



## Assumptions:

- The quadrotor is modeled as a rigid body.
- Only vertical translation is considered for translational motion.
- 1st order motor model.
- Sensor noises are included.
- Wind disturbances are applied both as vertical force and torques.

## Rotor Thrust Mapping

$$A = \begin{bmatrix} 1 & 1 & 1 & 1 \\ 0 & -l & 0 & l \\ l & 0 & -l & 0 \\ c & -c & c & -c \end{bmatrix}, \quad b = \begin{bmatrix} T_{total} \\ \tau_\phi \\ \tau_\theta \\ \tau_\psi \end{bmatrix}$$

$$T_{cmd} = A^{-1}b$$

$$\dot{T} = \frac{T_{cmd} - T_{act}}{\tau_{motor}}$$

$$T_{act} = T_{act} + \dot{T} \times dt$$

$\tau$  : motor time constant

## Total Force and Moments

$$F_z = \sum T_{act} + F_{z,wind}$$

$$\tau_\phi^{act} = l \times (-T_{act}(2) + T_{act}(4)) + \tau_{\phi,wind}$$

$$\tau_\theta^{act} = l \times (T_{act}(1) - T_{act}(3)) + \tau_{\theta,wind}$$

$$\tau_\psi^{act} = c \times (T_{act}(1) - T_{act}(2) + T_{act}(3) - T_{act}(4)) + \tau_{\psi,wind}$$

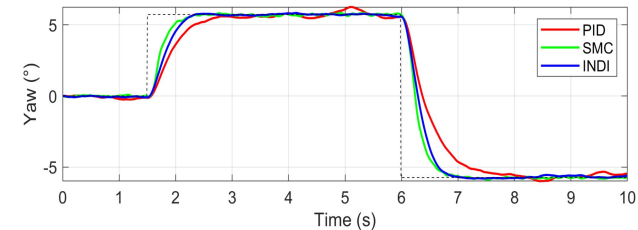
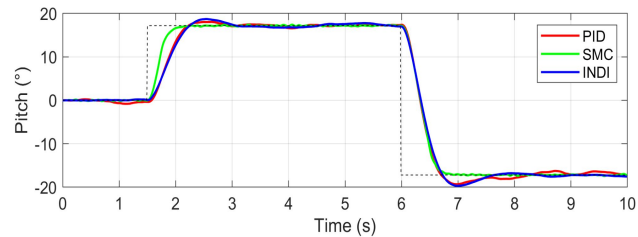
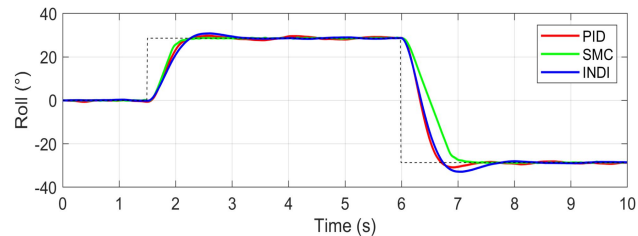
## Accelerations

$$a_z = F_z/m - g$$

$$\dot{p} = \tau_\phi^{act} / I_{xx}$$

$$\dot{q} = \tau_\theta^{act} / I_{yy}$$

$$\dot{r} = \tau_\psi^{act} / I_{zz}$$





## Outer loop for angle control

$$p_{\text{des}} = K_p^\phi (\phi_{\text{des}} - \phi_{\text{meas}})$$

$$q_{\text{des}} = K_p^\theta (\theta_{\text{des}} - \theta_{\text{meas}})$$

$$r_{\text{des}} = K_p^\psi (\psi_{\text{des}} - \psi_{\text{meas}})$$

## Inner loop for angular velocity control

$$\tau_\phi = I_{xx} \left( K_p^p e_p + K_i^p \int e_p dt + K_d^p \dot{e}_p \right)$$

$$\tau_\theta = I_{yy} \left( K_p^q e_q + K_i^q \int e_q dt + K_d^q \dot{e}_q \right)$$

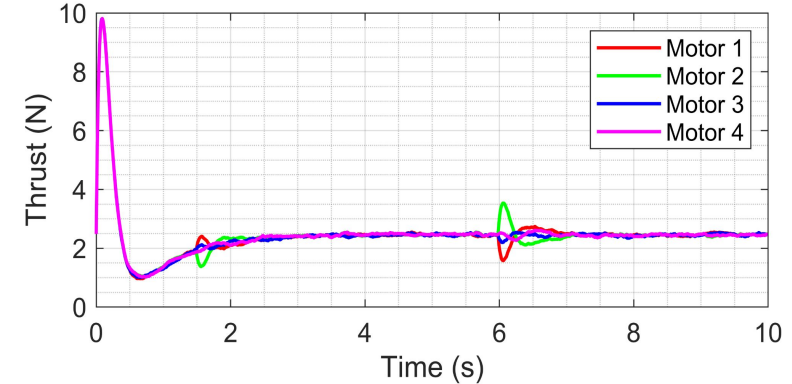
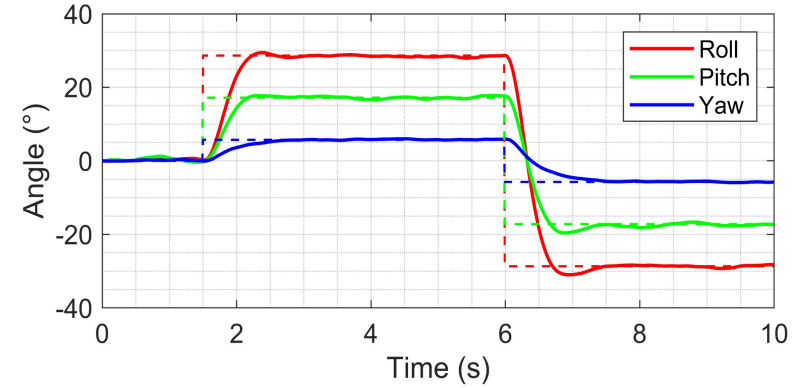
$$\tau_\psi = I_{zz} \left( K_p^r e_r + K_i^r \int e_r dt + K_d^r \dot{e}_r \right)$$

$$e_p = p_{\text{des}} - p_{\text{meas}}$$

$$\text{where, } e_q = q_{\text{des}} - q_{\text{meas}}$$

$$e_r = r_{\text{des}} - r_{\text{meas}}$$

Results - PID Controller





## Description:

Sliding mode control (SMC) is a robust control method that works by pulling the system dynamics to a predetermined “sliding surface (s)”. When the system remains on this surface, it shows high resistance to uncertainties and disturbances.

## Pros and cons:

Its advantages include its simple structure, fast response and robustness to parameter changes. However, it can lead to some disadvantages in practice due to the generation of high-frequency chattering and the continuous switching of the control signal.

## Angle Control

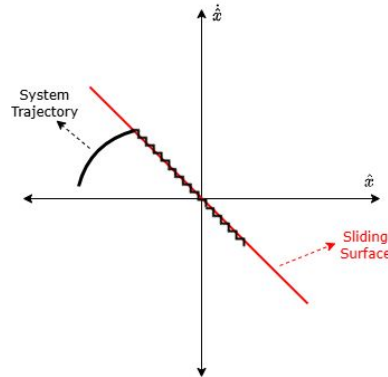
$$e_\phi = \phi_{\text{des}} - \phi_{\text{meas}}$$

$$\dot{e}_\phi = -p_{\text{meas}}$$

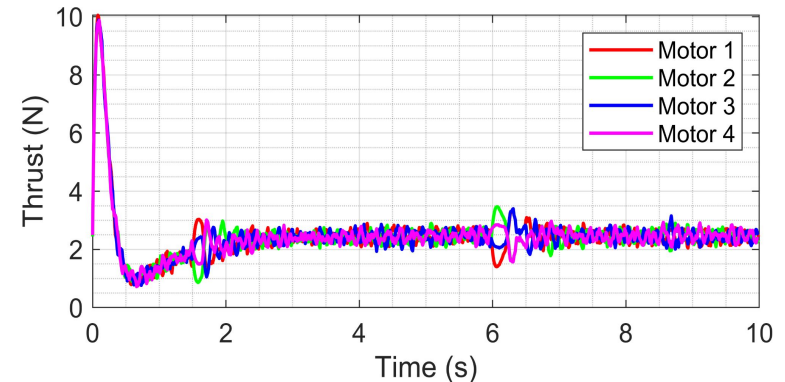
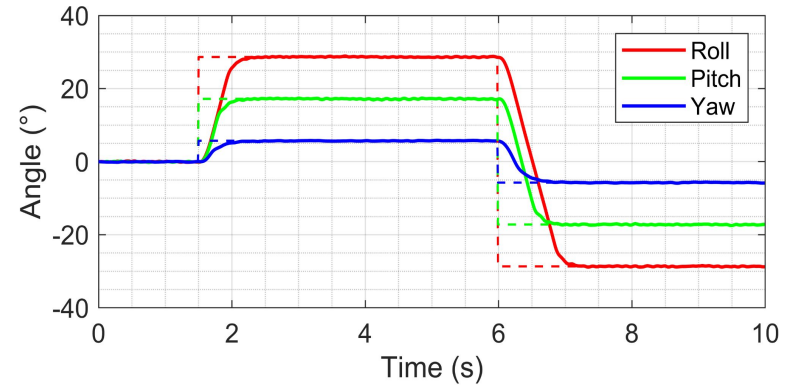
$$s_\phi = \lambda_\phi e_\phi + \dot{e}_\phi$$

$$\tau_\phi = I_{xx}(\lambda_\phi \dot{e}_\phi + K_\phi \text{sat}(s_\phi, 0.1))$$

... Same for pitch and yaw angles.



Results - SMC Controller



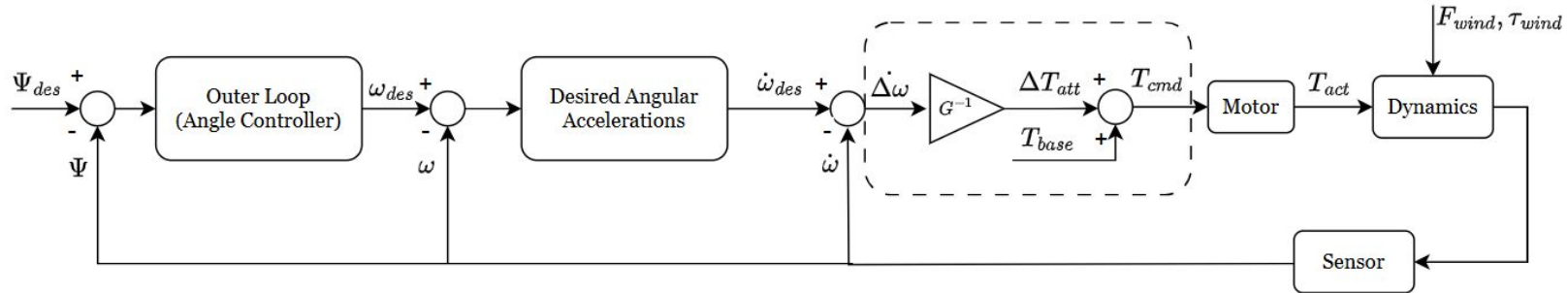


## Description:

Incremental Nonlinear Dynamic Inversion (INDI) is a variation of the traditional Nonlinear Dynamic Inversion (NDI) method and is an advanced control technique used especially in aircrafts.

The basic principle of INDI is to incrementally compute the control action based on measurable quantities such as acceleration without the need for a full dynamic model of the system. This provides high robustness against model uncertainties and external disturbances.

While in the NDI method, the control input is inversely computed over the full model of the system, in INDI only the variation of the dynamics is taken into account, which simplifies the controller design and reduces the problems caused by model errors.





## 1. Outer Loop

$$p_{\text{des}} = K_p \phi (\phi_{\text{des}} - \phi_{\text{meas}})$$

$$q_{\text{des}} = K_p \theta (\theta_{\text{des}} - \theta_{\text{meas}})$$

$$r_{\text{des}} = K_p \psi (\psi_{\text{des}} - \psi_{\text{meas}})$$

## 2. Desired Angular Accelerations

$$\dot{p}_{\text{des}} = K_p e_p + K_i \int e_p dt + K_d \dot{e}_p$$

$$\dot{q}_{\text{des}} = K_p e_q + K_i \int e_q dt + K_d \dot{e}_q$$

$$\dot{r}_{\text{des}} = K_p e_r + K_i \int e_r dt + K_d \dot{e}_r$$

## 3. Calculate Control Increments

$$\Delta \dot{p} = \dot{p}_{\text{des}} - \dot{p}$$

$$\Delta \dot{q} = \dot{q}_{\text{des}} - \dot{q}$$

$$\Delta \dot{r} = \dot{r}_{\text{des}} - \dot{r}$$

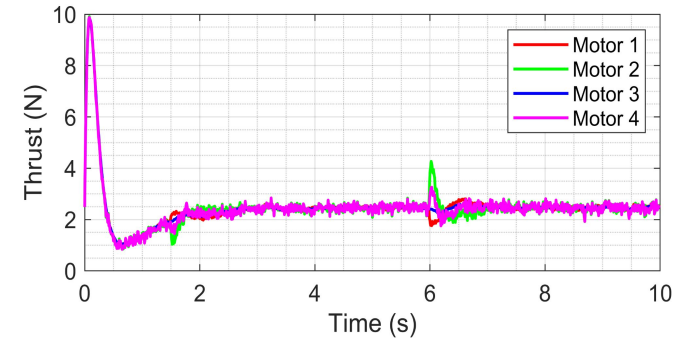
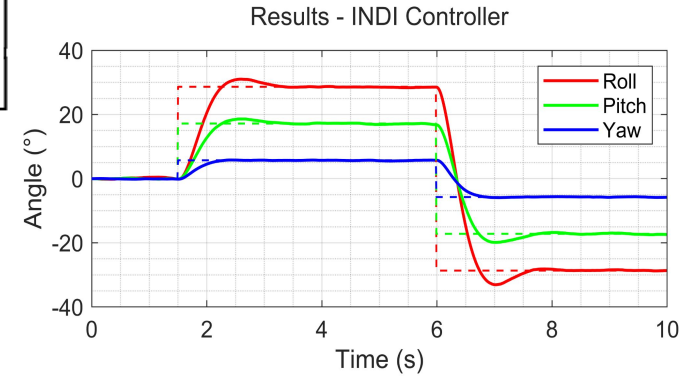
## 4. Calculate Thrust Increments

$$\Delta T_{\text{att}} = \underbrace{\begin{bmatrix} 0 & -\frac{l}{I_{zz}} & 0 & \frac{l}{I_{zz}} \\ \frac{l}{I_{yy}} & 0 & -\frac{l}{I_{yy}} & 0 \\ \frac{c}{I_{zz}} & -\frac{c}{I_{zz}} & \frac{c}{I_{zz}} & -\frac{c}{I_{zz}} \end{bmatrix}^{-1}}_G \begin{bmatrix} \Delta \dot{p} \\ \Delta \dot{q} \\ \Delta \dot{r} \end{bmatrix}$$

## 5. Combine Control Law

$$T_{\text{base}} = \frac{T_{\text{total}}}{4} \times [1; 1; 1; 1]$$

$$T_{\text{cmd}} = T_{\text{base}} + \Delta T_{\text{att}}$$





The aim of this study was to give an overview of PID, SMC and INDI control algorithms and to show their main differences in a simulation environment. The simulations gave stable results under nominal conditions, but extensive testing was not done, especially for fault scenarios and larger external disturbances.

In the future, the performance of these algorithms in more challenging and fault conditions can also be investigated.

## References

Herrera, Marco & Chamorro, William & Gómez, Alejandro & Camacho, Oscar. (2015). Sliding Mode Control: An Approach to Control a Quadrotor. 10.1109/APCASE.2015.62.

Coelho, Rodrigo & Moutinho, Alexandra & Azinheira, José. (2017). Quadrotor Attitude Control using Incremental Nonlinear Dynamics Inversion. 98-109. 10.5220/00064359000980109.

Da Silva, E. L. (2015). Incremental Nonlinear Dynamic Inversion for Quadrotor Control (Doctoral dissertation, Master's thesis, Delft University of Technology).