Z-transform:

In mathematics and signal processing, the **Z-transform** converts a discrete-time signal, which is a sequence of real or complex numbers, into a complex frequency-domain (**z-domain** or **z-plane**) representation.

- F.T. [xct)] =
$$\int_{-\infty}^{\infty} xct$$
 = $\int_{-\infty}^{-i\omega t} dt$
- D.F.T. [x(n)] = $\sum_{n=-\infty}^{\infty} x(n)$ e $\sum_{n=-\infty}^{\infty} x(n)$

The Z-transform can be defined as either a one-sided or two-sided transform.

Bilateral Z-transform

The bilateral or two-sided Z-transform of a discrete-time signal x[n] is the formal power series X(z) defined as

$$X(z)=\mathcal{Z}\{x[n]\}=\sum_{n=-\infty}^{\infty}x[n]z^{-n}$$

where n is an integer and z is, in general, a complex number:

$$z = Ae^{j\phi} = A \cdot (\cos \phi + j\sin \phi)$$

where A is the magnitude of z, j is the imaginary unit, and ϕ is the *complex argument* (also referred to as *angle* or *phase*) in radians.

Unilateral Z-transform

Alternatively, in cases where x[n] is defined only for $n \geq 0$, the single-sided or unilateral Z-transform is defined as

$$X(z)=\mathcal{Z}\{x[n]\}=\sum_{n=0}^{\infty}x[n]z^{-n}.$$

- Relation Let. DET & Z.T.

-
$$\chi(z) = \frac{\omega}{2\pi} \chi(z) = \frac{\omega}{2\pi$$

| Inverse
$$Z = tounytoun$$
 | $= | \chi(z) | = DFT [\chi(n) | x^n]$ | $= | Jz | = | Jz |$

Region of Convergence

Region of Convergence is the range of complex variable Z in the Z-plane. The Z-transformation of the signal is finite or convergent. So, ROC represents those set of values of Z, for which X(Z) has a finite value.

Properties of ROC

- ROC does not include any pole.
- For right-sided signal, ROC will be outside the circle in Z-plane.
- For left sided signal, ROC will be inside the circle in Z-plane.
- For stability, ROC includes unit circle in Z-plane.
- For Both sided signal, ROC is a ring in Z-plane.
- For finite-duration signal, ROC is entire Z-plane.

Example

Let us find the Z-transform and the ROC of a signal given as $x(n)=\{7,3,4,9,5\}$, where origin of the series is at 3.

Solution - Applying the formula we have -

$$X(z) = \sum_{n=-\infty}^{\infty} x(n)Z^{-n}$$

$$= \sum_{n=-1}^{3} x(n)Z^{-n}$$

$$= x(-1)Z + x(0) + x(1)Z^{-1} + x(2)Z^{-2} + x(3)Z^{-3}$$

$$= 7Z + 3 + 4Z^{-1} + 9Z^{-2} + 5Z^{-3}$$

ROC is the entire Z-plane excluding Z = 0, ∞

Problem based on ROC in z-transform Problem 01: Determine z-transform and their Roc. of the following signal. $\chi[n] = \begin{cases} 1, 2, 3, 4, 5 \end{cases}$

Solution:
$$\chi[n] = \{ 1, 2, 3, 4, 5 \}$$

$$\chi[n] = \chi(z) = \sum_{n=-\infty}^{\infty} \chi(n) \cdot z^{-n}.$$

$$\chi(z) = \sum_{n=0}^{4} \chi(n) \cdot z^{-n}.$$

$$\chi(z) = \sum_{n=0}^{4} \chi(n) \cdot z^{-n}.$$

$$\chi(z) = \chi(z) \cdot \chi(z) \cdot z^{-1} + \chi(z) \cdot z^{-2} + \chi(z) \cdot z^{-2} + \chi(z) \cdot z^{-3} + \chi(z) \cdot z^{-4}.$$

$$\chi(z) = 1 + 2z^{-1} + 3z^{-2} + 4z^{-3} + 5z^{-4}$$

$$X(z) = 1 + 2z^{-1} + 3z^{-2} + 4z^{-3} + 5z^{-4}$$
. Keeda

 $X(z) = 1 + \frac{2}{z} + \frac{3}{z^2} + \frac{4}{z^3} + \frac{5}{z^4}$

at $z = 0$, $X(z) = 0$.

At $z = 0$, $X(z) = 1 + 0 + 0 + 0 + 0 = 1$.

The ROC of $X(z)$ is available over the entire region of z -plane, except $z = 0$.

Problem 02:-

Determine z-transform and their Roc of the following discrete time signal,

$$\chi[n] = \{1, 2, 3, 4, 5\}$$



Solution: - x[n] = 31,2,3,4,53 By the definition of z-transform, $z[\chi(n)] = \chi(z) = Z^{\infty} \chi(n).z^{-n}$ n= -4 to 0. $\therefore \times (z) = \overline{z}^{\circ} \pi(n) \cdot \overline{z}^{\eta}.$ = $\chi(-4) \cdot z^4 + \chi(-3) \cdot z^3 + \chi(-2) \cdot z^2 +$ $\chi(-1) z' + \chi(0).$ $\times(z) = 1 \cdot z^4 + 2z^3 + 3z^2 + 4z + 5$

$$X(z) = z^4 + 2z^3 + 3z^2 + 4z + 5$$

$$R \cdot 0.C.$$
At $z = 0$,
$$X(z) = 5$$
At $z = \infty$

$$X(z) = \infty$$
The Roc of $X(z)$ is available over entire region of z -plane, except $z = \infty$.

A finite sequence
$$\chi(n)$$
 is defined as $\chi(n) = \begin{cases} \frac{10}{5}, \frac{10}{3}, \frac{10}{3}, \frac{10}{3}, \frac{10}{3} \end{cases}$ find $\chi(z)$

$$\Xi T \left[\chi(n) \right] = \chi(\Xi) = \frac{3}{5} \chi(n) \Xi^{n}$$

$$= \frac{5}{5} \chi(n) \Xi^{n}$$

$$= \chi(0) \Xi + \chi(1) \Xi^{1} + \chi(2) \Xi^{2} + \chi(3) \Xi^{3} + \chi(4) \Xi^{4}$$

$$+ \chi(5) \Xi^{5}.$$

$$\chi(\Xi) = 5 + 3 \Xi^{1} - 3 \Xi^{2} + 4 \Xi^{4} - 9 \Xi^{5}$$

Z-tomsform for finite sequence

A finite sequence
$$\alpha(n)$$
 is defined as $\beta(n) = \{5,3,-3,0,4,-2\}$

find $\alpha(n) = \{5,3,-3,0,4,-2\}$
 $\alpha(n) = \{6,3,2,1,4,1,2\}$
 $\alpha(n) = \{6,3,3,1,4,1,2\}$
 $\alpha(n) = \{6,3,3,1,4,$

Finite duation Sequence
$$x(n) = \{5, 3, 0, 1, 2, 4\}$$

Find z -transform of $x(n)$.

$$x(0) x(1) x(2) x(3) x(4) x(5)$$

$$-x(n) = \{5, 3, 0, 1, 2, 4\}$$

$$1$$

$$-x(2) = \{x(n)z^{-1} = \{x(n)z^{-1} + x(2)z^{-1} + x(3)z^{-1} + x(4)z^{-1} + x(5)z^{-1}\}$$

$$= x(0)z^{0} + x(1)z^{-1} + x(2)z^{-1} + x(3)z^{-1} + x(4)z^{-1} + x(5)z^{-1}$$

$$= 5z^{0} + 3z^{-1} + 0z^{-2} + 1 \times z^{-3} + 2 \times z^{-1} + 4 \times z^{-5}$$

$$= 5 + 3z^{-1} + z^{-3} + 2z^{-1} + 4z^{-5}$$

Linearity

It states that when two or more individual discrete signals are multiplied by constants, their respective Z-transforms will also be multiplied by the same constants.

Mathematically,

$$a_1x_1(n) + a_2x_2(n) = a_1X_1(z) + a_2X_2(z)$$

Proof - We know that,

$$egin{align} X(Z) &= \sum_{n=-\infty}^{\infty} x(n) Z^{-n} \ &= \sum_{n=-\infty}^{\infty} (a_1 x_1(n) + a_2 x_2(n)) Z^{-n} \ &= a_1 \sum_{n=-\infty}^{\infty} x_1(n) Z^{-n} + a_2 \sum_{n=-\infty}^{\infty} x_2(n) Z^{-n} \ &= a_1 X_1(z) + a_2 X_2(z) \ &HenceProved \ \end{array}$$

Here, the ROC is $\ ROC_1 \bigcap ROC_2$.

Time Shifting

Time Shifting property of 2 transform

- If
$$x(n) \stackrel{2T}{\rightleftharpoons} x(z)$$

- The time Shifting property State that
$$x(n-m) \stackrel{2T}{\rightleftharpoons} z^m x(z)$$

$$x(n+m) \stackrel{2T}{\rightleftharpoons} z^m x(z)$$

$$x(n+m) \stackrel{2T}{\rightleftharpoons} z^m x(z)$$

$$- x(z) = z.T. [x(n)]$$

$$= \stackrel{E}{\rightleftharpoons} x(n) z^n$$

$$= z.T. [x(n-m)] = \stackrel{E}{\rightleftharpoons} x(n-m) z^n$$

$$= z.T. [x(n-m)] = \frac{z}{z}$$

$$-1f n-m=p =) n=p+m$$

$$= \sum_{p_2-\infty}^{\infty} x(p) Z^{-(p+m)}$$

$$= \sum_{p_2-\infty}^{\infty} x(p) Z^{-p} Z^{-m}$$

$$= Z^{-m} \sum_{p_2-\infty}^{\infty} x(p) Z^{-p}$$

$$= Z^{-m} \times (z)$$

example find
$$z$$
-Toomstorm of $6(n-k)$

$$6(n) = 2T$$

$$6(n) = 2T$$

Time shifting property depicts how the change in the time domain in the discrete signal will affect the Z-domain, which can be written as;

$$x(n-n_0)\longleftrightarrow X(Z)Z^{-n}$$

$$x(n-1) \longleftrightarrow Z^{-1}X(Z)$$

Proof -

Let
$$y(P)=X(P-K)$$

$$Y(z)=\sum_{p=-\infty}^{\infty}y(p)Z^{-p}$$

$$=\sum_{p=-\infty}^{\infty}(x(p-k))Z^{-p}$$

Let s = p-k

$$\begin{split} &= \sum_{s=-\infty}^{\infty} x(s) Z^{-(s+k)} \\ &= \sum_{s=-\infty}^{\infty} x(s) Z^{-s} Z^{-k} \\ &= Z^{-k} [\sum_{s=-\infty}^{\infty} x(m) Z^{-s}] \\ &= Z^{-k} X(Z) \end{split}$$

Hence Proved

Time Scaling

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Scaling property of 2 Transform

- \alpha(n) \stackrel{ZT}{\rightleftharpoons} x(z)
Then scaling property of 2 transform started that \alpha^n x(n) \stackrel{ZT}{\rightleftharpoons} x [2/\alpha]

Proof
x(z) = z.T. [x(m)]
= \stackrel{Z}{\rightleftharpoons} x(n) \stackrel{Z}{\rightleftharpoons}
2.T. [\alpha^n (x(n))] = \stackrel{Z}{\rightleftharpoons} \alpha^n x(n) \stackrel{Z}{\rightleftharpoons}
n_{-\infty}
```

2.T.
$$[a^{n}(x(n))] = \sum_{n=-\infty}^{\infty} a^{n}x(n) z^{-n}$$
 $1 - \infty$
 $2 = \sum_{n=-\infty}^{\infty} x(n) (az^{-1})^{n}$
 $2 = \sum_{n=-\infty}^{\infty} x(n) (az^{-1})^{-n}$
 $2 = \sum_{n=-\infty}^{\infty} x(n) (a^{-1}z^{-n})^{-n}$
 $2 = \sum_{n=-\infty}^{\infty} x(n) (a^{-1}z^{-n})^{-n}$

e.g.
$$\chi(n) = a^{2} \chi(n)$$
, find $\chi(2)$

$$\chi(n) \stackrel{\text{ZT}}{\leftrightarrow} \frac{2}{Z-1}$$

$$\chi(n) \stackrel{\text{ZT}}{\leftrightarrow} \frac{2}{Z-1}$$

$$\chi(n) \stackrel{\text{ZT}}{\leftrightarrow} \frac{(2/q)}{(\frac{Z}{q})-1} = \frac{Z}{Z-q}$$

Time Scaling property tells us, what will be the Z-domain of the signal when the time is scaled in its discrete form, which can be written as;

$$a^n x(n) \longleftrightarrow X(a^{-1}Z)$$

Proof -

Let
$$y(p)=a^px(p)$$

$$Y(P)=\sum_{p=-\infty}^\infty y(p)Z^{-p}$$

$$=\sum_{p=-\infty}^\infty a^px(p)Z^{-p}$$

$$=\sum_{p=-\infty}^\infty x(p)[a^{-1}Z]^{-p}$$

$$=X(a^{-1}Z)$$
 $Henceproved$

Convolution

This depicts the change in Z-domain of the system when a convolution takes place in the discrete signal form, which can be written as –

$$x_1(n) * x_2(n) \longleftrightarrow X_1(Z). X_2(Z)$$

Proof -

$$egin{aligned} X(Z) &= \sum_{n=-\infty}^{\infty} x(n) Z^{-n} \ &= \sum_{n=-\infty}^{\infty} [\sum_{k=-\infty}^{\infty} x_1(k) x_2(n-k)] Z^{-n} \ &= \sum_{k=-\infty}^{\infty} x_1(k) [\sum_{n=-\infty}^{\infty} x_2(n-k) Z^{-n}] \ &= \sum_{k=-\infty}^{\infty} x_1(k) [\sum_{n=-\infty}^{\infty} x_2(n-k) Z^{-(n-k)} Z^{-k}] \end{aligned}$$

Let n-k = I, then the above equation cab be written as -

$$egin{aligned} X(Z) &= \sum_{k=-\infty}^{\infty} x_1(k) [Z^{-k} \sum_{l=-\infty}^{\infty} x_2(l) Z^{-l}] \ &= \sum_{k=-\infty}^{\infty} x_1(k) X_2(Z) Z^{-k} \ &= X_2(Z) \sum_{k=-\infty}^{\infty} x_1(Z) Z^{-k} \ &= X_1(Z). \, X_2(Z) \end{aligned}$$

Time reversal property of 2 Tranform

- If
$$x(n) = \frac{2T}{2} \times (2)$$

Then time reversal property states that $x(-n) = \frac{2T}{2} \times (z^{-1})$

Poset

- $x(z) = z \cdot T \cdot [x(n)]$

= $\frac{z}{2} \times (z^{-1}) = \frac{z}{2} \times (z^{-1}) =$

$$-2.T[\chi(-n)] = \underbrace{\mathbb{E}}_{\chi(-n)} \chi(-n) = \underbrace{\mathbb{E}}_{\chi(-n)$$

e.g.
$$x(n) = \alpha^{n} u(-n)$$
. $+ ind x(2)$

$$- x(n) = \alpha^{n} u(-n)$$

$$= (\frac{1}{q})^{-n} u(-n)$$

$$x(n) > (\frac{1}{q})^{n} u(n) \xrightarrow{2T} \frac{Z}{Z^{-1} u(-n)}$$

$$(\frac{1}{q})^{-n} u(-n) \xrightarrow{2T} \frac{Z^{-1}}{Z^{-1} - Va} = \frac{1}{2} \frac{a}{a-Z}$$

Multiplication of Convolution property of 2-Transform

If $x_1(n) \neq \frac{2T}{2T} \times x_1(2)$ $x_2(n) \neq \frac{2T}{2T} \times x_2(2)$ - Multiplication property states that $x_1(n) \times x_2(n) \neq \frac{2T}{2T} \times x_1(2) + x_2(2)$ - convolution property states that $x_1(n) \times x_2(n) \neq \frac{2T}{2T} \times x_1(2) \times x_2(2)$.

e.g.
$$u(n-1) \neq 6(n)$$
, $find z - Team form$

$$- x_1(n) = u(n-1) \stackrel{2T}{\longleftrightarrow} x_1(2) = \frac{1}{Z-1}$$

$$- x_2(n) = 6(n) \stackrel{2T}{\longleftrightarrow} x_2(2) = 1$$

$$- x(2) = ZT [x_1(n) \neq x_2(n)]$$

$$= x_1(2) x_2(2)$$

$$= \frac{1}{Z-1}$$

$$\frac{e.g.}{-x_{1}(n)} = x_{1}(n-2) + \frac{5(n-3)}{2}, \text{ find } z - \frac{1}{2} = \frac{1}{2} \left(\frac{1}{2-1}\right) = \frac{1}{2} \left(\frac{1}{2-1}\right)$$

$$= x_{1}(n) = x_{1}(n-2) + \frac{1}{2} \times x_{1}(2) = \frac{1}{2} \left(\frac{1}{2-1}\right) = \frac{1}{2} \left(\frac{1}{2-1}\right)$$

$$= x_{2}(n) = \frac{1}{2} (n-1) + \frac{1}{2} \times x_{2}(n)$$

$$= x_{1}(n) + x_{2}(n)$$

$$= x_{1}(n) + x_{2}(n)$$

$$= x_{1}(n) + x_{2}(n)$$

$$= \frac{1}{2} \left(\frac{1}{2-1}\right) \left(\frac{1}{2}\right)$$

$$= \frac{1}{2} \left(\frac{1}{2-1}\right) \left(\frac{1}{2}\right)$$

Successive Differentiation

Successive Differentiation property shows that Z-transform will take place when we differentiate the discrete signal in time domain, with respect to time. This is shown as below.

$$\frac{dx(n)}{dn} = (1 - Z^{-1})X(Z)$$

Proof -

Consider the LHS of the equation – $\frac{dx(n)}{dn}$

$$= \frac{[x(n) - x(n-1)]}{[n - (n-1)]}$$

$$=x(n)-X(n-1)$$
 $=x(Z)-Z^{-1}x(Z)$ $=(1-Z^{-1})x(Z)$ $HenceProved$

Multiplication in Time

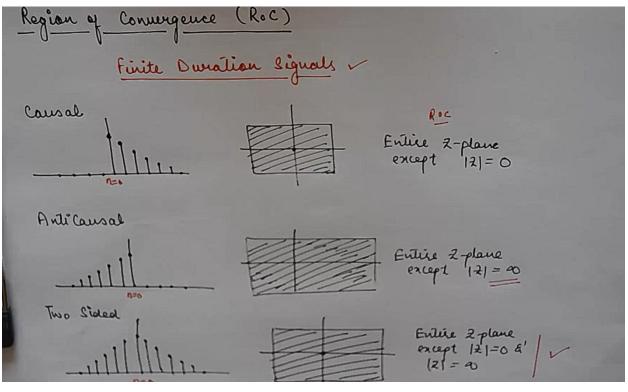
It gives the change in Z-domain of the signal when multiplication takes place at discrete signal level.

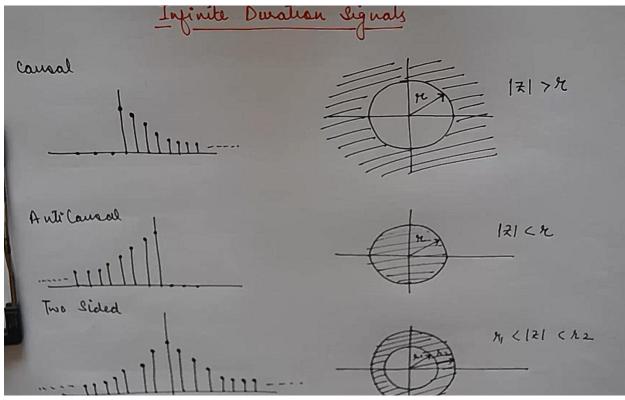
$$x_1(n).\,x_2(n) \longleftrightarrow (\tfrac{1}{2\Pi j})[X1(Z)*X2(Z)]$$

Conjugation in Time

This depicts the representation of conjugated discrete signal in Z-domain.

$$X^*(n) \longleftrightarrow X^*(Z^*)$$





RELATIONSHIP BETWEEN Z-TRANSFORM AND FOURIER TRANSFORM Definition of z-transform, $z[x(n)] = X(z) = \sum_{n=-\infty}^{\infty} x(n) \cdot z^{-n}.$ Definition of Discrete Time Fourier Transform, $z[x(n)] = x[e^{j\omega}] = \sum_{n=-\infty}^{\infty} x[n] \cdot e^{j\omega n}.$ $z[z(n)] = x[e^{j\omega}] = x[e^{j\omega}] = \sum_{n=-\infty}^{\infty} x[n] \cdot e^{j\omega n}.$ $z[z(n)] = x[z(n)] = x[e^{j\omega}] = \sum_{n=-\infty}^{\infty} x[n] \cdot e^{j\omega n}.$

57. Determine the z-transform of the signal $x(n) = \alpha^n u(n)$ and also the ROC and pole & zero locations of X(z) in the z-plane.

Solution:

Given $x(n) = \alpha^n u(n)$

By definition
$$X(z) = \sum_{n=-\infty}^{\infty} x(n)z^{-n}$$
$$= \sum_{n=-\infty}^{\infty} \alpha^n u(n)z^{-n}$$
$$= \sum_{n=0}^{\infty} (\alpha z^{-1})^n$$

Using geometric series,

$$\sum_{n=0}^{\infty} a^n = \frac{1}{1-a}$$

$$\therefore X(z) = \frac{1}{1-\alpha z^{-1}}; |z| > |\alpha|$$

$$= \frac{z}{z-\alpha}; |z| > |\alpha|$$

There is a pole at $z = \alpha \& zero$ at z = 0.

```
POC of Impulse tunction in 2-Tourstoom

1. 6(n) 4^{2T}; |

POC! Entire Z plane, including z=0 & z=\infty

2. 6(n-k) 4^{2T}; z^{-k}

POC! Entire Z plane, including z=\infty & excluding z=0

3. 6(n+k) 4^{2T}; z^{-k}

POC! Entire Z plane, including z=\infty & excluding z=\infty

6(n-3) - Entire Z plane.

Includes z=\infty

Gychide z=\infty
```

ROC of dispete time segumee in 2 Transform

- 1. If sequence is purely right sided or (ausal, then ROC: Entire Z plane except Z=0 x(n) = 21, 2, 3, 4 $\uparrow x(2) = 1+2z^{-1}+3z^{-2}+4$
- 2. If Sequence is purely Left sided or Anticousal. Then

 Check may be

 Check may be

 Then

 Check may be

 Then

 Check may be

 Then

 Then

 Check may be

 Then

 Then

 Check may be

 Then

 The
- 3. If sequence is a two sided, Then

 ROC! Entire Z plane except of 2=0 4 z=00 $x(n) = \{1,2,1,3,1\} \xrightarrow{2T} x(2) = 22 + 22 + 1 + 32^{-1} + 2^{-2}$

* Z-Toransform of Discrete unit step function. Le(n): *

ZT[u(n)].

ZT[u(n)] =
$$\times(z) = \underset{n=-\infty}{\overset{\infty}{\sum}} \chi(n) z^{n}$$
.

ZT[u(n)] = $\underset{n=-\infty}{\overset{\infty}{\sum}} u(n) z^{n}$.

 $= \underset{n=0}{\overset{\infty}{\sum}} z^{n}$
 $= \underset{n=0}{\overset{\infty}{\sum}} (z^{n})^{n} = \underset{n=0}{\overset{\infty}{\sum}} z^{n}$

2- Tourstorm of Unit Impulse function $- \chi(n) = 6cn$ $- \chi(2) = 2.T. [\chi(n)]$ $= \frac{\varepsilon}{2} \chi(n) z^{-n}$ $= \frac{\varepsilon}{2} \chi(n$

Z-Tourstorm of Standard basic Signals

- Find the Z-tourstorm of the signals of
$$u(n)$$
 is $a^n u(n)$

- $u(n) = a^n u(n)$

- $u(n) = a^n u(n)$

- $u(n) = a^n u(n)$

= $u(n) = a^n u(n)$

= $u(n) = a^n u(n)$

= $u(n) = a^n u(n)$
 $u(n) = a^n u(n)$
 $u(n) = a^n u(n)$

$$- \times (2) = \underbrace{\mathbb{E}}_{(q^{n})} (q^{n}) (z^{n})$$

$$= \underbrace{\mathbb{E}}_{(qz^{1})^{n}} (qz^{1})^{n}$$

$$J(n) = \overline{a^n} u(n)$$

$$- y(2) = 2.T. [Y(n)]$$

$$= \sum_{n=0}^{\infty} Y(n) \overline{z^n}$$

$$= \sum_{n=0}^{\infty} \overline{a^n} u(n) \overline{z^n}$$

$$= \sum_{n=0}^{\infty} \overline{a^n} u(n) \overline{z^n}$$

Z-Toursform of Standard basic Signal

Find the Z-toursform of the Sequence
$$x(n) = -a^n u(-n-1)$$

$$x(z) = z.T. [x(n)]$$

$$= \sum_{n=-\infty}^{\infty} x(n) z^{-n}$$

$$= \sum_{n=-\infty}^{\infty} -a^n u(-n-1) z^{-n}$$

$$\chi(2) = \underbrace{\xi}_{n,-\infty} - a^{n}z^{-n}$$

$$= -\underbrace{\xi}_{n,-\infty} (qz^{1})^{n} \qquad \left[\underbrace{\xi}_{n,0} - a^{n}z^{-n} \right]$$

$$= -\underbrace{\xi}_{n,-\infty} (qz^{1})^{n} \qquad \left[\underbrace{\xi}_{n,0} - a^{n}z^{-n} \right]$$

$$= -\underbrace{\xi}_{n,-\infty} (qz^{1})^{-n}$$

$$= -\underbrace{\xi}_{n,-\infty} (zq^{1})^{n}$$

$$= -\underbrace{\xi}_{n,-\infty} (zq^{1})^{n}$$

$$= -\underbrace{\xi}_{n,-\infty} (zq^{1})^{n}$$

 $\begin{pmatrix}
 a u(n) & 2T \\
 -a^{n}u(-n-1) & 2T \\
 -a^{n}u(-n-1) & 2T \\
 \hline
 a^{n}u(n) & 2T \\
 \hline
 a^{n}u(-n-1) & 2T \\
 a^{n}u(-n-1) & 2T \\
 \hline
 a^{n}u(-n-1) & 2T \\
 \hline
 a^{n}u(-n-1) & 2T \\
 a^{n}u(-n-1) & 2T \\$

$$\gamma(z) = \sum_{n=0}^{\infty} \frac{-n}{q^{2}}$$

$$= \sum_{n=0}^{\infty} (az)^{n}$$

$$we know = \sum_{n=0}^{\infty} \frac{1}{1-a}$$

$$-\gamma(z) = \frac{1}{1-(az)^{-1}} = \frac{1}{1-\sqrt{az}}$$

$$= \frac{az}{az-1}$$

$$\gamma(n) = \frac{-n}{q} \gamma(n)$$

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Examples of Inverse Z tours form

Find the Signal x(n) for which the Z tours form is

x(2) = 4z^4 - z^3 - 3z + 4z^1 + 3z^2
-x(n) = z^1 [x(z)] = \frac{1}{zn} \int x(z) z^{n-1} dz
x(2) = 4z^4 - 1z^3 + 0z^2 - 3z + 0z^0 + 4z^1 + 3z^2
x(2) = 4z^4 - 1z^3 + 0z^2 - 3z + 0z^0 + 4z^1 + 3z^2
x(2) = 2z^2 - 2z(n)z^{n-1}
x(2) = 2z^2 - 2z(n)z^{n-2}
x(3) = 2z^2 - 2z(n)z^{n-2}
x(4) = 2z^2 - 2z(n)z^{n-2}
x(6) = 2z^2 - 2z(n)z^{n-2}
x(1) = 2z^2 - 2z(n)z^{n-2}
x(2) = 2z^2 - 2z(n)z^{n-2}
```

If
$$\alpha x(z) = 2z^2 + 3$$
, find $x(n)$

$$-x(z) = 2z^2 + 0 \times z + 3 \times z$$

$$-x(2) = \frac{2}{2}x(n)z^{-n-1}$$

$$-x(2) = \frac{2}{2}x(n)z^{-n-1}$$

$$-x(n) = \frac{2}{2}x(n)z^{-n}$$

$$-x(n) = \frac{2}{2}x(n)z^{-n}$$

$$-x(n) = \frac{2}{2}x(n)z^{-n}$$

If
$$x(2) = 3 + 2z^{-1} + 3z^{-3}$$
, find $x(n)$

$$-x(2) = 3 \times z^{0} + 2 \times z^{-1} + 0 \times z^{-2} + 3 \times z^{-3}$$

$$-x(2) = \frac{3 \times z^{0} + 2 \times z^{-1} + 0 \times z^{-2} + 3 \times z^{-3}}{x(2)}$$

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4.18. Using the power series expansion technique, find the inverse z-transform of the following X(z):

(a)
$$X(z) = \frac{z}{2z^2 - 3z + 1}$$
 $|z| < \frac{1}{2}$
(b) $X(z) = \frac{z}{2z^2 - 3z + 1}$ $|z| > 1$

(b)
$$X(z) = \frac{z}{2z^2 - 3z + 1}$$
 $|z| >$

(a) Since the ROC is $|z| < \frac{1}{2}$, x[n] is a left-sided sequence. Thus, we must divide to obtain a series in power of z. Carrying out the long division, we obtain

$$z + 3z^{2} + 7z^{3} + 15z^{4} + \cdots$$

$$1 - 3z + 2z^{2} \begin{vmatrix} z \\ z - 3z^{2} + 2z^{3} \\ \hline 3z^{2} - 2z^{3} \\ 3z^{2} - 9z^{3} + 6z^{4} \\ \hline 7z^{3} - 6z^{4} \\ 7z^{3} - 21z^{4} + 14z^{5} \\ \hline 15z^{4} + \cdots$$

Thus,

$$X(z) = \cdots + 15z^4 + 7z^3 + 3z^2 + z$$

and so by definition (4.3) we obtain

$$x[n] = \{..., 15, 7, 3, 1, 0\}$$

(b) Since the ROC is |z| > 1, x[n] is a right-sided sequence. Thus, we must divide so as to obtain a series in power of z^{-1} as follows:

$$2z^{2} - 3z + 1 \boxed{z}$$

$$\underbrace{\frac{z - \frac{3}{2} - \frac{1}{2}z^{-1}}{\frac{\frac{3}{2} - \frac{1}{2}z^{-1}}{\frac{\frac{3}{2} - \frac{9}{4}z^{-1} + \frac{3}{4}z^{-2}}}_{\frac{7}{4}z^{-1} - \frac{3}{4}z^{-2}}$$

Thus,

$$X(z) = \frac{1}{2}z^{-1} + \frac{3}{4}z^{-2} + \frac{7}{8}z^{-3} + \cdots$$

and so by definition (4.3) we obtain

$$x[n] = \{0, \frac{1}{2}, \frac{3}{4}, \frac{7}{8}, \dots\}$$