## Queensland University of Technology

## **EGB120**

Foundations of Electrical Engineering

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## **Electrical Circuits**

#### 1.1 Fundamental Units

Quantity	Definition	Symbol	SI Unit
Charge	Physical property of matter that causes it to experience a force when placed in an electromagnetic field	q	Coulomb (C)
Current	$i(t) = \frac{\mathrm{d}q}{\mathrm{d}t}$	i	Ampere (A)
Voltage	$v(t) = \frac{\mathrm{d}w}{\mathrm{d}q}$	v	Volt (V)
Power	$p(t) = \frac{\mathrm{d}w}{\mathrm{d}t} = \frac{\mathrm{d}q}{\mathrm{d}t} \times \frac{\mathrm{d}w}{\mathrm{d}q} = vi$	p	Watt (W)
Energy	$w(\tau) = \int_{\tau}^{0} p(t)dt$	e	Joule (J)

#### 1.2 Basic Circuit Elements

#### 1.2.1 Voltage Source

Produces or dissipates power at a specific voltage with whatever current is required

#### 1.2.2 Current Source

Produces or dissipates power at a specific current with whatever voltage is required

#### 1.2.3 Resistor

Dissipates power so that the voltage across the terminals is proportional to the current

$$v = Ri$$
 (Ohm's Law)

Following this

$$p = vi = Ri^2 = \frac{v^2}{R}$$

## Simple Circuits

#### 2.1 Physics Ignored

- Electrical effects are instantaneous
- Net charge on every component is zero
- No magnetic coupling between components

#### 2.2 Series

Elements are connected end to end and have the same current flowing through them



Figure 2.1: Series Circuit

#### 2.3 Parallel

Both ends of one element are connected directly and have the same voltage across them

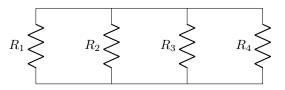


Figure 2.2: Parallel Circuit

#### 2.4 Kirchoff's Laws

#### 2.4.1 Kirchoff's Current Law

The sum of currents entering a node is equal to the sum of currents leaving a node

$$\sum_{k=1}^{n} i_k = 0$$

$$i_1 + i_2 + i_3 = 0$$

$$i_3$$

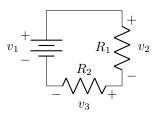
$$i_2$$

Figure 2.3: Kirchoff's Current Law

#### 2.4.2 Kirchoff's Voltage Law

The sum of voltages around a closed loop is zero





$$v_1 - v_2 - v_3 = 0$$
$$v_1 = v_2 + v_3$$

Figure 2.4: Kirchoff's Voltage Law

#### 2.5 Component Behaviours in Series and Parallel

Component	Series	Parallel
Voltage Source	$v_s = v_1 + v_2 + v_3$	$v_s = v_1 = v_2 = v_3$
Current Source	$i_s = i_1 = i_2 = i_3$	$i_s = i_1 + i_2 + i_3$
Resistor	$R_s = R_1 + R_2 + R_3$	$\frac{1}{R_s} = \frac{1}{R_1} + \frac{1}{R_2} + \frac{1}{R_3}$
Inductor	$L_s = L_1 + L_2 + L_3$	$\frac{1}{L_s} = \frac{1}{L_1} + \frac{1}{L_2} + \frac{1}{L_3}$
Capacitor	$\frac{1}{C_s} = \frac{1}{C_1} + \frac{1}{C_2} + \frac{1}{C_3}$	$C_s = C_1 + C_2 + C_3$

Figure 2.5: Component Behaviours in Series and Parallel

#### 2.6 Voltage Divider

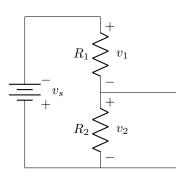


Figure 2.6: Voltage Divider

Current through resistors is

$$i = \frac{v_s}{R_1 + R_2}$$

Voltage across resistors is

$$v_1 = iR_1 = \frac{R_1}{R_1 + R_2} v_s$$

For more resistors

$$v_k = \frac{R_j}{R_{eq}} v_s$$

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#### 2.7 Current Divider

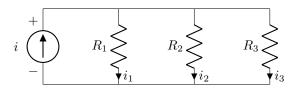


Figure 2.7: Current Divider

Voltage across resistors is

$$v = i(R_1||R_2||R_3)$$

Current through resistors is

$$i_j = \frac{v}{R_j} = \frac{R_1||R_2||R_3}{R_j}i$$

For more resistors

$$i_j = \frac{R_{eq}}{R_j}i$$

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## **Diodes**

Current flows from the anode to the cathode and the voltage across the diode is positive. A diode requires voltage to be applied across it to conduct current. This voltage is called the **forward voltage**.

#### 3.1 Voltage-Current Characteristics

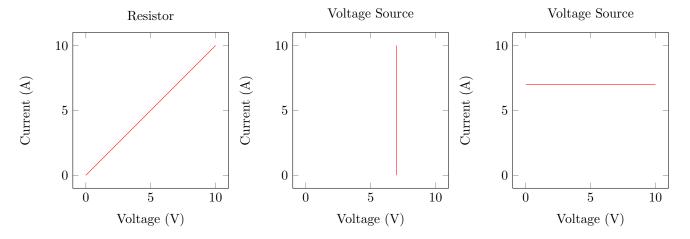


Figure 3.1: Diode VI Characteristics

#### 3.1.1 Simplified Diode Characteristics

Diodes have non-linear characteristics. Therefore, they are simplified to make calculations easier.

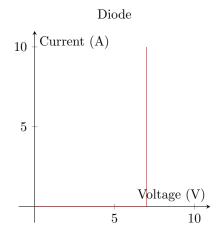


Figure 3.2: Simplified Diode Characteristics

#### 3.1.2 Full Diode Characteristics

Diodes can be accurately modelled using Shcokley's equation

$$i_D = I_s \left( e^{\frac{v_D}{V_T} - 1} \right) \qquad V_T = \frac{kT}{q}$$

where  $I_s$  is the saturation current,  $V_T$  is the thermal voltage, k is Boltzmann's constant, T is the temperature in Kelvin and q is the charge of an electron.

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## RL and RC Circuits

#### 4.1 Natural Response

Is a decaying response for an RL and RC circuit

$$v(t) = V_0 e^{-\frac{t}{RC}}$$

Figure 4.1: RC Natural Response

Figure 4.2: RL Natural Response

#### 4.2 Step Response

$$v(t) = I_s R + (V_0 - I_s R) e^{-\frac{t}{RC}}$$

(a) Equation for RC Circuit Step Response

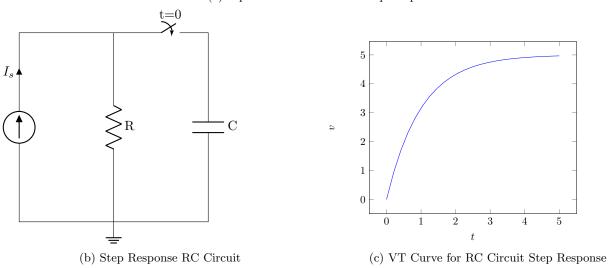


Figure 4.3: RC Step Response

$$i(t) = \frac{V_s}{R} + \left(I_o - \frac{V_s}{R}\right)e^{-\frac{Rt}{L}}$$

(a) Equation for RL Circuit Step Response

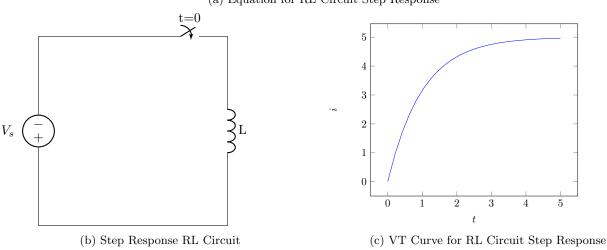


Figure 4.4: RL Step Response

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### $\mathbf{AC}$

#### 5.1 AC Circuit Analysis

Element	Impedance	Diagram
Resistor	R	
		$\mathbf{C}$
Capacitor	$\frac{1}{i\omega C}$ or $\omega L \angle 90^{\circ}$	
Inductor	$j\omega L$ or $\frac{1}{\omega L}\angle - 90^{\circ}$	

Figure 5.1: Impedance of Circuit Elements

#### 5.2 Bode Plots

Plot frequency on x axis using the log of frequency

- $\bullet$  1, 10, 100, 1000, 10000 are equally spaced
- 3, 30, 300, 3000, 30000 are midpoints

Plot the log of gain (decibels) on one y axis and the phase (degrees) on another y axis

$$\omega = 2\pi f \qquad \text{Gain}_{\text{dB}} = 20 \log_{10} \left( \text{Gain} \right) \qquad H(\omega) = \frac{V_o(\omega)}{V_i(\omega)} \qquad \text{Phase} = \angle H(\omega) \\ = \arctan \frac{\mathrm{imag}(H(\omega))}{\mathrm{real}(H(\omega))} \\ \text{(a) Hz to Radians per Second} \qquad \text{(b) Gain to Decibels} \qquad \text{(c) Transfer Function} \qquad \text{(d) Transfer Function to Phase} \\ \text{Gain} = |(H(\omega))| \\ = \sqrt{\left(\mathrm{real}(H(\omega))\right)^2 + \left(\mathrm{imag}(H(\omega))\right)^2} \\ \text{(e) Transfer Function to Gain} \\ \text{Figure 5.2: Bode Plot Equations}$$

5.3 Transfer Function  $H(\omega)$  - The 3dB Point

The 3dB point is the frequency at which the gain is  $\frac{1}{\sqrt{2}}$  of the maximum gain. Also called the **half power point** or **half power frequency** or **break frequency** or **corner frequency** 

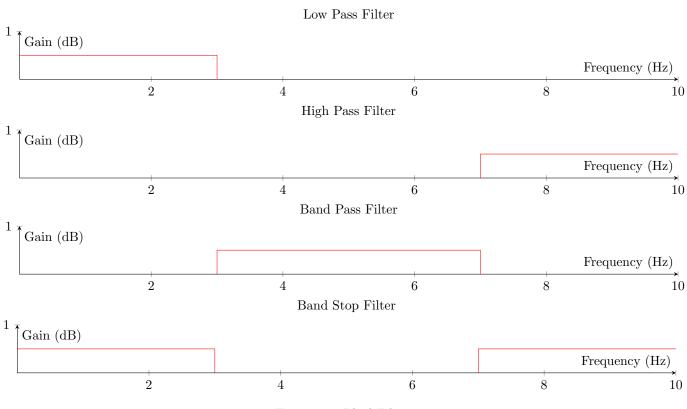
## Filters and Rectifiers

Filters keep the desired frequency components of a signal and remove the unwanted frequency components. They do this in the frequency domain

#### 6.1 Applications of Filters

- Audio Signals
  - Remove high frequency hiss (magnetic tape)
  - Remove low frequency rumble (vinyl)
- Medical Signals
  - EEG alpha waves are 8-12Hz
  - EEG beta waves are 12-30Hz
  - ECG waves are 1 40Hz
- 50Hz interference
  - Remove all signals near 50Hz
- Signal Processing

#### 6.2 Ideal Filters

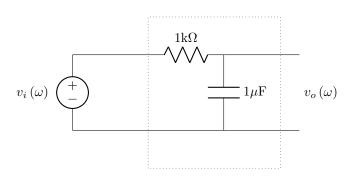


#### 6.3 Passive Filters

#### 6.3.1 Low Pass Filter

#### 

Figure 6.2: Low Pass Filter



$$v_o(\Omega) = \frac{\frac{1}{j\omega C}}{R + \frac{1}{j\omega C}} v_i(\omega)$$
$$\frac{v_0(\omega)}{v_i(\omega)} = \frac{1}{j\omega RC + 1}$$

For  $R = 1 \mathrm{k}\Omega$  and  $C = 1 \mu F$ 

$$H(\omega) = \frac{1000}{j\omega + 1000}$$
 
$$H(62.8) \approx 1 \qquad \qquad H(62800) = 0.016 \angle - 89^{\circ}$$

#### Plotting the Response

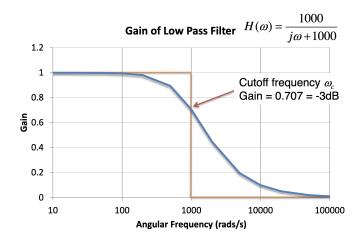
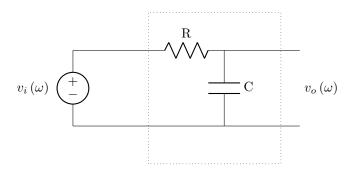


Figure 6.3: Low Pass Filter Bode Plot

#### Designing a Low Pass Filter

Find the required **cutoff frequency**  $\omega_c$  in radians per second (remember  $\omega_c = 2\pi f$ ) Choose a value for R and calculate C such that  $\omega_c = \frac{1}{RC}$ 

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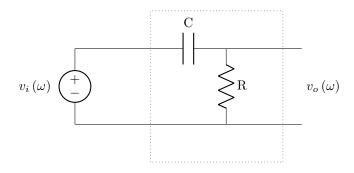


$$H(\omega) = \frac{\frac{1}{RC}}{j\omega + \frac{1}{RC}}$$
$$\omega_c = \frac{1}{RC}$$

#### 6.3.2 High Pass Filter

Similar to a low pass filter, but the capacitor and resistor are swapped

#### Desigining a High Pass Filter



$$v_o(\omega) = \frac{R}{R + \frac{1}{j\omega C}} v_i(\omega)$$
$$\frac{v_0(\omega)}{v_i(\omega)} = H(\omega) = \frac{j\omega RC}{j\omega RC + 1}$$
$$\omega_c = \frac{1}{RC}$$

#### Plotting the Response

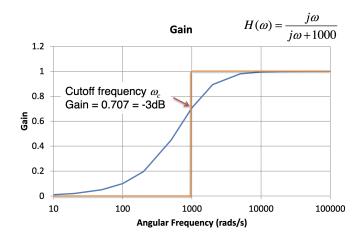


Figure 6.4: High Pass Filter Bode Plot

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#### 6.4 Active Filters

#### 6.4.1 Op-Amps

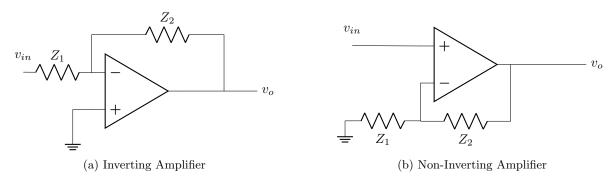
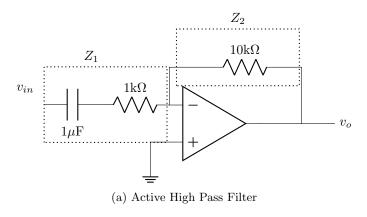


Figure 6.5: Active Filters

#### 6.4.2 Using an Op Amp

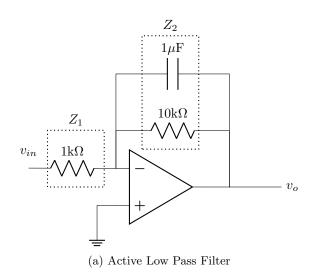
#### Active High Pass Filter



# $$\begin{split} v_o &= -\frac{Z_2}{Z_1} v_{in} \\ \frac{v_o}{v_{in}} &= -\frac{R_2}{R_1 + \frac{1}{j\omega C_1}} = -\frac{R_2}{R_1} \frac{j\omega}{j\omega + \frac{1}{R_1 C_1}} \end{split}$$

High pass filter with  $\omega_c = \frac{1}{R_1 C_1} = \frac{1}{10^3 \times 10^{-6}} = 1000 \text{rad/s}$  Gain of  $-\frac{R_2}{R_1} = -10$  in passband

#### Active Low Pass Filter



$$\begin{aligned} v_o &= -\frac{Z_2}{Z_1} v_{in} \\ &= \frac{\frac{1}{\frac{1}{R_2} + j\omega C_2}}{R_1} v_{in} \\ &\frac{v_o}{v_{in}} = H(\omega) = -\frac{R_2}{R_1} \frac{1}{1 + j\omega R_2 C_2} \\ &= -\frac{R_2}{R_1} \frac{\frac{1}{R_2 C_2}}{j\omega + \frac{1}{R_2 C_2}} \end{aligned}$$

Low pass filter with  $\omega_c = \frac{1}{R_2C} = \frac{1}{10^4 \times 10^{-6}} = 100 \text{rad/s}$  Gain of  $-\frac{R_2}{R_1} = -10$  in passband

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#### 6.5 Band Pass and Band Stop Filters

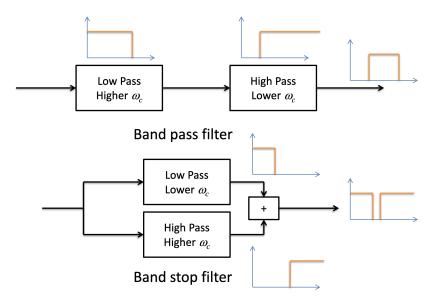


Figure 6.8: Band Pass and Band Stop Filters

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