

# A Design Study Approach to Classical Control

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Updated: May 17, 2016

## Homework B.11

The objective of this problem is to implement state feedback controller using the full state. Start with the simulation files developed in Homework ??.??.

- (a) Using the values for  $\omega_{n_z}$ ,  $\zeta_z$ ,  $\omega_{n_\theta}$ , and  $\zeta_\theta$  selected in Homework ??.??, find the desired closed loop poles.
- (b) Add the state space matrices  $A$ ,  $B$ ,  $C$ ,  $D$  derived in Homework ??.?? to your param file.
- (c) Verify that the state space system is controllable by checking that  $\text{rank}(\mathcal{C}) = n$ .
- (d) Find the feedback gain  $K$  so that the eigenvalues of  $(A - BK)$  are equal to desired closed loop poles. Find the reference gain  $k_r$  so that the DC-gain from  $z_r$  to  $z$  is equal to one.
- (e) Implement the state feedback scheme and tune the closed loop poles to get good response. You should be able to get much faster response using state space methods.

## Solution

From HW ??, the state space equations for the inverted pendulum are given by

$$\begin{aligned}\dot{x} &= \begin{pmatrix} 0 & 0 & 1.0000 & 0 \\ 0 & 0 & 0 & 1.0000 \\ 0 & -2.4500 & -0.0500 & 0 \\ 0 & 24.5000 & 0.1000 & 0 \end{pmatrix} x + \begin{pmatrix} 0 \\ 0 \\ 1 \\ -2 \end{pmatrix} u \\ y &= \begin{pmatrix} 1 & 0 & 0 & 0 \\ 0 & 1 & 0 & 0 \end{pmatrix} x,\end{aligned}\tag{1}$$

where Equation (1) represents the measured outputs. The reference output is the position  $z$ , which implies that

$$y_r = \begin{pmatrix} 1 & 0 & 0 & 0 \end{pmatrix} x.$$

**Step 1.** The controllability matrix is therefore

$$\mathcal{C}_{A,B} = [B, AB, A^2B, A^3B] = \begin{pmatrix} 0 & 1.0000 & -0.0500 & 4.9025 \\ 0 & -2.0000 & 0.1000 & -49.0050 \\ 1.0000 & -0.0500 & 4.9025 & -0.4901 \\ -2.0000 & 0.1000 & -49.0050 & 2.9403 \end{pmatrix}.$$

The determinant is  $\det(\mathcal{C}_{A,B}) = -1536 \neq 0$ , implying that the system is controllable.

**Step 2.** The open loop characteristic polynomial is

$$\Delta_{ol}(s) = \det(sI - A) = s^4 + 0.0500s^3 - 24.5000s^2 - 0.9800s$$

which implies that

$$\begin{aligned}\mathbf{a}_A &= (0.05, -24.5, 0.98, 0) \\ \mathcal{A}_A &= \begin{pmatrix} 1 & 0.05 & -24.5 & 0.98 \\ 0 & 1 & 0.05 & -24.5 \\ 0 & 0 & 1 & 0.05 \\ 0 & 0 & 0 & 1 \end{pmatrix}.\end{aligned}$$

**Step 3.** The desired closed loop polynomial is

$$\begin{aligned}\Delta_{cl}^d(s) &= (s^2 + 2\zeta_\theta\omega_{n_\theta}s + \omega_{n_\theta}^2)(s^2 + 2\zeta_z\omega_{n_z}s + \omega_{n_z}^2) \\ &= s^4 + 4.4280s^3 + 9.5248s^2 + 9.2280s + 4.0000\end{aligned}$$

which implies that

$$\boldsymbol{\alpha} = (4.4280, 9.5248, 9.2280, 4.0000).$$

**Step 4.** The gains are therefore given as

$$\begin{aligned}K &= (\boldsymbol{\alpha} - \mathbf{a}_A)\mathcal{A}_A^{-1}\mathcal{C}_{A,B}^{-1} \\ &= (-0.2041 - 17.1144 - 0.5208 - 2.4494)\end{aligned}$$

The feedforward reference gain  $k_r = -1/C_r(A - BK)^{-1}B$  is computed using  $C_r = (1, 0, 0, 0)$ , which gives

$$\begin{aligned}k_r &= \frac{-1}{C_r(A - BK)^{-1}B} \\ &= -0.2041.\end{aligned}$$

Alternatively, we could have used the following Matlab script

```
1 % state space model
2 A = [...
3     0, 0, 1, 0;...
4     0, 0, 0, 1;...
5     0, -P.m1*P.g/P.m2, -P.b/P.m2, 0;...
6     0, (P.m1+P.m2)*P.g/P.m2/P.e11, P.b/P.m2/P.e11, 0;...
7 ];
8 B = [0; 0; 1/P.m2; -1/P.m2/P.e11];
9 Cr = [1, 0, 0, 0];
10
11 % gains for pole locations
12 wn_th = 2;
13 zeta_th = 0.707;
14 wn_z = 1;
15 zeta_z = 0.8;
16 char_poly = conv([1, 2*zeta_z*wn_z, wn_z^2], ...
17                 [1, 2*zeta_th*wn_th, wn_th^2]);
18 des_poles = roots(char_poly);
```

```

19
20 % is the system controllable?
21 if rank(ctrb(A,B))≠4, disp('System Not Controllable'); end
22 P.K = place(A,B,des_poles);
23 P.kr = -1/(Cr*inv(A-B*P.K)*B);

```

The Matlab code for the controller is given by

```

1 function F=pendulum_ctrl(in,P)
2     z_d   = in(1);
3     z     = in(2);
4     theta = in(3);
5     t     = in(4);
6
7     % use a digital differentiator to find zdot and thetadot
8     persistent zdot
9     persistent z_d1
10    persistent thetadot
11    persistent theta_d1
12    % reset persistent variables at start of simulation
13    if t<P.Ts,
14        zdot      = 0;
15        z_d1      = 0;
16        thetadot  = 0;
17        theta_d1  = 0;
18    end
19    zdot = (2*P.tau-P.Ts)/(2*P.tau+P.Ts)*zdot...
20          + 2/(2*P.tau+P.Ts)*(z-z_d1);
21    thetadot = (2*P.tau-P.Ts)/(2*P.tau+P.Ts)*thetadot...
22             + 2/(2*P.tau+P.Ts)*(theta-theta_d1);
23    z_d1 = z;
24    theta_d1 = theta;
25
26    % construct the state
27    x = [z; theta; zdot; thetadot];
28    % compute the state feedback controller
29    F = sat( -P.K*x + P.kr*z_d, P.F_max);
30 end

```