

6.52. Show that if $x[n]$ is real, then its DFT $X[k]$ satisfies the relation

$$X[N-k] = X^*[k] \quad (6.204)$$

where * denotes the complex conjugate.

From Eq. (6.92)

$$X[N-k] = \sum_{n=0}^{N-1} x[n] W_N^{(N-k)n} = \sum_{n=0}^{N-1} x[n] e^{-j(2\pi/N)(N-k)n}$$

$$\text{Now } e^{-j(2\pi/N)(N-k)n} = e^{-j2\pi n} e^{j(2\pi/N)kn} = e^{j(2\pi/N)kn}$$

Hence, if $x[n]$ is real, then $x^*[n] = x[n]$ and

$$X[N-k] = \sum_{n=0}^{N-1} x[n] e^{j(2\pi/N)kn} = \left[\sum_{n=0}^{N-1} x[n] e^{-j(2\pi/N)kn} \right]^* = X^*[k]$$

6.53. Show that

$$x[n] = \text{IDFT}\{X[k]\} = \frac{1}{N} [\text{DFT}\{X^*[k]\}]^* \quad (6.205)$$

where * denotes the complex conjugate and

$$X[k] = \text{DFT}\{x[n]\}$$

We can write Eq. (6.94) as

$$x[n] = \frac{1}{N} \left[\sum_{k=0}^{N-1} X[k] e^{j(2\pi/N)kn} \right] = \frac{1}{N} \left[\sum_{k=0}^{N-1} X^*[k] e^{-j(2\pi/N)nk} \right]^*$$

Noting that the term in brackets in the last term is the DFT of $X^*[k]$, we get

$$x[n] = \text{IDFT}\{X[k]\} = \frac{1}{N} [\text{DFT}\{X^*[k]\}]^*$$

which shows that the same algorithm used to evaluate the DFT can be used to evaluate the IDFT.

Radix-2 DIT - FFT Algorithm:

DIT → Decimation in Time

FFT → Fast Fourier Transform.

$x(n) \rightarrow$ length N

$$x(n) = \{x(0), x(1), x(2), x(3), \dots, x(N-2), x(N-1)\}$$

even indexed Seq.: $\{x(0), x(2), x(4), \dots, x(N-2)\}$

odd indexed Seq.: $\{x(1), x(3), x(5), \dots, x(N-1)\}$

W.K.T. $\frac{N}{N-1}$ -Point DFT

$$X(k) = \sum_{n=0}^{N-1} x(n) W_N^{kn}; \quad 0 \leq k \leq N-1 \rightarrow ①$$

\therefore

even, odd Seq.

eqn ① → Decimation → even, odd Seq.

$$X(k) = \sum_{n=0}^{N-2} x(n) W_N^{kn} + \sum_{n=1}^{N-1} x(n) W_N^{kn} \rightarrow ②$$

$n \rightarrow \text{even}$

$n \rightarrow \text{odd}$

Put $n = 2\tau$ in 1st term, $n = 2\tau+1$ in 2nd term.

$$X(k) = \sum_{\tau=0}^{N/2-1} x(2\tau) W_N^{2\tau k} + \sum_{\tau=0}^{N/2-1} x(2\tau+1) W_N^{k(2\tau+1)} \rightarrow ③$$

$$X(k) = \sum_{\tau=0}^{N/2-1} g(\tau) W_N^{2\tau k} + \sum_{\tau=0}^{N/2-1} h(\tau) W_N^{k(2\tau+1)}$$

$$g(\tau) = \sum_{\tau=0}^{N/2-1} x(2\tau) W_N^{2\tau k}$$

$$\therefore W_N = e^{-j\frac{2\pi}{N}} \Rightarrow W_N^2 = e^{-j\frac{4\pi}{N}} = e^{-j\frac{2\pi}{N/2}} = W_{N/2}$$

Rearrange,

$$x(k) = \sum_{r=0}^{N/2-1} g(r) w_{N/2}^{kr} + w_N^k \sum_{r=0}^{N/2-1} h(r) w_{N/2}^{kr} \quad (1)$$

$\frac{N}{2}$ DFT eqn

$$x(k) = G(k) + w_N^k H(k); \quad 0 \leq k \leq \frac{N}{2} - 1 \rightarrow (2)$$

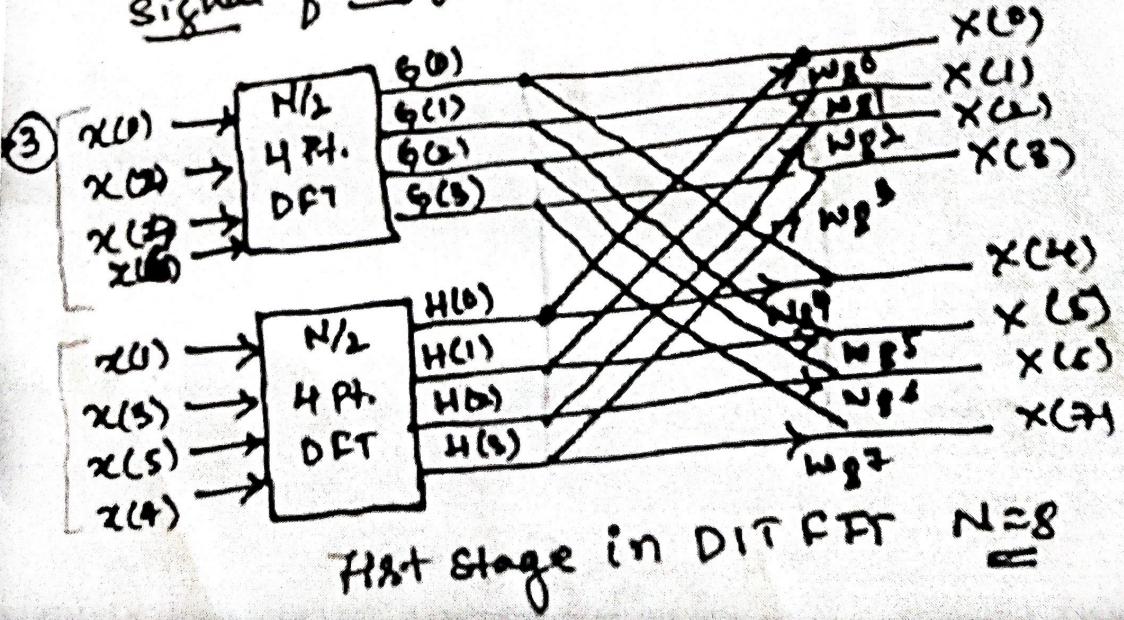
$G(k) \& H(k) \rightarrow$ periodic with period $\frac{N}{2}$

$$x(k) = G\left(k - \frac{N}{2}\right) + w_N^k H\left(k - \frac{N}{2}\right); \quad \frac{N}{2} \leq k \leq N-1 \rightarrow (3)$$

$$\text{Ex:- } N=8 \quad \therefore k \rightarrow 0+0^+ \quad k=0+0^+ \rightarrow (4) \quad k=4+0^+ \rightarrow (5) \quad k=4+0^+ \rightarrow (6)$$

$$(5) \Rightarrow \begin{aligned} k=0; x(0) &= g(0) + w_8^0 H(0) & k=4; x(4) &= g(0) + w_8^4 H(0) \\ k=1; x(1) &= g(1) + w_8^1 H(1) & k=5; x(5) &= g(1) + w_8^5 H(1) \\ k=2; x(2) &= g(2) + w_8^2 H(2) & k=6; x(6) &= g(2) + w_8^6 H(2) \\ k=3; x(3) &= g(3) + w_8^3 H(3) & k=7; x(7) &= g(3) + w_8^7 H(3) \end{aligned}$$

$k=3; x(3) \rightarrow$
signal flow graph



$g(k) \& h(k) \rightarrow \frac{N}{2}$ point

combination of two $\frac{N}{4}$ points.

(8)

$$G(k) = \sum_{r=0}^{N_2-1} g(r) W_{N/2}^{kr} \rightarrow (7)$$

$$G(k) = \sum_{r=0}^{N_2-2} g(r) W_{N/2}^{kr} + \sum_{r=0}^{N_2-1} g(r) W_N^{kr} \quad \begin{matrix} r+1 = \frac{N}{2}-1 \\ r+1 = \frac{N}{2}-2 \end{matrix} \quad \begin{matrix} r+1 = \frac{N}{2}-1 \\ r+1 = \frac{N}{2}-2 \end{matrix}$$

(9) =

$$g(r) = \{ g(0), g(1), g(2) \dots g\left(\frac{N}{2}-2\right), g\left(\frac{N}{2}-1\right) \}$$

Put $r = 2d$ in 1st term, $r = 2d+1$ in 2nd (7)

$$G(k) = \sum_{d=0}^{N/4-1} g(2d) W_{N/2}^{2kd} + \sum_{d=0}^{N/4-1} g(2d+1) \boxed{W_{N/2}^{k(2d+1)}} \quad \begin{matrix} b(d) \\ b(d+1) \end{matrix} \quad W_{N/2}^{2kd} \cdot W_{N/2}^k$$

$$\Rightarrow G(k) = \sum_{d=0}^{N/4-1} c(d) W_{N/4}^{2kd} + W_{N/2}^k \sum_{d=0}^{N/4-1} b(d) W_{N/4}^{3kd} \quad \text{eqn (10)}$$

$$G(k) = \boxed{A(k)} + W_{N/2}^k \boxed{B(k)} ; \quad 0 \leq k \leq \frac{N}{4}-1 \rightarrow (8)$$

$$H(k) = C(k) + W_{N/2}^k D(k) ; \quad 0 \leq k \leq \frac{N}{4}-1 \rightarrow (9)$$

$A(k), B(k), C(k) \& D(k) \rightarrow \text{periodic } \frac{N}{4}$.

$$⑧ \Rightarrow G(K) = \frac{A(K-\frac{N}{4})}{A(K-2)} + \frac{w_{N/2}^K B(K-\frac{N}{4})}{B(K-2)} ; \frac{N}{4} \leq K \leq \frac{N}{2}-1 \rightarrow ⑩$$

②

$$⑨ \Rightarrow H(K) = \frac{C(K-\frac{N}{4})}{C(K-2)} + \frac{w_{N/2}^K D(K-\frac{N}{4})}{D(K-2)} ; \frac{N}{4} \leq K \leq \frac{N}{2}-1 \rightarrow ⑪$$

1)

④

$\omega_{N/4}$

$$K=0 \text{ & } 1 \rightarrow \text{eqn } ⑧ \text{ & } ⑨$$

$$2 \text{ & } 3 \rightarrow \text{eqn } ⑩ \text{ & } ⑪$$

eqn ⑥

$$K=0; G(0) = A(0) + w_4^0 B(0) ; H(0) = C(0) + w_4^0 D(0)$$

$\omega_{N/4}$

$$K=1; G(1) = A(1) + w_4^1 B(1) ; H(1) = C(1) + w_4^1 D(1)$$

$\omega_{N/4}$

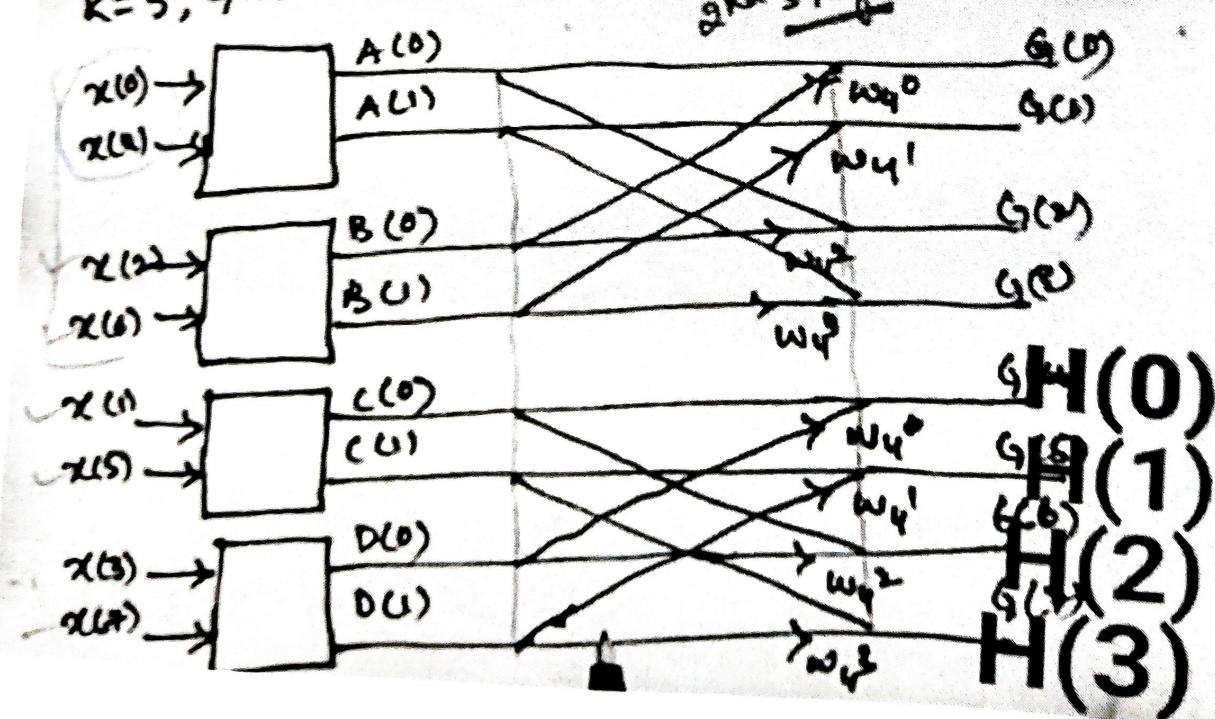
$$\text{eqn } ⑩ \quad K=2; G(2) = A(0) + w_4^2 B(0) ; H(2) = C(0) + w_4^2 D(0)$$

$\omega_{N/4}$

$$\text{eqn } ⑪ \quad K=3; G(3) = A(1) + w_4^3 B(1) ; H(3) = C(1) + w_4^3 D(1)$$

$$K=3; G(3) = A(1) + w_4^3$$

grid stage



Radix-2 DIT-FFT Algorithm:

Each $\frac{N}{4}$ DFT as two $\frac{N}{8}$ point DFTs.

the 2-point DFT of $x(0)$ & $x(4)$

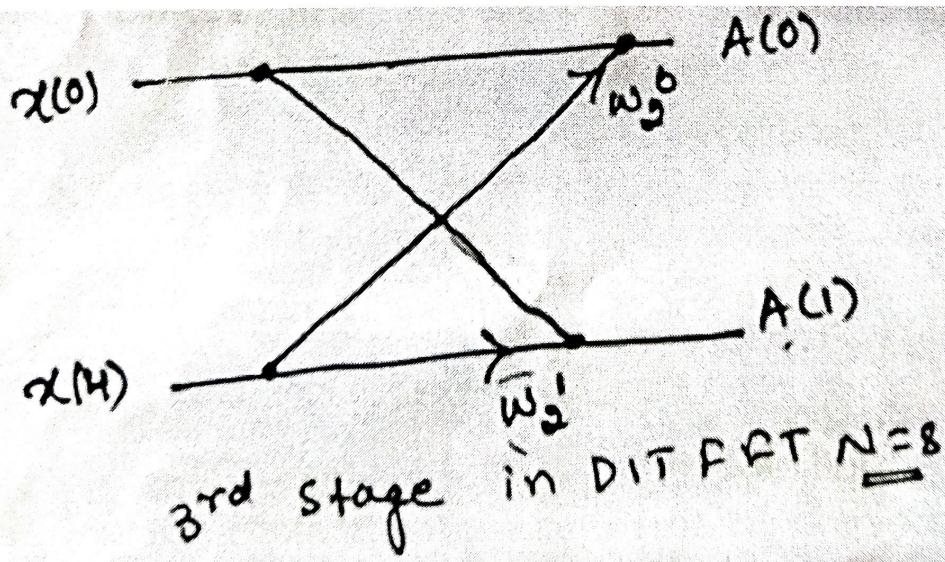
$$A(k) = \sum_{n=0}^{N/4-1} x(n) W_{N/4}^{kn}; 0 \leq k \leq \frac{N}{4}-1$$

$\boxed{N=8}$

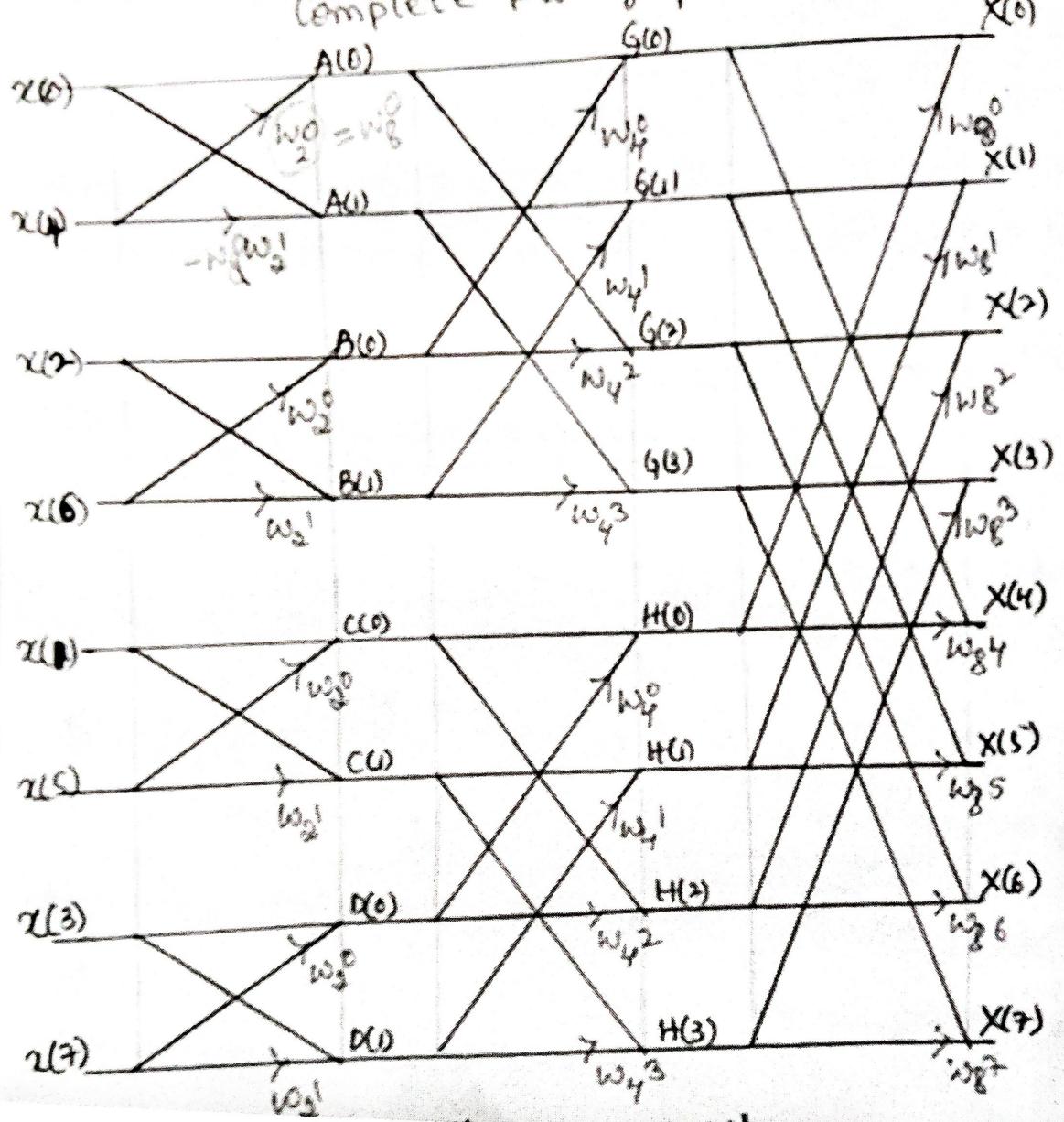
$$A(k) = \sum_{n=0}^1 x(n) W_2^{kn}; 0 \leq k \leq 1$$

$$\text{For } k=0 \Rightarrow A(0) = x(0) + w_2^0 x(4)$$

$$\text{For } k=1 \Rightarrow A(1) = x(1) + w_2^1 x(4)$$



Complete flow graph DIT-FFT ; $N=8$



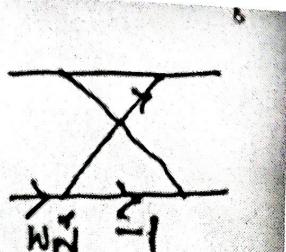
$$ab \xrightarrow{w_N^r} a + w_N^r b$$

$$ab \xrightarrow{w_N^{r+\frac{N}{2}}} a + w_N^{r+\frac{N}{2}} b$$

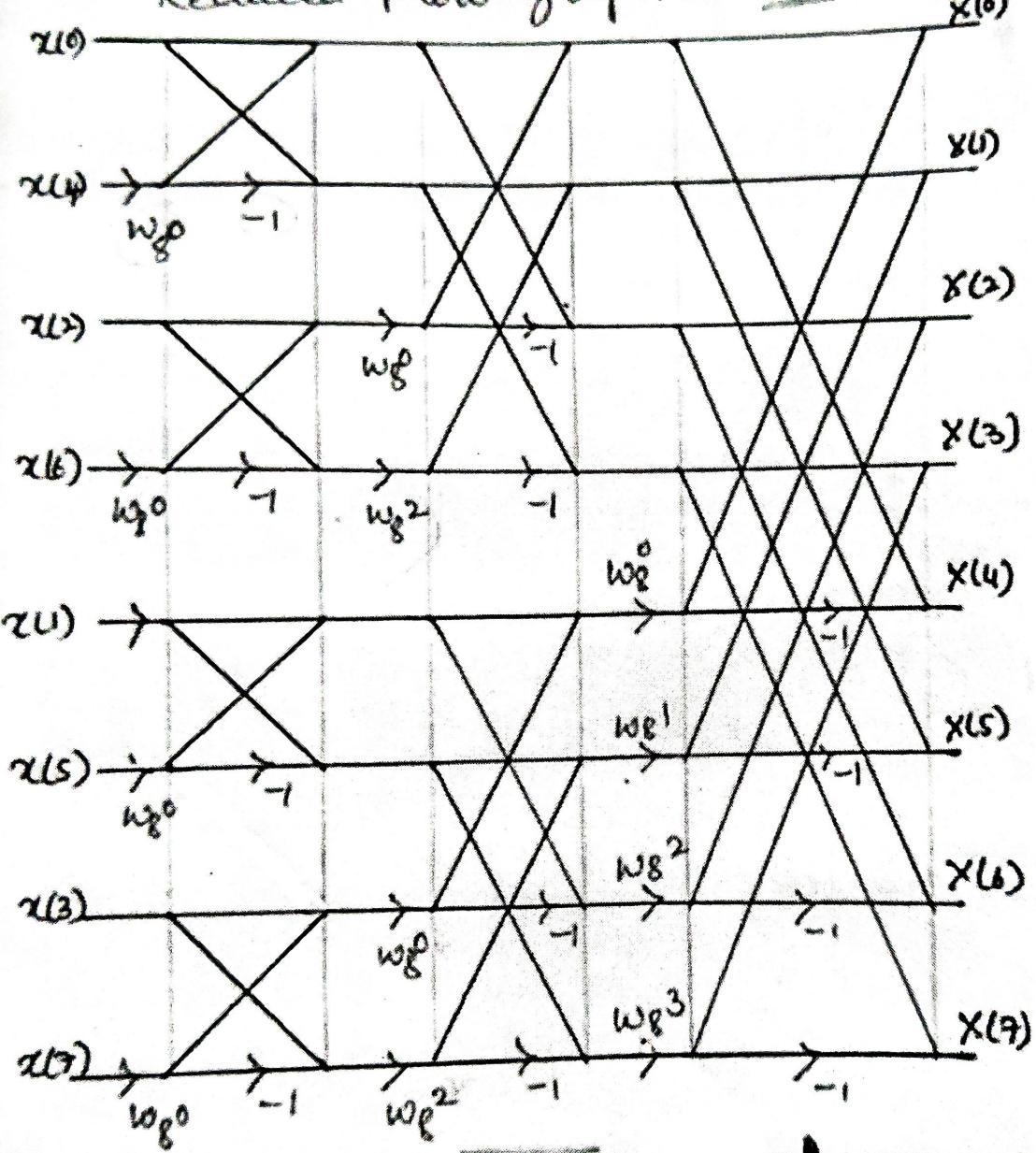
$$w_N^{r+\frac{N}{2}} = -w_N^r$$

$$w_N^{r+\frac{N}{2}} = -w_N^r$$

↳ Symmetry



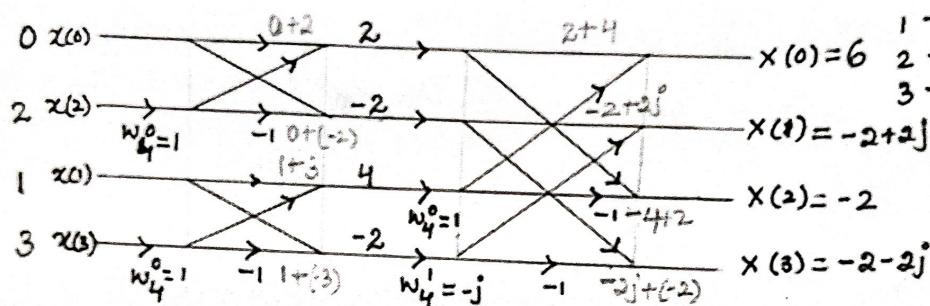
Reduced Flow graph. DIT-PFT $N=8$



Given $x(n) = \{0, 1, 2, 3\}$, find $X(k)$ using DIT-FFT Algorithm.

$$\therefore N=4$$

$$N_4^0 = 1 \quad N_4^1 = -j$$



bit reversal

$$4 = 2^2$$

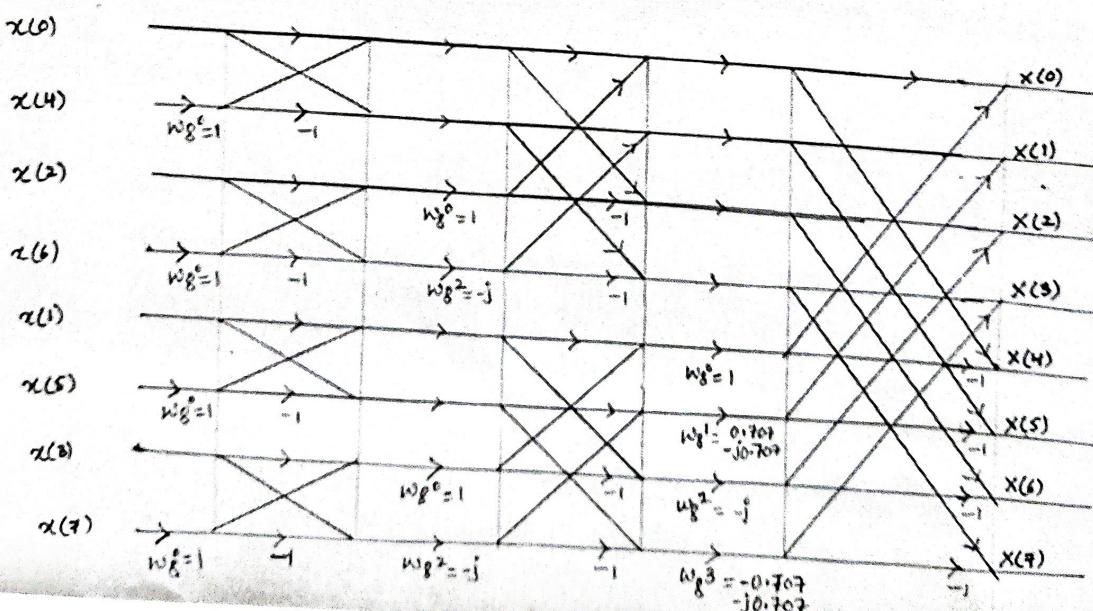
BR	BR
0 → 00	00 → 0v
1 → 01	10 → 2v
2 → 10	01 → 1v
3 → 11	11 → 3v

Flow-graph for DIT-FFT: $N=8$

$$\therefore X(k) = \{6, -2+2j, -2, -2-2j\}$$

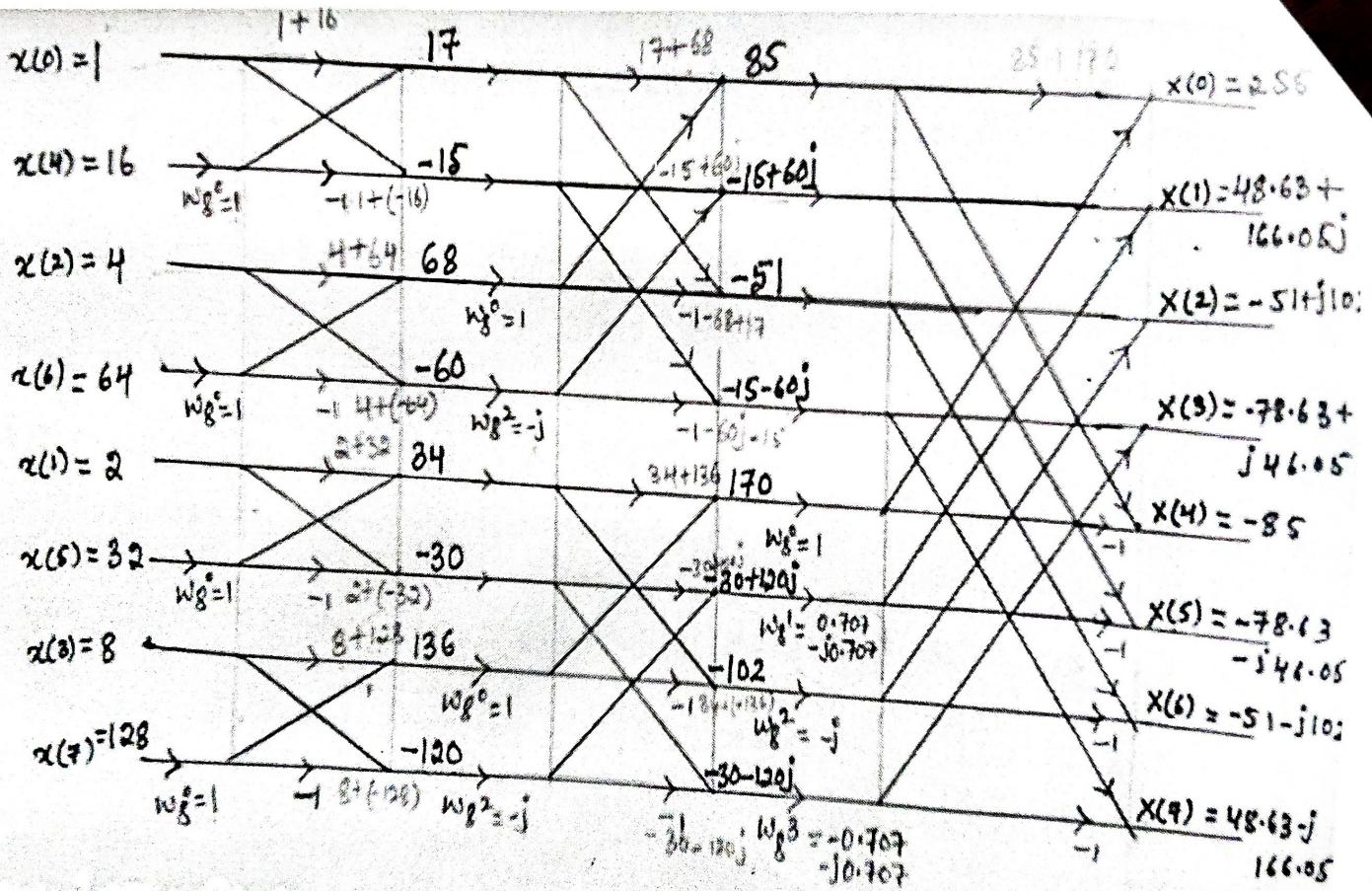
Given $x(n) = \{1, 2, 4, 8, 16, 32, 64, 128\}$

Find $X(k)$ using DIT-FFT.



BIT REVERSAL

000	-->	000	-->	0
001	-->	100	-->	4
010	-->	010	-->	2
011	-->	110	-->	6
100	-->	001	-->	1
101	-->	101	-->	5
110	-->	011	-->	3
111	-->	111	-->	7



$$\therefore X(k) = \{256, 48.63 + j166.05, -51 + j102, -78.63 + j46.05, -85, -78.63 - j46.05, -51 - j102, 48.63 - j166.05\}$$

6.57. Consider a sequence

$$x[n] = \{1, 1, -1, -1, -1, 1, 1, -1\}$$

Determine the DFT $X[k]$ of $x[n]$ using the decimation-in-time FFT algorithm.

From Figs. 6-38(a) and (b), the phase factors W_4^n and W_8^n are easily found as follows:

$$W_4^0 = 1 \quad W_4^1 = -j \quad W_4^2 = -1 \quad W_4^3 = j$$

and

$$W_8^0 = 1 \quad W_8^1 = \frac{1}{\sqrt{2}} - j \frac{1}{\sqrt{2}} \quad W_8^2 = -j \quad W_8^3 = -\frac{1}{\sqrt{2}} - j \frac{1}{\sqrt{2}}$$

$$W_8^4 = -1 \quad W_8^5 = -\frac{1}{\sqrt{2}} + j \frac{1}{\sqrt{2}} \quad W_8^6 = j \quad W_8^7 = \frac{1}{\sqrt{2}} + j \frac{1}{\sqrt{2}}$$

Next, from Eqs. (6.215a) and (6.215b)

$$f[n] = x[2n] = \{x[0], x[2], x[4], x[6]\} = \{1, -1, -1, 1\}$$

$$g[n] = x[2n+1] = \{x[1], x[3], x[5], x[7]\} = \{1, -1, 1, -1\}$$

Then, using Eqs. (6.206) and (6.212), we have

$$\begin{bmatrix} F[0] \\ F[1] \\ F[2] \\ F[3] \end{bmatrix} = \begin{bmatrix} 1 & 1 & 1 & 1 \\ 1 & -j & -1 & j \\ 1 & -1 & 1 & -1 \\ 1 & j & -1 & -j \end{bmatrix} \begin{bmatrix} 1 \\ -1 \\ -1 \\ 1 \end{bmatrix} = \begin{bmatrix} 0 \\ 2+j2 \\ 0 \\ 2-j2 \end{bmatrix}$$

$$\begin{bmatrix} G[0] \\ G[1] \\ G[2] \\ G[3] \end{bmatrix} = \begin{bmatrix} 1 & 1 & 1 & 1 \\ 1 & -j & -1 & j \\ 1 & -1 & 1 & -1 \\ 1 & j & -1 & -j \end{bmatrix} \begin{bmatrix} 1 \\ -1 \\ -1 \\ 1 \end{bmatrix} = \begin{bmatrix} 0 \\ 0 \\ 4 \\ 0 \end{bmatrix}$$

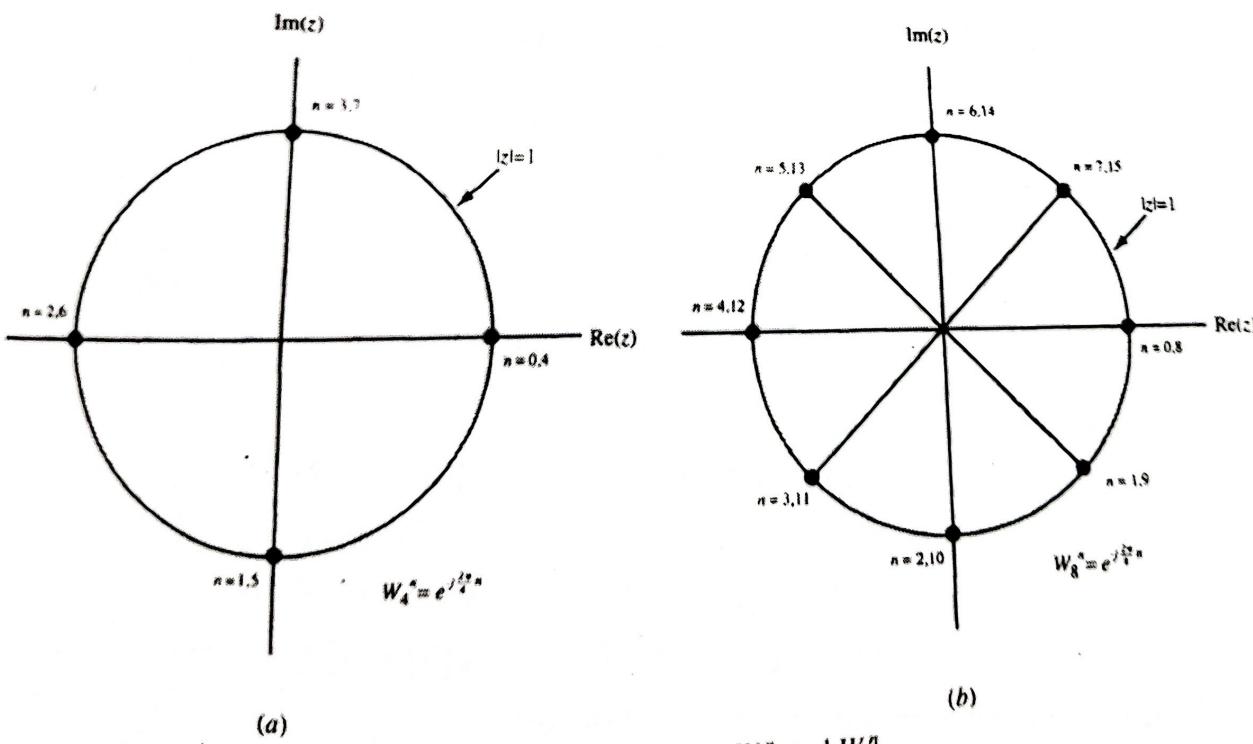


Fig. 6-38 Phase factors W_4^n and W_8^n .

and by Eqs. (6.217a) and (6.217b) we obtain

$$\begin{aligned} X[0] &= F[0] + W_8^0 G[0] = 0 \\ X[1] &= F[1] + W_8^1 G[1] = 2 + j2 \\ X[2] &= F[2] + W_8^2 G[2] = -j4 \\ X[3] &= F[3] + W_8^3 G[3] = 2 - j2 \end{aligned}$$

$$\begin{aligned} X[4] &= F[4] + W_8^4 G[0] = 0 \\ X[5] &= F[5] + W_8^5 G[1] = 2 + j2 \\ X[6] &= F[6] + W_8^6 G[2] = j4 \\ X[7] &= F[7] + W_8^7 G[3] = 2 - j2 \end{aligned}$$

Noting that since $x[n]$ is real and using Eq. (6.204), $X[7]$, $X[6]$, and $X[5]$ can be easily obtained by taking the conjugates of $X[1]$, $X[2]$, and $X[3]$, respectively.

6.59. Using the decimation-in-frequency FFT technique, redo Prob. 6.57.

From Prob. 6.57

$$x[n] = \{1, 1, -1, -1, -1, 1, 1, -1\}$$

By Eqs. (6.225a) and (6.225b) and using the values of W_8^n obtained in Prob. 6.57, we have

$$\begin{aligned} p[n] &= x[n] + x\left[n + \frac{N}{2}\right] \\ &= \{(1-1), (1+1), (-1+1), (-1-1)\} = \{0, 2, 0, 2\} \\ q[n] &= \left(x[n] - x\left[n + \frac{N}{2}\right]\right) W_8^n \\ &= \{(1+1)W_8^0, (1-1)W_8^1, (-1-1)W_8^2, (-1+1)W_8^3\} \\ &= \{2, 0, j2, 0\} \end{aligned}$$

Then using Eqs. (6.206) and (6.212), we have

$$\begin{aligned} \begin{bmatrix} P[0] \\ P[1] \\ P[2] \\ P[3] \end{bmatrix} &= \begin{bmatrix} 1 & 1 & 1 & 1 \\ 1 & -j & -1 & j \\ 1 & -1 & 1 & -1 \\ 1 & j & -1 & -j \end{bmatrix} \begin{bmatrix} 0 \\ 2 \\ 0 \\ -2 \end{bmatrix} = \begin{bmatrix} 0 \\ -j4 \\ 0 \\ j4 \end{bmatrix} \\ \begin{bmatrix} Q[0] \\ Q[1] \\ Q[2] \\ Q[3] \end{bmatrix} &= \begin{bmatrix} 1 & 1 & 1 & 1 \\ 1 & -j & -1 & j \\ 1 & -1 & 1 & -1 \\ 1 & j & -1 & -j \end{bmatrix} \begin{bmatrix} 2 \\ 0 \\ j2 \\ 0 \end{bmatrix} = \begin{bmatrix} 2+j2 \\ 2-j2 \\ 2+j2 \\ 2-j2 \end{bmatrix} \end{aligned}$$

and by Eqs. (6.226a) and (6.226b) we get

$$\begin{aligned} X[0] &= P[0] = 0 & X[4] &= P[2] = 0 \\ X[1] &= Q[0] = 2 + j2 & X[5] &= Q[2] = 2 + j2 \\ X[2] &= P[1] = -j4 & X[6] &= P[3] = j4 \\ X[3] &= Q[1] = 2 - j2 & X[7] &= Q[3] = 2 - j2 \end{aligned}$$

which are the same results obtained in Prob. 6.57.