

SIO 207A: Fundamentals of Digital Signal Processing

Class 4

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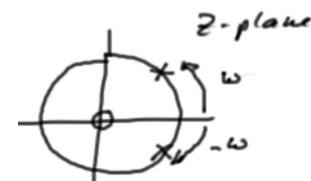
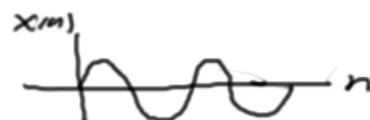
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Z-Transform of Sinusoidal Sequence

- We consider the sinusoidal sequence $x[n] = (\sin \omega n)u[n]$; note that

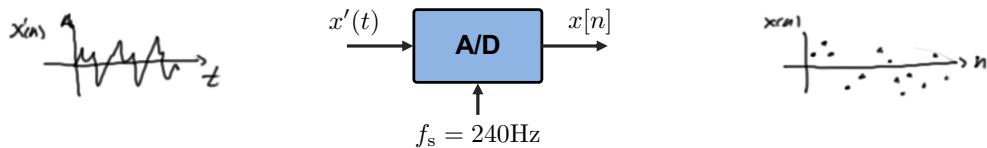
$$\sin \omega n = (e^{j\omega n} - e^{-j\omega n})/2j$$

$$\begin{aligned} X(z) &= \sum_{n=0}^{\infty} \sin[\omega n] z^{-n} \\ &= \sum_{n=0}^{\infty} \left(\frac{e^{j\omega n} - e^{-j\omega n}}{2j} \right) z^{-n} \\ &= \frac{1}{2j} \sum_{n=0}^{\infty} e^{j\omega n} z^{-n} - \frac{1}{2j} \sum_{n=0}^{\infty} e^{-j\omega n} z^{-n} \\ &= \frac{1}{2j} \left(\frac{z}{z - e^{j\omega}} \right) - \frac{1}{2j} \left(\frac{z}{z - e^{-j\omega}} \right) \\ &= \frac{z \sin \omega}{(z - e^{j\omega})(z - e^{-j\omega})} \end{aligned}$$

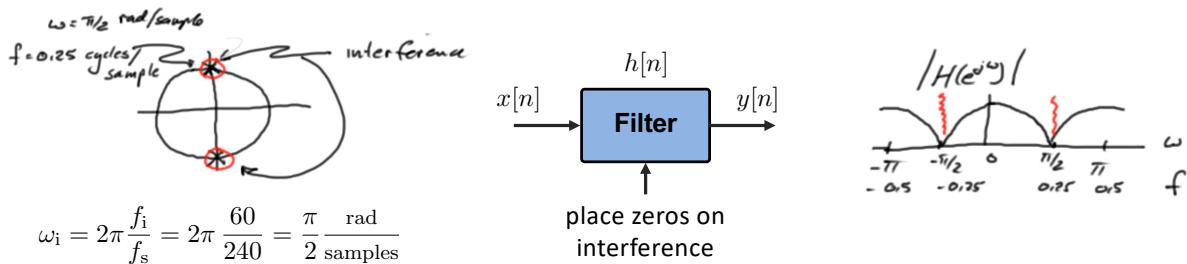


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Interference Cancellation



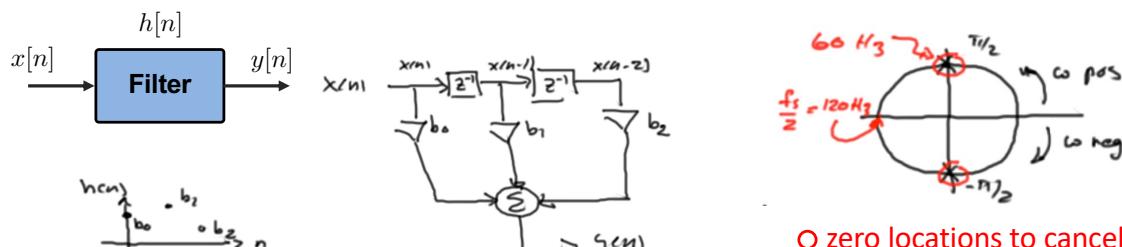
- Problem: $x'(t)$ is contaminated with $f_i = 60\text{Hz}$ interference
- For $f_s = 240\text{Hz}$, 60Hz interference is sampled 4 times per cycle, i.e., $f = 0.25 \frac{\text{cycles}}{\text{samples}}$



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Frequency Domain Representation



O zero locations to cancel interference poles

$$y[n] = b_0x[n] + b_1x[n-1] + b_2x[n-2]$$

$$\begin{aligned} Y(z) &= b_0X(z) + b_1z^{-1}X(z) + b_2z^{-2}X(z) \\ &= (b_0 + b_1z^{-1} + b_2z^{-2})X(z) \end{aligned}$$

$$z_1, z_2 = \frac{-b_1 \pm \sqrt{b_1^2 - 4b_0b_2}}{2b_0}$$

$$H(z) = \frac{Y(z)}{X(z)} = b_0 + b_1z^{-1} + b_2z^{-2} = \frac{b_0z^2 + b_1z + b_2}{z^2} = \frac{(z - z_1)(z - z_2)}{z^2}$$

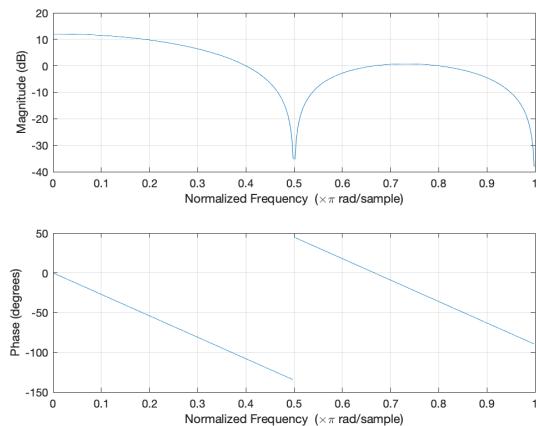
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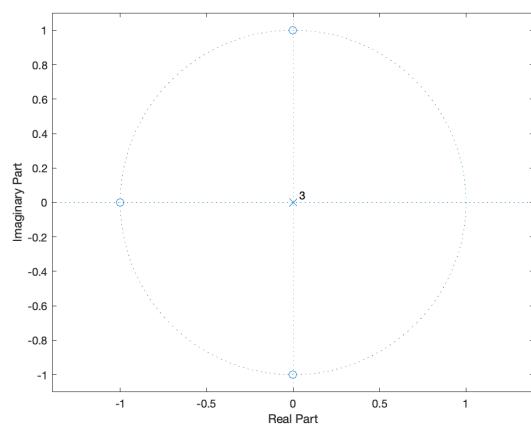
Z-Transform Examples

$$H(z) = \frac{z^3 + z^2 + z + 1}{z^3}$$

`freqz([1, 1, 1, 1])`



`zplane([1, 1, 1, 1])`

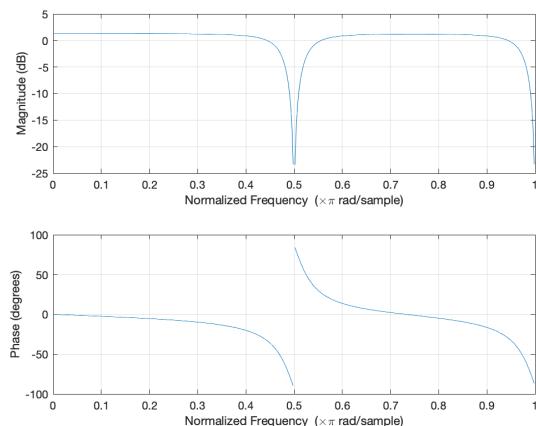


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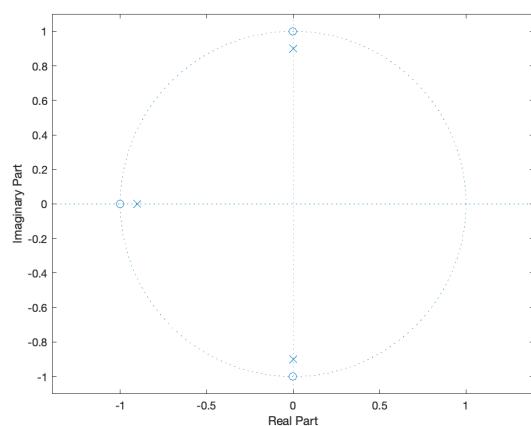
Z-Transform Examples

$$H(z) = \frac{z^3 + z^2 + z + 1}{z^3 + 0.9z^2 + 0.81z + 0.729}$$

`freqz([1, 1, 1, 1], [1, 0.9, 0.81, 0.729])`



`zplane([1, 1, 1, 1], [1, 0.9, 0.81, 0.729])`



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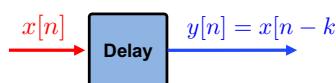
Z-Transform Pairs and Properties

- **A:** Linearity

$$x[n] = a_1x_1[n] + a_2x_2[n]$$

$$X(z) = \sum_{n=-\infty}^{\infty} \{a_1x_1[n] + a_2x_2[n]\} z^{-n}$$

- **B:** Shifting
(delay/advance)



$$\begin{aligned} Y(z) &= \sum_{n=-\infty}^{\infty} y[n]z^{-n} = \sum_{n=-\infty}^{\infty} x[n-k]z^{-n} \\ &= \sum_{l=-\infty}^{\infty} x[l]z^{-(l+k)} = z^{-k} \sum_{l=-\infty}^{\infty} x[l]z^{-l} \\ &= z^{-k} X(z) \end{aligned}$$

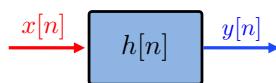
let $l = n - k$
 $\rightarrow n = l + k$

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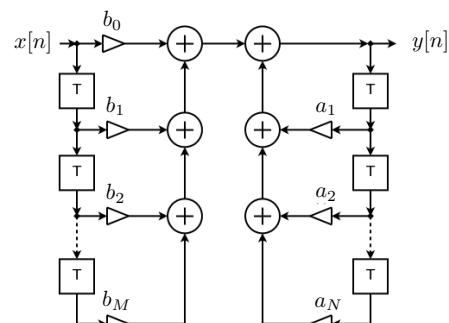
System Input/Output Description

- **C:** System Input/Output Description



$$\begin{aligned} Y(z) &= \sum_{k=1}^N a_k z^{-k} Y(z) + \sum_{r=0}^M b_r z^{-r} X(z) \\ &= H(z)X(z) \end{aligned}$$

$$\begin{aligned} H(z) &= \frac{Y(z)}{X(z)} = \frac{\sum_{r=0}^M b_r z^{-r}}{1 - \sum_{k=1}^N a_k z^{-k}} \\ &= \frac{b_0 z^M + b_1 z^{M-1} + \dots + b_M}{z^N + a_1 z^{N-1} + \dots + a_N} \end{aligned}$$



$$\begin{aligned} y[n] &= \sum_{k=1}^N a_k y[n-k] + \sum_{r=0}^M b_r x[n-r] \\ &= \sum_{k=-\infty}^{\infty} h[k] x[n-k] \quad \text{convolution summation} \end{aligned}$$

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Multiplication

- **D:** Multiplication

$$y[n] = x_1[n]x_2[n]$$

$$Y(z) = \sum_{n=-\infty}^{\infty} x_1[n]x_2[n]z^{-n}$$

- We restrict our interest to the unit circle, i.e., $z = e^{j\omega}$ and assume a Fourier transform

$$\begin{aligned} Y(e^{j\omega}) &= \sum_{n=-\infty}^{\infty} x_1[n]x_2[n]e^{-j\omega n} \\ &= \sum_{n=-\infty}^{\infty} x_1[n] \left\{ \frac{1}{2\pi} \int_{-\pi}^{\pi} X_2(e^{j\omega'}) e^{j\omega' n} d\omega' \right\} e^{-j\omega n} \\ &= \frac{1}{2\pi} \int_{-\pi}^{\pi} X_2(e^{j\omega'}) \sum_{n=-\infty}^{\infty} x_1[n] e^{-j(\omega-\omega')n} d\omega' \\ &= \frac{1}{2\pi} \int_{-\pi}^{\pi} X_2(e^{j\omega'}) X_1(e^{j(\omega-\omega')}) d\omega' \end{aligned}$$

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Multiplication

- Multiplication in time domain is convolution in frequency domain, i.e.,

$$Y(e^{j\omega}) = \frac{1}{2\pi} \int_{-\pi}^{\pi} X_2(e^{j\omega'}) X_1(e^{j(\omega-\omega')}) d\omega'$$

- Note:

1. This is a periodic or circular convolution (not linear)

$$Y(e^{j\omega}) = X_1(e^{j\omega}) \oplus X_2(e^{j\omega})$$

2. Think of circular convolution as a divided cylinder with Fourier transforms pointed on them

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Multiplication

- E. Multiplication by a^n

$$x_1[n] = a^n x[n]$$

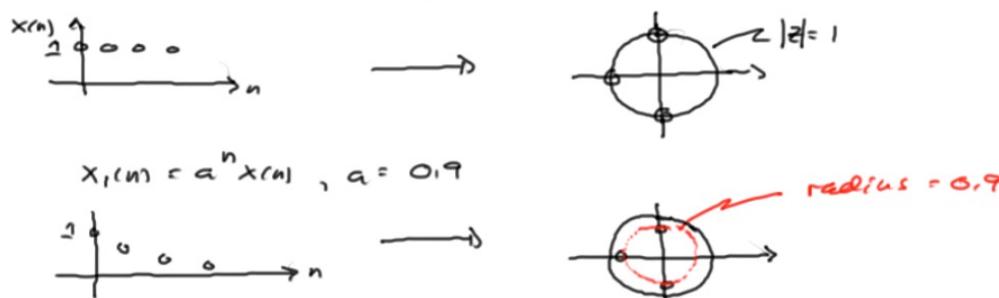
$$\begin{aligned} X_1(z) &= \sum_{n=-\infty}^{\infty} (a^n x[n]) z^{-n} \\ &= \sum_{n=-\infty}^{\infty} x[n] (a^{-1}z)^{-n} \quad \text{let } z' = a^{-1}z \\ &= \sum_{n=-\infty}^{\infty} x[n] z'^{-n} \\ &= X(z') \end{aligned}$$

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Examples

- Let a be real and positive with $a \leq 1$
- This has the effect of drawing the roots inward on radial paths

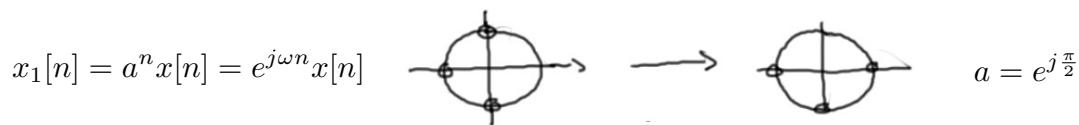


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Examples

- Let a be complex and on the unit circle, i.e., $a = e^{j\omega_c}$
- This has the effect of rotating the original z-transform



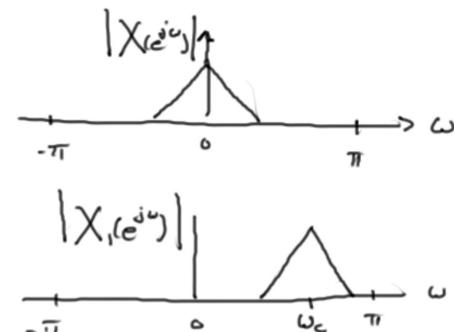
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Modulation

$x[n]$ is a low pass audio process

$$x_1[n] = a^n x[n] \quad \text{where} \quad a = e^{j\omega_c}$$



amplitude modulation $x_1[n] = \cos(\omega_c n) x[n]$



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