

SIO 209: Signal Processing for Ocean Sciences

Class 10

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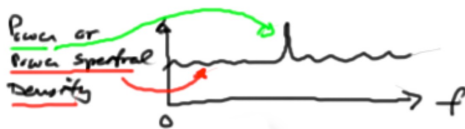
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Recovering Sinusoid Power from FFT



Power of Sinusoid: $A^2/2$

$$\begin{aligned} x[n] &= A \sin \omega n \\ &= A \left\{ \frac{e^{j\omega n} - e^{-j\omega n}}{2j} \right\} \end{aligned}$$

$$\text{let } \omega = \frac{2\pi}{M} k'$$

M length of FFT

$$\begin{aligned} X(k) &= \sum_{n=0}^{M-1} w[n] x[n] e^{-j \frac{2\pi}{M} nk} \\ &= A \sum_{n=0}^{M-1} w[n] \left\{ \frac{e^{j \frac{2\pi}{M} nk'} - e^{-j \frac{2\pi}{M} nk'}}{2j} \right\} e^{-j \frac{2\pi}{M} nk} \\ &= \frac{A}{2j} \sum_{n=0}^{M-1} w[n] \quad \text{for } k = k' \text{ (sinusoid is bin centered)} \end{aligned}$$

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Recovering Sinusoid Power from FFT

Amplitude and Power of Bin-Centered Sinusoid:

$$X(k) = \frac{A}{2j} \sum_{n=0}^{M-1} w[n] \quad k = k'$$

$$\text{sinusoid amplitude} \quad A = \frac{2}{\sum_{n=0}^{M-1} w[n]} |X(k')|$$

True regardless of the phase of sinusoid, since we use $|X(k')|$ which is insensitive to phase

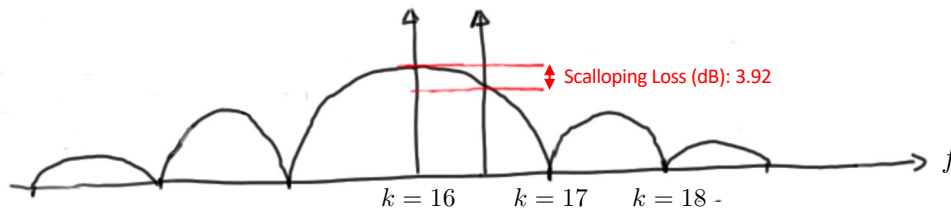
$$\text{sinusoid power} \quad \frac{A^2}{2} = \frac{\boxed{2}}{\left(\sum_{n=0}^{M-1} w[n]\right)^2} |X(k')|^2$$

Positive and negative frequency power

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Scalloping Loss (dB)

Example: Rectangular Window



- Comparison between sinusoid at bin center (e.g., $k = 16$) vs. bin boundary (e.g., halfway between $k = 16$ and $k = 17$)
- Scalloping loss of Kaiser-Bessel window is 1.2 dB for $\alpha = 2.5$ ($\beta = \pi\alpha = 7.85$)

F. J. Harris, "On the use of windows for harmonic analysis with the discrete Fourier transform." Proc. IEEE, 1978

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Recovering Sinusoid Power from PSD Estimate

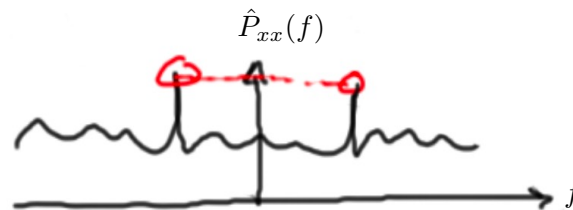
- Recall 2-sided power spectral density (power per unit frequency)

$$\hat{P}_{xx}(f_k) = \frac{1}{MUf_s} \overline{|X(k)|^2}$$

averaged over potentially
overlapping and windowed records
in "Power"/Hz

$$U = \frac{1}{M} \sum_{n=0}^{M-1} w^2[n]$$

normalization constant



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Recovering Sinusoid Power from PSD Estimate

Sinusoid Power

$$\frac{A^2}{2} = \frac{2}{\left(\sum_{n=0}^{M-1} w[n]\right)^2} \overline{|X(k)|^2}$$

$$= \frac{2 f_s MU \hat{P}_{xx}(f_k)}{\left(\sum_{n=0}^{M-1} w[n]\right)^2}$$

Positive and negative
frequency power

$$= \frac{2 f_s}{M} \frac{M \sum_{n=0}^{M-1} w^2[n]}{\left(\sum_{n=0}^{M-1} w[n]\right)^2} \hat{P}_{xx}(f_k)$$

Equivalent noise
bandwidth (see table
in F. Harris, 1978)

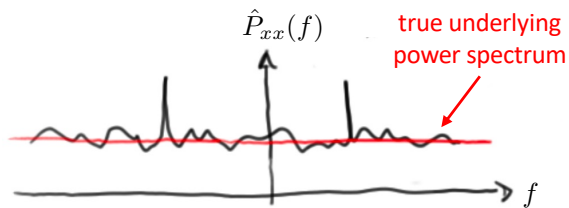
2-sided power
spectral density

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Confidence Intervals

- 90 % confidence intervals for conventional power spectral density estimation



| | Number of Averages K | | | | |
|------------------|------------------------|------|------|------|------|
| | 4 | 8 | 16 | 32 | 64 |
| Upper Limit (dB) | +4.7 | +3.0 | +2.0 | +1.4 | +1.0 |
| Lower Limit (dB) | -2.9 | -2.2 | -1.6 | -1.2 | -0.8 |

- Note that each sample of the power spectral density, follows a Chi-squared distribution with $2K$ degrees of freedom