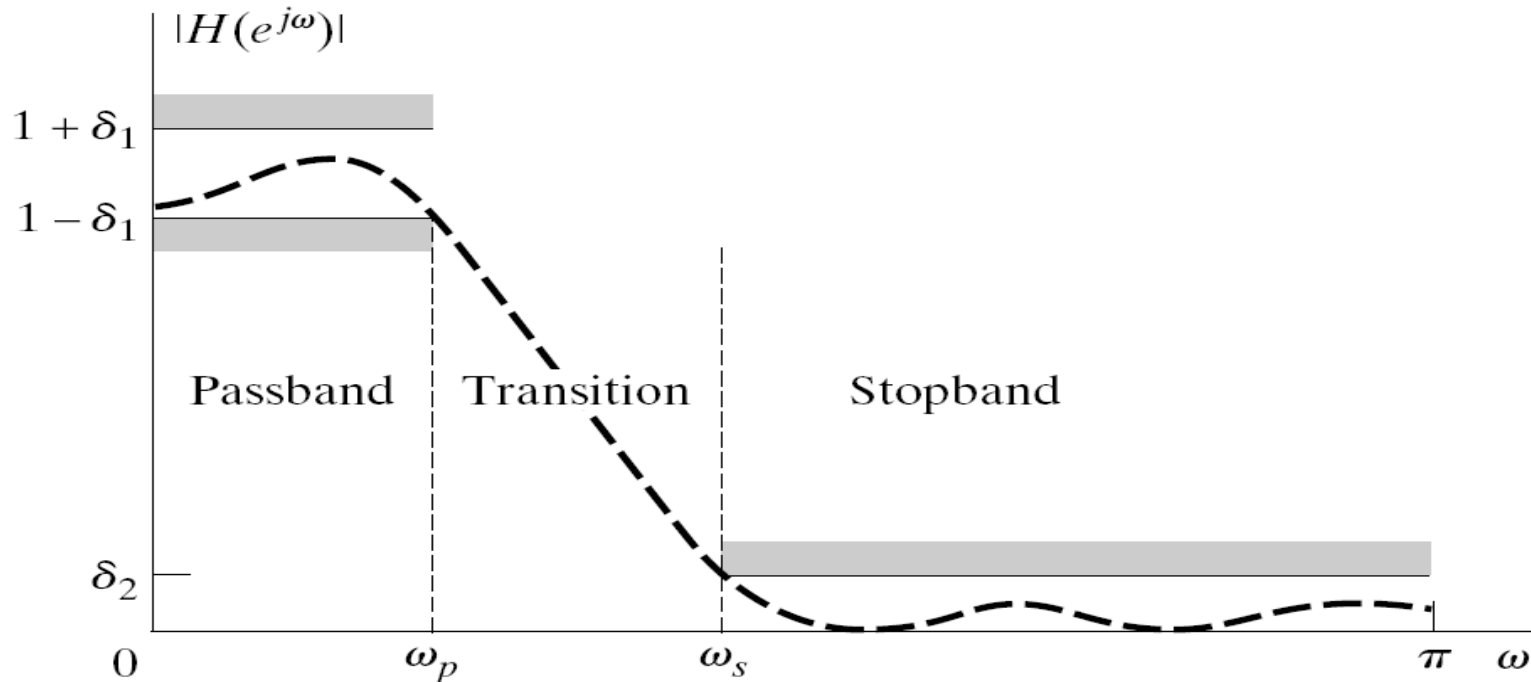


IIR Filter Design

Lecture By Prof. Meyer

ECE 161A

Filter Design: Low Pass Filter Design



Problem: Given $\delta_1, \delta_2, \omega_p$ and ω_s , find the lowest complexity filter that meets specification

Two Choices

1. IIR Filters (Infinite Impulse Response, $H(z) = \frac{B(z)}{A(z)}$)
2. FIR Filters (Finite Impulse Response, $H(z) = B(z)$)

IIR Filter Design

- Approach: Transform a continuous time filter to a discrete time filter

Rationale:

- Continuous time IIR filter design highly advanced.
- Many continuous time IIR filters have closed form design formulas
- Approximating directly in the digital domain not easy.

IIR Filter Design Methods

1. Impulse Invariance
2. Bilinear Transformation

Another useful Tool: Frequency Transformation.
It allows design of other filter types (High Pass, Band Pass...) by transforming a low pass filter.

Impulse Invariance Design

Simply applies the sampling theorem and hopes that aliasing is small

$$h[n] = T_d h_c(nT_d)$$

Then

$$H(e^{j\omega}) = \sum_{k=-\infty}^{\infty} H_c(j\frac{\omega}{T_d} - j\frac{2\pi}{T_d}k)$$

More generally

$$H(z) \Big|_{z=e^{sT_d}} = \sum_{k=-\infty}^{\infty} H_c(s - j\frac{2\pi}{T_d}k)$$

$$H_c(s) = \sum_{k=1}^N \frac{A_k}{s - s_k}$$

Then

$$h_c(t) = \sum_{k=1}^N A_k e^{s_k t}, \quad t \geq 0$$

$$h[n] = T_d h_c(nT_d) = T_d \sum_{k=1}^N A_k e^{s_k T_d n}, \quad n \geq 0$$

$$H(z) = \sum_{k=1}^N \frac{T_d A_k}{1 - e^{s_k T_d} z^{-1}}$$

Pole of $H_c(s)$ at s_k gets transformed to a pole of $H(z)$ at $e^{s_k T_d}$

Stability is preserved

Problem: How to account for aliasing

Similar approach can be used for step invariance or other waveform invariance criteria

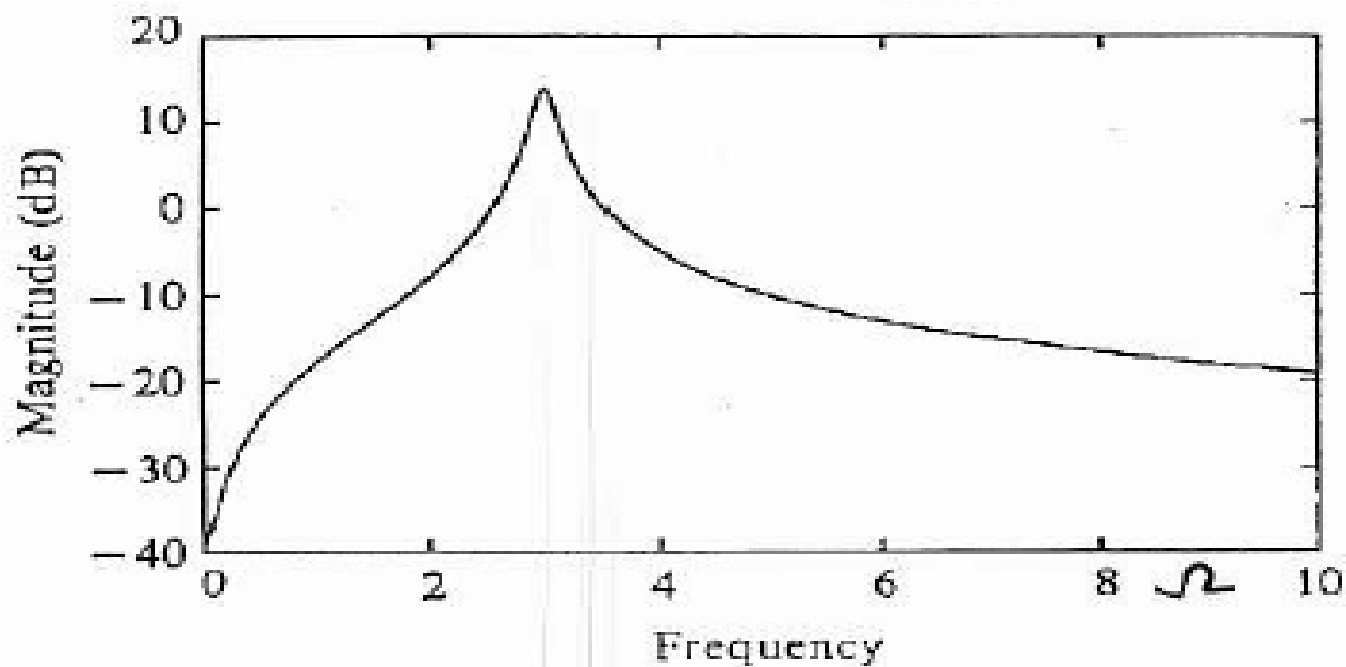
Impulse Variance Design (Cont.)

Example

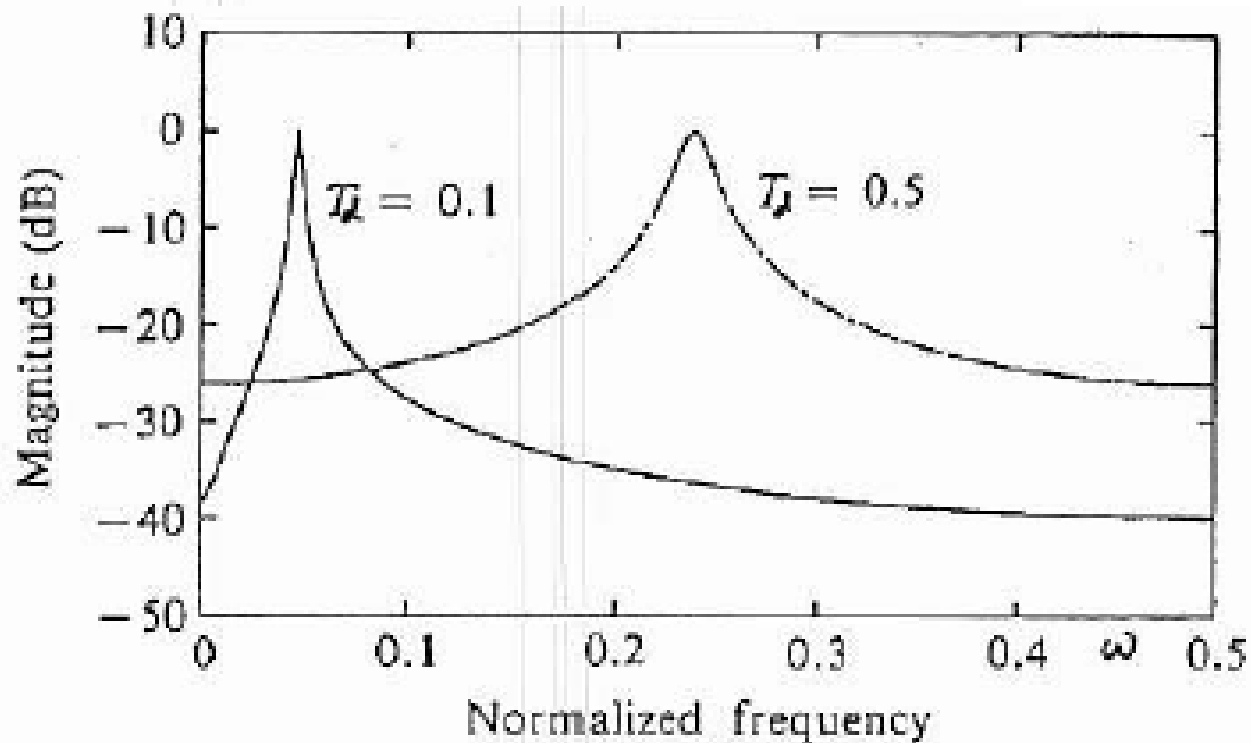
$$H_c(s) = \frac{s + .1}{(s + .1)^2 + 9} = \frac{.5}{s + .1 - j3} + \frac{.5}{s + .1 + j3}$$

$$H(z) = \frac{.5T_d}{1 - e^{-.1T_d} e^{j3T_d} z^{-1}} + \frac{.5T_d}{1 - e^{-.1T_d} e^{-j3T_d} z^{-1}}$$

Analog
Filter



Impulse Invariance
Design



Bi-Linear Transformation

$$s = \frac{2}{T_d} \frac{1 - z^{-1}}{1 + z^{-1}}$$

The digital filter is obtained by a change of variables

$$H(z) = H_c \left[\frac{2}{T_d} \frac{1 - z^{-1}}{1 + z^{-1}} \right]$$

$$z = \frac{1 + \frac{T_d}{2}s}{1 - \frac{T_d}{2}s} = \frac{1 + \frac{\sigma T_d}{2} + j \frac{\Omega T_d}{2}}{1 - \frac{\sigma T_d}{2} - j \frac{\Omega T_d}{2}} \quad \text{or} \quad |z|^2 = \frac{\left(1 + \frac{\sigma T_d}{2}\right)^2 + \frac{\Omega^2 T_d^2}{4}}{\left(1 - \frac{\sigma T_d}{2}\right)^2 + \frac{\Omega^2 T_d^2}{4}}$$

Bilinear Transformation

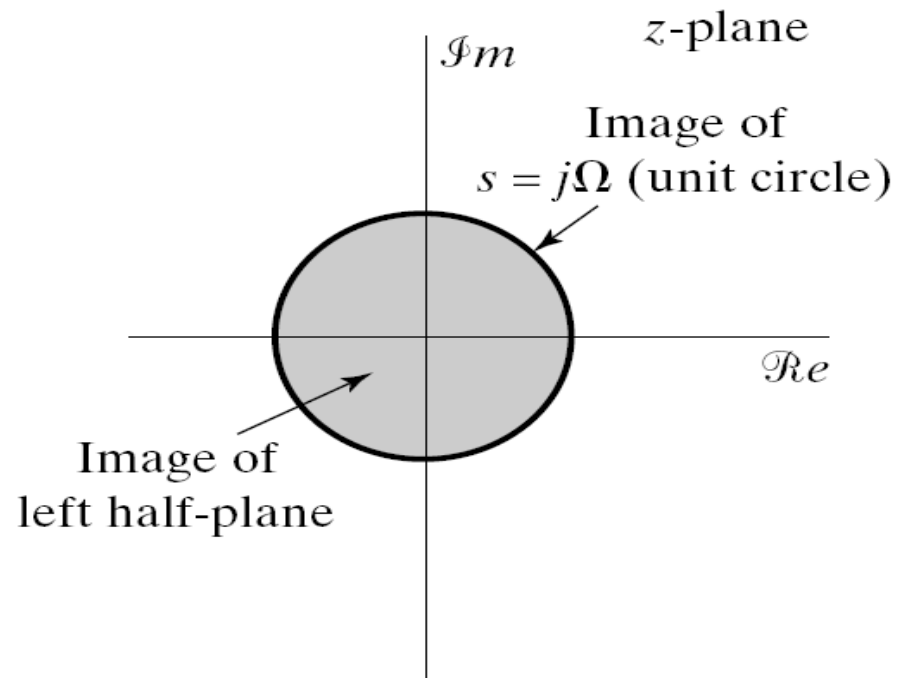
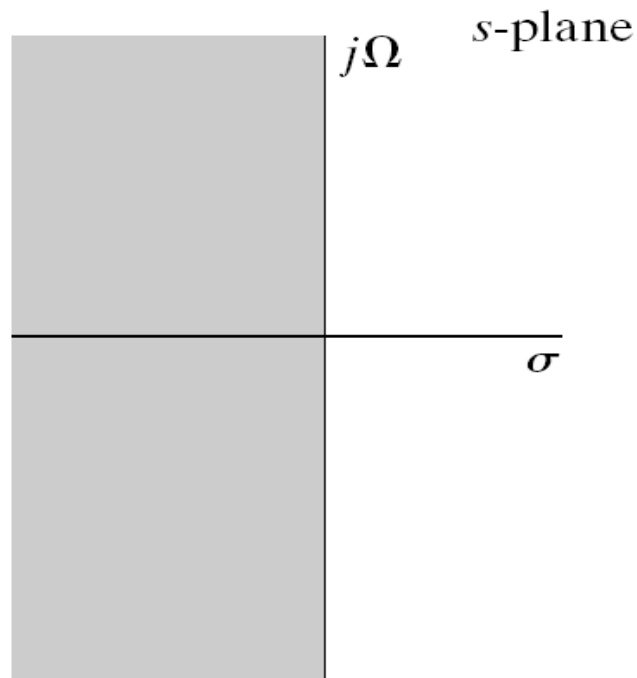
Properties of Mapping

$$\sigma < 0 \rightarrow |z| < 1$$

$$\sigma = 0 \rightarrow |z| = 1$$

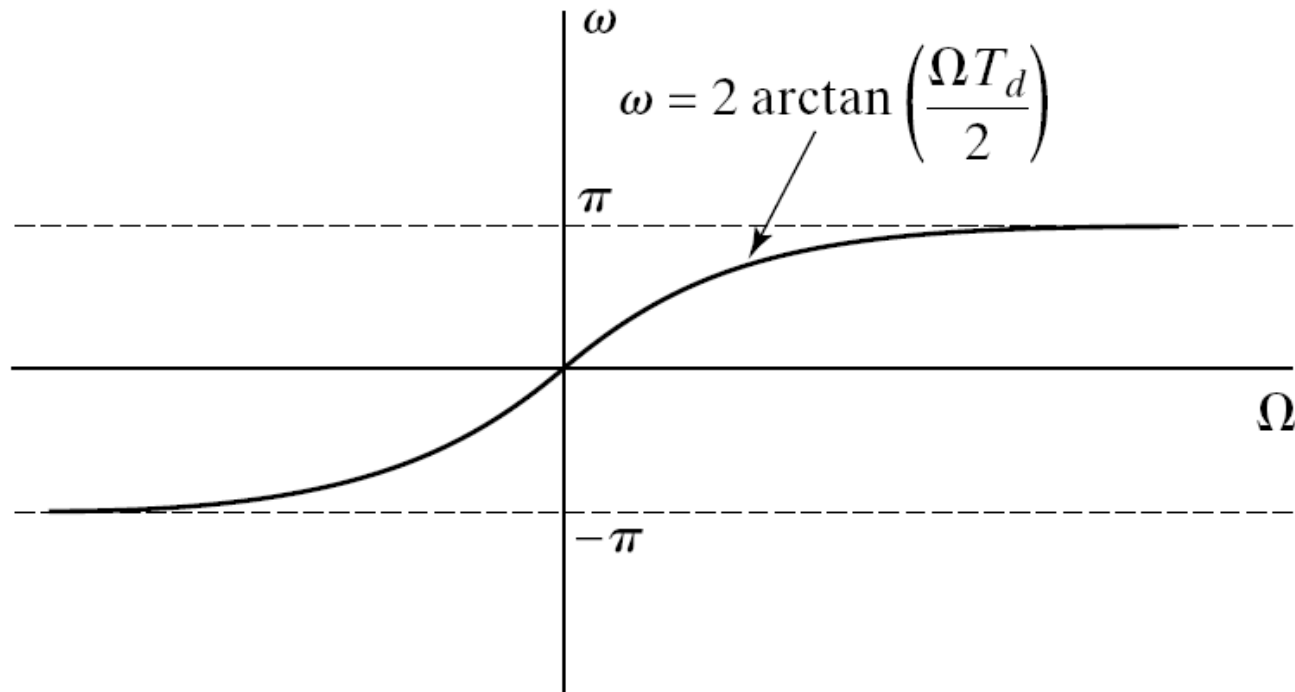
$$\sigma > 0 \rightarrow |z| > 1$$

$$|z|^2 = \frac{\left(1 + \frac{\sigma T_d}{2}\right)^2 + \frac{\Omega^2 T_d^2}{4}}{\left(1 - \frac{\sigma T_d}{2}\right)^2 + \frac{\Omega^2 T_d^2}{4}}$$

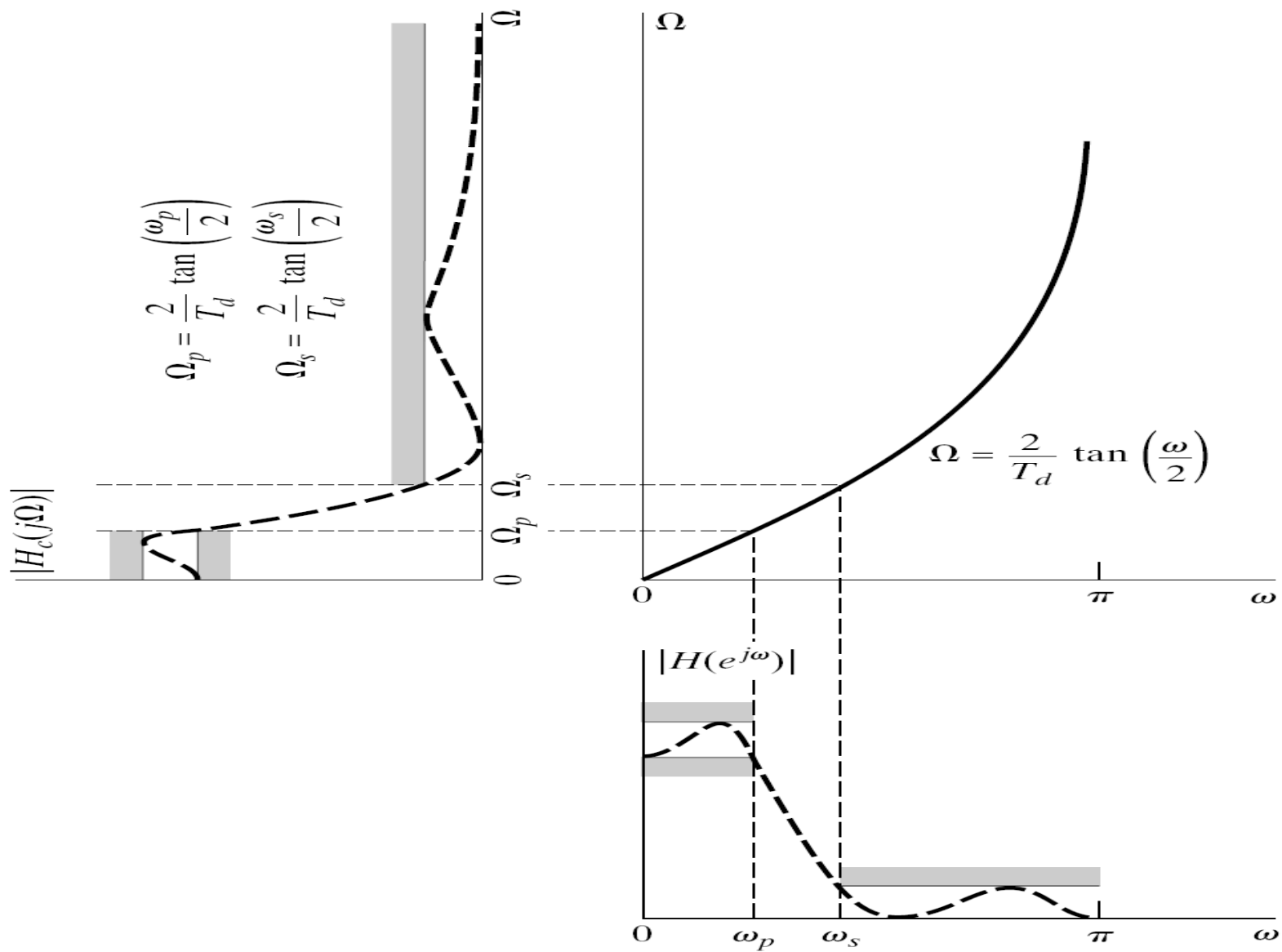


Mapping and its Role in Filter Design

$$e^{j\omega} = \frac{1 + j\frac{\Omega T_d}{2}}{1 - j\frac{\Omega T_d}{2}} \quad \text{or} \quad \Omega = \frac{2}{T_d} \tan\left(\frac{\omega}{2}\right) \quad \text{or} \quad \omega = 2 \arctan\left(\frac{\Omega T_d}{2}\right)$$



Mapping between the frequency ω and Ω resulting from the Bilinear transformation



Frequency warping inherent in the bi-linear transformation of a continuous-time lowpass filter into a discrete-time lowpass filter.

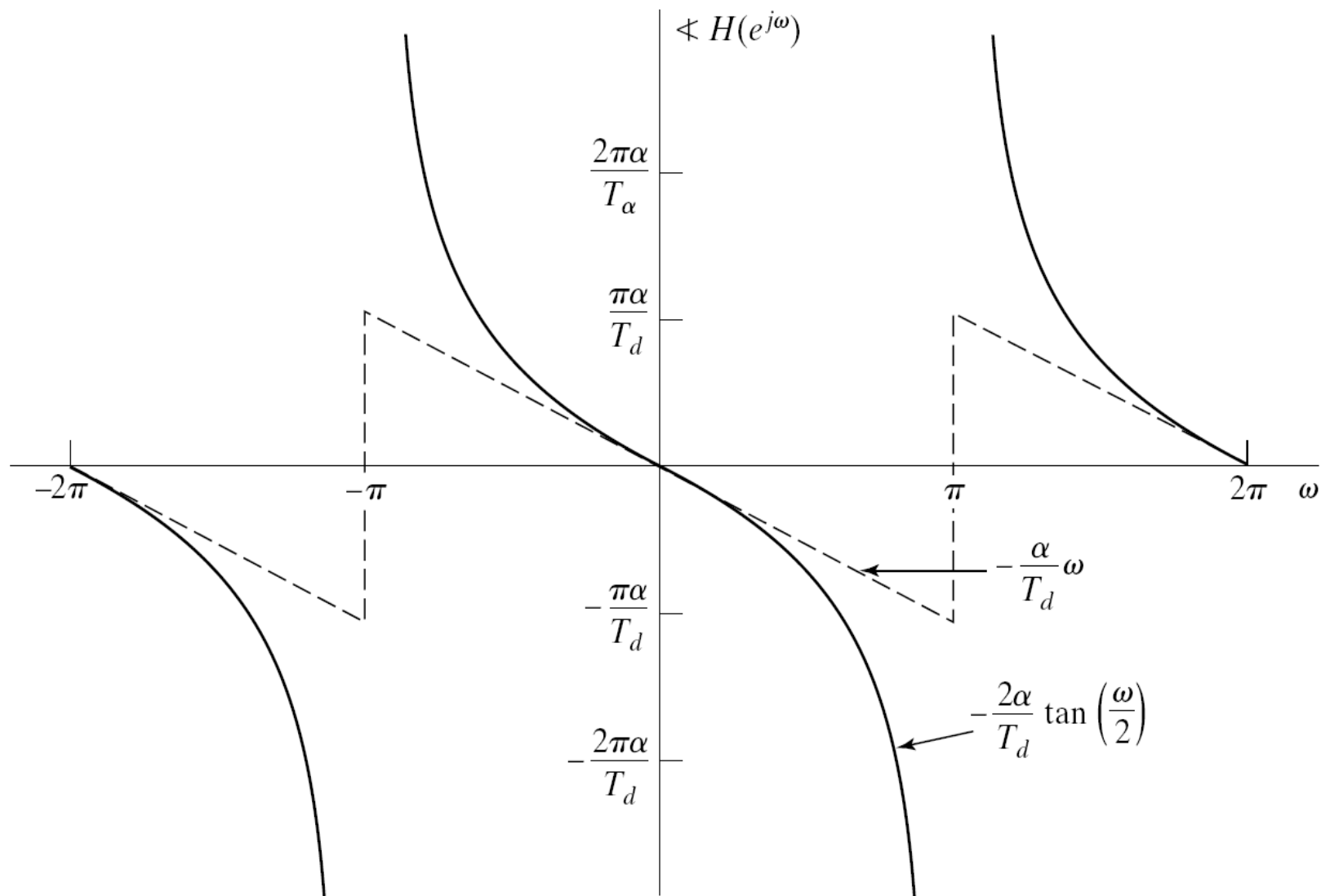


Illustration of the effect of the bi-linear transformation on a linear phase characteristic

Overall Design Steps

Given $\delta_1, \delta_2, \omega_p, \omega_s$

1. Pre-warp to get $\delta_1, \delta_2, \Omega_p, \Omega_s$

where $\Omega_i = \frac{2}{T_d} \tan\left(\frac{\omega_i}{2}\right)$

2. Using these specs design Analog Filter $H_c(j\Omega)$ or $H_c(s)$

3. $H(z) = H_c \left[\frac{2}{T_d} \frac{1 - z^{-1}}{1 + z^{-1}} \right]$

