

Semi-Blind Decorrelating Multiuser Detection for Synchronous CDMA

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Introduction

- CDMA techniques have attracted attention for their spectral efficiency, interference resistance and flexible traffic control.
- MAI is the dominant impairment for CDMA systems and exists even with perfect power control.
- Multiuser detection promising approach to mitigate MAI: classic/non-blind multiuser detection and blind multiuser detection
- With exploring a new data model and the trade-off between classic multiuser detection and blind multiuser detection, we propose semi-blind decorrelating multiuser detection.
 - ★ Least squares semi-blind decorrelating multiuser detection (LS-DD)
 - ★ Total least squares semi-blind decorrelating multiuser detection (TLS-DD)
 - ★ Mixed LS/TLS semi-blind decorrelating multiuser detection (MLS-DD)

Data Model

- Consider a synchronous CDMA system with L chips per bit ($T_b = LT_c$).
- Received signal, after chip-matched filter followed by chip-rate sampling:

$$\mathbf{r} = \sum_{k=1}^K A_k b_k \mathbf{s}_k + \mathbf{n} = \mathbf{S} \mathbf{A} \mathbf{b} + \mathbf{n}$$

where $\mathbf{A} = \text{diag} \{A_1 \ A_2 \ \dots \ A_K\}$, $\mathbf{S} = [\mathbf{s}_1 \ \mathbf{s}_2 \ \dots \ \mathbf{s}_K]$ and $\mathbf{b} = [b_1 \ b_2 \ \dots \ b_K]^T$ with $b_i \in \{+1, -1\}$.

- Linear multiuser detectors can be written in the form

$$\mathbf{b} = \text{sgn} \{ \mathbf{W} \mathbf{r} \}$$

or for the k th user $b_k = \text{sgn} \{ \mathbf{w}_k^T \mathbf{r} \}$ where \mathbf{w}_k^T is the k th row of \mathbf{W} .

- We consider detection of user 1's data only.
- Maintain the restriction that $L > K$.

New Data Model w.r.t. User 1 – I

- In terms of user 1, a new semi-blind signature matrix \mathcal{S} is constructed by

$$\begin{aligned}\mathcal{S} &= [A_1 \mathbf{s}_1 \quad \mathbf{r}_1 \quad \mathbf{r}_2 \quad \dots \quad \mathbf{r}_{M-1}] \\ &= \mathbf{S} \mathbf{A} [\mathbf{e}_1 \quad \mathbf{D}] + \mathbf{N} \\ &= \mathbf{S} \mathbf{A} \mathbf{B} + \mathbf{N}\end{aligned}$$

where $\mathbf{B} = \begin{bmatrix} 1 & \bar{\mathbf{d}}^T \\ \mathbf{0} & \tilde{\mathbf{D}} \end{bmatrix} = \begin{bmatrix} \mathbf{c}^T & \\ \mathbf{0} & \tilde{\mathbf{D}} \end{bmatrix}$.

- The relationship between \mathbf{r} and \mathcal{S} is

$$\mathbf{r} = \mathcal{S} \mathbf{d} + \mathbf{z}$$

where \mathbf{d} is named the detection vector.

$$\mathbf{d} = \mathbf{B}^+ \mathbf{b} = \begin{bmatrix} 1 & \bar{\mathbf{d}}^T \\ \mathbf{0} & \tilde{\mathbf{D}} \end{bmatrix}^+ \begin{bmatrix} b_1 \\ \tilde{\mathbf{b}} \end{bmatrix}$$

New Data Model w.r.t. User 1 – II

- \mathbf{N} is a noise matrix of the form, $\mathbf{N} = [\mathbf{0} \ \tilde{\mathbf{N}}]$. \mathbf{z} is the new noise vector defined as $\mathbf{z} = \mathbf{n} - \mathbf{N}\mathbf{B}^+\mathbf{b}$.
- The bit sent by the first user, b_1 , during the time interval $t \in [(n-1)T, nT]$ can be estimated as

$$b_1 = \mathbf{c}^T \mathbf{d}$$

where $\mathbf{c} = [1 \ \bar{\mathbf{d}}^T]^T$ and $\mathbf{d} = [d_1 \ \tilde{\mathbf{d}}]^T$.

- Now, the semi-blind multiuser detection problem becomes how to estimate \mathbf{d} .

Least-Square Semi-blind Multiuser Detection (LS-DD)

- The LS estimation problem is

$$\mathbf{d}_{\text{LS}} = \min_{\mathbf{x}} \|\mathbf{S}\mathbf{x} - \mathbf{r}\|_2$$

- The LS solution is

$$\mathbf{d}_{\text{LS}} = \mathbf{S}^+ \mathbf{r}$$

and the bit sent by the first user, b_1 , in the n th signaling interval is detected with

$$\begin{aligned} \hat{b}_1^{\text{LS}} &= \text{sgn}\{\mathbf{w}_{\text{LS}}^T \mathbf{r}\} \\ &= \text{sgn}\{b_1 + \mathbf{c}^T \mathbf{S}^+ \tilde{\mathbf{n}}\} . \end{aligned}$$

- In LS estimation, it assumes that \mathbf{S} is error-free. However, this assumption is not entirely accurate according since there is a noise term, \mathbf{N} .

Total Least-Square Semi-blind Multiuser Detection (TLS-DD)

- \mathbf{r} can also be expressed as $\mathbf{r} = \hat{\mathbf{S}}\mathbf{d} + \mathbf{n}$, where $\hat{\mathbf{S}} = \mathbf{S} - \mathbf{N} = \mathbf{SAB}$.
- The LS problem can then be transformed into the following TLS problem:

$$\mathbf{d}_{\text{TLS}} = \min_{\hat{\mathbf{S}}, \mathbf{x}} \|[\mathbf{S} \quad \mathbf{r}] - [\bar{\mathbf{S}} \quad \bar{\mathbf{S}}\mathbf{x}]\|_2 .$$

- The TLS solution is

$$\mathbf{d}_{\text{TLS}} = \left(\mathbf{S}^T \mathbf{S} - \sigma_{K+1}^2 \mathbf{I} \right)^{-1} \mathbf{S}^T \mathbf{r}$$

and the n th bit sent by the first user, b_1 , can be detected with

$$\begin{aligned} \hat{b}_1^{\text{TLS}} &= \text{sgn}\{\mathbf{w}_{\text{TLS}}^T \mathbf{r}\} \\ &= \text{sgn}\left\{ \mathbf{c}^T \left(\mathbf{S}^T \mathbf{S} - \sigma_{K+1}^2 \mathbf{I} \right)^{-1} \mathbf{S}^T \mathbf{r} \right\} . \end{aligned}$$

Mixed LS/TLS Semi-blind Multiuser Detection (MLS-DD) I

- In the LS problem, it assumed the semi-blind signature matrix \mathcal{S} is error-free. Again, this assumption is not completely accurate.
- In the TLS problem, it assumed that in each column of the semi-blind signature matrix, \mathcal{S} , some noise or error exists. This assumption also is not complete.
- Hence, to maximize the estimation accuracy of the detection vector \mathbf{d} , it is natural to require that the corresponding columns of \mathcal{S} be unperturbed since they are known exactly. The estimation problem becomes the following MLS problem.

$$\mathbf{d}_{\text{MLS}} = \min_{\bar{\mathcal{S}}, \mathbf{x}} \| [\tilde{\mathcal{S}} \quad \mathbf{r}] - [\bar{\mathcal{S}} \quad [A_1 \mathbf{s}_1 \quad \bar{\mathcal{S}}] \mathbf{x}] \|_2$$

Mixed LS/TLS Semi-blind Multiuser Detection (MLS-DD) II

- The MLS solution is given by

$$\mathbf{d}_{\text{MLS}} = \left(\mathbf{S}^T \mathbf{S} - \sigma^2 \begin{bmatrix} 0 & \mathbf{0} \\ \mathbf{0} & \mathbf{I}_{M-1} \end{bmatrix} \right)^{-1} \mathbf{S}^T \mathbf{r} .$$

and the bit sent by the first user, b_1 , can be detected with

$$\begin{aligned} \hat{b}_1^{\text{MLS}} &= \text{sgn}\{\mathbf{w}_{\text{MLS}}^T \mathbf{r}\} \\ &= \text{sgn} \left\{ \mathbf{c}^T \left(\mathbf{S}^T \mathbf{S} - \sigma^2 \begin{bmatrix} 0 & \mathbf{0} \\ \mathbf{0} & \mathbf{I}_{M-1} \end{bmatrix} \right)^{-1} \mathbf{S}^T \mathbf{r} \right\} . \end{aligned}$$

Performance Analysis I

- When $\mathbf{N} = \mathbf{0}$, the proposed LS-DD and the classic decorrelating detector share the same signature space and

$$\mathbf{w}_{\text{LS}} = A_k^{-1} \mathbf{w}_{\text{DD}} \text{ .}$$

Hence, they share the same performance for user 1, too.

- It is straightforward to develop the following relationship between the estimation errors of b_1 and \mathbf{d} :

$$\Delta b_1 \leq \|\Delta \mathbf{d}\|_1$$

where $\Delta b_1 = \hat{b}_1 - b_1$, $\Delta \mathbf{d} = \hat{\mathbf{d}} - \mathbf{d}$ and $\|\mathbf{m}\|_1$ denotes the 1-norm of the vector \mathbf{m} .

Performance Analysis II

- The mean of the semi-blind noise item \mathbf{z} is

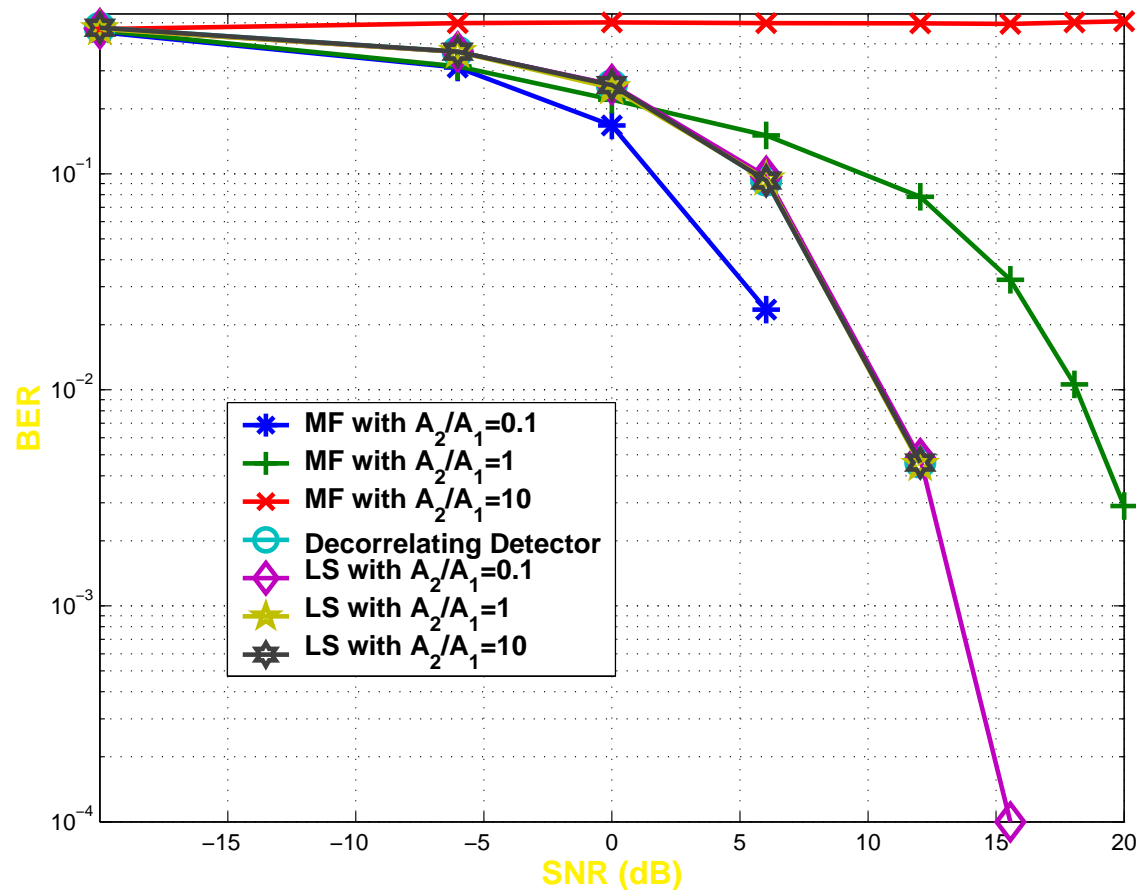
$$\mu = E\{\mathbf{z}\} = 0 .$$

- The variance of the semi-blind noise item \mathbf{z} satisfies the following inequality

$$\begin{aligned} \max\{var\{\mathbf{z}\}\} &= \max\{E\{(\mathbf{z} - \mu)^2\}\} \\ &\leq \sigma^2 + (K - 1)\|\tilde{\mathbf{D}}^+\|_2^2\sigma_N^2 \end{aligned}$$

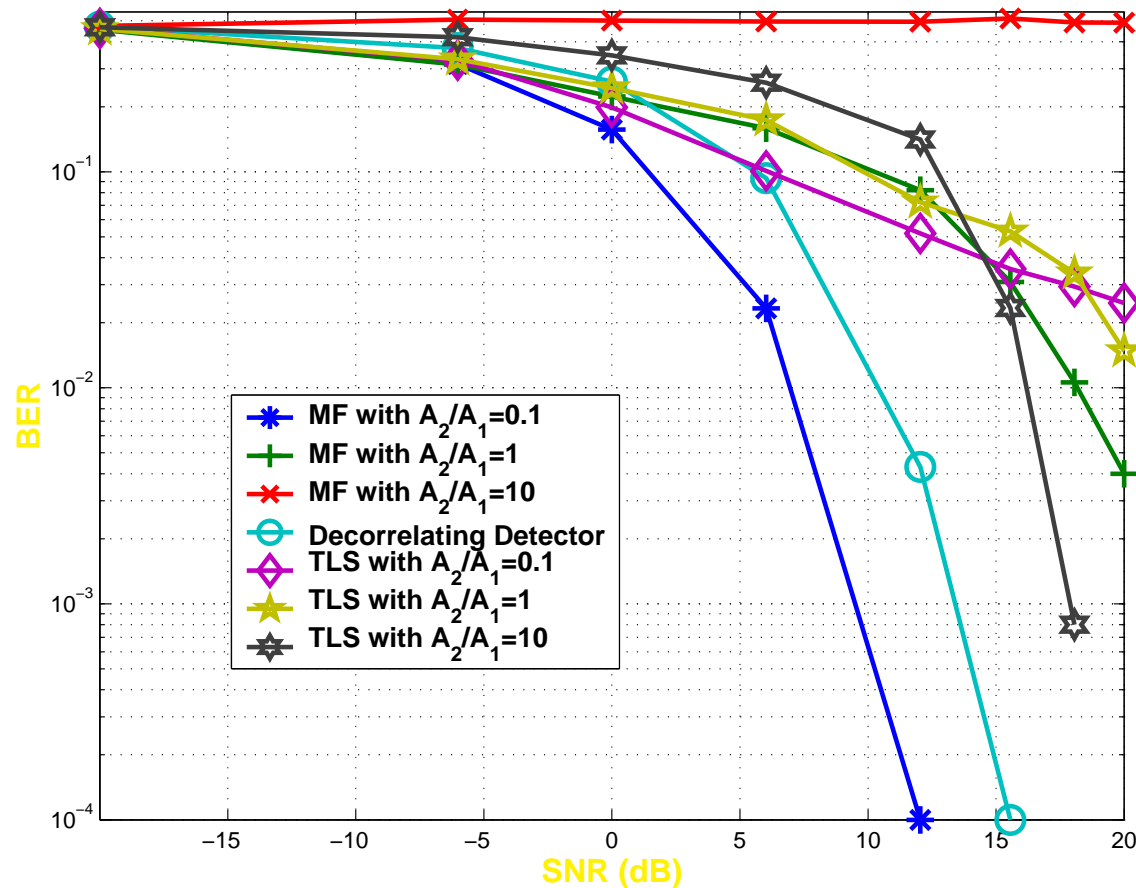
where $\max\{\mathbf{m}\}$ denotes the maximum item in the vector \mathbf{m} and σ_N^2 is the power of the noise item \mathbf{N} in the semi-blind signature matrix \mathbf{S} .

Computer Simulations I



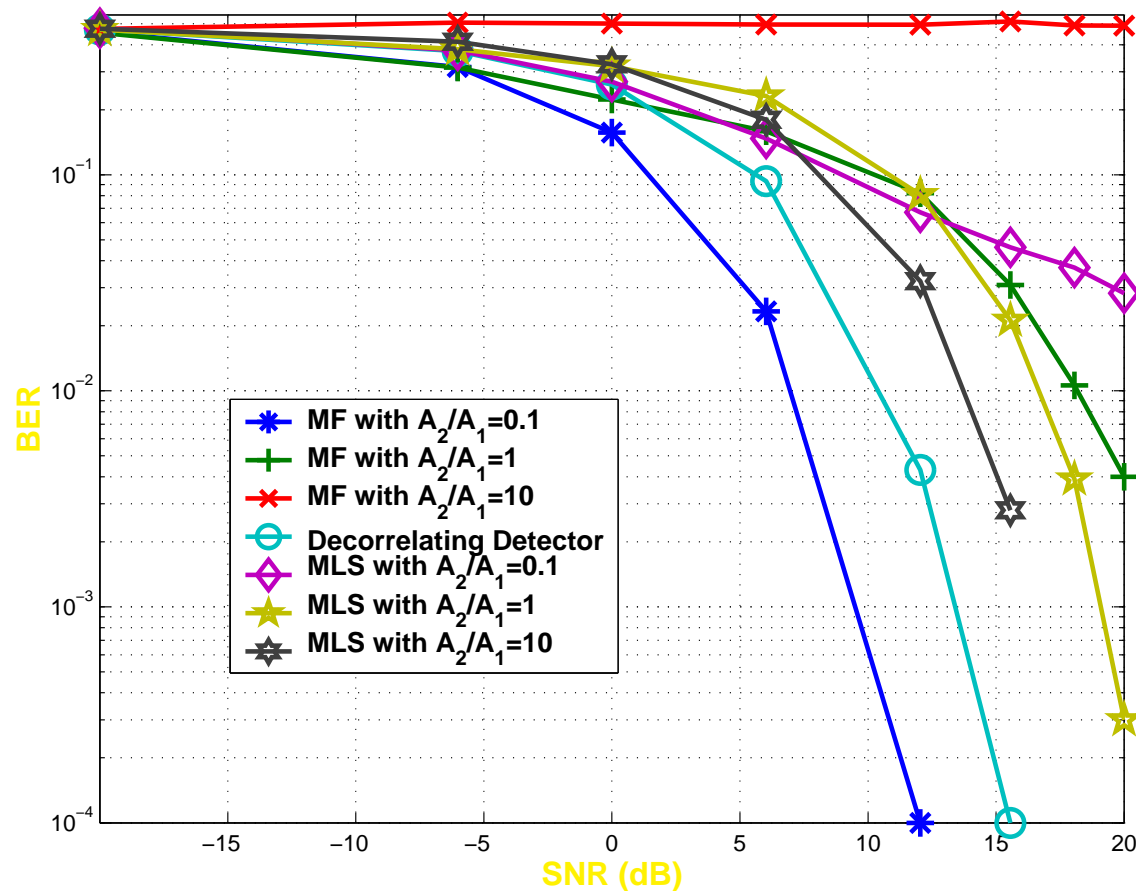
BER comparison of the single-user matched filter, decorrelating detector, and the LS semi-blind detector for the first user in a two user system with $\rho = 0.75$.

Computer Simulations II



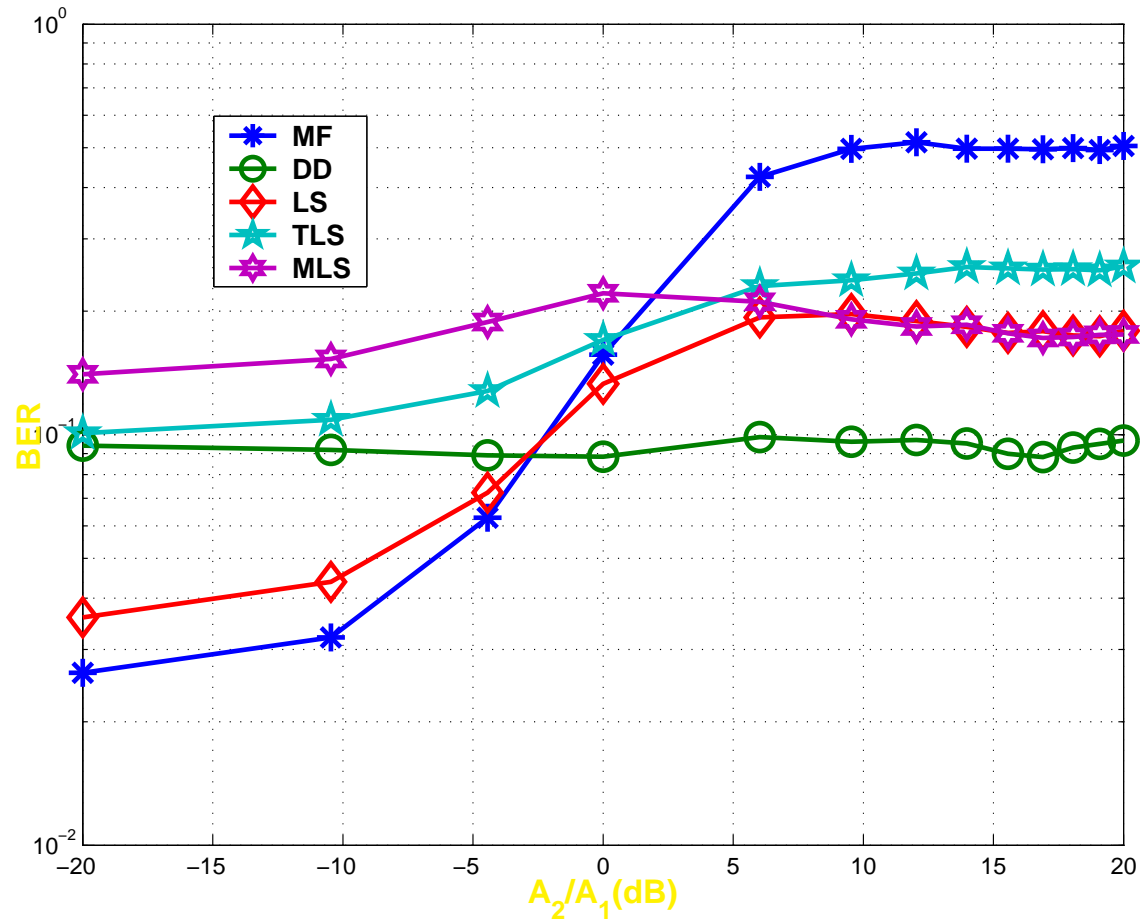
BER comparison of the single-user matched filter, decorrelating detector, and the TLS semi-blind detector for the first user in a two user system with $\rho = 0.75$.

Computer Simulations III



BER comparison of the single-user matched filter, decorrelating detector, and the MLS semi-blind detector for the first user in a two user system with $\rho = 0.75$.

Computer Simulations IV



Near-far resistance comparison of the single-user matched filter, decorrelating detector, and the LS, TLS and MLS semi-blind detectors for the first user in a two user system with $\rho = 0.75$ and $SNR = 6dB$.

Conclusion & Future Directions

- Multiuser precoding provides another alternative to solve the near-far problem in multiuser systems.
- With linear system theory, it can be shown that many linear multiuser precoders and detectors can achieve the same performance.
- With considering the input information bits, nonlinear multiuser precoding can be developed to further enhance system performance.
- Additional performance enhancement may be achievable by jointly optimizing transmitter multiuser precoding and adaptive receivers/transmitters.
- For unknown or time-variable channels, adaptive multiuser precoding with feedback from receivers and multiuser precoding with channel coding can be interesting topics.