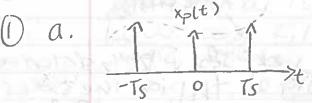
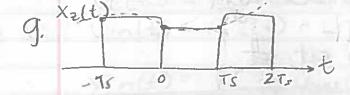
ps08



e. Use an Ideal low pass fitter with a cutoff frequency of wm to get X (w) which is x(t) because of the one-to-one Fourier transform property



$$\begin{array}{ll}
h. & \chi_{2}(w) = \int_{-\infty}^{\infty} z(t) e^{-jwt} dt \\
&= \int_{0}^{\infty} z(t) e^{-jwt} dt + \int_{0}^{\infty} z(t) e^{-jwt} dt + \int_{0}^{\infty} z(u) e^{-jwt} dt \\
&= \int_{0}^{\infty} z(t) e^{-jwt} dt + \int_{0}^{\infty} z(u) e^{-jwt} dt \\
&= \int_{0}^{\infty} z(t) e^{-jwt} dt + \int_{0}^{\infty} z(u) e^{-jwt} dt \\
&= \int_{0}^{\infty} z(t) e^{-jwt} dt + \int_{0}^{\infty} z(u) e^{-jwt} dt \\
&= \int_{0}^{\infty} z(t) e^{-jwt} dt + \int_{0}^{\infty} z(u) e^{-jwt} dt + \int_{0}^{\infty} z(u) e^{-jwt} dt \\
&= \int_{0}^{\infty} z(t) e^{-jwt} dt + \int_{0}^{\infty} z(u) e^{-jwt} dt + \int_{0}^{\infty} z(u) e^{-jwt} dt \\
&= \int_{0}^{\infty} z(t) e^{-jwt} dt + \int_{0}^{\infty} z(u) e^{-jwt} dt + \int_{0}^{\infty} z(u) e^{-jwt} dt \\
&= \int_{0}^{\infty} z(t) e^{-jwt} dt + \int_{0}^{\infty} z(u) e^{-jwt} dt + \int_{0}^{\infty} z(u) e^{-jwt} dt \\
&= \int_{0}^{\infty} z(t) e^{-jwt} dt + \int_{0}^{\infty} z(u) e^{-jwt} du + \int_{0}^{\infty} z(u) e^{-jw$$

Sin(图下等), TS = TS·Sinc(Ts·大·空前)

T·岩前 TS + S(t-誓) | X → TS·Sinc(Ts·岩)
X e-jw誓

LEFT , e-jw誓

Zw

Xz(w) time delay Signtly distarted cont. on the sides X(w), X(w) is slightly distorted, at a height of 1, and has a time delay of e-juits X (wis a perfect copy, at a beignt of ts, and has not ime delay They're both zero at 75 FFT of cos(wt) = a.y(w) = { (w-w,)+= 5(w+w,) added together Ь. -w2-w1 -w2 tw1 2w1

Multiply by cos and then apply an ideal Low pass filter to isolate the signal and C. divide by II to get the scaling right to produce the original signal Vinlt) = Velt) + VL(t) + Vout(t) blc KUL

V(t) = L d i(t) a. UP(t) = Rolt), bolo mornovodollos Vinlt) = Rilt) + L d i(t) + vout(t) Vin(t) = RC at Vout(t) + (Late Vout(t) + Vout(t) Vin(w) = RC jwVout(w) + CL(jw)2 Vout(w) + Vout(w) Vin(w) = Vout(w) RCjw-CLw2+1] Voutlw) = 1 Vinlw) RCjw-CLw2+1 4- H(w) V(RCW)2+(CLW2+1)2 = |H(W)| magnitude of Hlw) Vimaginary2 + real2

 $\frac{1}{dw} \left[\sqrt{(RCw)^2 + (CLw^2 + 1)^2} \right] = 0$ $\frac{-2C^2R^2w + 4CLw(CLw^2 + 1)}{2(C^2R^2w^2 + (CLw^2 + 1)^2)^3/2} = 0$ $w = \pm \sqrt{-CR^2 - 2L}$ $\sqrt{2}\sqrt{c} L$

e. See MATLAB code + graphs at end of assignment (colloboration credit to Jacob Riedle)

A Company (w) and (w)

- = 1 (with = = = (1+1 + (100) + (100)

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```
R = 400;
L = 10e-2;
C = 10e-7;

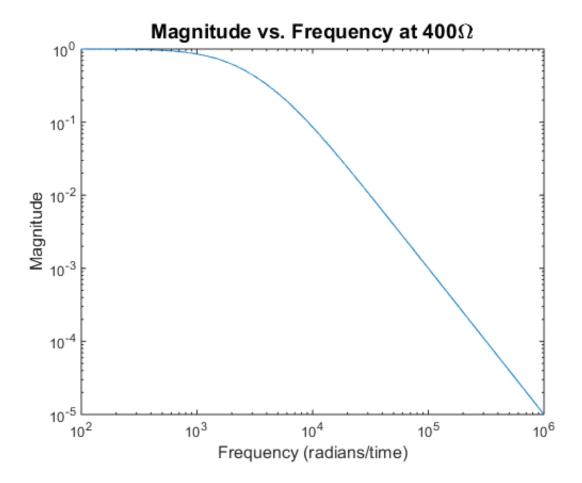
w = logspace (2,6,10000);

mag = 1./(sqrt((R.*C.*w).^2+(w.^2.*C.*L+1).^2));
phase = mag.*exp(-1i.*atan(w.*R.*C));

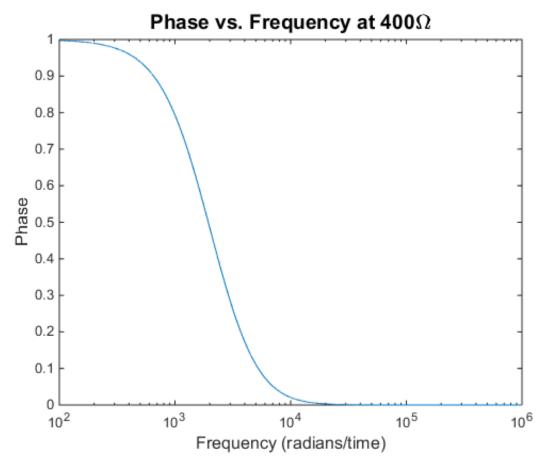
loglog (w,mag)
title('Magnitude vs. Frequency at 400\Omega','FontSize',14)
xlabel('Frequency (radians/time)','FontSize',12)
ylabel('Magnitude','FontSize',12)

figure;
semilogx(w,phase)
title('Phase vs. Frequency at 400\Omega','FontSize',14)
xlabel('Frequency (radians/time)','FontSize',12)
ylabel('Phase','FontSize',12)
```

Warning: Imaginary parts of complex X and/or Y arguments ignored



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```
R = 50;
L = 10e-2;
C = 10e-7;
w = logspace (-6,6,10000);
mag = 1./(sqrt((R.*C.*w).^2+(w.^2.*C.*L+1).^2));
phase = mag.*exp(-1i.*atan(w.*R.*C));
loglog(w,mag,'r')
title('Magnitude vs. Frequency at 50\Omega','FontSize',14)
xlabel('Frequency (radians/time)','FontSize',12)
ylabel('Magnitude','FontSize',12)
figure;
semilogx(w,phase,'r')
title('Phase vs. Frequency at 50\Omega','FontSize',14)
xlabel('Frequency (radians/time)','FontSize',12)
ylabel('Phase','FontSize',12)
```

Warning: Imaginary parts of complex X and/or Y arguments ignored

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