# ECE 161A: The Discrete Fourier Transform (DFT)

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### Discrete Time Fourier Transform of Periodic Sequences

$$\tilde{x}[n] \stackrel{?}{\longleftrightarrow} \tilde{X}(e^{j\omega})$$

We will use the following observation about exponential sequences

$$e^{j\omega_0 n} \longleftrightarrow 2\pi \sum_{r=-\infty}^{\infty} \delta(\omega - \omega_0 + 2\pi r)$$

and so

$$W_N^{-nk} = e^{j\frac{2\pi}{N}kn} \longleftrightarrow 2\pi \sum_{r=-\infty}^{\infty} \delta(\omega - \frac{2\pi}{N}k + 2\pi r)$$

Hence

$$\tilde{x}[n] = \frac{1}{N} \sum_{k=0}^{N-1} \tilde{X}[k] e^{j\frac{2\pi}{N}kn} \stackrel{linearity}{\longleftrightarrow} \frac{2\pi}{N} \sum_{k=0}^{N-1} \tilde{X}[k] \sum_{r=-\infty}^{\infty} \delta(\omega - \frac{2\pi}{N}k + 2\pi r)$$

$$= \frac{2\pi}{N} \sum_{k=-\infty}^{\infty} \tilde{X}[k] \delta(\omega - \frac{2\pi}{N}k)$$

$$\tilde{x}[n] = \frac{1}{N} \sum_{k=0}^{N-1} \tilde{X}[k] e^{j\frac{2\pi}{N}kn} \stackrel{linearity}{\longleftrightarrow} = \frac{2\pi}{N} \sum_{k=0}^{\infty} \tilde{X}[k] \delta(\omega - \frac{2\pi}{N}k)$$

DTFT of periodic sequences

$$\tilde{x}[n] \longleftrightarrow \tilde{X}[k] \longleftrightarrow \frac{2\pi}{N} \sum_{k=-\infty}^{\infty} \tilde{X}[k] \delta(\omega - \frac{2\pi}{N}k)$$

A DTFT which has delta functions uniformly spaced implies a periodic time domain sequence.

$$\frac{2\pi}{N} \sum_{k=0}^{\infty} \tilde{X}[k]\delta(\omega - \frac{2\pi}{N}k) \longleftrightarrow \tilde{X}[k] \longleftrightarrow \tilde{x}[n]$$

### **DFS Summary**

With  $W_N = e^{-j\frac{2\pi}{N}}$ 

$$\begin{split} \tilde{x}[n] &= \frac{1}{N} \sum_{k=0}^{N-1} \tilde{X}[k] e^{j\frac{2\pi}{N}kn} = \frac{1}{N} \sum_{k=0}^{N-1} \tilde{X}[k] W_N^{-nk} \quad \text{(Synthesis Equation)} \\ \tilde{X}[k] &= \sum_{k=0}^{N-1} \tilde{x}[n] e^{-j\frac{2\pi}{N}kn} = \sum_{k=0}^{N-1} \tilde{x}[n] W_N^{nk} \quad \text{(Analysis Equation)} \end{split}$$

Fourier Transform (DTFT of  $\tilde{x}[n]$ )

$$\tilde{x}[n] = \frac{1}{N} \sum_{k=0}^{N-1} \tilde{X}[k] e^{j\frac{2\pi}{N}kn} \stackrel{linearity}{\longleftrightarrow} = \frac{2\pi}{N} \sum_{k=-\infty}^{\infty} \tilde{X}[k] \delta(\omega - \frac{2\pi}{N}k)$$

Procedure for computing DTFT of periodic sequences

$$\tilde{x}[n] \longleftrightarrow \tilde{X}[k] \longleftrightarrow \frac{2\pi}{N} \sum_{k=-\infty}^{\infty} \tilde{X}[k] \delta(\omega - \frac{2\pi}{N}k)$$

A DTFT which has delta functions uniformly spaced implies a periodic time domain sequence.

$$\frac{2\pi}{N} \sum_{k=0}^{\infty} \tilde{X}[k] \delta(\omega - \frac{2\pi}{N}k) \longleftrightarrow \tilde{X}[k] \longleftrightarrow \tilde{x}[n]$$



### Sampling the DTFT

$$X(e^{j\omega}) = \sum_{n=-\infty}^{\infty} x[n]e^{-j\omega n}$$

Sample using a train of impulses in the frequency domain

$$P(e^{j\omega}) = \frac{2\pi}{N} \sum_{k=-\infty}^{\infty} \delta(\omega - \frac{2\pi}{N}k)$$

N controls the spacing  $\frac{2\pi}{N}$  of the frequency samples.The sampled DTFT is given by

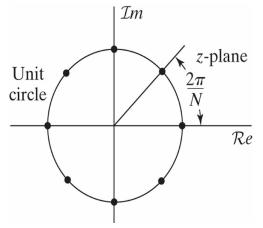
$$X_{s}(e^{j\omega}) = X(e^{j\omega})P(e^{j\omega}) = X(e^{j\omega})\frac{2\pi}{N}\sum_{k=-\infty}^{\infty}\delta(\omega - \frac{2\pi}{N}k)$$
$$= \frac{2\pi}{N}\sum_{k=-\infty}^{\infty}X(e^{j\frac{2\pi k}{N}})\delta(\omega - \frac{2\pi}{N}k) = \frac{2\pi}{N}\sum_{k=-\infty}^{\infty}\tilde{X}[k]\delta(\omega - \frac{2\pi}{N}k)$$

where  $\tilde{X}[k] = X(e^{j\frac{2\pi k}{N}})$ , the samples of the DTFT of x[n] on the unit circle at a spacing of  $\frac{2\pi}{N}$ .



$$X(e^{j\omega}) = \sum_{n=-\infty}^{\infty} x[n]e^{-j\omega n} \xrightarrow{N} \frac{2\pi}{N} \sum_{k=-\infty}^{\infty} \tilde{X}[k]\delta(\omega - \frac{2\pi}{N}k) \leftrightarrow \left(N, \tilde{X}[k] = X(e^{j\frac{2\pi k}{N}}), k = 0, 1, \dots, N-1\right)$$

Figure 8.7 Points on the unit circle at which X(z) is sampled to obtain the periodic sequence  $\tilde{X}[k]$  (N = 8).



### Frequency Sampling

$$x[n] \leftrightarrow X(e^{j\omega}) = \sum_{n=-\infty}^{\infty} x[n]e^{-j\omega n} \xrightarrow{N} X_{s}(e^{j\omega}) = \frac{2\pi}{N} \sum_{k=-\infty}^{\infty} \tilde{X}[k]\delta(\omega - \frac{2\pi}{N}k)$$

$$\leftrightarrow \left(N, \tilde{X}[k] = X(e^{j\frac{2\pi k}{N}}), k = 0, 1, \dots, N-1\right) \leftrightarrow \tilde{x}[n]$$

#### Important questions

- 1. What is the relationship  $\tilde{x}[n]$  and x[n]?
- 2. Can we recover x[n] from  $\tilde{x}[n]$ ?

Significance: If we can answer yes to the second question, then  $\tilde{X}[k]$  has all the information needed to determine  $X(e^{j\omega})$ .

### Frequency Sampling: Pictorial Depiction

Suppose x[n] is of duration L, i.e. x[n] is nonzero for n = 0, 1, ..., L - 1.

The frequency sampling pictorially is shown below.

$$x[n] \leftrightarrow X(e^{j\omega}) = \sum_{n=0}^{L-1} x[n] e^{-j\omega n} \stackrel{N}{\longleftrightarrow} X_s(e^{j\omega}) = \frac{2\pi}{N} \sum_{k=-\infty}^{\infty} \tilde{X}[k] \delta(\omega - \frac{2\pi}{N} k)$$
Duration L

Only N≥L

$$\tilde{X}[k] = X(e^{j\omega})|_{\omega = \frac{2\pi}{N} k, k = 0, 1, \dots, N-1}$$

DFS N

$$\tilde{x}[n], n = 0, 1, \dots, N-1$$

# Frequency Sampling Theorem (FST)

$$x[n] \leftrightarrow X(e^{j\omega}) \xrightarrow{N} X_s(e^{j\omega}) = \frac{2\pi}{N} \sum_{k=-\infty}^{\infty} \tilde{X}[k] \delta(\omega - \frac{2\pi}{N}k) \leftrightarrow \tilde{X}[k] = X(e^{j\frac{2\pi k}{N}}) \leftrightarrow \tilde{x}[n]$$

#### Theorem:

- 1.  $\tilde{x}[n] = \sum_{r=-\infty}^{\infty} x[n+rN]$  (time domain aliasing)
- 2. If x[n] is of finite duration L, then for  $N \ge L x[n]$  can be recovered from  $\tilde{x}[n]$  and

$$x[n] = \tilde{x}[n]w[n]$$

where w[n] is rectangular window

$$w[n] = \begin{cases} 1, & 0 \le k \le N - 1 \\ 0, & \text{otherwise} \end{cases}$$

# Recovery of x[n] from $\tilde{x}[n]$

$$\tilde{x}[n] = \sum_{r=-\infty}^{\infty} x[n+rN] = \ldots + x[n-N] + x[n] + x[n+N] + \ldots$$

If x[n] is of duration L, then x[n] is nonzero between  $0 \le n \le L-1$ . x[n-N] is nonzero between  $N \le n \le N+L-1$ . x[n+N] is nonzero between  $-N \le n \le -N+L-1$ .

No overlap and hence no aliasing for  $N \ge L$ .

x[n] can be obtained by keeping samples of  $\tilde{x}[n], 0 \le n \le N-1$ . Mathematically

$$x[n] = \tilde{x}[n]w[n],$$

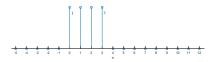
where w[n] is rectangular window

$$w[n] = \begin{cases} 1, & 0 \le k \le N - 1 \\ 0, & \text{otherwise} \end{cases}$$



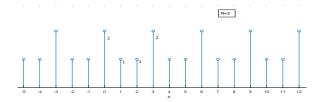
### Example





Three samples (N = 3)in the frequency domain.

$$\tilde{x}[n] = \sum_{r=-\infty}^{\infty} x[n-r3] = \begin{cases} x[0] + x[3], & n = .., -3, 0, 3, ... \\ x[1], & n = .., -2, 1, 4, ... \\ x[2], & n = , .. -1, 2, 5, ... \end{cases}$$



### Example Cont'd

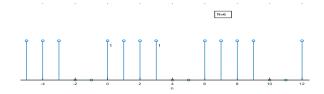




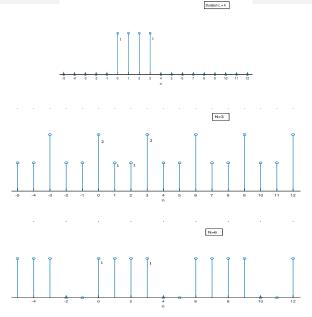
Six samples (N = 6)in the frequency domain.

$$\tilde{x}[n] = \sum_{r=-\infty}^{\infty} x[n-r6]$$

A period of  $\tilde{x}[n]$  is x[n] padded with 2 (6 – 4) zeros.



### Example: Summary



# Discrete Time Fourier Transform (DFT)

Define

$$X[k] = \left\{ egin{array}{ll} ilde{X}[k] = X(e^{jrac{2\pi}{N}k}), & k = 0, 1, \ldots, N-1 \ 0, & ext{otherwise}. \end{array} 
ight.$$

We can now rewrite the equations involving only x[n] and X[k].

$$X[k] = \sum_{n=0}^{N-1} x[n] e^{-j\frac{2\pi}{N}kn} = \sum_{n=0}^{N-1} x[n] W_N^{nk}, k = 0, 1, \dots, N-1 \text{ (DFT)}$$

$$x[n] = \frac{1}{N} \sum_{k=0}^{N-1} X[k] e^{j\frac{2\pi}{N}kn} = \frac{1}{N} \sum_{k=0}^{N-1} X[k] W_N^{-nk}, n = 0, 1, \dots, N-1 \text{ (IDFT)}$$

### DFS versus DFT

DFS:

$$\begin{split} \tilde{X}[k] &= \sum_{n=0}^{N-1} \tilde{x}[n] e^{-j\frac{2\pi}{N}kn} = \sum_{n=0}^{N-1} \tilde{x}[n] W_N^{nk} \\ \tilde{x}[n] &= \frac{1}{N} \sum_{k=0}^{N-1} \tilde{X}[k] e^{j\frac{2\pi}{N}kn} = \frac{1}{N} \sum_{k=0}^{N-1} \tilde{X}[k] W_N^{-nk} \end{split}$$

DFT:

$$\begin{split} X[k] & = \sum_{n=0}^{N-1} x[n] e^{-j\frac{2\pi}{N}kn} = \sum_{n=0}^{N-1} x[n] W_N^{nk}, \, k = 0, 1, \dots, N-1 \text{ (DFT)} \\ x[n] & = \frac{1}{N} \sum_{k=0}^{N-1} X[k] e^{j\frac{2\pi}{N}kn} = \frac{1}{N} \sum_{k=0}^{N-1} X[k] W_N^{-nk}, \, n = 0, 1, \dots, N-1 \text{ (IDFT)} \end{split}$$

- 1. DFS applies to periodic signals and DFT to finite duration signals.
- 2. In DFS, N is the periodicity and in DFT  $N \ge L$ , where L is the duration of the sequence.