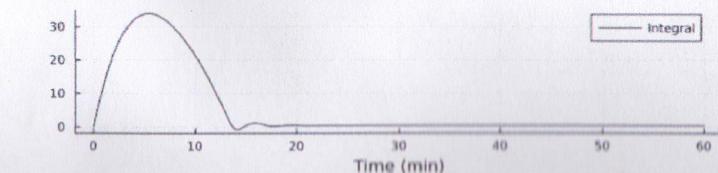
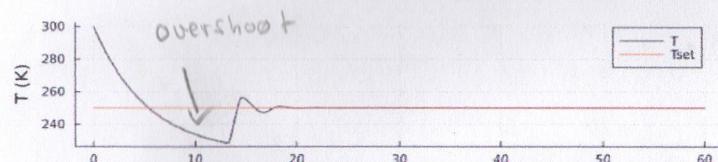
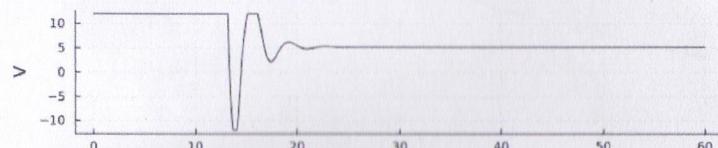
$$\int_0^t e(\tau) d\tau$$

$$f(t) = \begin{cases} -1 & g(t) < -1 \\ 1 & g(t) > 1 \\ g(t) & \text{else} \end{cases}$$

Solution is sensitive  
to integral gain

$T_{set}$  250

$T_{span}$  20  $K_p$  1.7  $K_i$  0.1



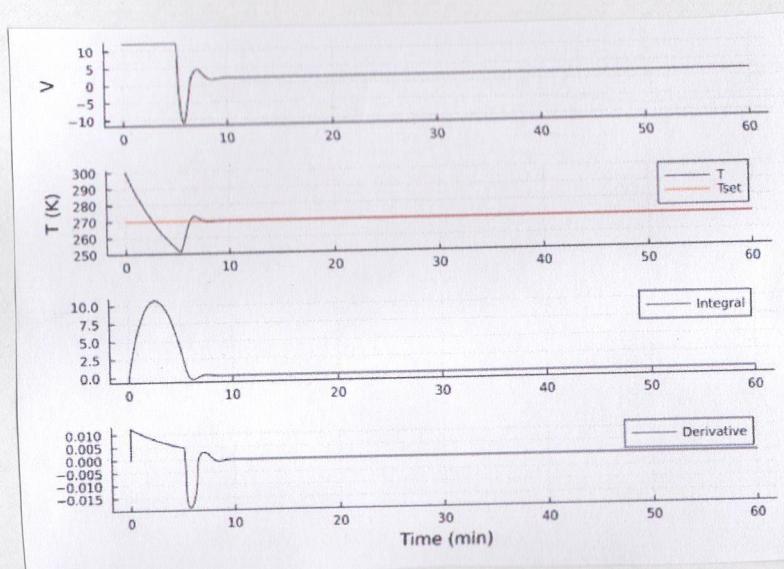
- oscillating solution that generally converges but not always converges
- no offset error

## Proportional - Integral - Derivative (PID) (cont'd)

same as PI except

$$g(t) = k_p e(t) + k_i \int_0^t e(\tilde{t}) d\tilde{t} + k_d \frac{de(t)}{dt}$$

↑  
derivative gain



D-term can

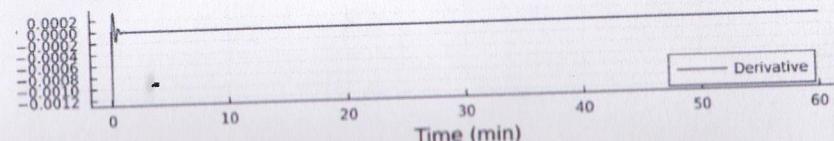
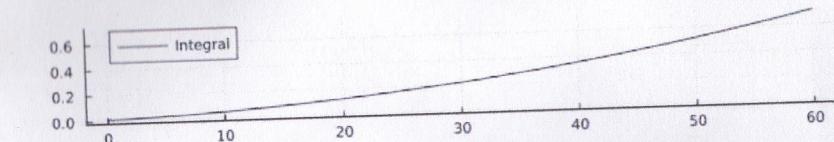
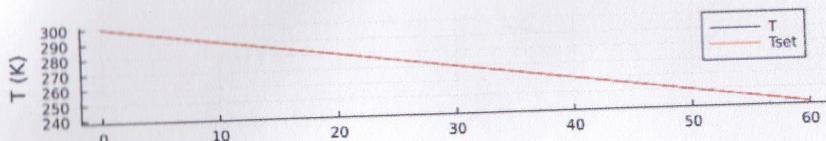
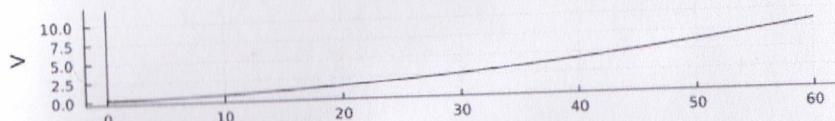
- increase convergence rates
- be susceptible to noise amplification

# Trajectory Control

$c_r [K \text{ min}^{-1}]$

$T_{span}$

$K_p$   $K_i$   $K_d$



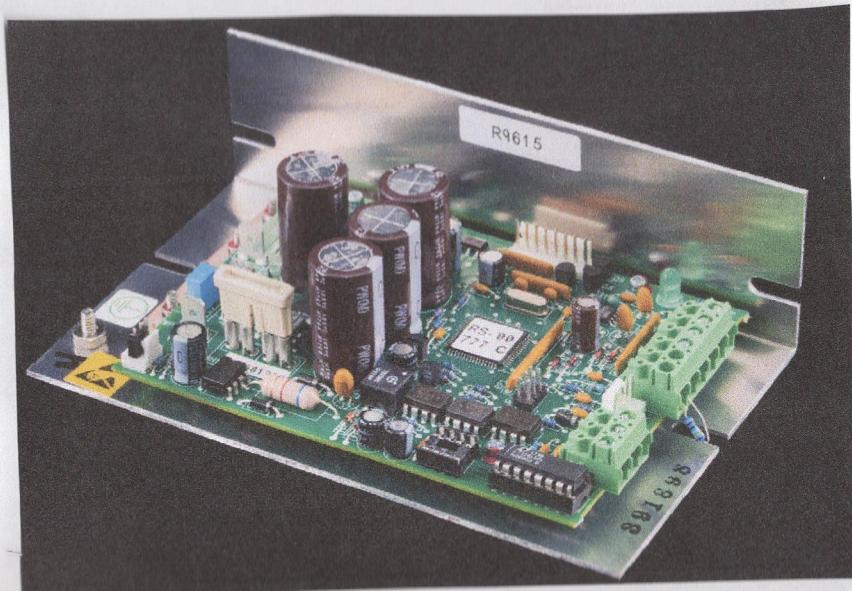
Setpoint  $r$  is treated as  $r(t)$

e.g. cooling rate  $c_r = \frac{dr}{dt} = \text{const}$

- avoids oscillations
- avoids overshoot
- controlled way to steer system

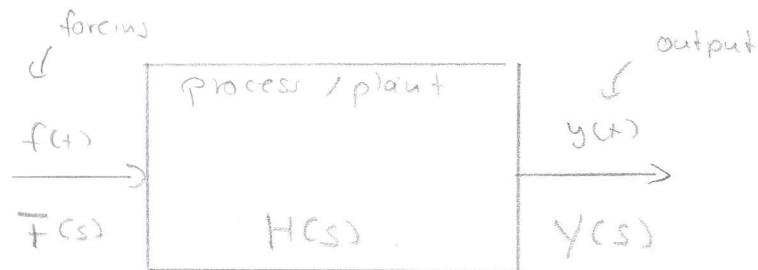
## Hardware Controllers (special domain)

### PID Temperature Controller



- programmable  $T_{set}$ ,  $T_{span}$ ,  $K_p$ ,  $K_i$ ,  $K_d$  and spans
- reads temp (thermistor)
- modulates input V to ± output V for thermoelectrics

# Formal Reasoning about Control Systems



$$\begin{aligned} \mathcal{L}f'' &= s^2\bar{f} + sf(0) - f'(0) \\ \mathcal{L}f' &= s\bar{f} - sf(0) \\ \mathcal{L}af &= a\bar{f} \end{aligned}$$

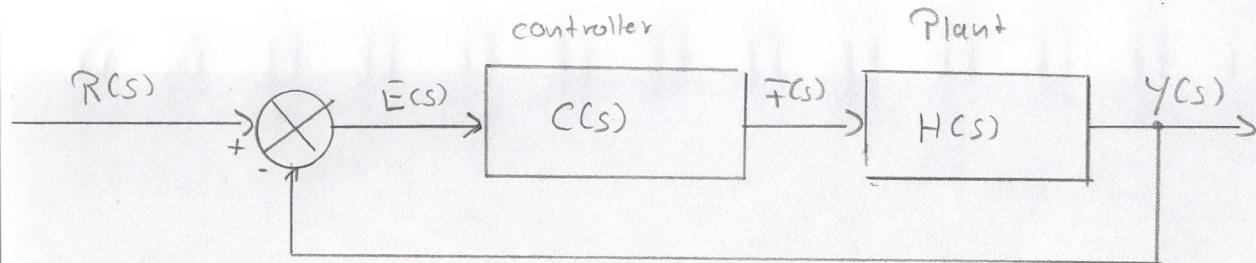
$$f(t) = a_n y^{(n)} + \dots + a_1 y' + a_0 y \quad (\text{zero initial conditions})$$

$$\mathcal{L} \bar{f}(s) = (a_n s^n + \dots + a_1 s + a_0) Y(s)$$

$$H(s) = \frac{Y(s)}{F(s)} = \frac{1}{a_n s^n + \dots + a_0}$$

↑  
transfer function

# Closed Loop Control System



$$\frac{Y(s)}{R(s)} = \underbrace{\frac{C(s)H(s)}{1 + C(s)H(s)}}_{\text{use this for stability analysis}}$$

↑  
System transfer function

Controller

$$f(t) = k_p e(t) + k_i \int_0^t e(\tau) d\tau + k_d \frac{de(t)}{dt}$$

$$C(s) = k_p + \frac{k_i}{s} + k_d s$$

use this for stability analysis

- roots in numerator  $\Rightarrow$  zeros  $\Rightarrow$  stabilize system
- roots in denominator  $\Rightarrow$  poles  $\Rightarrow$  destabilize system
- can add noise filter to  $C(s)$

$\mathcal{L}^{-1}(Y(s)) \Rightarrow y(t)$  : analytical analog to simulation

## Controllability

$$\vec{q}' = A\vec{q} + Bf$$

→ reach any state  $\vec{q}'$  using control system       $\vec{q}'$  is  $n \times 1$

$$R = [B \ AB \ A^2B \ \dots \ A^{n-1}B]$$



controllability matrix

→ system is controllable if R is full row rank, i.e.  $\text{rank}(R) = n$

↳ derive through  $\mathcal{L}\vec{q}'$